

LM833 Dual Audio Operational Amplifier

General Description

The LM833 is a dual general purpose operational amplifier designed with particular emphasis on performance in audio systems.

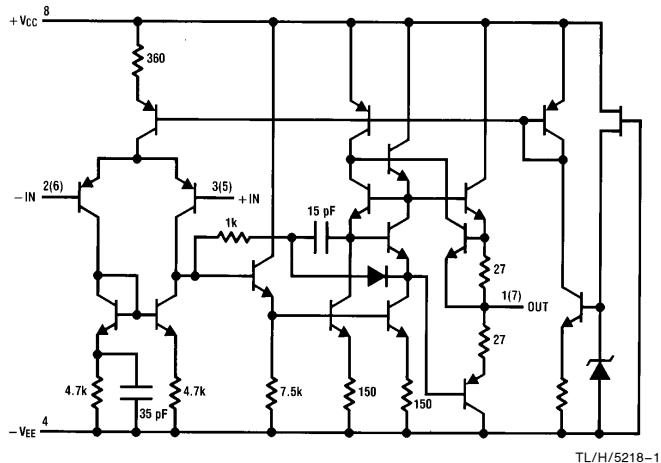
This dual amplifier IC utilizes new circuit and processing techniques to deliver low noise, high speed and wide bandwidth without increasing external components or decreasing stability. The LM833 is internally compensated for all closed loop gains and is therefore optimized for all preamp and high level stages in PCM and HiFi systems.

The LM833 is pin-for-pin compatible with industry standard dual operational amplifiers.

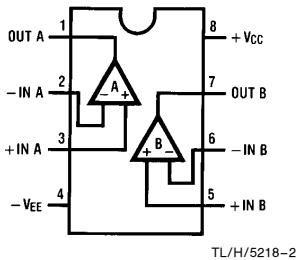
Features

- Wide dynamic range >140 dB
- Low input noise voltage 4.5 nV/Hz
- High slew rate 7 V/μs (typ)
- High gain bandwidth product 5 V/μs (min)
- 15 MHz (typ)
- 10 MHz (min)
- Wide power bandwidth 120 kHz
- Low distortion 0.002%
- Low offset voltage 0.3 mV
- Large phase margin 60°

Schematic Diagram (1/2 LM833)

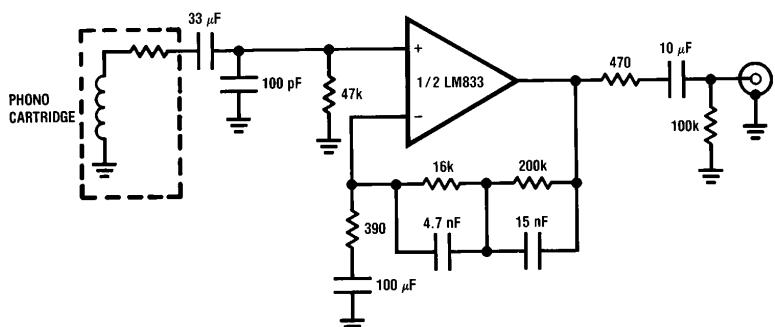


Connection Diagram



Order Number LM833M or LM833N
See NS Package Number
M08A or N08E

Typical Application RIAA Preamp



$A_v = 35 \text{ dB}$ $f = 1 \text{ kHz}$
 $E_n = 0.33 \mu\text{V}$ A Weighted
 $S/N = 90 \text{ dB}$ A Weighted, $V_{IN} = 10 \text{ mV}$
 @ $f = 1 \text{ kHz}$

Absolute Maximum Ratings

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/Distributors for availability and specifications.

Supply Voltage	$V_{CC} - V_{EE}$	36V	Soldering Information	
Differential Input Voltage (Note 1)	V_{ID}	$\pm 30V$	Dual-In-Line Package	260°C
Input Voltage Range (Note 1)	V_{IC}	$\pm 15V$	Soldering (10 seconds)	215°C
Power Dissipation (Note 2)	P_D	500 mW	Small Outline Package	220°C
Operating Temperature Range	T_{OPR}	$-40 \sim 85^\circ C$	Vapor Phase (60 seconds)	
Storage Temperature Range	T_{STG}	$-60 \sim 150^\circ C$	Infrared (15 seconds)	
			See AN-450 "Surface Mounting Methods and Their Effect on Product Reliability" for other methods of soldering surface mount devices.	
			ESD tolerance (Note 3)	1600V

DC Electrical Characteristics ($T_A = 25^\circ C$, $V_S = \pm 15V$)

Symbol	Parameter	Conditions	Min	Typ	Max	Units
V_{OS}	Input Offset Voltage	$R_S = 10\Omega$		0.3	5	mV
I_{OS}	Input Offset Current			10	200	nA
I_B	Input Bias Current			500	1000	nA
A_V	Voltage Gain	$R_L = 2 k\Omega$, $V_O = \pm 10V$	90	110		dB
V_{OM}	Output Voltage Swing	$R_L = 10 k\Omega$ $R_L = 2 k\Omega$	± 12 ± 10	± 13.5 ± 13.4		V
V_{CM}	Input Common-Mode Range		± 12	± 14.0		V
CMRR	Common-Mode Rejection Ratio	$V_{IN} = \pm 12V$	80	100		dB
PSRR	Power Supply Rejection Ratio	$V_S = 15 \sim 5V, -15 \sim -5V$	80	100		dB
I_Q	Supply Current	$V_O = 0V$, Both Amps		5	8	mA

AC Electrical Characteristics ($T_A = 25^\circ C$, $V_S = \pm 15V$, $R_L = 2 k\Omega$)

Symbol	Parameter	Conditions	Min	Typ	Max	Units
SR	Slew Rate	$R_L = 2 k\Omega$	5	7		$V/\mu s$
GBW	Gain Bandwidth Product	$f = 100 kHz$	10	15		MHz

Design Electrical Characteristics ($T_A = 25^\circ C$, $V_S = \pm 15V$)

The following parameters are not tested or guaranteed.

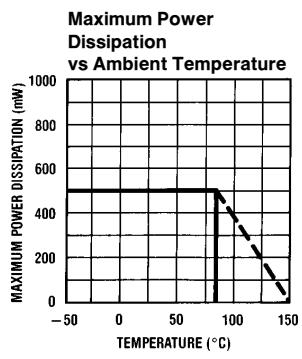
Symbol	Parameter	Conditions	Typ	Units
$\Delta V_{OS}/\Delta T$	Average Temperature Coefficient of Input Offset Voltage		2	$\mu V/^{\circ}C$
THD	Distortion	$R_L = 2 k\Omega$, $f = 20 \sim 20 kHz$ $V_{OUT} = 3 V_{rms}$, $A_V = 1$	0.002	%
e_n	Input Referred Noise Voltage	$R_S = 100\Omega$, $f = 1 kHz$	4.5	nV/\sqrt{Hz}
i_n	Input Referred Noise Current	$f = 1 kHz$	0.7	pA/\sqrt{Hz}
PBW	Power Bandwidth	$V_O = 27 V_{pp}$, $R_L = 2 k\Omega$, THD $\leq 1\%$	120	kHz
f_U	Unity Gain Frequency	Open Loop	9	MHz
ϕ_M	Phase Margin	Open Loop	60	deg
	Input Referred Cross Talk	$f = 20 \sim 20 kHz$	-120	dB

Note 1: If supply voltage is less than $\pm 15V$, it is equal to supply voltage.

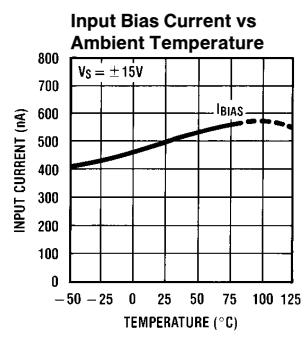
Note 2: This is the permissible value at $T_A \leq 85^\circ C$.

Note 3: Human body model, 1.5 $k\Omega$ in series with 100 pF.

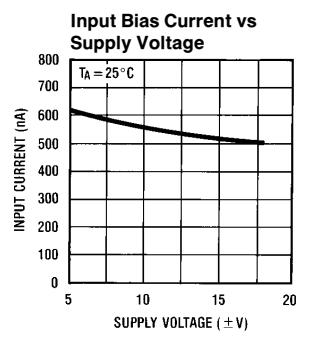
Typical Performance Characteristics



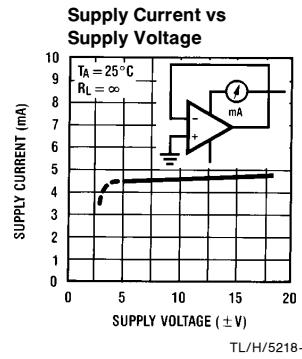
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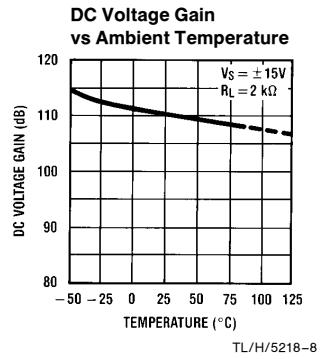
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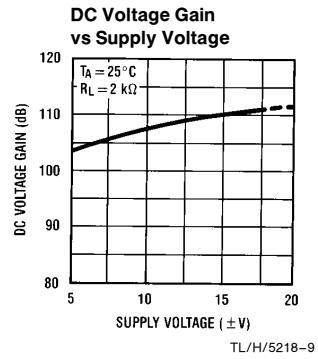
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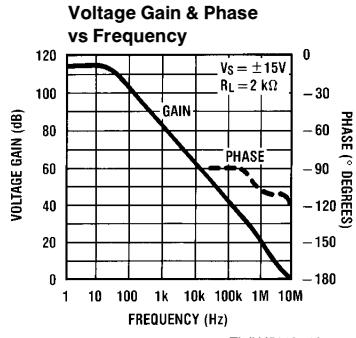
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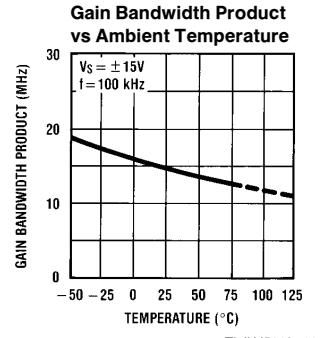
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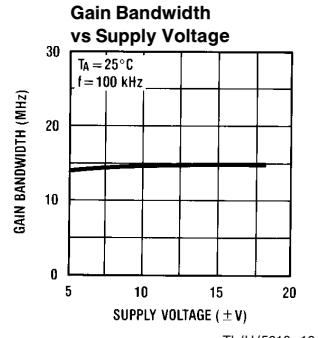
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TL/H/5218-10

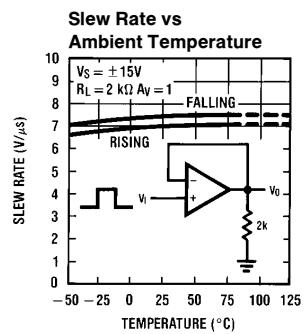


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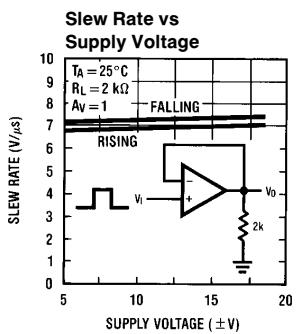


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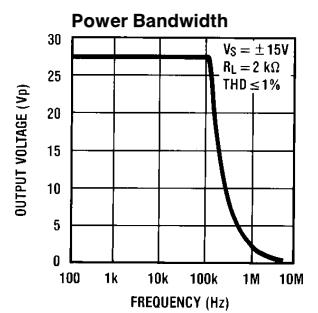
Typical Performance Characteristics (Continued)



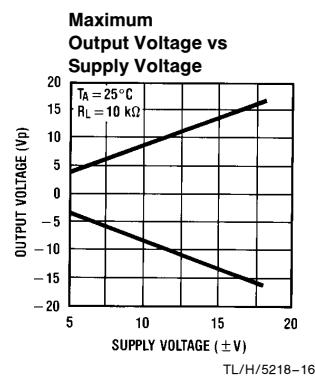
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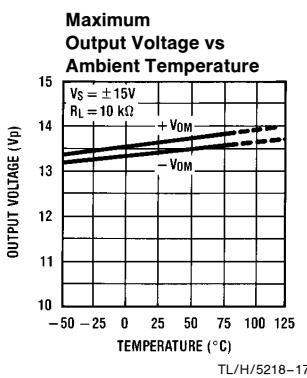
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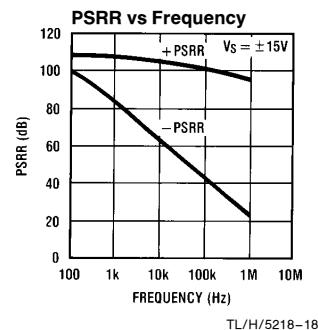
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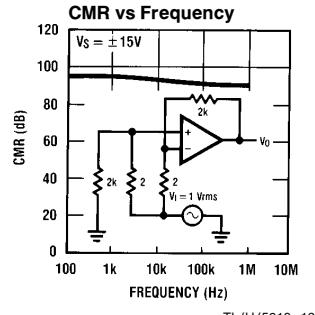
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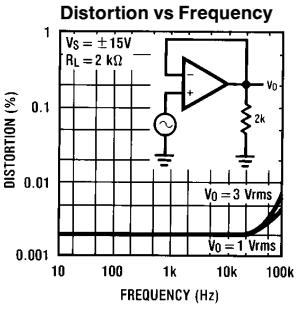
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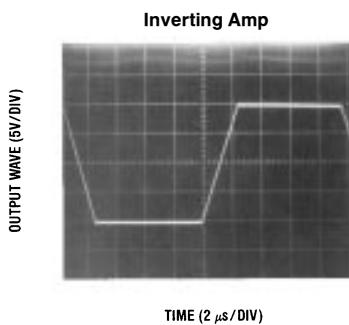
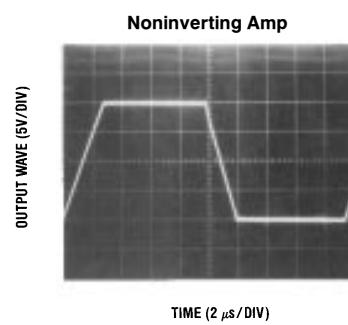
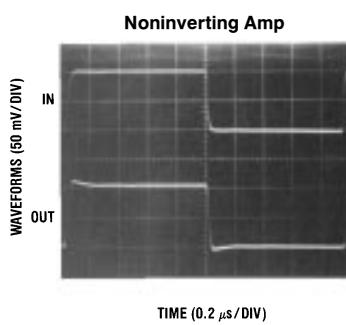
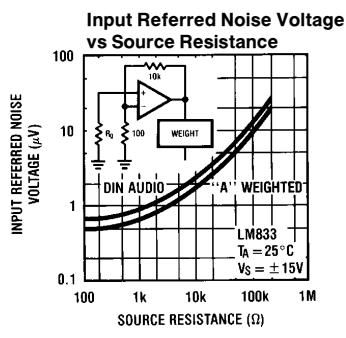
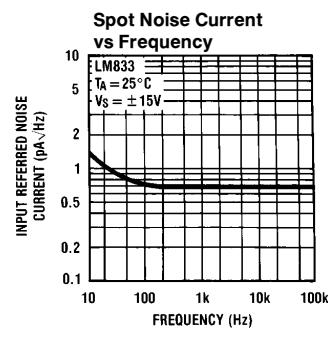
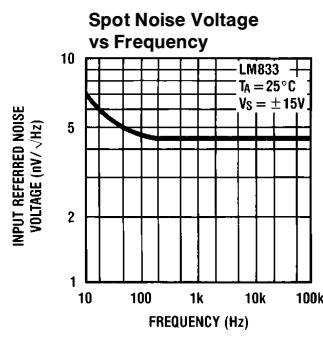


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Typical Performance Characteristics (Continued)



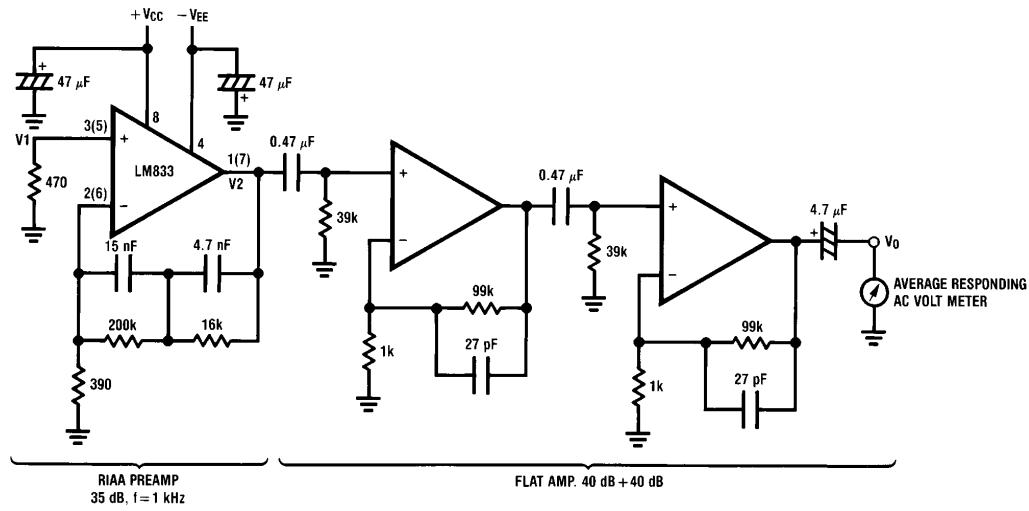
Application Hints

The LM833 is a high speed op amp with excellent phase margin and stability. Capacitive loads up to 50 pF will cause little change in the phase characteristics of the amplifiers and are therefore allowable.

Capacitive loads greater than 50 pF must be isolated from the output. The most straightforward way to do this is to put a resistor in series with the output. This resistor will also prevent excess power dissipation if the output is accidentally shorted.

Noise Measurement Circuit

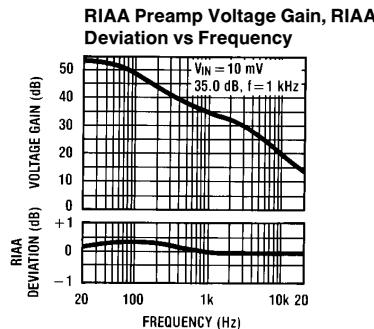
Complete shielding is required to prevent induced pick up from external sources. Always check with oscilloscope for power line noise.



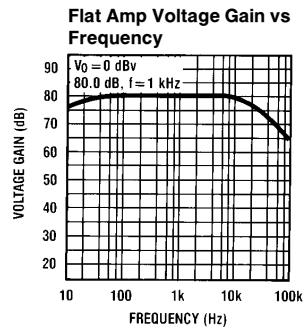
TL/H/5218-27

$$\text{Total Gain: } 115 \text{ dB} @ f = 1 \text{ kHz}$$

$$\text{Input Referred Noise Voltage: } e_n = V_0 / 560,000 \text{ (V)}$$



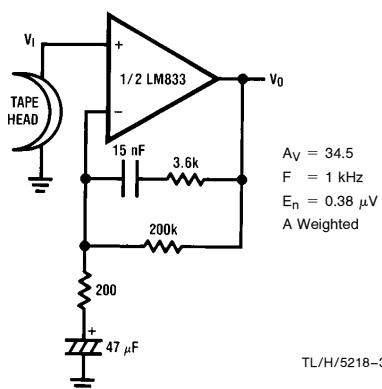
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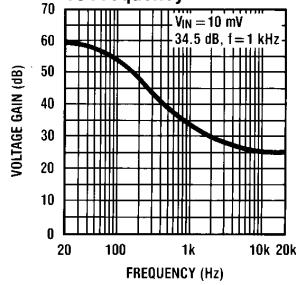
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Typical Applications

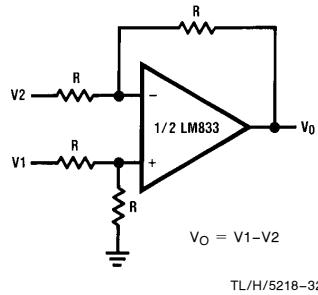
NAB Preamp



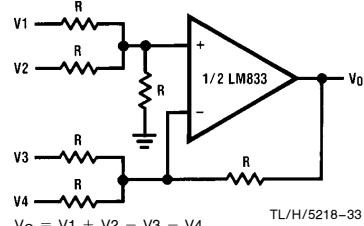
NAB Preamp Voltage Gain vs Frequency



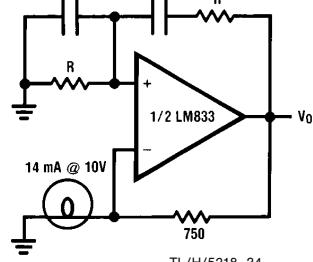
Balanced to Single Ended Converter



Adder/Subtractor

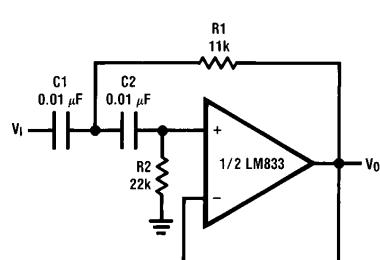


Sine Wave Oscillator



$$f_o = \frac{1}{2\pi RC}$$

Second Order High Pass Filter (Butterworth)



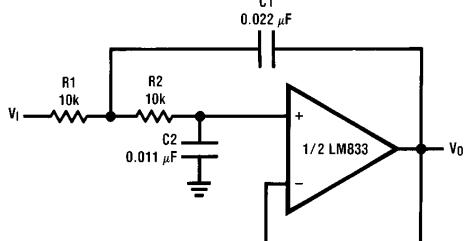
if $C_1 = C_2 = C$

$$R_1 = \frac{\sqrt{2}}{2\omega_0 C}$$

$$R_2 = 2 \cdot R_1$$

Illustration is $f_0 = 1 \text{ kHz}$

Second Order Low Pass Filter (Butterworth)



if $R_1 = R_2 = R$

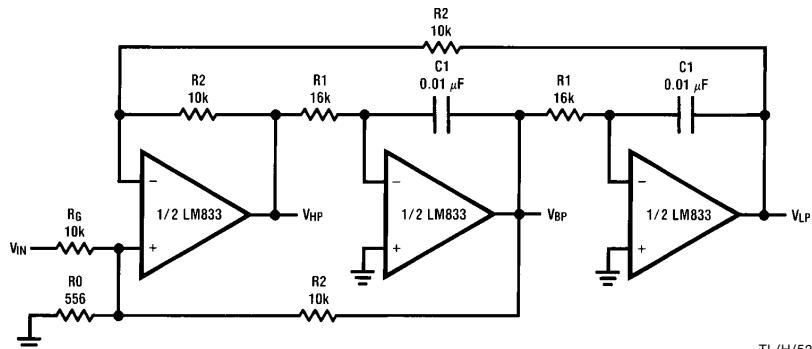
$$C_1 = \frac{\sqrt{2}}{\omega_0 R}$$

$$C_2 = \frac{C_1}{2}$$

Illustration is $f_0 = 1 \text{ kHz}$

Typical Applications (Continued)

State Variable Filter

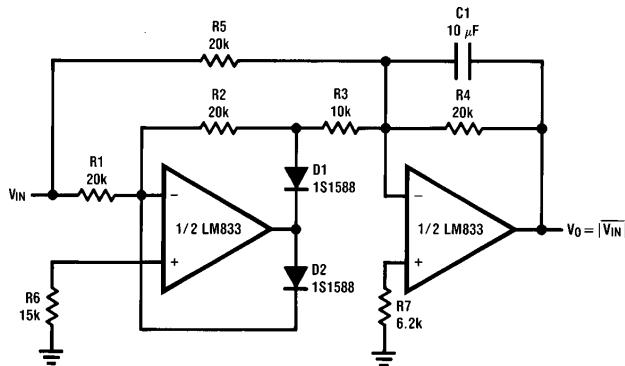


TL/H/5218-37

$$f_0 = \frac{1}{2\pi C_1 R_1}, Q = \frac{1}{2} \left(1 + \frac{R_2}{R_0} + \frac{R_2}{R_G} \right), A_{BP} = Q A_{LP} = Q A_{LH} = \frac{R_2}{R_G}$$

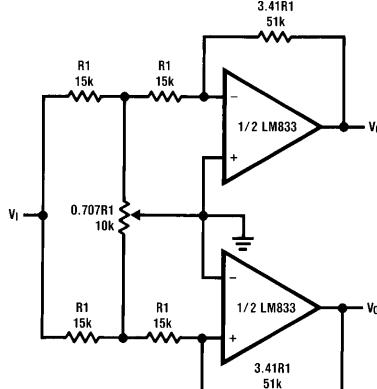
Illustration is $f_0 = 1$ kHz, $Q = 10$, $A_{BP} = 1$

AC/DC Converter



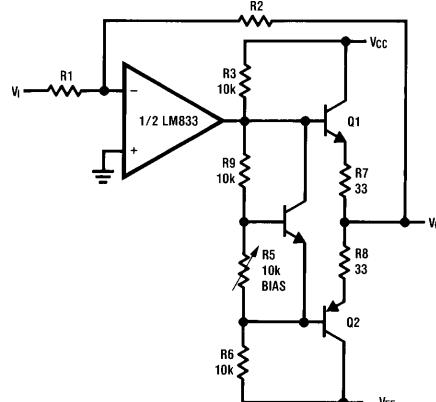
TL/H/5218-38

2 Channel Panning Circuit (Pan Pot)



TL/H/5218-39

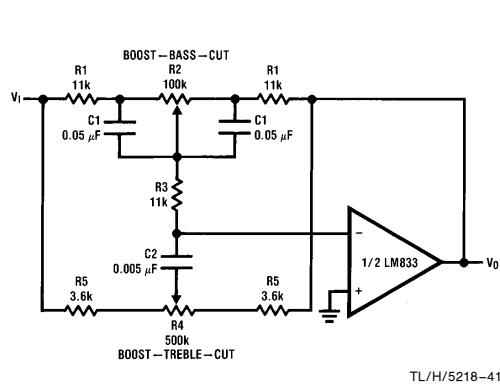
Line Driver



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Typical Application (Continued)

Tone Control



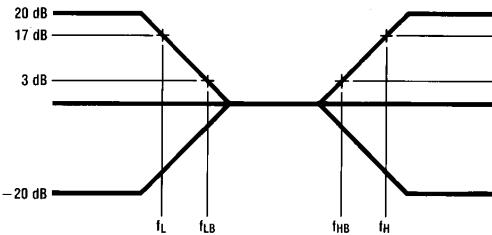
$$f_L = \frac{1}{2\pi R_2 C_1}, f_{LB} = \frac{1}{2\pi R_1 C_1}$$

$$f_H = \frac{1}{2\pi R_5 C_2}, f_{HB} = \frac{1}{2\pi(R_1 + R_5 + 2R_3)C_2}$$

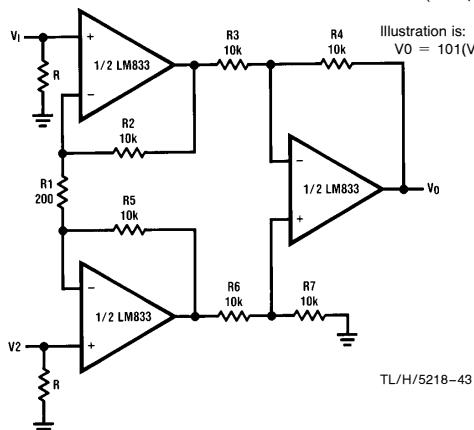
Illustration is:

$$f_L = 32 \text{ Hz}, f_{LB} = 320 \text{ Hz}$$

$$f_H = 11 \text{ kHz}, f_{HB} = 1.1 \text{ kHz}$$



Balanced Input Mic Amp

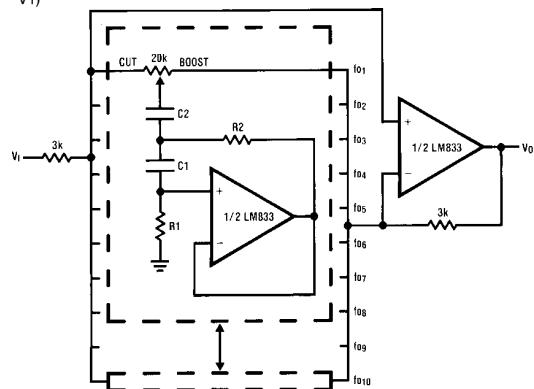


If $R_2 = R_5, R_3 = R_6, R_4 = R_7$

$$V_0 = \left(1 + \frac{2R_2}{R_1}\right) \frac{R_4}{R_3} (V_2 - V_1)$$

Illustration is:
 $V_0 = 10(V_2 - V_1)$

10 Band Graphic Equalizer



$f_o(\text{Hz})$	C_1	C_2	R_1	R_2
32	0.12μF	4.7μF	75kΩ	500Ω
64	0.056μF	3.3μF	68kΩ	510Ω
125	0.033μF	1.5μF	62kΩ	510Ω
250	0.015μF	0.82μF	68kΩ	470Ω
500	8200pF	0.39μF	62kΩ	470Ω
1k	3900pF	0.22μF	68kΩ	470Ω
2k	2000pF	0.1μF	68kΩ	470Ω
4k	1100pF	0.056μF	62kΩ	470Ω
8k	510pF	0.022μF	68kΩ	510Ω
16k	330pF	0.012μF	51kΩ	510Ω

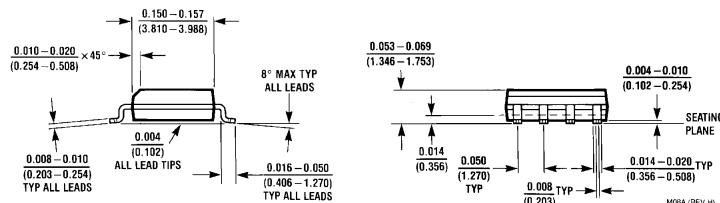
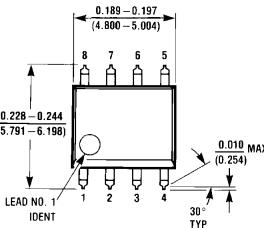
At volume of change = $\pm 12 \text{ dB}$

$Q = 1.7$

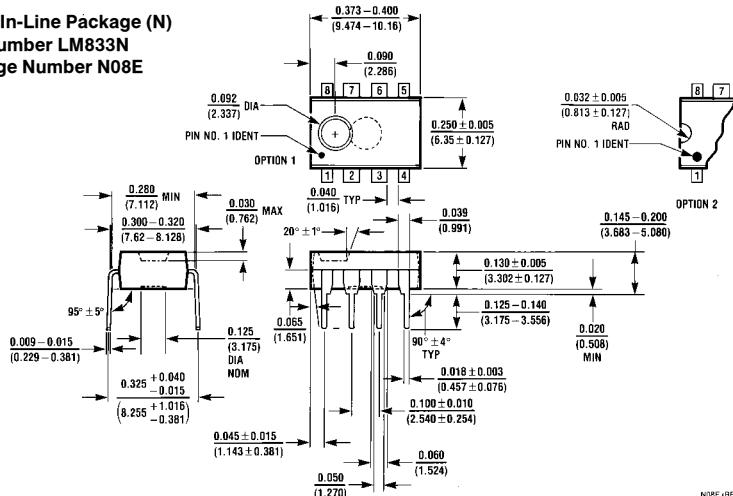
Reference: "AUDIO/RADIO HANDBOOK", National Semiconductor, 1980, Page 2-61

Physical Dimensions inches (millimeters)

Small Outline Package (M)
Order Number LM833M
NS Package Number M08A



Molded Dual-In-Line Package (N)
Order Number LM833N
NS Package Number N08E



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