EL 6 Output Pentode

This is another 18 W, indirectly-heated, high conductance output pentode, the need for which arose from a demand for a "larger" output valve which, fully excited, would take about the same grid input as the EL 3. The advantage of this valve is that receivers having a 9 W or 18 W output stage, apart from the rectifier, may be developed along exactly the same lines. At the working point the EL 6 has the unusually high mutual conductance of 14.5 mA/V. With 10 % distortion the maximum obtainable output is 8 W. The peak alternating grid voltage for this output is only 4.8 V_(eff) whilst the sensitivity (for 50 mW output) is 0.3 V_(eff).

The valve can also figure in balanced output stages, although the output obtainable is then not so high as in the case of two EL 5 type valves. On the other hand, the EL 6 has the advantage of a higher mutual conductance. The optimum output power is 14.5 W with 2.2 % distortion at an alternating grid voltage of 7.3 $V_{(eff)}$ per grid. Taking into account an average voltage drop of 15 V across the output transformer, the output at $V_u = 250$ V with $V_{g2} = 265$ V is somewhat higher, viz. 16 W, with 1.4 % distortion with a grid g2 input of 8.5 $V_{(eff)}$. The maximum distortion is roughly 3 %, which moccurs at approximately 10 W output.

The very high mutual conductance is due to the special construction of the cathode, with its relatively low heater power: at 6.3 V the current consumed is 1.2 A.

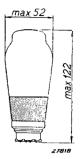


Fig. 1 Dimensions in mm.

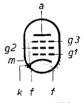




Fig. 2
Arrangement of electrodes and base connections.

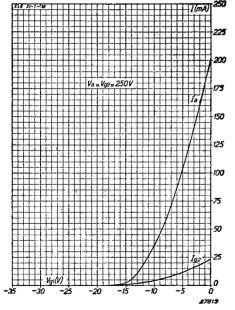


Fig. 3 Anode and screen current as a function of the grid bias, at $Va = Vg_1 = 250$ V.

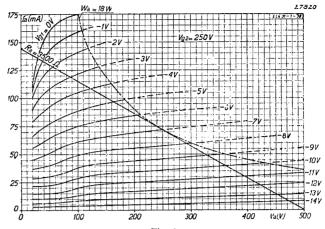


Fig. 4 Anode current as a function of the anode voltage at $\,Vg_2=250\,$ V with $\,Vg_1$ as parameter.

HEATER RATINGS

Heating	g: indire	$^{ m ct}$	b;	у.	Α.	С.,	p	ar	all	$_{\mathrm{el}}$	Sŧ	ıpı	ply	٠.											
Heater	voltage																				T,	f ==	6.3	; '	V
Heater	current																				I_f	=	: 1.2	1	A

CAPACITANCES

Anode-grid .													Com	<1	0.7 m	uF	

OPERATING DATA: EL 6 used as a normal output valve (single valve)

Anode voltage							 V_a	=	250 V
Screen-grid voltage .							 V_{g_2}	***	250 V
Grid bias							 V_{g_1}	=	_7 V
Cathode resistor							 R_k		90 ohms
Anode current							 I_a		72 mA
Screen-grid current .							 I_{g_2}	=	8.0 mA
Mutual conductance.							 \mathcal{S}	_	14.5 mA/V
Internal resistance .							 R_i	=	20,000 ohms
Load resistor							 R_{tt}	=	3,500 ohms
Output power with 10	0 % d	istort	ion				 W_o	_	8 W
Alternating input volu	tage fo	or W	0 = 3	8 W			 V_i	==	$4.8 V_{ m eff}$
Sensitivity ($W_o = 50$	mW).						 V_{i}	==	$0.3~{ m V_{eff}}$
Amplification factor,	screen	with	rest	ect to	o grid	11.	 $\mu_{q_2q_1}$	=	20

OPERATING DATA: EL 6 used as an output valve in balanced circuits (two valves) with automatic grid bias.

Anode voltage $\dots \dots V_a$		250 V	250 V
Screen-grid voltage V_{g_2}	27	250 V	265 V
Cathode resistor R_k		90 ohms	97 ohms
Anode current (without signal) I_{ao}		2 imes45	$2 imes 45 \; \mathrm{mA}$
Anode current at max. modulation $I_{a \text{ n}}$		2 imes53	2 imes 54 mA
Screen current (without signal) I_{g20}	,	2 imes 5.1	$2 \times 5.1 \text{ mA}$
Screen current at max. modulation I_{g_2}	max =	2 imes 8.5	$2 imes 9.9 \; \mathrm{mA}$
Load resistor between anodes R_{aa}		5,000 ohms	5,000 ohms
Output power W_o	<u></u>	14.5 W	16 W
Distortion d_{tot}	,	2.2 %	1.7 %
Alternating grid voltage per grid V_i	127	$7.3~\mathrm{V_{eff}}$	$8.2~\mathrm{V_{eff}}$

MAXIMUM RATINGS

Anode voltage in cold condition V_{ao}	==	max. 550 V								
Anode voltage $\dots \dots \dots$										
Anode dissipation W_a		max. 18 W								
Screen voltage in cold condition $V_{g_{20}}$	=	max. 550 V								
Screen voltage V_{g_2}	=	max. 275 V								
Screen dissipation $(V_i = 0) \dots W_{g_2}$		max. 2 W								
Screen dissipation $(W_0 - \max)$, W_{g_2}	=	max. 3 W								
Cathode current I_k	=	max. 90 mA								
Grid voltage at grid current start $(I_{g1} = +0.3 \mu A) V_{g1}$	=	max. —1.3 V								
External resistance between grid and cathode R_{g_1k}		max. 0.7 M ohm								
External resistance between heater and cathode R_{fk}	=	max. 5,000 ohms								
Voltage between heater and cathode (D.C. voltage										
or effective value of alternating voltage) V_{fk}	=	max. 50 V								

Fig. 6 gives a number of useful data plotted against the screen voltage in the range 250—275 V. With an anode voltage of 250 V by means of these characteristics any voltage drop in the output transformer from 0 to 25 V can be taken into account in investigating the operation of the valve. Dynamic characteristics of the EL 6 as a function of the screen voltage, in the case of receivers in which the available anode voltage is less than 250 V and whereby the anode voltage is less than that of the screen by 15 V, are given in Fig. 8. Allowance is made for an average voltage drop of 15 V across the output transformer.

In the case of Class A and A/B amplification the grid bias must be automatic (cathode resistor); semi-automatic bias may be employed so long as the cathode current of the EL 6 is in excess of 50 % of the total current flowing through the biasing resistor. The maximum value of the grid leak, as indicated in the Maximum Ratings should then be reduced in accordance with the following:

Cathode current of the output valve

Total current passing through the resistor producing the voltage drop

It should be noted, further, that the current of those valves to which automatic gain control is applied will affect the bias on the output valve, so that when the control voltage rises the bias quickly becomes too low and the anode current too high. The high mutual conductance of this valve should be taken into consideration in the design of receiver circuits, in view of the resultant tendency towards R.F. feedback and oscillation. Leads to the valve contacts should therefore be as short as possible, and a resistor of about 1,000 ohms in the grid lead is indispensable.

For the use of the valve in balanced circuits employing automatic bias the necessary data will be found in Figs 8 and 9: the former gives the distortion and alternating grid voltage at $V_a = 250$ V and $V_{g2} = 250$ V, whilst Fig. 9 shows various data, such as the biasing resistor, output power, etc. as functions of the screen voltage when the anode current is 2×24 mA with a constant voltage of 250 V on the anode. Using the curves it is possible for the designer to obtain the appropriate operating conditions with respect to almost any voltage drop across the output transformer. In balanced output stages care should be taken, if the anode current (without signal) is more than 45 mA per valve, to see that each valve has its own biasing resistor. This precaution is advisable in all cases where a possibility exists that one of the valves may be removed while the set is in operation, as this will otherwise result inevitably in damage to the other valve.

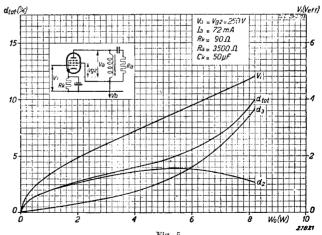


Fig. 5 Total distortion, and 2nd and 3rd harmonic distortion; EL 6 used as normal output pentode with auto, bias and decoupling capacitor in the cathode circuit ($Va = Vg_2 = 250 \text{ V}$)



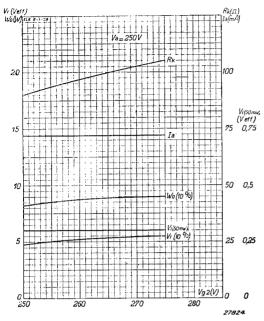


Fig. 6

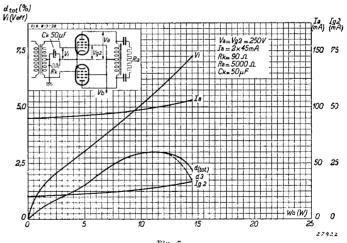
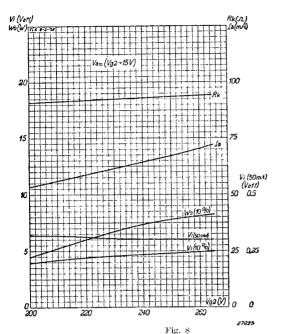


Fig. 7 Total anode current Ia, total screen current Ig_2 , total distortion dtot, 3rd harmonic distortion and alternating grid voltage per grid Vi, as functions of the output power Wo when using two EL 6 valves in a balanced circuit with $Va = Vg_2 = 250$ V.



Output power with 10 % distortion. Wo (10 %) Alternating grid voltage with 10 % distortion. Vi (10 %) Sensitivity . Vi (50 mW) Cathode resistor . Rk Anode current . Ia

as functions of the screen-grid voltage (in the range 200–265 V) where the voltage of the anode is 15 V lower than that of the screen.

