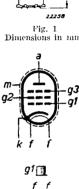
## EF 5 Variable MU R.F. pentode

The EF 5 is a variable-mu R. F. or I.F. pentode. Special care has been devoted in the design of this valve to the greatest possible reduction in cross-modulation and modulation hum. At a screengrid voltage of 100 V the anode current of the EF 5 is 8 mA, when the mutual conductance is 1.7 mA/V, the control range being from -3 to -46.5 V. The control range is capable of modification by means of the screen voltage; at lower screen potentials, for the same grid bias, the mutual conductance drops sharply, but the crossmodulation conditions are then not so very good. With a screen voltage of 85 V the control range extends from —2 to —39 V only. Obviously, a lower screen potential will result in a lower screen Dimensions in mm, current as well as a lower anode current and it is thus possible to reduce the bias at the working point from -3 to -2 V to increase the slope; the working value of the mutual conductance is then 1.85 mA/V.

With 60 V screen potential the conductance is still further reduced, q2 to -2/-29 V.

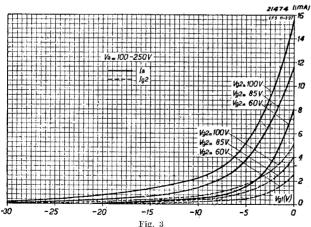
The very greatly diminished modulation hum in this valve is of first importance in A.C./D.C. receivers, where alternating voltages at mains frequency can easily occur between heater and grid. The EF 5 is notable for its low inter-electrode capacitances and high internal resistance; excellent results are obtained on the short-wave range. Although on short waves the circuit magnification is usually only fair, the excellent properties of the EF 5 make it possible to achieve extremely good amplification in this range. On short waves, too, the mutual conductance is the same as on the other ranges (e.g. 200 m). The high impedance of anode and grid with respect to earth in the 12 to 60 metre band, as compared with the impedance values of practical tuned circuits, enables the EF 5 to produce in that range amplification values equal to the product of mutual conductance and



max 32



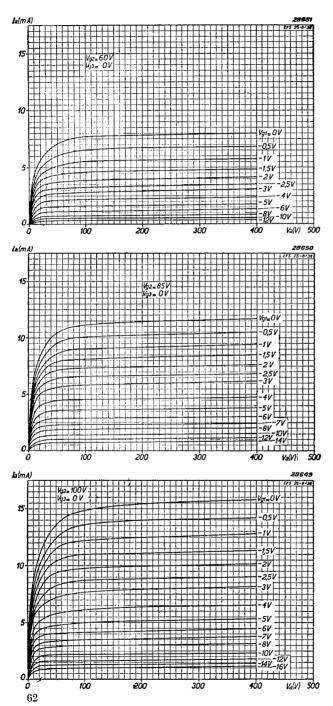
Fig. 2 Arrangement of electrodes and base connections.



Anode current and screen current as a function of the grid bias, for different values of screen potential, at 250 V anode. The curves also apply as an approximation to anode voltages of 100-250 V.

impedance. On the shortwave bands the (feedback) impedance, which takes the place of the anode-to-grid capacitance on the long waves, is unusually high and there is therefore no risk of parasitic oscillation, even with maximum permissible amount of gain.

A factor contributing in no small degree towards the high properties of this valve is the use of side contacts (P-type



base). The suppressor grid and metallizing, each with their own individual contacts, can be connected direct to earth to give the best possible results with short-wave reception.

Fig. 4
Anode current as a function of the anode voltage, for different values of screen potential and grid bias.

HEATER RATINGS								
Heating: indirect, A.C. or D.C., series or parallel supply Heater voltage								
CAPACITANCES								
$C_{ag_1} < 0.003 \ \mu\mu { m F}$ $C_{g_1} = 5.4 \ \mu\mu { m F}$ $C_u = 6.9 \ \mu\mu { m F}$								
OPERATING DATA	1: valv	e empl	oyed as	R.F.	or I.F. a	amplifi	er	
Anode voltage								
$V_a$ (V) Screen-grid voltage	100			200			250	
$V_{g_2}$ (V)	100			100			100	
Suppressor-grid voltag $V_{q_3}$ (V)	,e 0			0			0	
Cathode resistor	_							
$R_k$ (ohms) Grid bias	170			180			180	
$V_{g_1}$ (V) -2.85 <sup>1</sup> ) Anode current	-342)	$-46.5^{3}$ )	$-2.95^{1}$ )	$-34^{2}$ )	$-46.5^{3}$ )	-31)	-342)	$-46.5^{3}$ )
$I_a$ (A) 8		_	8		-	8		
Screen current			0.0			5 C		
$I_{g_2}$ (mA) 2.6 Mutual conductance			2.6		-	2.6		
$S(\mu A/V) = 1700$	17	2	1700	17	2	1700	17	2
Amplification factor u 500			1600			2000		
$\mu$ 500 Internal resistance			1600			2000	_	_
$R_i  (\mathrm{M  ohms}) = 0.3$	> 10	> 10	0.95	> 10	> 10	1.2	> 10	> 10
4 3 1,								
Anode voltage $F_{\mu}(V)$	100			200			250	
Screen-grid voltage	100			200			200	
$V_{g_2}(\mathbf{V})$	85			85			85	
Suppressor-grid voltag							75	
$V_{g_3}$ (V) Cathode resistor	0			0			0	
$R_k$ (ohms)	190			195			200	
Grid bias								
$V_{g_1}\left(\mathrm{V} ight)$ -1.91) Anode current	-292)	-39 <sup>3</sup> )	-1.951)	$-29^{2}$ )	-39 <sup>3</sup> )	-21)	$-29^{2}$ )	-393)
$I_a \text{ (mA)}$ 7.5		-	7.5			7.5		
Screen current $I_{g_2} \text{ (mA)} \qquad 2.45$			2.45		_	2.45		
Mutual conductance $S(\mu A/V)$ 1850	18	2	1850	18	2	1850	18	2
Amplification factor		-		- * *	=		- * ·	****
μ 550			1750	-		2200		<u> </u>
Internal resistance $R_i$ (M ohms) 0.3	> 10	> 10	0.95	> 10	> 10	1.2	~ 10	> 10

Anode voltage									
$V_a$ (V)		100			200			250	
Screen-grid vol	ltage								
$V_{g_2}\left(\mathbf{V}\right)$		60			60			60	
Supressor-grid	voltage	:							
$V_{g_3}(\nabla)$		0			0			0	
Cathode resisto	r								
$R_k$ (ohms)		360			370			380	
Grid bias									
$V_{g_1}(\mathbf{V})$	$-1.9^{1}$	$-22^{2}$ )	$-29^{3}$ )	$-2.95^{1}$ )	$-22^{2}$ )	$-29^{3}$ )	$-2^{1}$ )	$-22^{2}$ )	$29^{3}$ )
Anode current									
$I_a$ (mA)	4	_		4			4	_	
Screen-grid current									
$I_{g_2} (\mathrm{mA})$	1.3			1.3			1.3		
Mutual conduc	tance								
$S(\mu A/V)$	1400	14	2	1400	14	<b>2</b>	1400	14	$^2$
Amplification factor									
$\mu$	1200			1900			2000		_
Internal resistance									
$R_i$ (M ohms)	0.85	> 10	> 10	1.35	> 10	> 10	1.4	> 10	> 10

1) Without control

<sup>2</sup>) Mutual conductance reduced to one-hundredth of uncontrolled value.

of uncontrolled value.

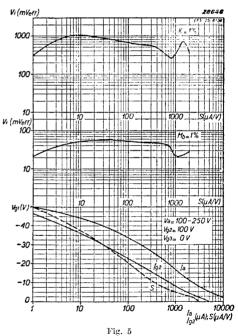
3) Extreme limit of control.

## MAXIMUM RATINGS

$V_{ao}$	= max. 550 V
$V_a$	= max. 250 V
$W_a$	= max. 2 W
$\Gamma_{g_{20}}$	= max. 400 V
$V_{\boldsymbol{g_2}}$	max. 125 V
$W_{g_2}$	max. 0.4 W
$I_k$	max. 15 mA
5	$-+0.3 \mu\text{A}) = \text{max.}1.3 \text{ V}$
$R_{g_1}$	max. 2.5 M ohms
$R_{fk}$	max. 20,000 ohms
$V_{fk}$	max. 100 V <sup>1</sup> )

 Direct voltage or effective value of alternating voltage.

Due to the curvature of the characteristic, the uses of the EF 5 are restricted to R.F. and I.F. amplification. It can be employed as amplifier with either automatic or manual control. It is preferable to feed the screen through a potential divider; in many cases it would be found when using a series resistor that the screen voltage would become too high on full control and that the amplification control would be far too tardy.



At 100-250 V anode and 100 V screen; Upper diagram. Alternating grid voltage as a function of the mutual conductance, with 1 % cross modulation.

Centre diagram. Alternating grid voltage as a function of the mutual conductance, with 1 % modulation hum.

Lower diagram. Mutual conductance S, anode current Ia and screen current  $Ig_2$  as a function of the voltage on the first grid.

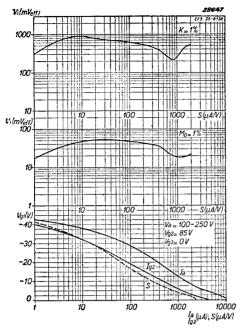
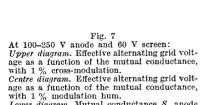
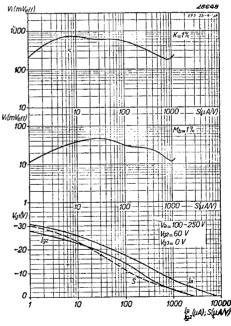
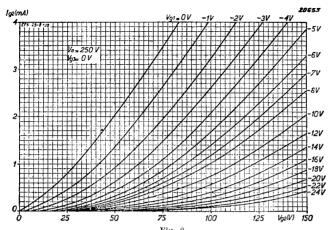


Fig. 6 At 100–250 V anode and 85 V screen; Upper diagram. Effective alternating grid voltage as a function of the mutual conductance, with 1 % cross modulation. Centre diagram. Effective alternating grid voltage as a function of the mutual conductance, with 1 % modulation hum. Lower diagram. Mutual conductance S, anode current Ia and screen current  $Ig_2$  as a function of the grid bias.



Lower diagram. Mutual conductance S, anode current Ia and screen current  $Ig_2$  as a function of the grid bias.





 ${\bf Fig.~8} \\ {\bf Screen-grid~current~as~a~function~of~the~screen~voltage,~for~different~values~of~grid~bias.}$ 

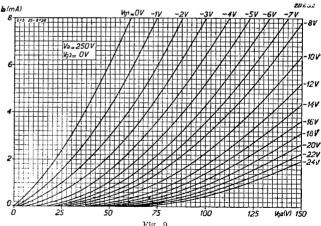


Fig. 9
Anode current as a function of the screen-grid voltage, for different values of grid bias.