

DATA SHEET



TDA10045H DVB-T channel receiver

Product specification
Supersedes data of 2000 Jun 21
File under Integrated Circuits, IC02

2001 Nov 08

DVB-T channel receiver**TDA10045H****FEATURES**

- 2 and 8 kbytes Coded Orthogonal Frequency Division Multiplexer (COFDM) demodulator (fully DVB-T compliant: ETSI 300-744)
- All modes supported, including hierarchical modes
- Fully automatic transmission parameters detection (including Fast Fourier Transformer (FFT) size and guard interval)
- Digital Signal Processor (DSP) based synchronization (software can be upgraded on the fly)
- No extra-host software required
- On-chip 10-bit Analog-to-Digital Converter (ADC)
- 2nd or 1st IF variable analog input
- Only fundamental crystal oscillator required (4 MHz typical ± 100 ppm)
- 6, 7 and 8 MHz channels with the same crystal
- Pulse killer algorithm to protect against impulse noise
- Digital frequency correction (± 90 kHz)
- Frequency offset ($\pm 1/6$ MHz) automatic estimator to speed-up the scan
- RF tuner input power measurement
- Parallel or serial transport stream interface
- BER measurement (before and after Viterbi decoder)
- Signal-to noise ratio estimation
- Constellation, CSI and channel frequency response outputs
- TPS bits I²C-bus readable (including spare ones)
- Controllable dedicated I²C-bus for the tuner (5 V tolerant)
- 3 low frequency spare DACs and 2 spare inputs
- CMOS 0.2 μ m technology.

APPLICATIONS

- DVB-T fully compatible
- Digital data transmission using COFDM modulation.

**GENERAL DESCRIPTION**

The TDA10045H is a single-chip channel receiver for 2 and 8 kbytes COFDM modulated signals based on the ETSI specification (ETSI 300-744). The device interfaces directly to an IF signal, which could be either 1st or 2nd IF and integrates a 10-bit Analog-to-Digital Converter (ADC), a Numerically Controlled Oscillator (NCO) and a Phase-Locked Loop (PLL), simplifying external logic requirements and limiting system costs.

The TDA10045H performs all the COFDM demodulation tasks from IF signal to the MPEG-2 transport stream. An internal DSP core manages the synchronization and the control of the demodulation process, and implements specially developed software for robustness against co-channel and adjacent channel interference, to deal with Single Frequency Network (SFN) echo situations, and to assist in a very fast scan of the bandwidth. After baseband conversion and FFT demodulation, the channel frequency response is estimated, which is based on the scattered pilots, and filtered in both time and frequency domains. This estimation is used as a correction on the signal, carrier by carrier. A common phase error and estimator is used to deal with the tuner phase noise. The Forward Error Correction (FEC) decoder is automatically synchronized by the frame synchronization algorithm that uses the TPS information included in the modulation. An embedded 'pulse killer' algorithm enables the bad effects of short and strong impulsive noise interference that could be caused by electrical domestic devices and/or car traffic to be greatly reduced.

This device is controlled via an I²C-bus (master). The chip provides 2 switchable I²C-buses derived from the master: a tuner I²C-bus to be disconnected from the I²C-bus master when not necessary and an EEPROM I²C-bus. The DSP software code can be fed to the chip via the master I²C-bus or via the dedicated EEPROM I²C-bus.

Designed in 0.2 μ m CMOS technology and housed in a 100 pin QFP package, the TDA10045H operates over the commercial temperature range.

DVB-T channel receiver**TDA10045H**

ORDERING INFORMATION

| TYPE NUMBER | PACKAGE | | VERSION |
|----------------|---------|--|----------|
| | NAME | DESCRIPTION | |
| TDA10045H | QFP100 | plastic quad flat package; 100 leads (lead length 1.95 mm); body 14 × 20 × 2.8 mm | SOT317-2 |

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BLOCK DIAGRAM

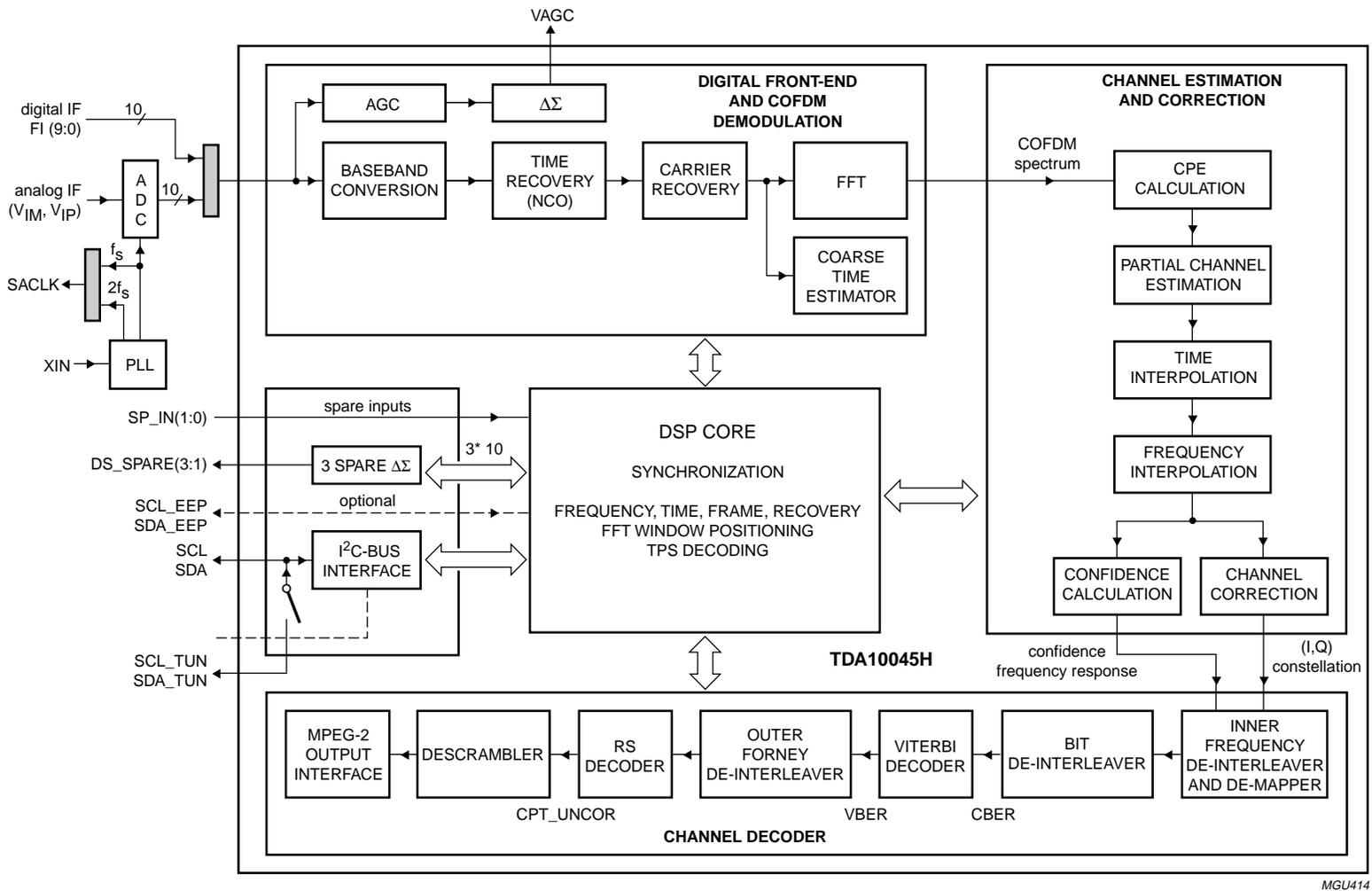


Fig.1 Block diagram.

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PINNING

| SYMBOL | PIN | TYPE | DESCRIPTION |
|--------------------|-----------|-------------------|--|
| V _{DDD33} | 1 | – | digital supply voltage for the pads (3.3 V typ.) |
| V _{SSD} | 2 | – | digital ground supply (0 V) |
| DS_SPARE3 | 3 | O | spare delta-sigma output; managed by the DSP to generate an analog level (after a RC low-pass filter) |
| VAGC | 4 | O | output value from the Delta-Sigma modulator, used to control a log-scaled amplifier (after analog filtering) |
| SCL_EEP | 5 | O | extra I ² C-bus clock to download DSP code from an external EEPROM (optional mode); can be connected to the master I ² C-bus |
| V _{DDD33} | 6 | – | digital supply voltage for the pads (3.3 V typ.) |
| V _{SSD} | 7 | – | digital ground supply (0 V) |
| SDA_EEP | 8 | I/OD | extra I ² C-bus data bus to download DSP code from an external EEPROM (optional mode). It can be connected to the master I ² C-bus; this pin is open-drain which requires an external pull-up resistor (to V _{DDD33} or V _{DDD50}), even if not used. |
| SCL_TUN | 9 | OD ⁽¹⁾ | tuner I ² C-bus serial clock signal; this signal is derived from the master SCL and is open-drain which requires an external pull-up resistor (to V _{DDD33} or V _{DDD50}), even if not used |
| SDA_TUN | 10 | I/OD | tuner I ² C-bus serial data signal; this signal is derived from the master SDA and is open-drain which requires an external pull-up resistor (to V _{DDD33} or V _{DDD50}), even if not used |
| SCL | 11 | I ⁽²⁾ | I ² C-bus master serial clock; up to 700 kbit/s |
| SDA | 12 | I/OD | I ² C-bus master serial data input/output, open-drain I/O pad, which requires an external pull-up resistor (to V _{DDD33} or V _{DDD50}) |
| n.c. | 13 | – | not connected |
| CLR# | 14 | I ⁽²⁾ | asynchronous reset signal; active LOW |
| EEPADDR | 15 | I ⁽²⁾ | EEPADDR is the LSB of the I ² C-bus address of the EEPROM. The MSBs are internally set to 101000. Therefore the complete I ² C-bus address of the EEPROM is (MSB to LSB): 1, 0, 1, 0, 0, 0, EEPADDR. |
| SADDR[1:0] | 16 and 17 | I ⁽²⁾ | SADDR[1:0] are the 2 LSBs of the I ² C-bus address of the TDA10045; the MSBs are internally set to 00010; therefore the complete I ² C-bus address of the TDA10045 is (MSB to LSB): 0, 0, 0, 1, 0, SADDR[1] and SADDR[0] |
| V _{DDD18} | 18 | – | digital supply voltage for the core (1.8 V typ.) |
| V _{SSD} | 19 | – | digital ground supply (0 V) |
| TM[3:0] | 20 to 23 | I ⁽²⁾ | test mode bus; for test purpose; must be set to '0000' |
| SCAN_EN | 24 | I ⁽²⁾ | scan enable for production test; connected to GND |
| V _{DDD50} | 25 | – | digital supply voltage (5 V typ.); can be set to 3.3 V (with caution) if the 5 V tolerant I/O is not required |
| V _{SSD} | 26 | – | digital ground supply (0 V) |
| DWNLOAD | 27 | I ⁽²⁾ | processor control, boot mode; if set to logic 0, the DSP downloads the software from an external EEPROM on the dedicated I ² C-bus (pins SDA_EEP and SCL_EEP). If set to logic 1 the software is downloaded in the I ² C-bus register CODE_IN from the host; in this case the external EEPROM is not needed. |
| SP_IN[1:0] | 28 and 29 | I ⁽²⁾ | spare inputs |

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| SYMBOL | PIN | TYPE | DESCRIPTION |
|--------------------|----------|-------------------|--|
| FFT_WIN | 30 | I/O | output or input signal indicating the start of the active data; equals 1 during complex sample 0 of the active FFT block; can be used to synchronize 2 chips |
| V _{DDD33} | 31 | – | digital supply voltage for the pads (3.3 V typ.) |
| V _{SSD} | 32 | – | digital ground supply (0 V) |
| SACLK | 33 | O | sampling frequency output; this output clock can be fed to an external (10-bit) ADC as a sampling clock; SACLK can also provide twice the sampling clock |
| FI[9:5] | 34 to 38 | I/O | input data from an external ADC, FI must be tied to ground when unused, positive notation (from 0 to 1023) or twos complement notation (from –512 to +511). In internal ADC mode, these outputs can be used to monitor extra demodulator output signals (constellation or frequency response). |
| V _{DDD18} | 39 | – | digital supply voltage for the core (1.8 V typ.) |
| V _{SSD} | 40 | – | digital ground supply (0 V) |
| FI[4:0] | 41 to 45 | IO | input data from an external ADC, FI must be tied to ground when unused, positive notation (from 0 to 1023) or twos complement notation (from –512 to +511). In internal ADC mode, these outputs can be used to monitor extra demodulator output signals (constellation or frequency response). |
| V _{DDD50} | 46 | – | digital supply voltage (5 V typ.); can be set to 3.3 V (with caution) if 5 V tolerant I/O is not required |
| V _{SSD} | 47 | – | digital ground supply (0 V) |
| IT | 48 | OD ⁽¹⁾ | interrupt line; this output interrupt line can be configured by the I ² C-bus interface. This pin is an open-drain output and therefore requires an external pull-up resistor (to V _{DDD33} or V _{DDD50}). |
| FEL | 49 | OD ⁽¹⁾ | front-end lock; FEL is an open-drain output and therefore requires an external pull-up resistor (to V _{DDD33} or V _{DDD50}) |
| n.c. | 50 | – | not connected |
| n.c. | 51 | – | not connected |
| TRSTN | 52 | I ⁽²⁾ | asynchronous reset signal for boundary scan; connected to GND if not used |
| TMS | 53 | I ⁽²⁾ | mode programming signal for boundary scan; connected to GND if not used |
| TDI | 54 | I ⁽²⁾ | input port for boundary scan; connected to GND if not used |
| TCK | 55 | I ⁽²⁾ | clock signal for boundary scan; connected to GND if not used |
| TDO | 56 | O | output port for boundary scan; not connected if not used |
| V _{DDD18} | 57 | – | digital supply voltage for the core (1.8 V typ.) |
| V _{SSD} | 58 | – | digital ground supply (0 V) |
| DS_SPARE2 | 59 | O | spare delta-sigma output; managed by the DSP or by an I ² C-bus register to generate an analog level (after a RC low-pass filter) |
| DS_SPARE1 | 60 | O | spare delta-sigma output; managed by the DSP to handle a low frequency DAC (automatic first stage tuner AGC measurement or 2nd AGC loop control as examples) |
| V _{DDD33} | 61 | – | digital supply voltage for the pads (3.3 V typ.) |
| V _{SSD} | 62 | – | digital ground supply (0 V) |
| UNCOR | 63 | O | RS error flag, active HIGH on one RS packet if the RS decoder fails to correct the errors |
| PSYNC | 64 | O | pulse synchro; this output signal goes HIGH on a rising edge of OCLK when a synchro byte is provided, then goes LOW until the next synchro byte |

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| SYMBOL | PIN | TYPE | DESCRIPTION |
|-----------------------|----------|------------------|--|
| DEN | 65 | O | output data validation signal; active HIGH during the valid and regular data bytes |
| OCLK | 66 | O | output clock; OCLK is the output clock for the parallel DO[7:0] outputs |
| DO[7:5] | 67 to 69 | O | output data carrying the current sample of the current MPEG2 packet (188 bytes), delivered on the rising edge of OCLK by default when the serial mode is selected. The output data is delivered by DO[0]. |
| V _{DDD18} | 70 | – | digital supply voltage for the core (1.8 V typ.) |
| V _{SSD} | 71 | – | digital ground supply (0 V) |
| DO[4:0] | 72 to 76 | O | output data carrying the current sample of the current MPEG2 packet (188 bytes), delivered on the rising edge of OCLK by default when the serial mode is selected. The output data is delivered by DO[0]. |
| V _{DDD33} | 77 | – | digital supply voltage for the pads (3.3 V typ.) |
| V _{SSD} | 78 | – | digital ground supply (0 V) |
| XIN | 79 | I ⁽²⁾ | crystal oscillator input pin |
| XOUT | 80 | O | crystal oscillator output pin; typically a fundamental crystal oscillator is connected between pins XIN and XOUT |
| V _{DDD18} | 81 | – | digital supply voltage for the core (1.8 V typ.) |
| V _{SSD} | 82 | – | digital ground supply (0 V) |
| n.c. | 83 | – | not connected |
| V _{CCD(PLL)} | 84 | – | power supply input for the digital circuits of the PLL module (1.8 V typ.) |
| DGND | 85 | – | ground return for the digital circuits of the PLL module |
| n.c. | 86 | – | not connected |
| PPLGND | 87 | – | ground return for the analog circuits of the PLL module |
| V _{CCA(PLL)} | 88 | – | power supply input for the analog circuits of the PLL module (3.3 V typ.) |
| V _{SSA3} | 89 | – | ground return for the analog circuits |
| V _{DDA3} | 90 | – | power supply input for the analog circuits; the DC voltage should be 3.3 V |
| V _{IP} | 91 | – | positive input to the ADC; this pin is DC biased to half supply through an internal resistor divider (2 × 20 kΩ resistors). In order to remain in the range of the ADC, the voltage difference between pins V _{IP} and V _{IM} should be between –0.5 and +0.5 V. |
| V _{IM} | 92 | – | negative input to the ADC; this pin is DC biased to half supply to remain in the range of the ADC, the voltage difference between pins V _{IP} and V _{IM} should be between –0.5 and +0.5 V through an internal resistor divider (2 × 20 kΩ resistors) |
| V _{ref(neg)} | 93 | – | negative reference voltage for the ADC |
| V _{ref(pos)} | 94 | – | positive reference voltage for the ADC |
| V _{DDA3} | 95 | – | power supply input for the analog circuits; the DC voltage should be 3.3 V |
| V _{SSA3} | 96 | – | ground return for analog circuits |
| V _{SSA2} | 97 | – | ground return for the analog clock drivers |
| V _{DDA2} | 98 | – | power supply input for the analog clock drivers; the DC voltage should be 3.3 V |
| V _{SSA1} | 99 | – | ground return for the digital switching circuitry |
| V _{DDD1} | 100 | – | power supply input for the digital switching circuitry; sensitive to the supply noise; the DC voltage should be 1.8 V |

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Notes

1. OD are open-drain outputs, so they must be connected to a pull-up resistor to either V_{DD33} or V_{DD50}
2. All inputs (I) are TTL, 5 V tolerant, (if V_{DD50} is set to 5 V).
3. Foundry test I/O inputs must be connected to GND.

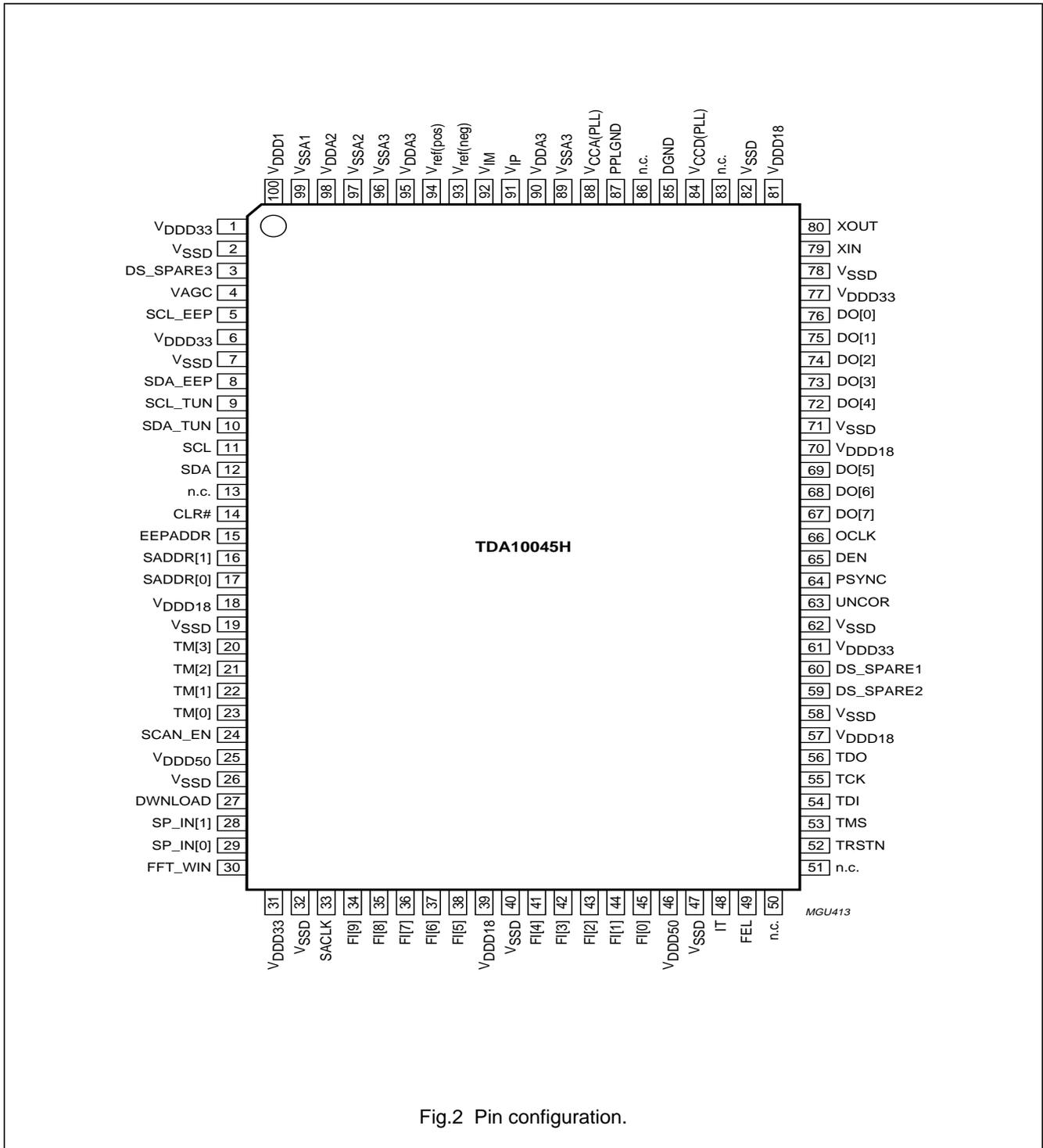


Fig.2 Pin configuration.

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LIMITING VALUES

In accordance with the Absolute Maximum Rate System (IEC 60134); note 1.

| SYMBOL | PARAMETER | MIN. | MAX. | UNIT |
|--------------------|-------------------------------------|------|------|------|
| V _{DDD18} | digital supply voltage for the core | -0.5 | +2.1 | V |
| V _{DDD33} | digital supply voltage for the pads | -0.5 | +3.8 | V |
| V _I | DC input voltage | -0.5 | +5.5 | V |
| I _I | DC input current | - | ±20 | mA |
| T _{lead} | lead temperature | - | 300 | °C |
| T _{stg} | storage temperature | -65 | +150 | °C |
| T _j | junction temperature | - | 150 | °C |
| T _{amb} | ambient temperature | 0 | 70 | °C |

Note

1. Stresses above the Absolute Maximum Ratings may cause permanent damage to the device. Exposure to Absolute Maximum Ratings conditions for extended periods may affect device reliability.

THERMAL CHARACTERISTICS

| SYMBOL | PARAMETER | CONDITIONS | VALUE | UNIT |
|----------------------|---|-------------|-------|------|
| R _{th(j-a)} | thermal resistance from junction to ambient | in free air | tbf | K/W |

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CHARACTERISTICS

| SYMBOL | PARAMETER | CONDITIONS | MIN. | TYP. | MAX. | UNIT |
|---------------------------------------|--|---|--------------|------|-------------------------|------|
| Core and pads | | | | | | |
| V _{DDD33} | digital supply voltage for the pads | V _{DDD} = 3.3 V ± 10% | 2.97 | 3.3 | 3.63 | V |
| V _{DDD18} | digital supply voltage for the core | V _{DDD} = 1.8 V ± 5% | 1.7 | 1.8 | 1.9 | V |
| V _{DDD50} | 5 V supply voltage | only for 5 V requirements; note 1 | 4.75 | 5.0 | 5.25 | V |
| T _{amb} | ambient temperature | | 0 | – | 70 | °C |
| V _{IH} | HIGH-level input voltage | TTL input; note 2 | 2 | – | V _{DDD50} | V |
| V _{IL} | LOW-level input voltage | TTL input | 0 | – | 0.8 | V |
| V _{OH} | HIGH-level output voltage | I _{OH} = ±2 mA | 2.4 | – | – | V |
| V _{OL} | LOW-level output voltage | I _{OL} = ±2 mA | – | – | 0.4 | V |
| C _i | input capacitance | | – | – | 5 | pF |
| C _o | output capacitance | | – | – | 5 | pF |
| PLL | | | | | | |
| V _{CCD(PLL)} | digital PLL supply voltage | V _{CCD} = 1.8 V ± 5% | 1.7 | 1.8 | 1.9 | V |
| V _{CCA(PLL)} | analog PLL supply voltage | V _{CCA} = 3.3 V ± 10% | 2.97 | 3.3 | 3.63 | V |
| ADC | | | | | | |
| V _{DDD1} | digital ADC supply voltage | V _{DDD} = 1.8 V ± 5% | 1.7 | 1.8 | 1.9 | V |
| V _{DDA2} , V _{DDA3} | analog ADC supply voltage | V _{DDA} = 3.3 V ± 10% | 2.97 | 3.3 | 3.63 | V |
| V _{i(ADC)} | analog ADC inputs pins V _{IP} and V _{IM} | | –0.5 | – | V _{DDD3} + 0.5 | V |
| V _i | signal input | I _R = V _{IP} – V _{IM} ; depending on SW register | –0.5 to –1.0 | – | +0.5 to +1.0 | V |
| V _{ref(pos)} | positive reference voltage | with SW register = 11 | 1.95 | 2.15 | 2.35 | V |
| V _{ref(neg)} | negative reference voltage | with SW register = 11 | 0.95 | 1.15 | 1.35 | V |
| V _{i(offset)} | input offset voltage | | –25 | – | +25 | mV |
| R _i | input resistance pin V _{IP} or V _{IM} | | – | 10 | | kΩ |
| C _i | input capacitance pin V _{IP} or V _{IM} | | – | 5 | 10 | pF |
| B _w | input full power bandwidth | 3 dB bandwidth | 40 | 50 | – | MHz |

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| SYMBOL | PARAMETER | CONDITIONS | MIN. | TYP. | MAX. | UNIT |
|--------------------------|---|--|------|------|------|------|
| Power consumption | | | | | | |
| I _{DDD} | digital supply current on pins: V _{DDD18} and V _{DDD1} | f _s = 29 Mhz; direct IF application | – | 140 | 160 | mA |
| | V _{DDD33} | | – | 3 | – | mA |
| | V _{CCD(PLL)} , V _{DDA2} and V _{DDA3} | | – | 35 | – | mA |
| | V _{DDD50} | | – | 5 | – | mA |
| P _{tot} | total power dissipation | | – | 400 | 470 | mW |

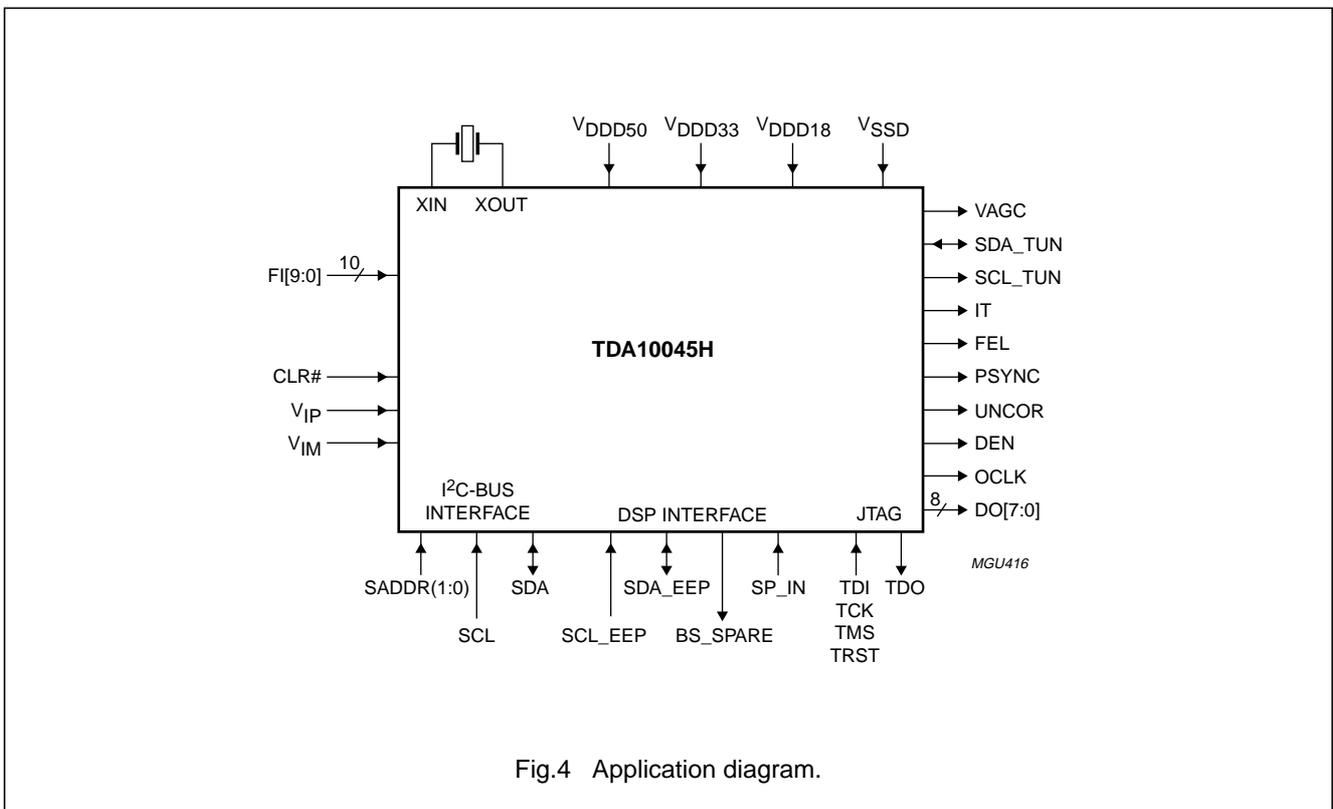
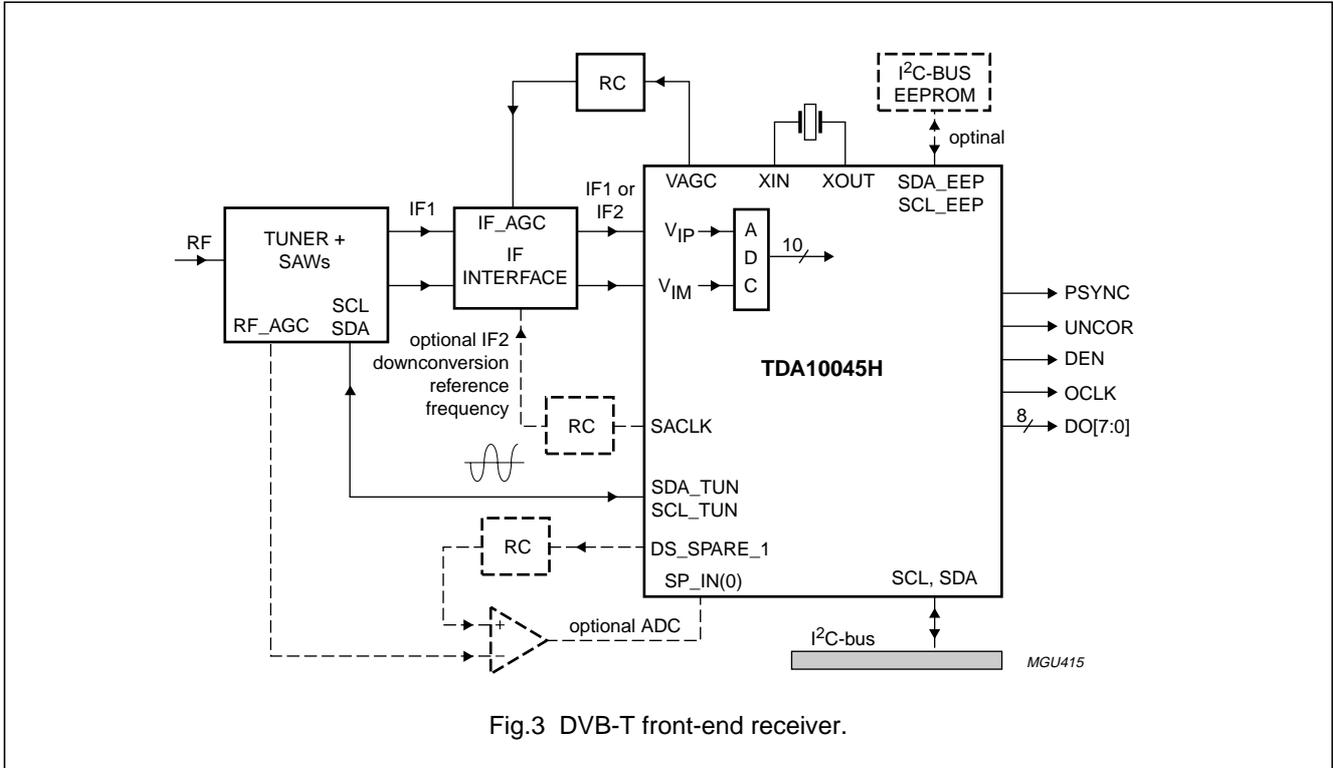
Notes

1. The voltage level of the 5 V supply must always exceed, or at least equal, the voltage level of the 3.3 V supply during power-up and down in order to guarantee protection against latch-up.
2. All inputs are 5 V tolerant.

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APPLICATION INFORMATION



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Tuner

- A RF tracking filter tracks the RF wanted frequency and suppresses the image
- A first local wideband AGC is usually done at RF level, the AGC level information could be provided externally and the chip offers facilities to measure this level by the optional ADC (this measurement is automatically made by the DSP, the host has just to read the result)
- A mixer oscillator and a PLL downconverts the RF signal to intermediate frequency IF1 (36.125 MHz typ.)
- SAW filters eliminate the power of the adjacent channels around IF1.

IF interface

- It is either an analog IF amplifier when IF1 is sampled (direct IF: digital downconversion concept) or an analog IF amplifier followed by a downconversion from IF1 to IF2 at a few MHz (e.g. 4.57 MHz)
- When this second solution is used, the ADC sampling clock could be used (after low-pass filtering) as a reference clock for downconversion (twice the ADC sampling clock could also be provided)
- The IF amplifier is controlled by the digital AGC of the chip. A simple RC circuit filters the single bit ($\Delta\Sigma$ modulated) AGC control (VAGC)
- The sampling clock could also be used to control an external ADC, the inputs to the chip will then be digital (FI[9:0]).

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- The chip is controlled by an I²C-bus and driven by an external low-cost crystal oscillator
- The software of the embedded DSP can be downloaded from the main I²C-bus or from a dedicated I²C-bus connected to an external slave I²C-bus EEPROM
- An internal bidirectional switch enables the tuner to be programmed through the chip and then switch-off the link in order to avoid phase noise distortions due to I²C-bus traffic.

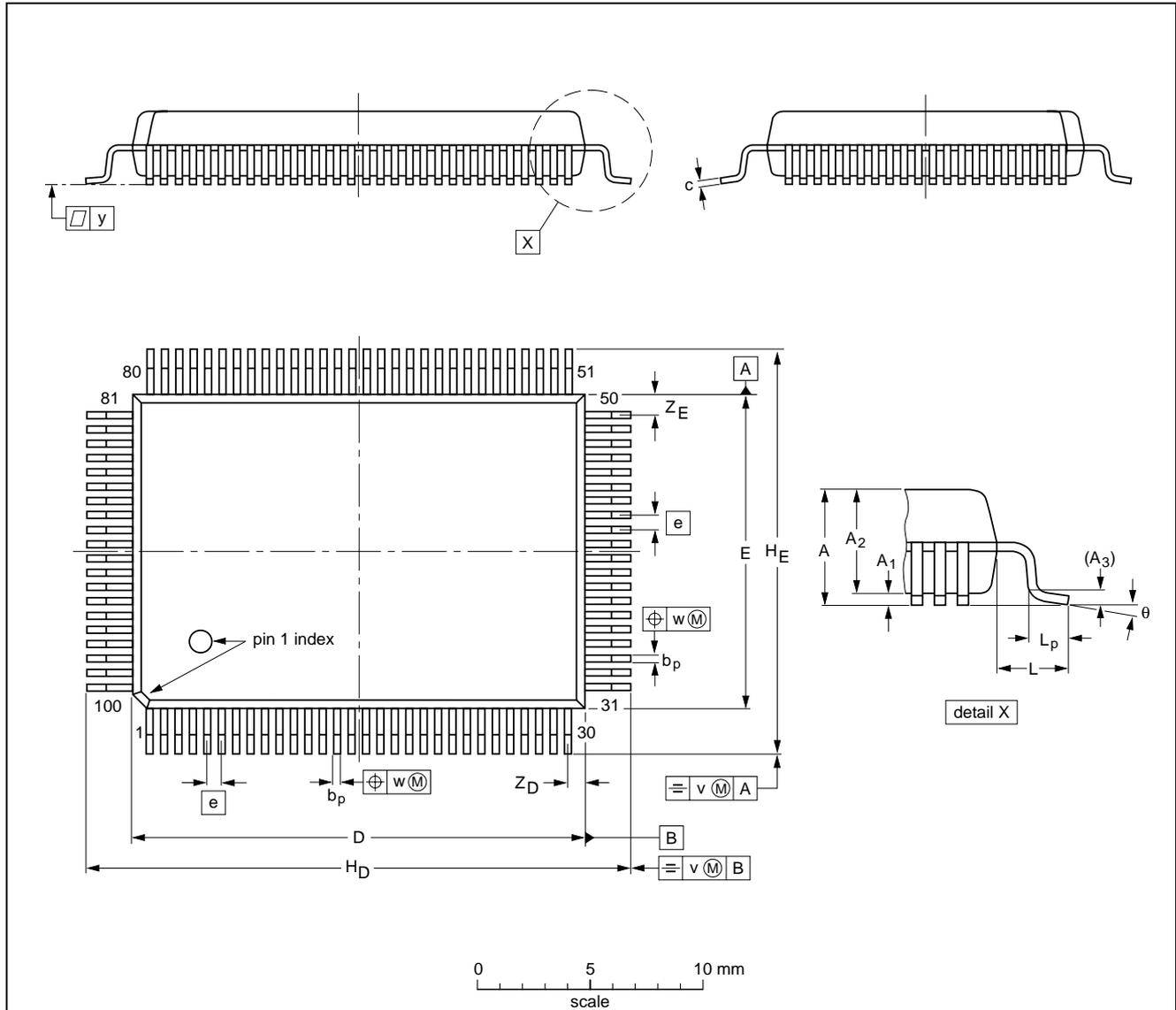
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PACKAGE OUTLINE

QFP100: plastic quad flat package; 100 leads (lead length 1.95 mm); body 14 x 20 x 2.8 mm

SOT317-2



DIMENSIONS (mm are the original dimensions)

| UNIT | A max. | A ₁ | A ₂ | A ₃ | b _p | c | D ⁽¹⁾ | E ⁽¹⁾ | e | H _D | H _E | L | L _p | v | w | y | Z _D ⁽¹⁾ | Z _E ⁽¹⁾ | θ |
|------|--------|----------------|----------------|----------------|----------------|--------------|------------------|------------------|------|----------------|----------------|------|----------------|-----|------|-----|-------------------------------|-------------------------------|----------|
| mm | 3.20 | 0.25 0.05 | 2.90 2.65 | 0.25 | 0.40 0.25 | 0.25 0.14 | 20.1 19.9 | 14.1 13.9 | 0.65 | 24.2 23.6 | 18.2 17.6 | 1.95 | 1.0 0.6 | 0.2 | 0.15 | 0.1 | 0.8 0.4 | 1.0 0.6 | 7° 0° |

Note

1. Plastic or metal protrusions of 0.25 mm maximum per side are not included.

| OUTLINE VERSION | REFERENCES | | | | EUROPEAN PROJECTION | ISSUE DATE |
|-----------------|------------|--------|------|--|---------------------|----------------------|
| | IEC | JEDEC | EIAJ | | | |
| SOT317-2 | | MO-112 | | | | 97-08-01 99-12-27 |

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SOLDERING**Introduction to soldering surface mount packages**

This text gives a very brief insight to a complex technology. A more in-depth account of soldering ICs can be found in our *"Data Handbook IC26; Integrated Circuit Packages"* (document order number 9398 652 90011).

There is no soldering method that is ideal for all surface mount IC packages. Wave soldering can still be used for certain surface mount ICs, but it is not suitable for fine pitch SMDs. In these situations reflow soldering is recommended.

Reflow soldering

Reflow soldering requires solder paste (a suspension of fine solder particles, flux and binding agent) to be applied to the printed-circuit board by screen printing, stencilling or pressure-syringe dispensing before package placement.

Several methods exist for reflowing; for example, convection or convection/infrared heating in a conveyor type oven. Throughput times (preheating, soldering and cooling) vary between 100 and 200 seconds depending on heating method.

Typical reflow peak temperatures range from 215 to 250 °C. The top-surface temperature of the packages should preferably be kept below 220 °C for thick/large packages, and below 235 °C for small/thin packages.

Wave soldering

Conventional single wave soldering is not recommended for surface mount devices (SMDs) or printed-circuit boards with a high component density, as solder bridging and non-wetting can present major problems.

To overcome these problems the double-wave soldering method was specifically developed.

If wave soldering is used the following conditions must be observed for optimal results:

- Use a double-wave soldering method comprising a turbulent wave with high upward pressure followed by a smooth laminar wave.
- For packages with leads on two sides and a pitch (e):
 - larger than or equal to 1.27 mm, the footprint longitudinal axis is **preferred** to be parallel to the transport direction of the printed-circuit board;
 - smaller than 1.27 mm, the footprint longitudinal axis **must** be parallel to the transport direction of the printed-circuit board.

The footprint must incorporate solder thieves at the downstream end.

- For packages with leads on four sides, the footprint must be placed at a 45° angle to the transport direction of the printed-circuit board. The footprint must incorporate solder thieves downstream and at the side corners.

During placement and before soldering, the package must be fixed with a droplet of adhesive. The adhesive can be applied by screen printing, pin transfer or syringe dispensing. The package can be soldered after the adhesive is cured.

Typical dwell time is 4 seconds at 250 °C.

A mildly-activated flux will eliminate the need for removal of corrosive residues in most applications.

Manual soldering

Fix the component by first soldering two diagonally-opposite end leads. Use a low voltage (24 V or less) soldering iron applied to the flat part of the lead. Contact time must be limited to 10 seconds at up to 300 °C.

When using a dedicated tool, all other leads can be soldered in one operation within 2 to 5 seconds between 270 and 320 °C.

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Suitability of surface mount IC packages for wave and reflow soldering methods

| PACKAGE | SOLDERING METHOD | |
|---|-----------------------------------|-----------------------|
| | WAVE | REFLOW ⁽¹⁾ |
| BGA, HBGA, LFBGA, SQFP, TFBGA | not suitable | suitable |
| HBCC, HLQFP, HSQFP, HSOP, HTQFP, HTSSOP, HVQFN, SMS | not suitable ⁽²⁾ | suitable |
| PLCC ⁽³⁾ , SO, SOJ | suitable | suitable |
| LQFP, QFP, TQFP | not recommended ⁽³⁾⁽⁴⁾ | suitable |
| SSOP, TSSOP, VSO | not recommended ⁽⁵⁾ | suitable |

Notes

- All surface mount (SMD) packages are moisture sensitive. Depending upon the moisture content, the maximum temperature (with respect to time) and body size of the package, there is a risk that internal or external package cracks may occur due to vaporization of the moisture in them (the so called popcorn effect). For details, refer to the Drypack information in the "Data Handbook IC26; Integrated Circuit Packages; Section: Packing Methods".
- These packages are not suitable for wave soldering as a solder joint between the printed-circuit board and heatsink (at bottom version) can not be achieved, and as solder may stick to the heatsink (on top version).
- If wave soldering is considered, then the package must be placed at a 45° angle to the solder wave direction. The package footprint must incorporate solder thieves downstream and at the side corners.
- Wave soldering is only suitable for LQFP, TQFP and QFP packages with a pitch (e) equal to or larger than 0.8 mm; it is definitely not suitable for packages with a pitch (e) equal to or smaller than 0.65 mm.
- Wave soldering is only suitable for SSOP and TSSOP packages with a pitch (e) equal to or larger than 0.65 mm; it is definitely not suitable for packages with a pitch (e) equal to or smaller than 0.5 mm.

DATA SHEET STATUS

| DATA SHEET STATUS ⁽¹⁾ | PRODUCT STATUS ⁽²⁾ | DEFINITIONS |
|----------------------------------|-------------------------------|--|
| Objective data | Development | This data sheet contains data from the objective specification for product development. Philips Semiconductors reserves the right to change the specification in any manner without notice. |
| Preliminary data | Qualification | This data sheet contains data from the preliminary specification. Supplementary data will be published at a later date. Philips Semiconductors reserves the right to change the specification without notice, in order to improve the design and supply the best possible product. |
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Notes

- Please consult the most recently issued data sheet before initiating or completing a design.
- The product status of the device(s) described in this data sheet may have changed since this data sheet was published. The latest information is available on the Internet at URL <http://www.semiconductors.philips.com>.

DVB-T channel receiver

TDA10045H

DEFINITIONS

Short-form specification — The data in a short-form specification is extracted from a full data sheet with the same type number and title. For detailed information see the relevant data sheet or data handbook.

Limiting values definition — Limiting values given are in accordance with the Absolute Maximum Rating System (IEC 60134). Stress above one or more of the limiting values may cause permanent damage to the device. These are stress ratings only and operation of the device at these or at any other conditions above those given in the Characteristics sections of the specification is not implied. Exposure to limiting values for extended periods may affect device reliability.

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DVB-T channel receiver

TDA10045H

NOTES

DVB-T channel receiver

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Contact information

For additional information please visit <http://www.semiconductors.philips.com>. Fax: +31 40 27 24825

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