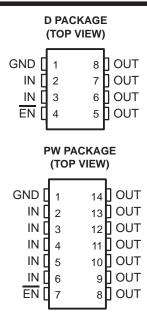
# TPS2010, TPS2011, TPS2012, TPS2013 POWER-DISTRIBUTION

SLVS097A - DECEMBER 1994 - REVISED AUGUST 1995

- 95-m $\Omega$  Max (5.5-V Input) High-Side MOSFET **Switch With Logic Compatible Enable Input**
- Short-Circuit and Thermal Protection
- Typical Short-Circuit Current Limits: 0.4 A, TPS2010; 1.2 A, TPS2011; 2 A, TPS2012; 2.6 A, TPS2013
- Electrostatic-Discharge Protection, 12-kV Output, 6-kV All Other Terminals
- **Controlled Rise and Fall Times to Limit Current Surges and Minimize EMI**
- **SOIC-8 Package Pin Compatible With the** Popular Littlefoot™ Series When GND Is Connected
- 2.7-V to 5.5-V Operating Range
- 10-μA Maximum Standby Current
- Surface-Mount SOIC-8 and TSSOP-14 **Packages**
- -40°C to 125°C Operating Junction **Temperature Range**



# description

The TPS201x family of power-distribution switches is intended for applications where heavy capacitive loads and short circuits are likely to be encountered. The high-side switch is a 95-m $\Omega$  N-channel MOSFET. Gate drive is provided by an internal driver and charge pump designed to control the power switch rise times and fall times to minimize current surges during switching. The charge pump operates at 100 kHz, requires no external components, and allows operation from supplies as low as 2.7 V. When the output load exceeds the current-limit threshold or a short circuit is present, the TPS201x limits the output current to a safe level by switching into a constant-current mode. Continuous heavy overloads and short circuits increase power dissipation in the switch and cause the junction temperature to rise. If the junction temperature reaches approximately 180°C, a thermal protection circuit shuts the switch off to prevent damage. Recovery from thermal shutdown is automatic once the device has cooled sufficiently.

The members of the TPS201x family differ only in short-circuit current threshold. The TPS2010 is designed to limit at 0.4-A load; the other members of the family limit at 1.2 A, 2 A, and 2.6 A (see the available options table). The TPS201x family is available in 8-pin small-outline integrated circuit (SOIC) and 14-pin thin shink small-outline (TSSOP) packages and operates over a junction temperature range of -40°C to 125°C. Versions in the 8-pin SOIC package are drop-in replacements for Siliconix's Littlefoot™ power PMOS switches, except that GND must be connected.



Please be aware that an important notice concerning availability, standard warranty, and use in critical applications of Texas Instruments semiconductor products and disclaimers thereto appears at the end of this data sheet.



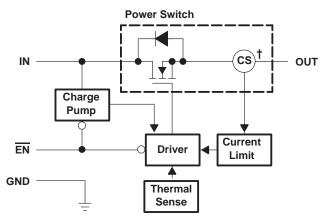
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#### **AVAILABLE OPTIONS**

ТЈ	RECOMMENDED MAXIMUM	TYPICAL SHORT-CIRCUIT	PACKAG	CHIP	
	CONTINUOUS LOAD CURRENT (A)	OUTPUT CURRENT LIMIT AT 25°C (A)	SOIC (D)†	TSSOP (PW) <sup>‡</sup>	FORM (Y)
	0.2	0.4	TPS2010D	TPS2010PWLE	TPS2010Y
40°C to 135°C	0.6	1.2	TPS2011D	TPS2011PWLE	TPS2011Y
-40°C to 125°C	1	2	TPS2012D	TPS2012PWLE	TPS2012Y
	1.5	2.6	TPS2013D	TPS2013PWLE	TPS2013Y

<sup>†</sup> The D package is available taped and reeled. Add an R suffix to device type (e.g., TPS2010DR).

# functional block diagram



†Current sense

# **Terminal Functions**

1	ERMINA	L				
NAME	N	NO.		DESCRIPTION		
INAIVIE	D	PW				
EN	4	7	I	Enable input. Logic low turns power switch on.		
GND	1	1	I	Ground		
IN	2, 3	2-6	I	Input voltage		
OUT	5-8	8-14	0	Power-switch output		

# detailed description

# power switch

The power switch is an N-channel MOSFET with a maximum on-state resistance of 95 m $\Omega$  (V<sub>I(IN)</sub> = 5.5 V), configured as a high-side switch.

#### charge pump

An internal 100-kHz charge pump supplies power to the driver circuit and provides the necessary voltage to pull the gate of the MOSFET above the source. The charge pump operates from input voltages as low as 2.7 V and requires very little supply current.



<sup>‡</sup>The PW package is only available left-end taped and reeled (indicated by the LE suffix on the device type; e.g., TPS2010PWLE).

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# detailed description (continued)

#### driver

The driver controls the gate voltage of the power switch. To limit large current surges and reduce the associated electromagnetic interference (EMI) produced, the driver incorporates circuitry that controls the rise times and fall times of the output voltage. The rise and fall times are typically in the 2-ms to 4-ms range instead of the microsecond or nanosecond range for a standard FET.

## enable (EN)

A logic high on the  $\overline{EN}$  input turns off the power switch and the bias for the charge pump, driver, and other circuitry to reduce the supply current to less than 10  $\mu$ A. A logic zero input restores bias to the drive and control circuits and turns the power on. The enable input is compatible with both TTL and CMOS logic levels.

#### current sense

A sense FET monitors the current supplied to the load. The sense FET is a much more efficient way to measure current than conventional resistance methods. When an overload or short circuit is encountered, the current-sense circuitry sends a control signal to the driver. The driver in turn reduces the gate voltage and drives the power FET into its linear region, which switches the output into a constant current mode and simply holds the current constant while varying the voltage on the load.

#### thermal sense

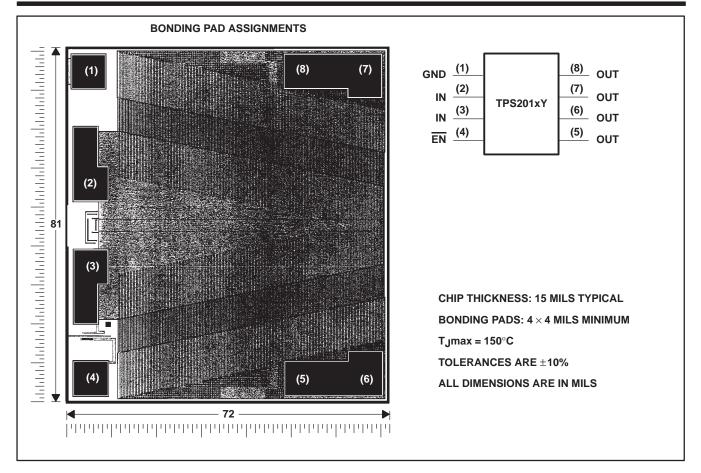
An internal thermal-sense circuit shuts the power switch off when the junction temperature rises to approximately 180°C. Hysteresis is built into the thermal sense, and after the device has cooled approximately 20 degrees, the switch turns back on. The switch continues to cycle off and on until the fault is removed.

# **TPS201xY** chip information

This chip, when properly assembled, displays characteristics similar to the TPS201xC. Thermal compression or ultrasonic bonding may be used on the doped aluminum bonding pads. The chip may be mounted with conductive epoxy or a gold-silicon preform.



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# absolute maximum ratings over operating free-air temperature range (unless otherwise noted)

Input voltage range, V <sub>I(IN)</sub> (see Note 1)	$-0.3\ V$ to 7 $V$
Output voltage range, VO (see Note 1)	$-0.3 \text{ V to V}_{I(IN)} + 0.3 \text{ V}$
Input voltage range, V <sub>I</sub> at EN	–0.3 V to 7 V
Continuous output current, IO	internally limited
Continuous total power dissipation	See Dissipation Rating Table
Operating virtual junction temperature range, T <sub>J</sub>	40°C to 125°C
Storage temperature range, T <sub>Stq</sub>	65°C to 150°C
Lead temperature soldering 1,6 mm (1/16 inch) from case for 10 seconds	

<sup>†</sup> Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability. NOTE 1: All voltages are with respect to GND.

#### **DISSIPATION RATING TABLE**

PACKAGE	$T_{\mbox{\scriptsize A}} \leq 25^{\circ}\mbox{\scriptsize C}$ POWER RATING	DERATING FACTOR ABOVE T <sub>A</sub> = 25°C	T <sub>A</sub> = 70°C POWER RATING	T <sub>A</sub> = 125°C POWER RATING
D	725 mW	5.8 mW/°C	464 mW	145 mW
PW	700 mW	5.6 mW/°C	448 mW	140 mW



# recommended operating conditions

		MIN	MAX	UNIT
Input voltage, VI(IN)		2.7	5.5	V
Input voltage, V <sub>I</sub> at EN		0	5.5	V
Continuous output current, IO	TPS2010	0	0.2	
	TPS2011	0	0.6	Α
	TPS2012	0	1	Α
	TPS2013	0	1.5	
Operating virtual junction temper	ature, TJ	-40	125	°C

# electrical characteristics over recommended operating junction temperature range, $V_{I(IN)} = 5.5 \text{ V}$ , $I_O = \text{rated current}$ , $\overline{EN} = 0 \text{ V}$ (unless otherwise noted)

# power switch

PARAMETER		TEST CONDITIONS <sup>†</sup>		TPS2010, TPS2011 TPS2012, TPS2013			UNIT
				MIN	TYP	MAX	
		$V_{I(IN)} = 5.5 \text{ V},$	T <sub>J</sub> = 25°C		75	95	
	On-state resistance	$V_{I(IN)} = 4.5 \text{ V},$	T <sub>J</sub> = 25°C		80	110	mΩ
	On-state resistance	$V_{I(IN)} = 3 V$	T <sub>J</sub> = 25°C		120	175	11152
		$V_{I(IN)} = 2.7 \text{ V},$	$T_J = 25^{\circ}C$		140	215	
	Output lookage ourrent	EN Veren	T <sub>J</sub> = 25°C		0.001	1	
	Output leakage current	$\overline{EN} = V_{I(IN)}$	-40°C ≤ T <sub>J</sub> ≤ 125°C			10	μΑ
	Output vice time	$V_{I(IN)} = 5.5 \text{ V},$	$T_J = 25^{\circ}C,  C_L = 1 \mu\text{F}$		4		
t <sub>r</sub>	Output rise time	$V_{I(IN)} = 2.7 \text{ V},$	$T_J = 25^{\circ}C$ , $C_L = 1 \mu F$		3.8		ms
	Output fall time	$V_{I(IN)} = 5.5 \text{ V},$	$T_J = 25^{\circ}C,  C_L = 1 \mu\text{F}$		3.9		
tf	Output fall time	$V_{I(IN)} = 2.7 \text{ V},$	$T_J = 25^{\circ}C,  C_L = 1 \mu\text{F}$		3.5		ms

<sup>†</sup> Pulse-testing techniques maintain junction temperature close to ambient temperature; thermal effects must be taken into account separately.

# enable input (EN)

PARAMETER		TEST CONDITIONS	TPS20 TPS20	UNIT		
			MIN	TYP	MAX	
	High-level input voltage	2.7 V ≤ V <sub>I(IN)</sub> ≤ 5.5 V	2			V
	Low level input voltage	$4.5 \text{ V} \le \text{V}_{\text{I(IN)}} \le 5.5 \text{ V}$			0.8	V
	Low-level input voltage	2.7 V ≤ V <sub>I(IN)</sub> < 4.5 V			0.4	V
	Input current	$\overline{EN} = 0 \ V \ or \ \overline{EN} = V_{I(IN)}$	-0.5		0.5	μΑ
tPLH	Propagation (delay) time, low-to-high-level output	C <sub>L</sub> = 1 μF			20	me
tPHL	Propagation (delay) time, high-to-low-level output	C <sub>L</sub> = 1 μF			40	ms

#### current limit

PARAMETER	TEST CONDITIONS <sup>†</sup>		TPS20 TPS20	UNIT		
			MIN	TYP	MAX	1
Short-circuit current	V <sub>I</sub> (IN) = 5.5 V, OUT connected to GND, device	TPS2010	0.22	0.4	0.6	
		TPS2011	0.66	1.2	1.8	^
		TPS2012	1.1	2	3	Α
		TPS2013	1.65	2.6	4.5	

<sup>†</sup> Pulse-testing techniques maintain junction temperature close to ambient temperature; thermal effects must be taken into account separately.



# TPS2010, TPS2011, TPS2012, TPS2013 POWER-DISTRIBUTION

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electrical characteristics over recommended operating junction temperature range,  $V_{I(IN)} = 5.5 \text{ V}$ ,  $I_O = \text{rated current}$ ,  $\overline{EN} = 0 \text{ V}$  (unless otherwise noted) (continued)

#### supply current

PARAMETER	TEST CONDITIONS		TPS20	UNIT		
			MIN	TYP	MAX	
Supply current, low-level output	$\overline{EN} = V_{I(IN)}$	T <sub>J</sub> = 25°C		0.015	1	μA
		$-40^{\circ}\text{C} \le \text{T}_{\text{J}} \le 125^{\circ}\text{C}$			10	μΑ
Supply ourront high lovel output	EN = 0 V	T <sub>J</sub> = 25°C		73	100	μA
Supply current, high-level output		-40°C ≤ T <sub>J</sub> ≤ 125°C			100	μΑ

electrical characteristics over recommended operating junction temperature range,  $V_{I(IN)}$  = 5.5 V,  $I_O$  = rated current,  $\overline{EN}$  = 0 V,  $T_J$  = 25°C (unless otherwise noted)

#### power switch

PARAMETER	TEST CONDITIONS†	TPS2010Y, TPS2011Y TPS2012Y, TPS2013Y	UNIT
	1201 201121110110	MIN TYP MAX	
On-state resistance	$V_{I(IN)} = 5.5 V,$	75	
	$V_{I(IN)} = 4.5 V,$	80	0
	$V_{I(IN)} = 3 V$	120	mΩ
	$V_{I(IN)} = 2.7 \text{ V},$	140	
Output leakage current	$\overline{EN} = V_{I(IN)}$	0.001	μΑ
Output rice time	$V_{I(IN)} = 5.5 \text{ V}, \qquad C_L = 1 \mu\text{F}$	4	ma
Output rise time	$V_{I(IN)} = 2.7 \text{ V}, \qquad C_L = 1 \mu F$	3.8	ms
Output fall time	$V_{I(IN)} = 5.5 \text{ V}, \qquad C_L = 1 \mu F$	3.9	ma
Output fall time	$V_{I(IN)} = 2.7 \text{ V}, \qquad C_L = 1 \mu F$	3.5	ms

<sup>†</sup> Pulse-testing techniques maintain junction temperature close to ambient temperature; thermal effects must be taken into account separately.

#### current limit

PARAMETER	TEST CONDITIONS†	TPS201 TPS201	UNIT		
		MIN	TYP	MAX	
Short-circuit current	V <sub>I(IN)</sub> = 5.5 V, OUT connected to GND, Device enabled into short circuit		0.4		А

<sup>†</sup> Pulse-testing techniques maintain junction temperature close to ambient temperature; thermal effects must be taken into account separately.

### supply current

PARAMETER	TEST CONDITIONS	TPS2010 TPS2012	UNIT		
		MIN	TYP	MAX	
Supply current, low-level output	$\overline{EN} = V_{I(IN)}$	(	0.015		μΑ
Supply current, high-level output	EN = 0 V	_	73		μΑ



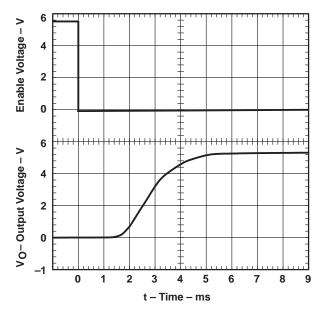


Figure 1. Propagation Delay and Rise Time With 1- $\mu$ F Load,  $V_{I(IN)}$  = 5.5 V

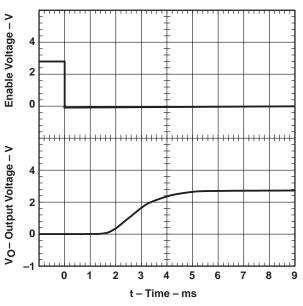


Figure 3. Propagation Delay and Rise Time With 1- $\mu$ F Load,  $V_{I(IN)}$  = 2.7 V

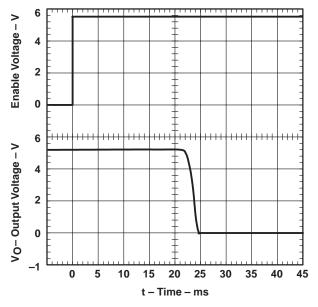


Figure 2. Propagation Delay and Fall Time With 1- $\mu$ F Load, V<sub>I(IN)</sub> = 5.5 V

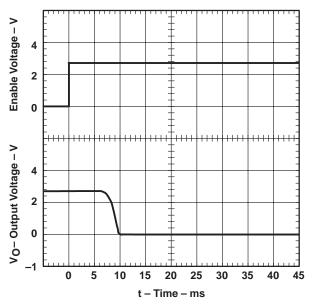


Figure 4. Propagation Delay and Fall Time With 1- $\mu$ F Load,  $V_{I(IN)}$  = 2.7 V

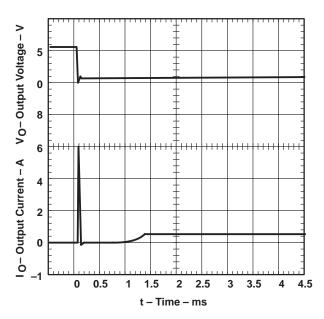


Figure 5. TPS2010, Short-Circuit Current. Short is Applied to Enabled Device,  $V_{I(IN)} = 5.5 \text{ V}$ 

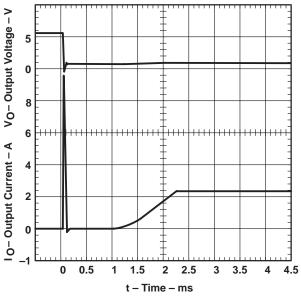


Figure 7. TPS2012, Short-Circuit Current. Short is Applied to Enabled Device,  $V_{I(IN)} = 5.5 \text{ V}$ 

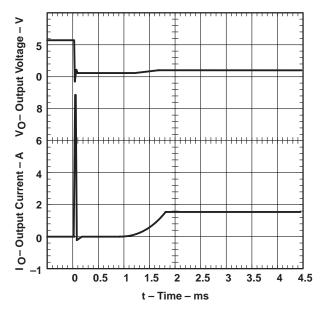


Figure 6. TPS2011, Short-Circuit Current. Short is Applied to Enabled Device,  $V_{I(IN)} = 5.5 \text{ V}$ 

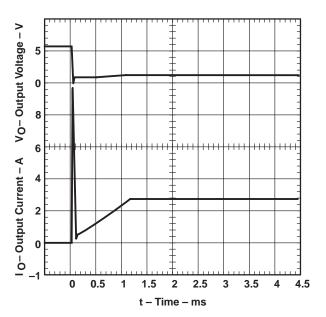


Figure 8. TPS2013 – Short-Circuit Current. Short is Applied to Enabled Device,  $V_{I(IN)} = 5.5 \text{ V}$ 

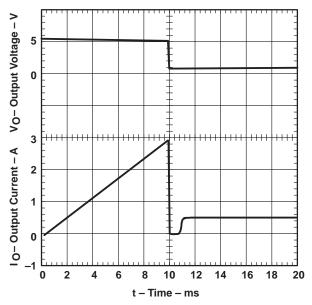


Figure 9. TPS2010 – Threshold Current,  $V_{I(IN)} = 5.5 \text{ V}$ 

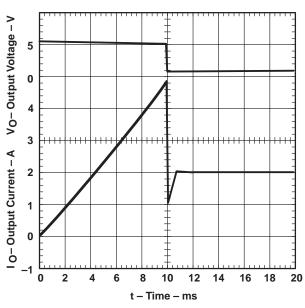


Figure 11. TPS2012 – Threshold Current,  $V_{I(IN)} = 5.5 \text{ V}$ 

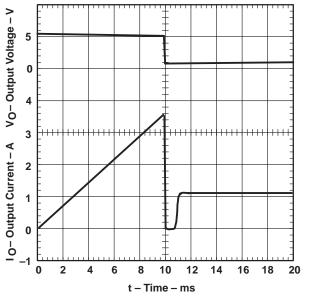


Figure 10. TPS2011 – Threshold Current,  $V_{I(IN)} = 5.5 \text{ V}$ 

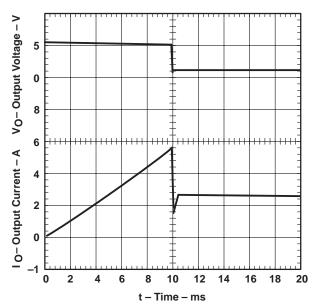


Figure 12. TPS2013 – Threshold Current,  $V_{I(IN)}$  = 5.5 V

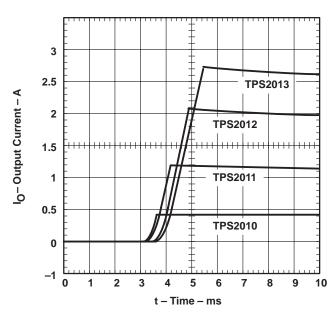


Figure 13. Turned-On (Enabled) Into Short Circuit,  $V_{I(IN)} = 5.5 \text{ V}$ 

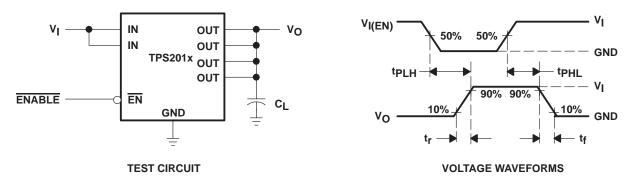
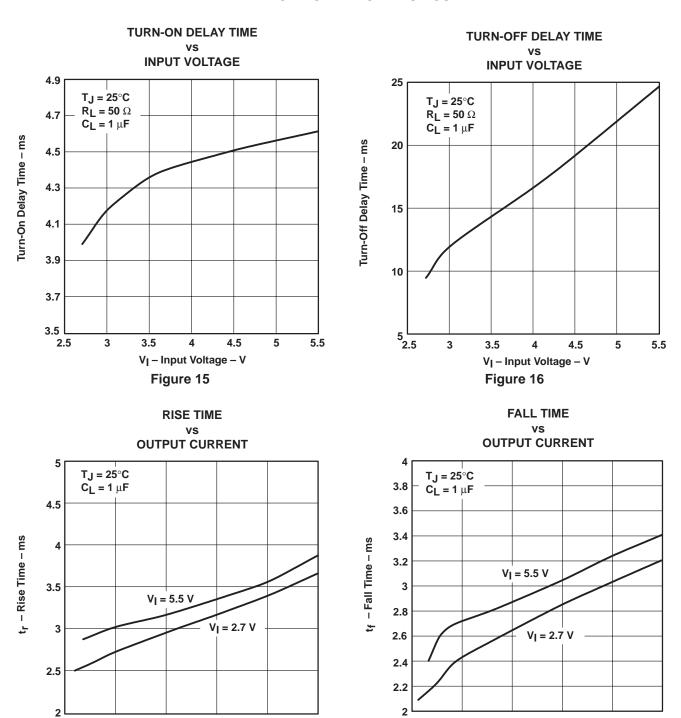


Figure 14. Test Circuit and Voltage Waveforms





1.5

0

0.3

0.9

I<sub>O</sub> – Output Current – A Figure 17 0.3

0.9

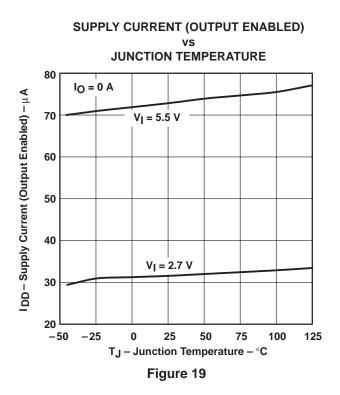
IO - Output Current - A

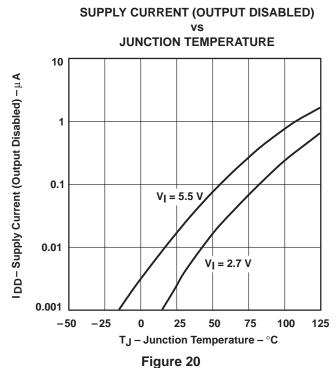
Figure 18

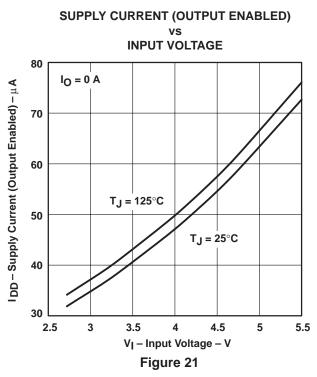
1.2

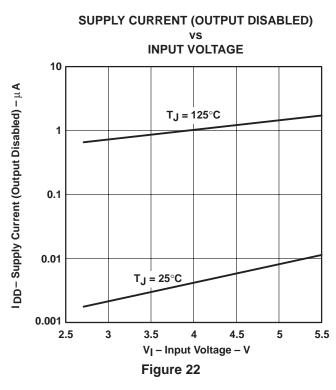
1.5

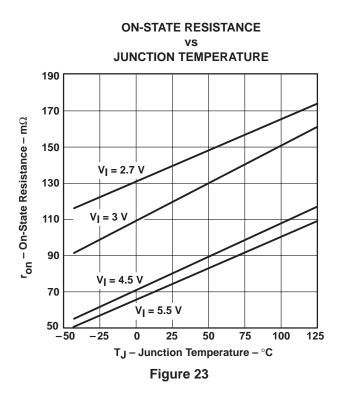
0

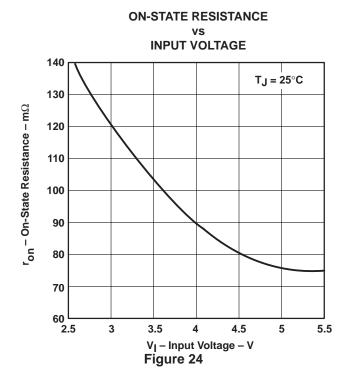


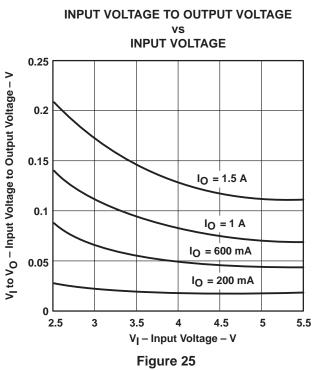


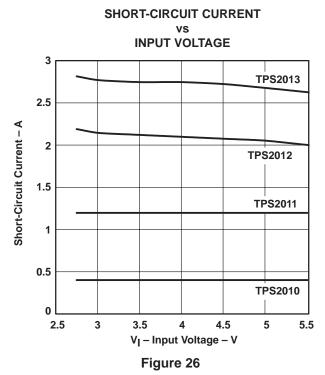


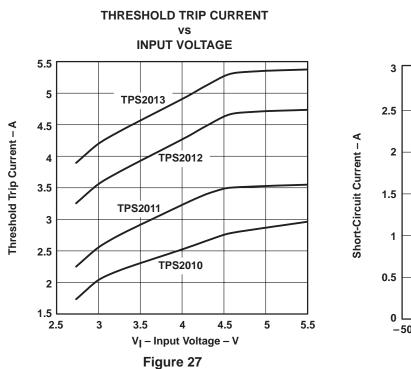


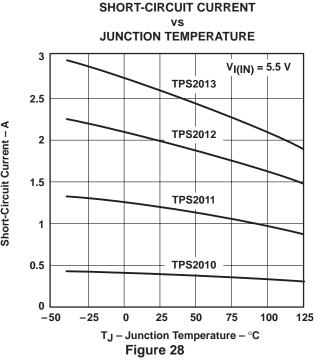












# **APPLICATION INFORMATION**

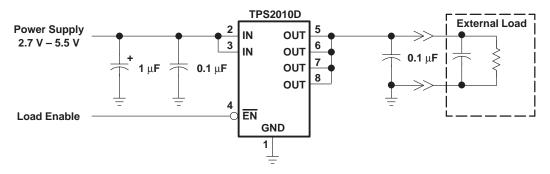


Figure 29. Typical Application

# power supply considerations

The TPS201x family has multiple inputs and outputs, which must be connected in parallel to minimize voltage drop and prevent unnecessary power dissipation.

A 0.047- $\mu F$  to 0.1- $\mu F$  ceramic bypass capacitor between IN and GND, close to the device, is recommended. A high-value electrolytic capacitor is also desirable when the output load is heavy or has large paralleled capacitors. Bypassing the output with a 0.1- $\mu F$  ceramic capacitor improves the immunity of the device to electrostatic discharge (ESD).



#### APPLICATION INFORMATION

#### overcurrent

A sense FET is employed to check for overcurrent conditions. Unlike sense resistors and polyfuses, sense FETs do not increase series resistance to the current path. When an overcurrent condition is detected, the device maintains a constant output current and reduces the output voltage accordingly. Shutdown only occurs if the fault is present long enough to activate thermal limiting.

Three possible overload conditions can occur. In the first condition, the output has been shorted before the device is enabled or before  $V_{I(IN)}$  has been applied (see Figure 30). The TPS201x senses the short and immediately switches into a constant-current output.

Under the second condition, the short occurs while the device is enabled. At the instant the short occurs, very high currents flow for a short time before the current-limit circuit can react (see Figures 5, 6, 7, and 8). After the current-limit circuit has tripped, the device limits normally.

Under the third condition, the load has been gradually increased beyond the recommended operating current. The current is permitted to rise until the current-limit threshold is reached (see Figures 9, 10, 11, and 12). The TPS201x family is capable of delivering currents up to the current-limit threshold without damage. Once the threshold has been reached, the device switches into its constant-current mode.

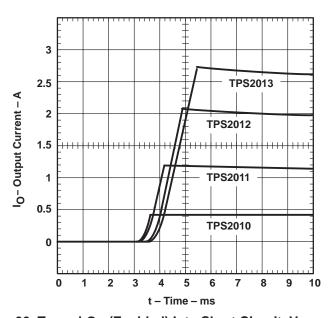


Figure 30. Turned-On (Enabled) Into Short Circuit,  $V_{I(IN)} = 5.5 \text{ V}$ 

#### **APPLICATION INFORMATION**

# power dissipation and junction temperature

The low on resistance of the N-channel MOSFET allows small surface-mount packages, such as SOIC or TSSOP to pass large currents. The thermal resistances of these packages are high compared to that of power packages; it is good design practice to check power dissipation and junction temperature. The first step is to find  $r_{on}$  at the input voltage and operating temperature. As an initial estimate, use the highest operating ambient temperature of interest and read  $r_{on}$  from Figure 23. Next calculate the power dissipation using:

$$P_D = r_{on} \times I^2$$

Finally, calculate the junction temperature:

$$T_J = P_D \times R_{\theta JA} + T_A$$

Where:

 $T_A$  = Ambient temperature

 $R_{\theta,IA}$  = Thermal resistance SOIC = 172°C/W, TSSOP = 179°C/W

Compare the calculated junction temperature with the initial estimate. If they do not agree within a few degrees, repeat the calculation using the calculated value as the new estimate. Two or three iterations are generally sufficient to get a reasonable answer.

#### thermal protection

Thermal protection is provided to prevent damage to the IC when heavy-overload or short-circuit faults are present for extended periods of time. The faults force the TPS201x into its constant current mode, which causes the voltage across the high-side switch to increase; under short-circuit conditions, the voltage across the switch is equal to the input voltage. The increased dissipation causes the junction temperature to rise to dangerously high levels. The protection circuit senses the junction temperature of the switch and shuts it off. The switch remains off until the junction has dropped approximately 20°C. The switch continues to cycle in this manner until the load fault or input power is removed.

#### **ESD** protection

All TPS201x terminals incorporate ESD-protection circuitry designed to withstand a 6-kV human-body-model discharge as defined in MIL-STD-883C. Additionally, the output is protected from discharges up to 12 kV.



#### **IMPORTANT NOTICE**

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