



TDA7401

DIGITALLY CONTROLLED AUDIO PROCESSOR WITH LOUDSPEAKERS EQUALIZER

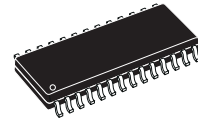
- FOUR HIGH PASS CHANNELS
- ONE STEREO LOW PASS CHANNEL WITH GAIN CONTROL
- DIRECT MUTE PIN
- FULLY PROGRAMMABLE VIA I²C BUS

DESCRIPTION

The TDA7401 is an upgrade of the TDA7435 audioprocessor.

Due to a highly linear signal processing, using CMOS-switching techniques very low distortion and very low noise are obtained.

Second order high pass and low pass filters with programmable corner frequencies provide the loudspeaker equalization.

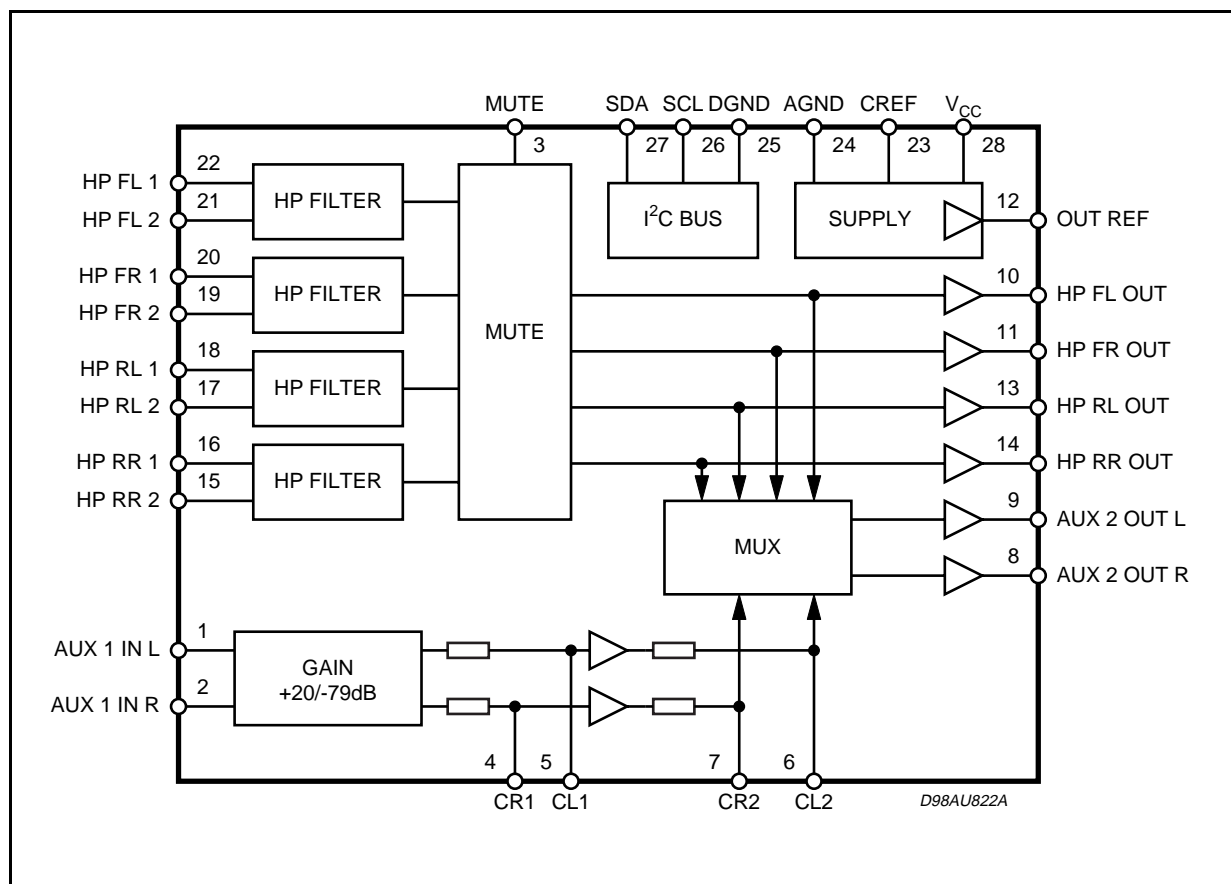


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ORDERING NUMBER: TDA7401D

Very low DC stepping is obtained by using a BICMOS technology.

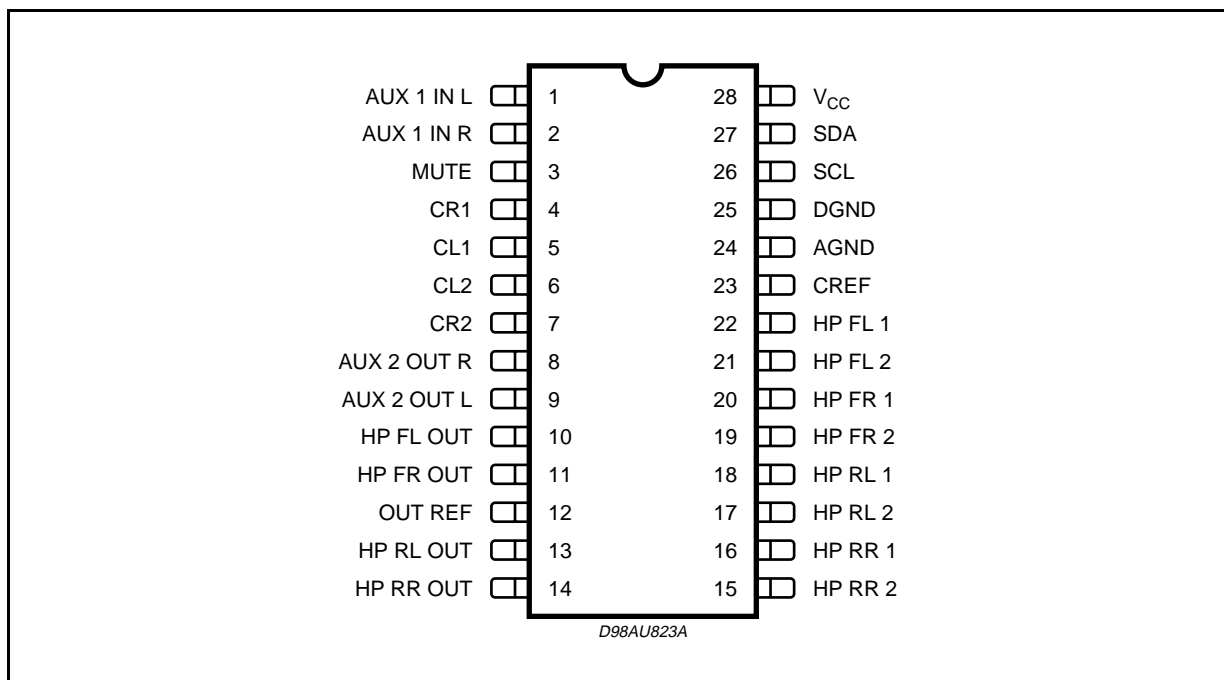
BLOCK DIAGRAM



ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Value	Unit
V_S	Operating Supply Voltage	10.5	V
T_{amb}	Operating Ambient Temperature	-40 to 85	°C
T_{stg}	Storage Temperature Range	-55 to 150	°C

PIN CONNECTION



THERMAL DATA

Symbol	Parameter	Value	Unit
$R_{th\ j-amb}$	Thermal Resistance Junction-pins	65	°C/W

QUICK REFERENCE DATA

Symbol	Parameter	Min.	Typ.	Max.	Unit
V_S	Supply Voltage	6	9	10.2	V
V_{CL}	Max. input signal handling	2.1	2.6		V _{rms}
THD	Total Harmonic Distortion $V = 1V_{rms}$ $f = 1KHz$		0.01	0.08	%
S/N	Signal to Noise Ratio		106		dB
S_C	Channel Separation $f = 1KHz$	-80	100		dB
V_{REF}	Reference Voltage Output (pin 12)	4.2	4.5	4.8	V

ELECTRICAL CHARACTERISTICS ($V_S = 9V$; $R_L = 10K\Omega$; $R_g = 50\Omega$; $T_{amb} = 25^\circ C$; all gains = 0dB; $f = 1KHz$. Refer to the test circuit, unless otherwise specified.)

Symbol	Parameter	Test Condition	Min.	Typ.	Max.	Unit
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INPUT STAGE: AUX1

R_I	Input Resistance		37.5	50	62.5	$K\Omega$
V_{CL}	Clipping Level	$d \leq 0.3\%$	2.1	2.6		V_{RMS}
S_I	Input Separation		80	100		dB

GAIN CONTROL

G_{MAX}	Maximum Input Gain			20		dB
A_{MAX}	Maximum Attenuation			79		dB
A_{STEP}	Step Resolution		0.5	1	1.5	dB
E_A	Attenuation Set Error	$G = -20$ to $+20$ dB	-1.25	0	+1.25	dB
		$G = -60$ to -20 dB	-4		3	dB
E_T	Tracking Error				2	dB
V_{DC}	DC Steps	Adjacent Attenuation Steps		0.1	3	mV
		From 0dB to G_{MIN}		0.5	5	mV

AUDIO OUTPUT (Pin 8 - 9, 10 - 14)

V_{clip}	Clipping Level	$d = 0.3\%$	2.1	2.6		V_{rms}
R_L	Output Load Resistance	AC coupled	2			$K\Omega$
R_O	Output Impedance			30	100	Ω
V_{DC}	DC Voltage Level		4.2	4.5	4.8	V

STAGE: HP FILTER

R_1	Resistance at pin HP1	HIGHPASS BYTE = XXXX1000	127.5	170	212.5	$K\Omega$
R_2	Resistance at pin HP2		1			$M\Omega$
V_{CL}	Clipping Level	$d \leq 0.3\%$	2.1	2.6		V_{rms}

MUTE

A_{MUTE}	Mute Attenuation		80	100		dB
V_{THM}	Mute Threshold		1.2	1.7	2.2	V
R_{INT}	Pullup Resistor (pin 3)	(note 1)	37.5	50	62.5	$K\Omega$

GENERAL

V_{CC}	Supply Voltage		6	9	10.2	V
I_{CC}	Supply Current		7	8	9	mA
PSRR	Power Supply Rejection Ratio	$f = 1KHz$	60	70		dB
e_{NO}	Output Noise	Non Inverting Output Muted ($B = 20$ to $20kHz$ flat)		3.5	15	μV
		All Gains 0dB ($B = 20$ to $20kHz$ flat)		5	15	μV
S/N	Signal to Noise Ratio	All Gains = 0dB; $V_O = 1V_{rms}$		106		dB
S_C	Channel Separation		80	100		dB
d	Distortion	$V_{IN} = 1V$		0.01	0.08	%

BUS INPUTS

V_{IL}	Input Low Voltage				0.8	V
V_{IH}	Input High Voltage		2.5			V
I_{IN}	Input Current	$V_{IN} = 0.4V$	-5		5	μA
V_O	Output Voltage SDA Acknowledge	$I_O = 1.6mA$	0.1		0.4	V

Note 1: Internal pullup resistor to 3.3V; "LOW" = mute active

Figure 1. HP Filter

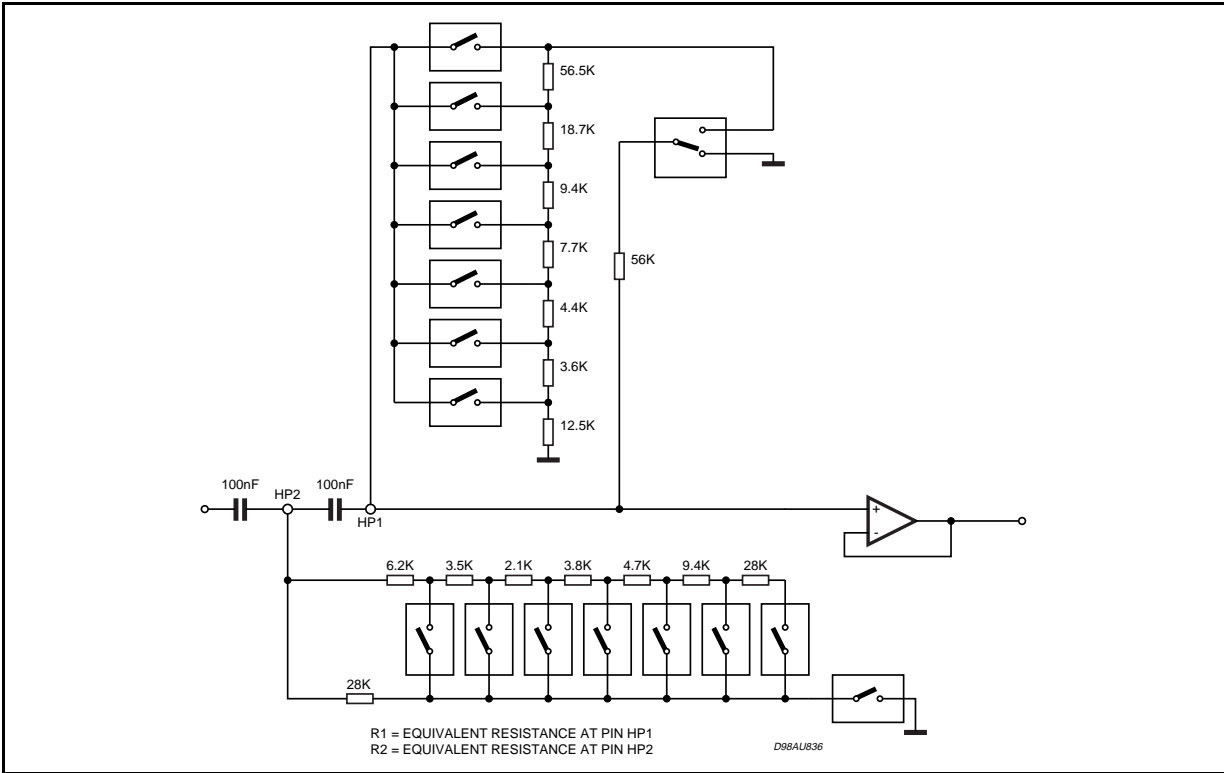
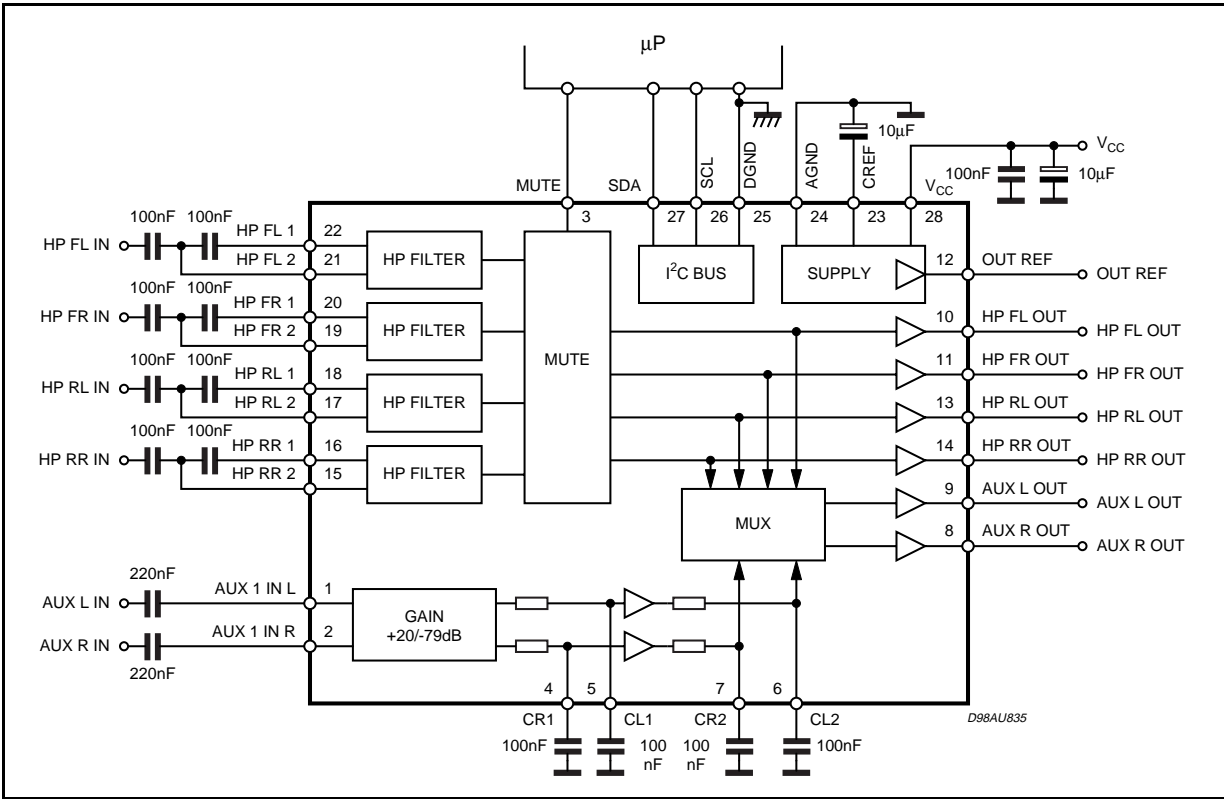


Figure 2. Application Circuit



I²C BUS INTERFACE

Data transmission from microprocessor to the TDA7401 and viceversa takes place thru the 2 wires I²C BUS interface, consisting of the two lines SDA and SCL (pull-up resistors to positive supply voltage must be externally connected).

Data Validity

As shown in fig. 2, the data on the SDA line must be stable during the high period of the clock. The HIGH and LOW state of the data line can only change when the clock signal on the SCL line is LOW.

Start and Stop Conditions

As shown in fig.3 a start condition is a HIGH to LOW transition of the SDA line while SCL is HIGH. The stop condition is a LOW to HIGH transition of the SDA line while SCL is HIGH. A STOP conditions must be sent before each START condition.

Byte Format

Every byte transferred to the SDA line must con-

tain 8 bits. Each byte must be followed by an acknowledge bit. The MSB is transferred first.

Acknowledge

The master (μ P) puts a resistive HIGH level on the SDA line during the acknowledge clock pulse (see fig. 4). The peripheral (audioprocessor) that acknowledges has to pull-down (LOW) the SDA line during the acknowledge clock pulse, so that the SDA line is stable LOW during this clock pulse.

The audioprocessor which has been addressed has to generate an acknowledge after the reception of each byte, otherwise the SDA line remains at the HIGH level during the ninth clock pulse time. In this case the master transmitter can generate the STOP information in order to abort the transfer.

Transmission without Acknowledge

Avoiding to detect the acknowledge of the audioprocessor, the μ P can use a simpler transmission: simply it waits one clock without checking the slave acknowledging, and sends the new data.

This approach of course is less protected from misworking and decreases the noise immunity.

Figure 3. Data Validity on the I²C BUS

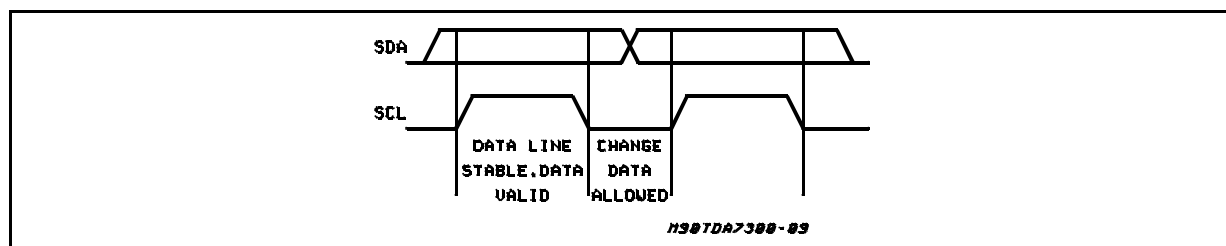


Figure 4. Timing Diagram of I²C BUS

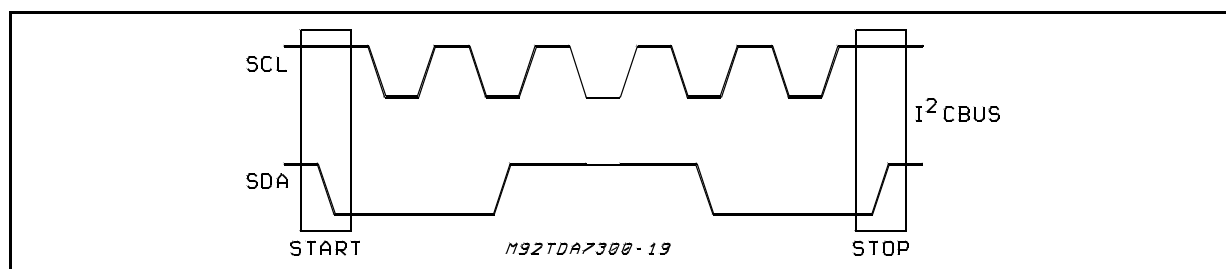
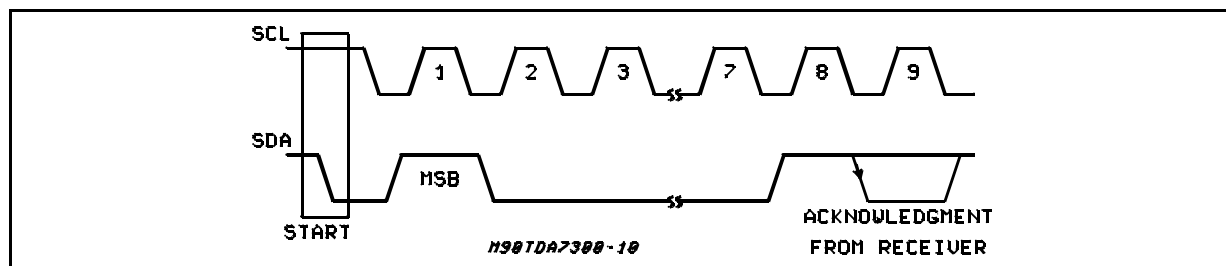


Figure 5. Acknowledge on the I²C BUS



SOFTWARE SPECIFICATION

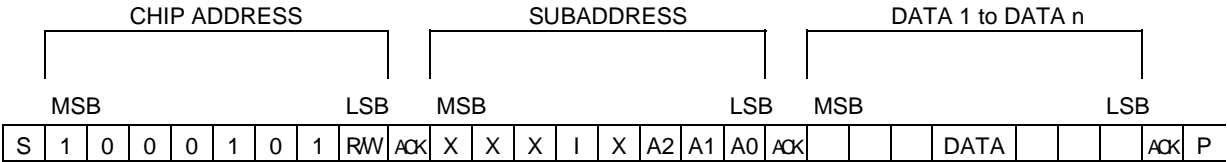
Interface Protocol

The interface protocol comprises:

- A start condition (s)
- A chip address byte,(the LSB bit determines

read/write transmission)

- A subaddress byte.
- A sequence of data (N-bytes + acknowledge)
- A stop condition (P)



ACK = Acknowledge

S = Start

P = Stop

I = Auto Increment

X = Not used

AUTO INCREMENT

If bit I in the subaddress byte is set to "1", the autoincrement of the subaddress is enabled

SUBADDRESS (receive mode)

MSB					LSB			FUNCTION
X	X	X	I	X	D2	D1	D0	
					0	0	0	Not used
					0	0	1	Mode
					0	1	0	Gain AUX 1 L
					0	1	1	Gain AUX 1 R
					1	0	0	High Pass Filter FL
					1	0	1	High Pass Filter FR
					1	1	0	High Pass Filter RL
					1	1	1	High Pass Filter RR

MODE

MSB							LSB	FUNCTION
D7	D6	D5	D4	D3	D2	D1	D0	
						0	X	High Pass Mute ON
						1		High Pass Mute OFF
					0			AUX1 Input Mute ON
					1			AUX1 Input Mute OFF
				0				AUX2 Inverted Output
				1				AUX2 Non Inv. Output
								AUX 2 Output Selection
		0	0					High Pass Filter Front
		0	1					High Pass Filter Rear
		1	0					Aux 1 Input
		1	1					Mute
								AUX1 Low Pass Filter (C1 = C2 = 100nF)
0	0							Flat
0	1							120Hz
1	0							80Hz
1	1							50Hz

GAIN AUX1L, AUX1R

MSB							LSB	GAIN AUX1L, R
D7	D6	D5	D4	D3	D2	D1	D0	
1	0	0	1	1	1	1	1	+31dB
:	:	:	:	:	:	:	:	:
1	0	0	1	0	0	0	1	+17dB
1	0	0	1	0	0	0	0	+16dB
1	0	0	0	1	1	1	1	+15dB
:	:	:	:	:	:	:	:	:
1	0	0	0	0	0	0	1	+1dB
1	0	0	0	0	0	0	0	0dB
0	0	0	0	0	0	0	0	0dB
0	0	0	0	0	0	0	1	-1dB
:	:	:	:	:	:	:	:	:
0	0	0	0	1	1	1	1	-15dB
0	0	0	1	0	0	0	0	-16dB
:	:	:	:	:	:	:	:	:
0	1	0	0	1	1	1	0	-78dB
0	1	0	0	1	1	1	1	-79dB
X	1	1	X	X	X	X	X	Mute

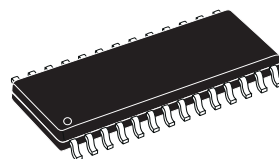
Note: Is is not recommended to use a gain more than 20dB for system performance reason. In general, the max. gain should be limited by software to the maximum value, which is needed for the system.

HIGH PASS FILTERS

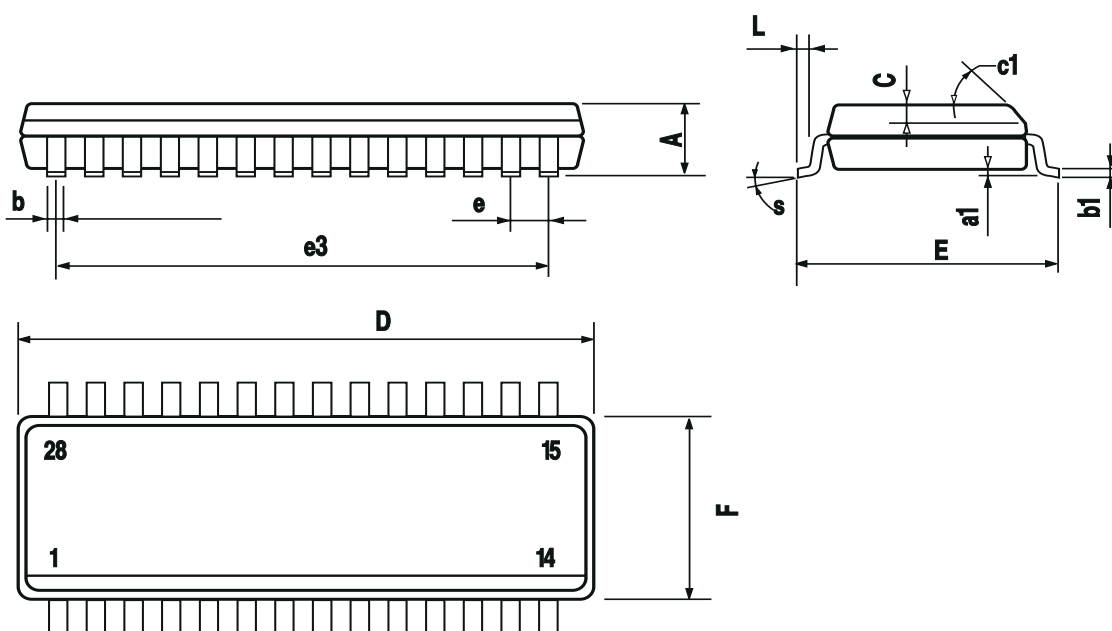
MSB								LSB	FL, FR, RL, RR
D7	D6	D5	D4	D3	D2	D1	D0		
									2nd order HP Filter Mode (C1 = C2 = 100nF)
X	X	X	X	0	0	0	0		$f_c = 40\text{Hz}$
				0	0	0	1		$f_c = 60\text{Hz}$
				0	0	1	0		$f_c = 80\text{Hz}$
				0	0	1	1		$f_c = 100\text{Hz}$
				0	1	0	0		$f_c = 120\text{Hz}$
				0	1	0	1		$f_c = 150\text{Hz}$
				0	1	1	0		$f_c = 180\text{Hz}$
				0	1	1	1		$f_c = 220\text{Hz}$
									First order HP Flat Mode
				1	0	0	0		$f_c = 9\text{Hz}$

DIM.	mm			inch		
	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
A			2.65			0.104
a1	0.1		0.3	0.004		0.012
b	0.35		0.49	0.014		0.019
b1	0.23		0.32	0.009		0.013
C		0.5			0.020	
c1	45° (typ.)					
D	17.7		18.1	0.697		0.713
E	10		10.65	0.394		0.419
e		1.27			0.050	
e3		16.51			0.65	
F	7.4		7.6	0.291		0.299
L	0.4		1.27	0.016		0.050
S	8° (max.)					

OUTLINE AND MECHANICAL DATA



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