

68HC705JB2

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SECTION 1

GENERAL DESCRIPTION

The MC68HC705JB2 is a member of the low-cost, high-performance MC68HC05 Family of 8-bit microcontroller units (MCUs). The MC68HC05 Family is based on the Customer-Specified Integrated Circuit (CSIC) design strategy. All MCUs in the family use the popular MC68HC05 central processing unit (CPU) and are available with a variety of subsystems, memory sizes and types, and package types. The MC68HC705JB2 is specifically designed to be used in applications where a Universal Serial Bus (USB) interface is required.

1.1 FEATURES

- Industry standard M68HC05 CPU core
- Memory-mapped Input/Output (I/O) registers
- 2048 Bytes of user EPROM
- 128 Bytes of user RAM
- Fully compliant Low Speed USB with 3 Endpoints:
 - 1 Control Endpoint (2 x 8-byte buffer)
 - 2 Interrupt Endpoints (1 x 8-byte buffer shared)
- 3.3Volt dc output pin for USB pullup resistors
- Multi-Function Timer
- 16-Bit Input Capture/Output Compare Timer
- 11 Bidirectional I/O pins with the following features:
 - 9 I/Os have software programmable pull-down capability
 - 2 open-drain I/Os have software programmable pull-up, 25mA current sink capability
 - 4 I/Os with external interrupt capability
- Low Voltage Reset (LVR) circuit
- Power saving STOP and WAIT Modes
- Available in 20-Pin PDIP and 20-pin SOIC packages

1.2 MASK OPTIONS

The mask options on the MC68HC705JB2 are handled with five EPROM bits in the Mask Option Register (\$01FF). These options are:

- External interrupt pins ($\overline{\text{IRQ}}$, PA0 to PA3):
[edge-triggered or edge-and-level-triggered]
- Port A and port B pull-down/pull-up resistors:
[connected or disconnected]
- PA0-PA3 external interrupt capability:
[enabled or disabled]
- OSC, crystal/ceramic resonator startup delay:
[4064 or 128 internal bus cycles]
- Low Voltage Reset (LVR):
[enabled or disabled]

To program the MOR, the MORON bit in the Program Control Register (bit 3 of \$3E) must be set to "1".

		bit-7	bit-6	bit-5	bit4	bit-3	bit-2	bit1	bit-0
MOR \$01FF	Read				IRQTRIG	PULLREN	PAINTEN	OSCDLY	LVREN
	Write								
	Erased	0	0	0	1	1	1	1	1
	Reset	Unaffected							

IRQTRIG – IRQ, PA0-PA3 Interrupt Options

- 1 = Edge-trigger only
- 0 = Edge-and-level-triggered

PULLREN – Port A and B Pullup/Pulldown Options

- 1 = Connected
- 0 = Disconnected

PAINTEN – PA0-PA3 External Interrupt Options

- 1 = Disabled
- 0 = Enabled

OSCDLY – Oscillator Delay Option

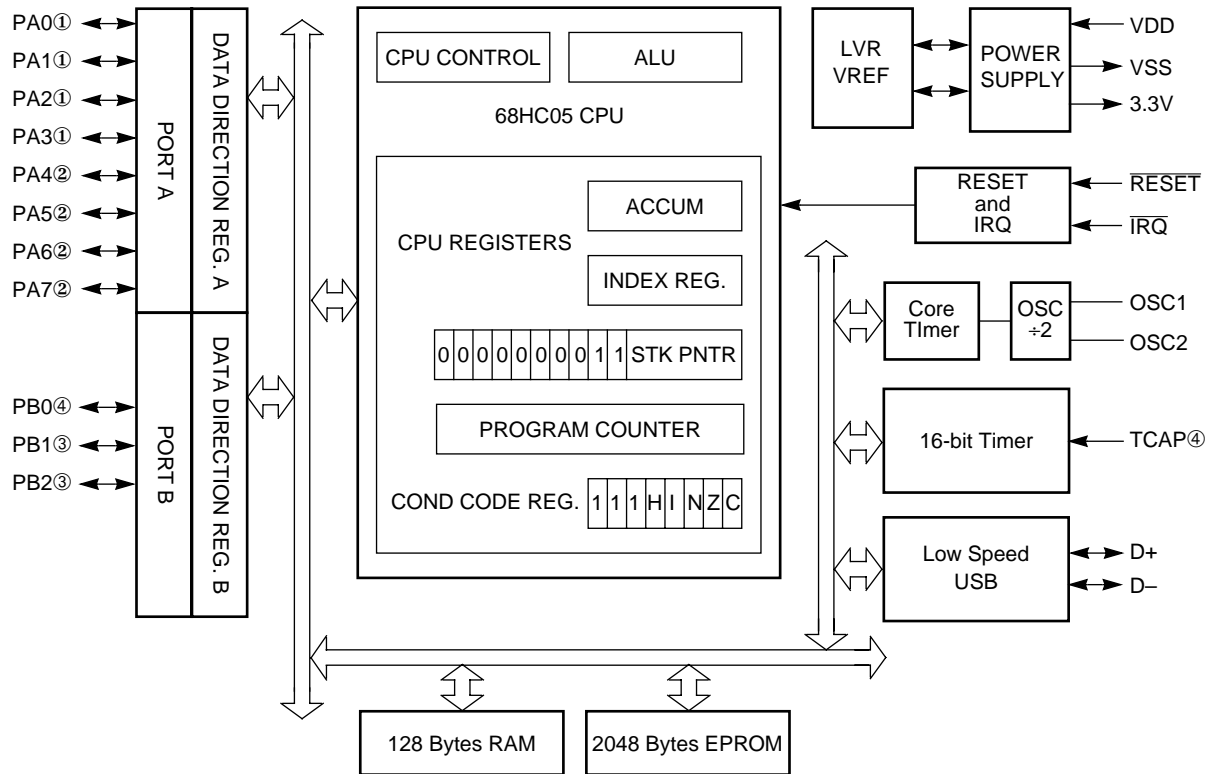
- 1 = 128 internal clock cycles
- 0 = 4064 internal clock cycles

LVREN – LVR Option

- 1 = Enabled
- 0 = Disabled

1.3 MCU STRUCTURE

Figure 1-1 shows the block diagram of the MC68HC705JB2.



- ①: External edge interrupt capability, with Schmitt trigger input
- ②: 8 mA current sink capability
- ③: 25 mA current sink, open-drained with internal pullup, slow transition O/P
- ④: TCAP is shared with PB0

Figure 1-1. MC68HC705JB2 Block Diagram

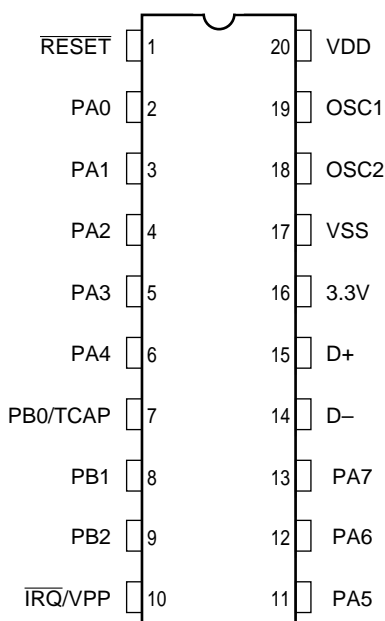


Figure 1-2. Pin Assignments for 20-Pin Package

1.4 FUNCTIONAL PIN DESCRIPTIONS

The following paragraphs give a description of the general function of each pin assigned in **Figure 1-2**.

1.4.1 V_{DD} , V_{SS}

Power is supplied to the MCU through V_{DD} and V_{SS} . V_{DD} is the positive supply, and V_{SS} is ground. The MCU operates from a single power supply.

Very fast signal transitions occur on the MCU pins. The short rise and fall times place very high short-duration current demands on the power supply. To prevent noise problems, special care should be taken to provide good power supply bypassing at the MCU by using bypass capacitors with good high-frequency characteristics that are positioned as close to the MCU as possible. Bypassing requirements vary, depending on how heavily the MCU pins are loaded.

1.4.2 OSC1, OSC2

The OSC1 and OSC2 pins are the connections for the on-chip oscillator. The OSC1 and OSC2 pins can accept the following sets of components:

1. A crystal as shown in **Figure 1-3(a)**
2. A ceramic resonator as shown in **Figure 1-3(a)**
3. An external clock signal as shown in **Figure 1-3(b)**

The frequency, f_{OSC} , of the oscillator or external clock source is divided by two to produce the internal operating frequency, f_{OP} . If the internal operating frequency is 3MHz, then the external oscillator frequency will be 6MHz. For LS USB 1.5MHz frequency clock can be derived from a divided by 4 circuit. The type of oscillator is selected by a mask option.

1.4.2.1 Crystal Oscillator

The circuit in **Figure 1-3(a)** shows a typical oscillator circuit for an AT-cut, parallel resonant crystal. The crystal manufacturer's recommendations should be followed, as the crystal parameters determine the external component values required to provide maximum stability and reliable start-up. The load capacitance values used in the oscillator circuit design should include all stray capacitances. The crystal and components should be mounted as close as possible to the pins for start-up stabilization and to minimize output distortion. An internal start-up resistor of typically $2M\Omega$ is provided between OSC1 and OSC2 for the crystal type oscillator.



Figure 1-3. Oscillator Connections

1.4.2.2 Ceramic Resonator Oscillator

In cost-sensitive applications, a ceramic resonator can be used in place of the crystal. The circuit in **Figure 1-3(a)** can be used for a ceramic resonator. The resonator manufacturer's recommendations should be followed, as the resonator parameters determine the external component values required for maximum stability and reliable starting. The load capacitance values used in the oscillator circuit design should include all stray capacitances. The ceramic resonator and

components should be mounted as close as possible to the pins for start-up stabilization and to minimize output distortion. An internal start-up resistor of $2\text{M}\Omega$ (typical) is provided between OSC1 and OSC2 for the ceramic resonator type oscillator.

1.4.2.3 External Clock

An external clock from another CMOS-compatible device can be connected to the OSC1 input, with the OSC2 input not connected, as shown in **Figure 1-3(b)**.

1.4.3 RESET

This is an I/O pin. This pin can be used as an input to reset the MCU to a known start-up state by pulling it to the low state. The $\overline{\text{RESET}}$ pin contains a steering diode to discharge any voltage on the pin to V_{DD} , when the power is removed. An internal pull-up is also connected between this pin and V_{DD} . The $\overline{\text{RESET}}$ pin contains an internal Schmitt trigger to improve its noise immunity as an input. This pin is an output pin if LVR triggers an internal reset.

1.4.4 $\overline{\text{IRQ}}/\text{VPP}$

This input pin drives the asynchronous IRQ interrupt function of the CPU. The IRQ interrupt function has a mask option to provide either only negative edge-sensitive triggering or both negative edge-sensitive and low level-sensitive triggering. If the option is selected to include level-sensitive triggering, the $\overline{\text{IRQ}}$ input requires an external resistor to V_{DD} for "wired-OR" operation, if desired. The $\overline{\text{IRQ}}$ pin contains an internal Schmitt trigger as part of its input to improve noise immunity.

NOTE

Each of the PA0 thru PA3 I/O pins may be connected as an OR function with the IRQ interrupt function by a mask option. This capability allows keyboard scan applications where the transitions or levels on the I/O pins will behave the same as the $\overline{\text{IRQ}}$ pin, except for the inverted phase. The edge or level sensitivity selected by a separate mask option for the $\overline{\text{IRQ}}$ pin also applies to the I/O pins OR'ed to create the IRQ signal.

In Bootloader mode, this pin (VPP) is used to supply the required programming voltage to the EPROM array.

1.4.5 PA0-PA7

These eight I/O lines comprise PortA. PA0 to PA7 are push-pull pins with pulldown devices. PA4 to PA7 are also capable of sinking 8 mA.

The state of any pin is software programmable and all Port A lines are configured as inputs during power-on or reset. The lower four I/O pins (PA0 thru PA3) can be connected via an internal OR gate to the IRQ interrupt function enabled by a mask option. See **Section 7** on Input/Output Ports for further details.

1.4.6 PB0-PB2

These three I/O lines comprise Port B. PB1 and PB2 are open-drain I/O lines with pullup devices, whereas PB0 (shared with TCAP), is a push-pull I/O line with pulldown device.

The state of any pin is software programmable and is configured as an input during power-on or reset. PB1 and PB2 are also slow transition outputs, each capable of sinking 25mA typical current at 0.5V V_{OL} Max. See **Section 7** on Input/Output Ports for further details.

1.4.7 D+, D–

D+ and D– are the differential data lines used by the USB module. See **Section 10** on Universal Serial Bus Module.

1.4.8 3.3V

This is the 3.3V output of the on-chip voltage regulator from the MCU. It is used to supply the voltage for the external pullup resistor required by the USB on D–. This regulator output is also used internally for the USB data driver circuitry. This 3.3V pin should be decoupled using a 1 μ F (or greater) capacitor and a 0.1 μ F bypass capacitor.

SECTION 2 MEMORY

2.1 MEMORY MAP

The MC68HC705JB2 has an 8-Kbyte memory map consisting of user EPROM, RAM, burn-in ROM, and input/output (I/O), as shown in **Figure 2-1**.

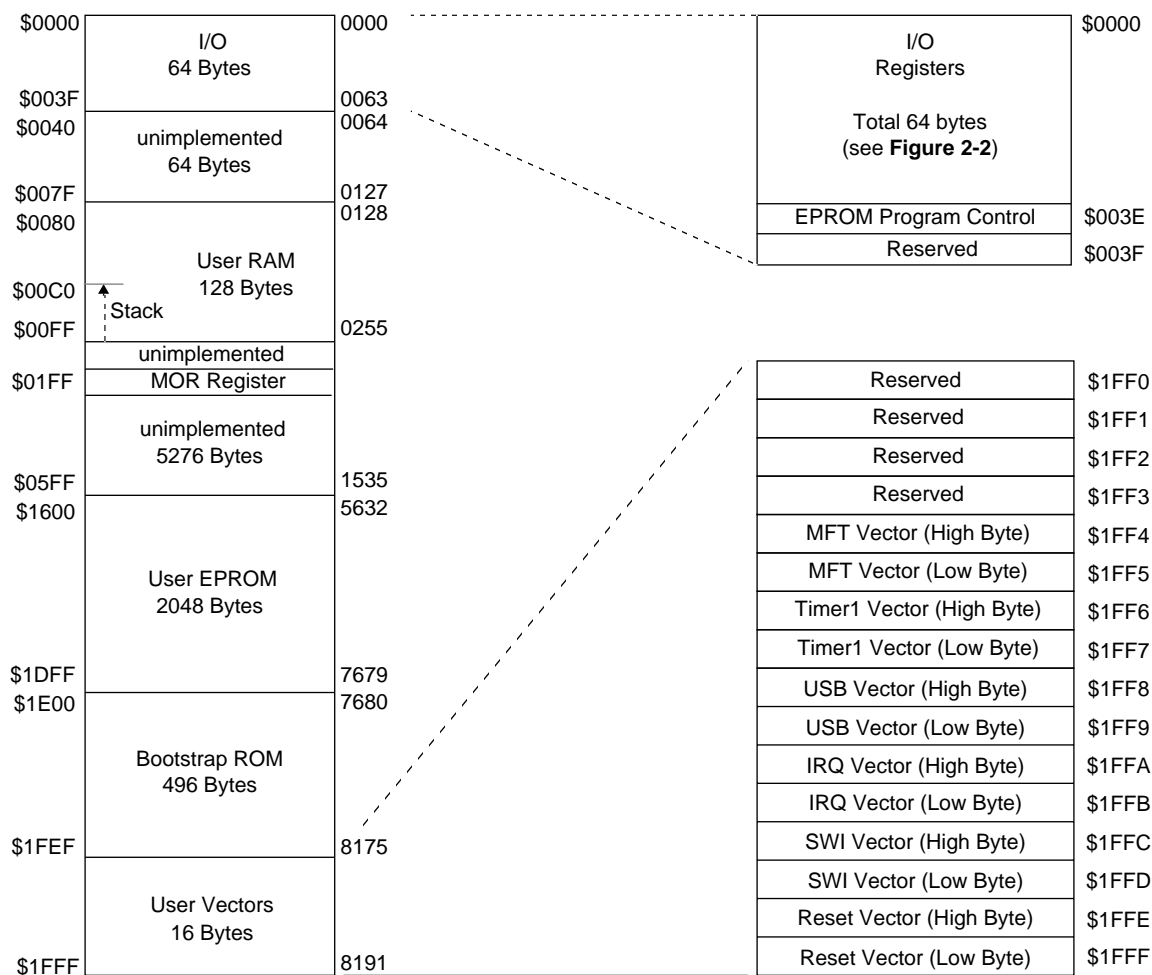


Figure 2-1. MC68HC705JB2 Memory Map

2.2 I/O AND CONTROL REGISTERS

The I/O and Control Registers reside at locations \$0000-\$003F. The outline of these registers are shown in **Figure 2-2**. The bit assignments for each register are shown in **Figure 2-3**, **Figure 2-4**, and **Figure 2-5**. Reading from unimplemented bits will return unknown states, and writing to unimplemented bits will be ignored.

Port A Data Register	\$0000
Port B Data Register	\$0001
unimplemented (2)	
Port A Data Direction Register	\$0004
Port B Data Direction Register	\$0005
unimplemented (2)	
Timer Control & Status Register	\$0008
Timer Counter Register	\$0009
IRQ Control & Status Register	\$000A
unimplemented (5)	
Port A Pulldown Register	\$0010
Port B Pulldown/up Register	\$0011
Timer1 Registers (10)	\$0012 to \$001B
unimplemented (4)	
USB Endpoint0 Data Registers (8)	\$0020 to \$0027
USB Endpoint1 Data Registers (8)	\$0028 to \$002F
USB Control2 Register	\$0037
USB Address Register	\$0038
USB Interrupt0 Register	\$0039
USB Interrupt1 Register	\$003A
USB Control0 Register	\$003B
USB Control1 Register	\$003C
USB Status Register	\$003D
EPROM Program Control Register	\$003E
Reserved	\$003F
Mask Option Register	\$01FF

Figure 2-2. I/O Registers

2.3 **RAM**

The user RAM consists of 128 bytes (including the stack) located from \$0080 to \$00FF. The stack begins at address \$00FF and proceeds down to \$00C0. Using the stack area for data storage or temporary work locations requires care to prevent it from being overwritten due to stacking from an interrupt or subroutine call.

2.4 **EPROM**

The on-chip user EPROM consists of 2048 bytes of EPROM from \$1600 to \$1DFF and 16 bytes of user vectors from \$1FF0 to \$1FFF.

The bootloader ROM and vectors are located from \$1E00 to \$1FEF.

12 of the user vectors, \$1FF4-\$1FFF, are dedicated to reset and interrupt vectors. The four remaining locations, \$1FF0-\$1FF3, are reserved for test functions. The Mask Option Register is located at \$01FF.

2.5 **BOOTLOADER ROM**

Addresses \$1E00 to \$1FEF are reserved ROM addresses that contain the instructions for the bootloader functions. (See **Section 11.**)

ADDR	REGISTER	R/W	BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
\$0000	PORT A DATA (PORTA)	R	PA7	PA6	PA5	PA4	PA3	PA2	PA1	PA0
		W								
\$0001	PORT B DATA (PORTB)	R						PB2	PB1	PB0
		W								
\$0002	UNIMPLEMENTED	R								
		W								
\$0003	UNIMPLEMENTED	R								
		W								
\$0004	PORT A DATA DIR (DDRA)	R	DDRA7	DDRA6	DDRA5	DDRA4	DDRA3	DDRA2	DDRA1	DDRA0
		W								
\$0005	PORT B DATA DIR (DDRB)	R	SLOWE					DDRB2	DDRB1	DDRB0
		W								
\$0006	UNIMPLEMENTED	R								
		W								
\$0007	UNIMPLEMENTED	R								
		W								
\$0008	MFT CONTROL & STATUS (TCSR)	R	TOF	RTIF	TOFE	RTIE	0	0	RT1	RT0
		W					TOFR	RTIFR		
\$0009	MFT COUNTER TCNT	R	TMR7	TMR6	TMR5	TMR4	TMR3	TMR2	TMR1	TMR0
		W								
\$000A	IRQ CONTROL & STATUS (ICSR)	R	IRQE	0	0	0	IRQF	0	0	0
		W							IRQR	
\$000B	UNIMPLEMENTED	R								
		W								
\$000C	UNIMPLEMENTED	R								
		W								
\$000D	UNIMPLEMENTED	R								
		W								
\$000E	UNIMPLEMENTED	R								
		W								
\$000F	UNIMPLEMENTED	R								
		W								

Figure 2-3. I/O Registers \$0000-\$000F

ADDR	REGISTER	R/W	BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
\$0010	PORT A PULLDOWN/UP (PDURA)	R								
		W	PDRA7	PDRA6	PDRA5	PDRA4	PDRA3	PDRA2	PDRA1	PDRA0
\$0011	PORT B PULLDOWN/UP (PDURB)	R								
		W						PURB2	PURB1	PURB0
\$0012	TIMER CONTROL (TCR)	R	ICIE	OCIE	TOIE	0	0	0	IEDG	0
		W								
\$0013	TIMER STATUS (TSR)	R	ICF	OCF	TOF	0	0	0	0	0
		W								
\$0014	INPUT CAPTURE HIGH (ICH)	R	BIT15	BIT14	BIT13	BIT12	BIT11	BIT10	BIT9	BIT8
		W								
\$0015	INPUT CAPTURE LOW (ICL)	R	BIT7	BIT6	BIT5	BIT4	BIT3	BIT2	BIT1	BIT0
		W								
\$0016	OUTPUT COMPARE HIGH (OCH)	R	BIT15	BIT14	BIT13	BIT12	BIT11	BIT10	BIT9	BIT8
		W								
\$0017	OUTPUT COMPARE LOW (OCL)	R	BIT7	BIT6	BIT5	BIT4	BIT3	BIT2	BIT1	BIT0
		W								
\$0018	TIMER HIGH (TCNTH)	R	BIT15	BIT14	BIT13	BIT12	BIT11	BIT10	BIT9	BIT8
		W								
\$0019	TIMER LOW (TCNTL)	R	BIT7	BIT6	BIT5	BIT4	BIT3	BIT2	BIT1	BIT0
		W								
\$001A	ALT COUNTER HIGH (ACNTH)	R	BIT15	BIT14	BIT13	BIT12	BIT11	BIT10	BIT9	BIT8
		W								
\$001B	ALT COUNTER LOW (ACNTL)	R	BIT7	BIT6	BIT5	BIT4	BIT3	BIT2	BIT1	BIT0
		W								
\$001C	UNIMPLEMENTED	R								
		W								
\$001D	UNIMPLEMENTED	R								
		W								
\$001E	UNIMPLEMENTED	R								
		W								
\$001F	UNIMPLEMENTED	R								
		W								

Figure 2-4. I/O Registers \$0010-\$001F

ADDR	REGISTER	R/W	BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
\$0020 to \$0027	USB ENDPOINT0 DATA REG. 0 TO 7 (UD0R0-7)	R	UE0RD7	UE0RD6	UE0RD5	UE0RD4	UE0RD3	UE0RD2	UE0RD1	UE0RD0
		W	UE0TD7	UE0TD6	UE0TD5	UE0TD4	UE0TD3	UE0TD2	UE0TD1	UE0TD0
\$0028 to \$002F	USB ENDPOINT1 DATA REG. 0 TO 7 (UD1R0-7)	R								
		W	UE1TD7	UE1TD6	UE1TD5	UE1TD4	UE1TD3	UE1TD2	UE1TD1	UE1TD0
\$0030 to \$0036	UNIMPLEMENTED	R								
		W								
\$0037	USB CONTROL2 (UCR2)	R		0	TX1ST	0	ENABLE2	ENABLE1	STALL2	STALL1
		W		TX1STR						
\$0038	USB ADDRESS (UADR)	R	USBEN	UADD6	UADD5	UADD4	UADD3	UADD2	UADD1	UADD0
		W								
\$0039	USB INTERRUPT0 (UIR0)	R	TXD0F	RXD0F	RSTF	SUSPND	TXD0IE	RXD0IE	0	0
		W	0	0	0				TXD0FR	RXD0FR
\$003A	USB INTERRUPT1 (UIR1)	R	TXD1F	EOPF	RESUMF	0	TXD1IE	EOPIE	0	0
		W	0	0	0	RESUMFR			TXD1FR	EOPFR
\$003B	USB CONTROL 0 (UCR0)	R	T0SEQ	STALL0	TX0E	RX0E	TP0SIZ3	TP0SIZ2	TP0SIZ1	TP0SIZ0
		W								
\$003C	USB CONTROL1 (UCR1)	R	T1SEQ	ENDADD	TX1E	FRESUM	TP1SIZ3	TP1SIZ2	TP1SIZ1	TP1SIZ0
		W								
\$003D	USB STATUS (USR)	R	RSEQ	SETUP			RPSIZ3	RPSIZ2	RPSIZ1	RPSIZ0
		W								
\$003E	PROG. CONTROL (PCR)	R						MORON	ELAT	PGM
		W								
\$003F	RESERVED	R								
		W								

Figure 2-5. I/O Registers \$0020-\$003F

ADDR	REGISTER	R/W	BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
\$01FF	MASK OPTION (MOR)	R				IRQTRIG	PULLREN	PAINTEN	OSCDLY	LVREN
		W								

Figure 2-6. Mask Option Register \$01FF

SECTION 3 CENTRAL PROCESSING UNIT

The MC68HC705JB2 has a 8Kbyte memory map. The stack has only 64 bytes. Therefore, the stack pointer has been reduced to only 6 bits and will only decrement down to \$00C0 and then wrap-around to \$00FF. All other instructions and registers behave as described in this chapter.

3.1 REGISTERS

The MCU contains five registers which are hard-wired within the CPU and are not part of the memory map. These five registers are shown in **Figure 3-1** and are described in the following paragraphs.

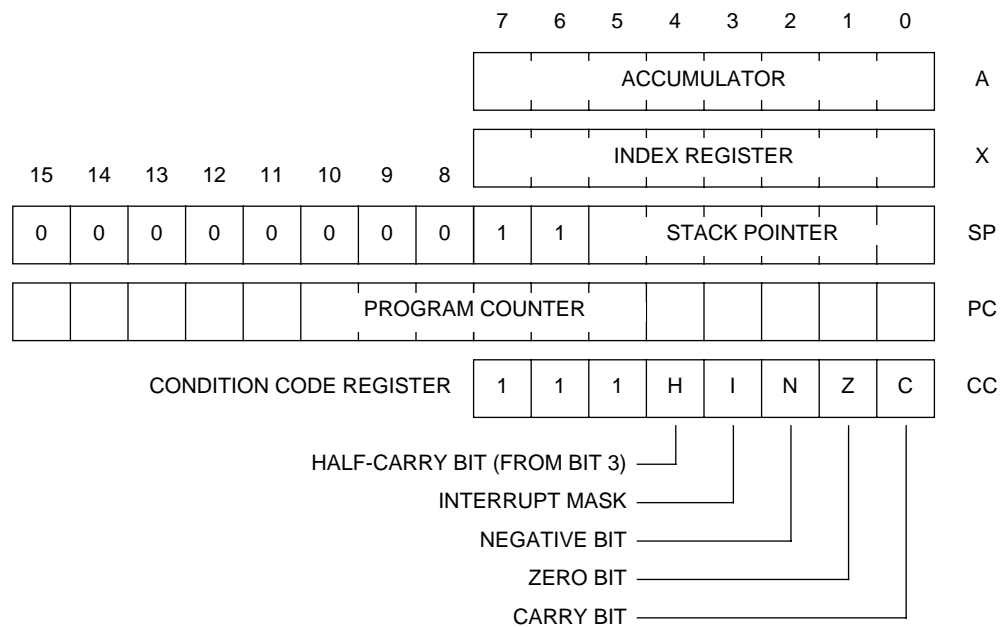


Figure 3-1. MC68HC05 Programming Model

3.2 ACCUMULATOR (A)

The accumulator is a general purpose 8-bit register as shown in **Figure 3-1**. The CPU uses the accumulator to hold operands and results of arithmetic calculations or non-arithmetic operations. The accumulator is not affected by a reset of the device.

3.3 INDEX REGISTER (X)

The index register shown in **Figure 3-1** is an 8-bit register that can perform two functions:

- Indexed addressing
- Temporary storage

In indexed addressing with no offset, the index register contains the low byte of the operand address, and the high byte is assumed to be \$00. In indexed addressing with an 8-bit offset, the CPU finds the operand address by adding the index register content to an 8-bit immediate value. In indexed addressing with a 16-bit offset, the CPU finds the operand address by adding the index register content to a 16-bit immediate value.

The index register can also serve as an auxiliary accumulator for temporary storage. The index register is not affected by a reset of the device.

3.4 STACK POINTER (SP)

The stack pointer shown in **Figure 3-1** is a 16-bit register. In MCU devices with memory space less than 64 Kbytes the unimplemented upper address lines are ignored. The stack pointer contains the address of the next free location on the stack. During a reset or the reset stack pointer (RSP) instruction, the stack pointer is set to \$00FF. The stack pointer is then decremented as data is pushed onto the stack and incremented as data is pulled off the stack.

When accessing memory, the ten most significant bits are permanently set to 0000000011. The six least significant register bits are appended to these ten fixed bits to produce an address within the range of \$00FF to \$00C0. Subroutines and interrupts may use up to 64(\$C0) locations. If 64 locations are exceeded, the stack pointer wraps around and overwrites the previously stored information. A subroutine call occupies two locations on the stack and an interrupt uses five locations.

3.5 PROGRAM COUNTER (PC)

The program counter shown in **Figure 3-1** is a 16-bit register. In MCU devices with memory space less than 64 Kbytes the unimplemented upper address lines are ignored. The program counter contains the address of the next instruction or operand to be fetched.

Normally, the address in the program counter increments to the next sequential memory location every time an instruction or operand is fetched. Jump, branch, and interrupt operations load the program counter with an address other than that of the next sequential location.

3.6 CONDITION CODE REGISTER (CCR)

The CCR shown in **Figure 3-1** is a 5-bit register in which four bits are used to indicate the results of the instruction just executed. The fifth bit is the interrupt mask. These bits can be individually tested by a program, and specific actions can be taken as a result of their states. The condition code register should be thought of as having three additional upper bits that are always ones. Only the interrupt mask is affected by a reset of the device. The following paragraphs explain the functions of the lower five bits of the condition code register.

3.6.1 Half Carry Bit (H-Bit)

When the half-carry bit is set, it means that a carry occurred between bits 3 and 4 of the accumulator during the last ADD or ADC (add with carry) operation. The half-carry bit is required for binary-coded decimal (BCD) arithmetic operations.

3.6.2 Interrupt Mask (I-Bit)

When the interrupt mask is set, the internal and external interrupts are disabled. Interrupts are enabled when the interrupt mask is cleared. When an interrupt occurs, the interrupt mask is automatically set after the CPU registers are saved on the stack, but before the interrupt vector is fetched. If an interrupt request occurs while the interrupt mask is set, the interrupt request is latched. Normally, the interrupt is processed as soon as the interrupt mask is cleared.

A return from interrupt (RTI) instruction pulls the CPU registers from the stack, restoring the interrupt mask to its state before the interrupt was encountered. After any reset, the interrupt mask is set and can only be cleared by the Clear I-Bit (CLI), or WAIT instructions.

3.6.3 Negative Bit (N-Bit)

The negative bit is set when the result of the last arithmetic operation, logical operation, or data manipulation was negative. (Bit 7 of the result was a logical one.)

The negative bit can also be used to check an often tested flag by assigning the flag to bit 7 of a register or memory location. Loading the accumulator with the contents of that register or location then sets or clears the negative bit according to the state of the flag.

3.6.4 Zero Bit (Z-Bit)

The zero bit is set when the result of the last arithmetic operation, logical operation, data manipulation, or data load operation was zero.

3.6.5 Carry/Borrow Bit (C-Bit)

The carry/borrow bit is set when a carry out of bit 7 of the accumulator occurred during the last arithmetic operation, logical operation, or data manipulation. The carry/borrow bit is also set or cleared during bit test and branch instructions and during shifts and rotates. This bit is neither set by an INC nor by a DEC instruction.

SECTION 4 INTERRUPTS

The MCU can be interrupted in six different ways:

- Non-maskable Software Interrupt Instruction (SWI)
- External Interrupt ($\overline{\text{IRQ}}$)
- Optional External Interrupt via IRQ on PA0-PA3 (mask option)
- USB Interrupt
- Timer1 Interrupt (16-bit Timer)
- Multi-Function Timer Interrupt

4.1 CPU INTERRUPT PROCESSING

Interrupts cause the processor to save register contents on the stack and to set the interrupt mask (I-bit) to prevent additional interrupts. Unlike RESET, hardware interrupts do not cause the current instruction execution to be halted, but are considered pending until the current instruction is complete.

If interrupts are not masked (I-bit in the CCR is clear) and the corresponding interrupt enable bit is set the processor will proceed with interrupt processing. Otherwise, the next instruction is fetched and executed. If an interrupt occurs the processor completes the current instruction, then stacks the current CPU register states, sets the I-bit to inhibit further interrupts, and finally checks the pending hardware interrupts. If more than one interrupt is pending following the stacking operation, the interrupt with the highest vector location shown in **Table 4-1** will be serviced first. The SWI is executed the same as any other instruction, regardless of the I-bit state.

When an interrupt is to be processed the CPU fetches the address of the appropriate interrupt software service routine from the vector table at locations \$1FF4 to \$1FFF as defined in **Table 4-1**.

Table 4-1. Reset/Interrupt Vector Addresses

Function	Source	Control Bit	Global Hardware Mask	Local Software Mask	Priority (1 = Highest)	Vector Address
Reset	Power-On Logic $\overline{\text{RESET}}$ Pin Low Voltage Reset Illegal Address Reset USB Reset	—	—	—	1	\$1FFE–\$1FFF
Software Interrupt (SWI)	User Code	—	—	—	Same Priority As Instruction	\$1FFC–\$1FFD
External Interrupt (IRQ)	$\overline{\text{IRQ}}$ Pin	—	I Bit	IRQE Bit	2	\$1FFA–\$1FFB
USB Interrupts	TXD0F TXD1F RESUMF	—	I Bit	TXD0IE TXD1IE —	3	\$1FF8–\$1FF9
Timer1 Interrupts	ICF Bit OCF Bit TOF Bit	—	I Bit	ICIE Bit OCIE Bit TOIE Bit	4	\$1FF6–\$1FF7
MFT Interrupts	TOF Bit RTIF	—	I Bit	TOFE Bit RTIE Bit	5	\$1FF4–\$1FF5
Reserved						\$1FF2–\$1FF3
Reserved						\$1FF0–\$1FF1

An RTI instruction is used to signify when the interrupt software service routine is completed. The RTI instruction causes the register contents to be recovered from the stack and normal processing to resume at the next instruction that was to be executed when the interrupt took place. **Figure 4-1** shows the sequence of events that occur during interrupt processing.

4.2 RESET INTERRUPT SEQUENCE

The RESET function is not in the strictest sense an interrupt; however, it is acted upon in a similar manner as shown in **Figure 4-1**. A low level input on the $\overline{\text{RESET}}$ pin or an internally generated RST signal causes the program to vector to its starting address which is specified by the contents of memory locations \$1FFE and \$1FFF. The I-bit in the condition code register is also set.

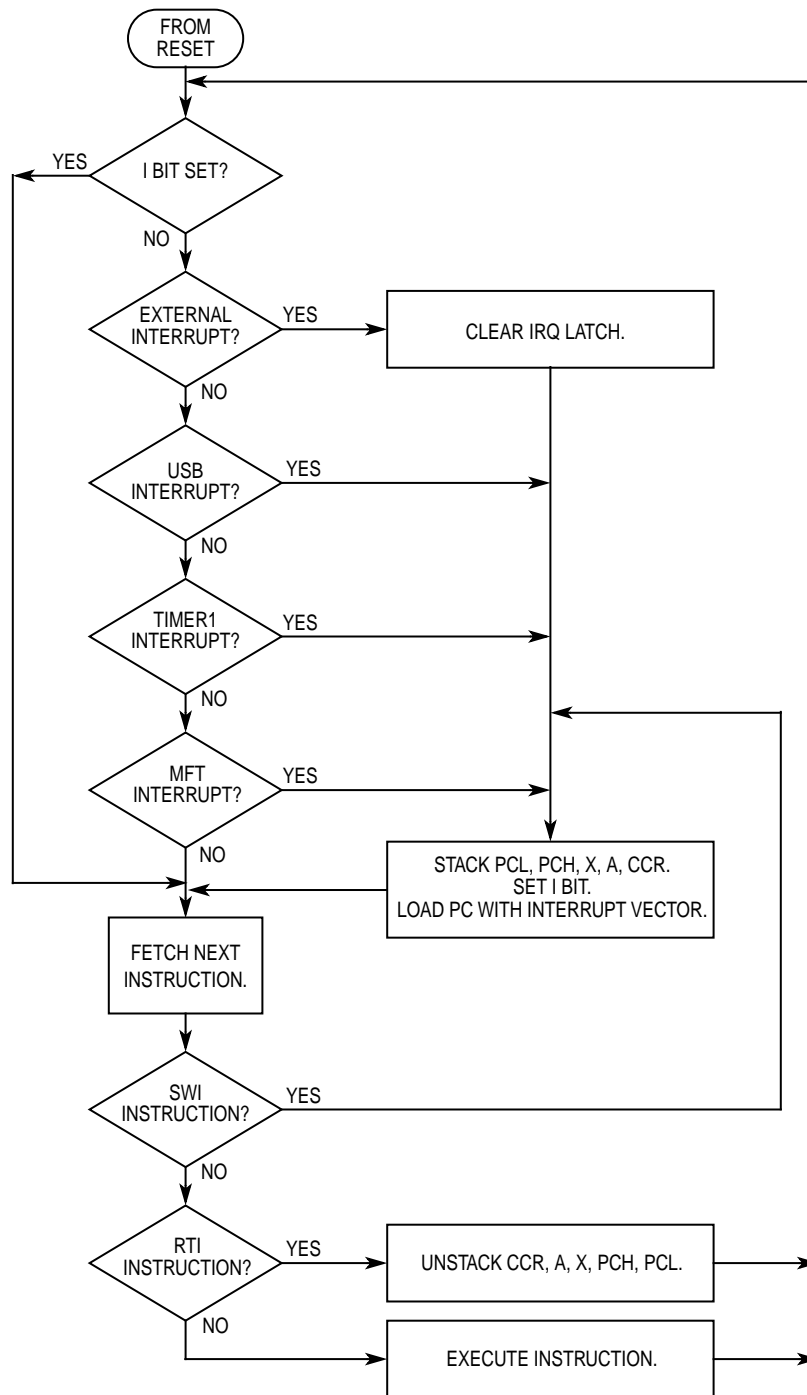


Figure 4-1. Interrupt Processing Flowchart

4.3 SOFTWARE INTERRUPT (SWI)

The SWI is an executable instruction and a non-maskable interrupt since it is executed regardless of the state of the I-bit in the CCR. As with any instruction, interrupts pending during the previous instruction will be serviced before the SWI opcode is fetched. The interrupt service routine address is specified by the contents of memory locations \$1FFC and \$1FFD.

4.4 HARDWARE INTERRUPTS

All hardware interrupts except RESET are maskable by the I-bit in the CCR. If the I-bit is set, all hardware interrupts (internal and external) are disabled. Clearing the I-bit enables the hardware interrupts. There are two types of hardware interrupts which are explained in the following sections.

4.4.1 External Interrupt ($\overline{\text{IRQ}}$)

The $\overline{\text{IRQ}}$ pin provides an asynchronous interrupt to the CPU. A block diagram of the IRQ function is shown in **Figure 4-2**.

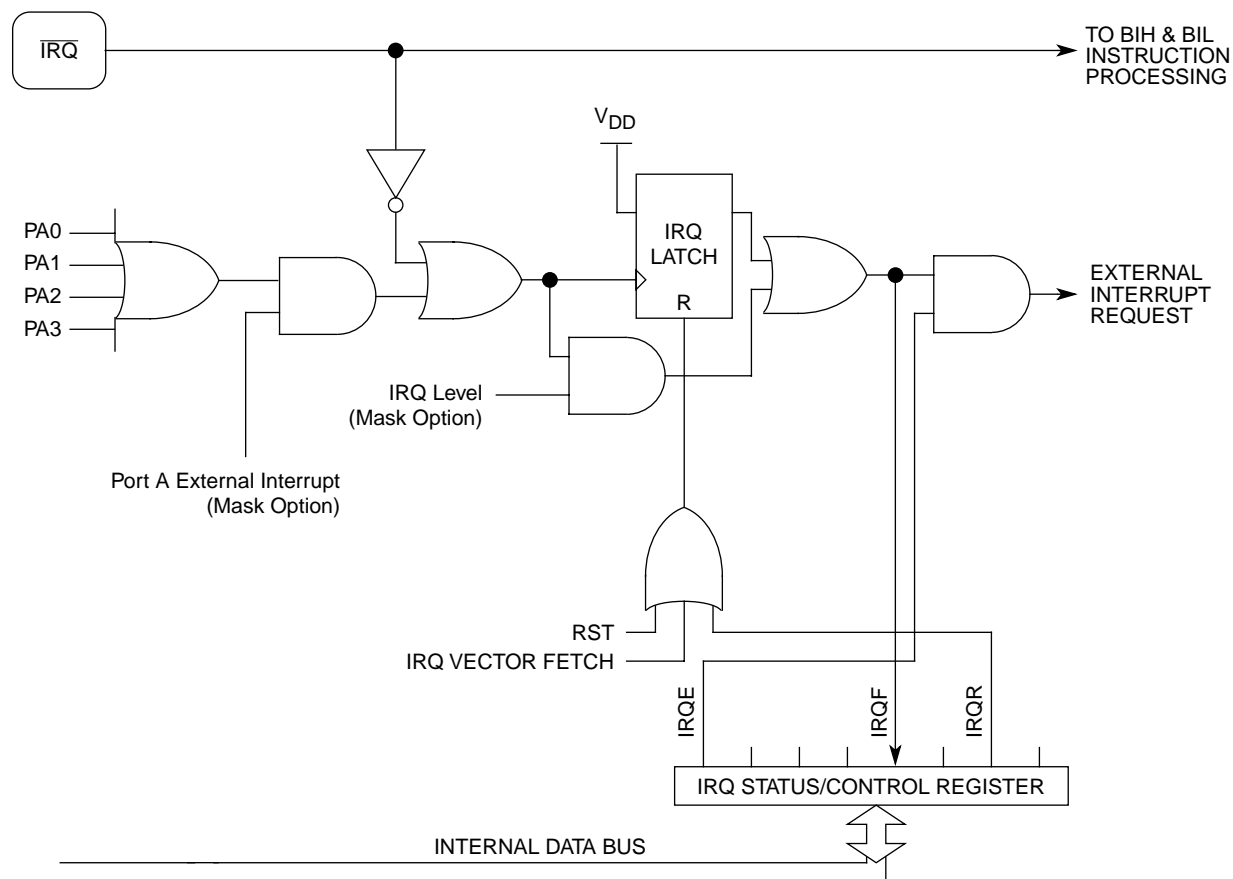


Figure 4-2. External Interrupt (IRQ) Logic

The $\overline{\text{IRQ}}$ pin is one source of an IRQ interrupt and a mask option can also enable the four lower Port A pins (PA0 thru PA3) to act as other IRQ interrupt sources.

Refer to **Figure 4-2** for the following descriptions. IRQ interrupt source comes from IRQ latch. The IRQ latch will be set on the falling edge of the $\overline{\text{IRQ}}$ pin or on any rising edge of PA0-3 pins if PA0-3 interrupts have been enabled. If "edge-only" sensitivity is chosen by a mask option, only the IRQ latch output can activate an IRQF flag which creates a request to the CPU to generate the IRQ interrupt sequence. This makes the IRQ interrupt sensitive to the following cases:

- Falling edge on the $\overline{\text{IRQ}}$ pin.
- Rising edge on any PA0-PA3 pin with IRQ enabled (via mask option).

If level sensitivity is chosen, the active high state the signal to the clock input of the IRQ latch can also activate an IRQF flag which creates an IRQ request to the CPU to generate the IRQ interrupt sequence. This makes the IRQ interrupt sensitive to the following cases:

- Low level on the $\overline{\text{IRQ}}$ pin.
- Falling edge on the $\overline{\text{IRQ}}$ pin.
- High level on any PA0- PA3 pin with IRQ enabled (via mask option).
- Rising edge on any PA0- PA3 pin with IRQ enabled (via mask option).

The IRQE enable bit controls whether an active IRQF flag can generate an IRQ interrupt sequence. This interrupt is serviced by the interrupt service routine located at the address specified by the contents of \$1FFA and \$1FFB.

The IRQ latch is automatically cleared by entering the interrupt service routine IF IRQE enable bit is cleared. If IRQE enable bit is also set, the only way of clearing IRQF is by writing a logic one to the IRQR acknowledge bit. Writing a logic one to the IRQR acknowledge bit in the ICSR is the other way of clearing IRQF flag. As long as the output state of the IRQF flag bit is active the CPU will continuously re-enter the IRQ interrupt sequence until the active state is removed or the IRQE enable bit is cleared.

4.4.2 IRQ Control/status Register (ICSR) \$0A

The IRQ interrupt function is controlled by the ICSR located at \$000A. All unused bits in the ICSR will read as logic zeros. The IRQF bit is cleared and IRQE bit is set by reset.

		BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
ICSR \$000A	R	IRQE	0	0	0	IRQF	0	0	0
	W							IRQR	
reset:		1	0	0	0	0	0	0	0

Figure 4-3. IRQ Control and Status Register (ICSR)

IRQR - IRQ Interrupt Acknowledge

The IRQR acknowledge bit clears an IRQ interrupt by clearing the IRQ latch. The IRQR acknowledge bit will always read as a logic zero.

- 1 = Writing a logic one to the IRQR acknowledge bit will clear the IRQ latch.
- 0 = Writing a logic zero to the IRQR acknowledge bit will have no effect on the IRQ latch.

IRQF - IRQ Interrupt Request Flag

Writing to the IRQF flag bit will have no effect on it. If the additional setting of IRQF flag bit is not cleared in the IRQ service routine and the IRQE enable bit remains set the CPU will re-enter the IRQ interrupt sequence continuously until either the IRQF flag bit or the IRQE enable bit is clear. The IRQF latch is cleared by reset.

- 1 = Indicates that an IRQ request is pending.
- 0 = Indicates that no IRQ request triggered by pins PA0-3 or $\overline{\text{IRQ}}$ is pending. The IRQF flag bit is cleared once the IRQ vector is fetched. The IRQF flag bit can be cleared by writing a logic one to the IRQR acknowledge bit to clear the IRQ latch and also conditioning the external IRQ sources to be inactive (if the level sensitive interrupts are enabled via mask option). Doing so before exiting the service routine will mask out additional occurrences of the IRQF.

IRQE - IRQ Interrupt Enable

The IRQE bit enables/disables the IRQF flag bit to initiate an IRQ interrupt sequence.

- 1 = Enables IRQF interrupt, that is, the IRQF flag bit can generate an interrupt sequence. Reset sets the IRQE enable bit, thereby enabling IRQ interrupts once the I-bit is cleared. Execution of the STOP or WAIT instructions causes the IRQE bit to be set in order to allow the external IRQ to exit these modes.
- 0 = The IRQF flag bit cannot generate an interrupt sequence.

4.4.3 Optional External Interrupts (PA0-PA3)

The IRQ interrupt can also be triggered by the inputs on the PA0 thru PA3 port pins if enabled by a single mask option. If enabled, the lower four bits of Port A can activate the IRQ interrupt function, and the interrupt operation will be the same as for inputs to the $\overline{\text{IRQ}}$ pin. This mask option of PA0-3 interrupt allow all of these input pins to be OR'ed with the input present on the $\overline{\text{IRQ}}$ pin. All PA0 thru PA3 pins must be selected as a group as an additional IRQ interrupt. All the PA0-3 interrupt sources are also controlled by the IRQE enable bit.

NOTE

The BIH and BIL instructions will only apply to the level on the $\overline{\text{IRQ}}$ pin itself, and not to the output of the logic OR function with the PA0 thru PA3 pins. The state of the individual Port A pins can be checked by reading the appropriate Port A pins as inputs.

NOTE

If enabled, the PA0 to PA3 pins will cause an IRQ interrupt regardless of whether these pins are configured as inputs or outputs.

4.4.4 Timer1 Interrupt (TIMER1)

The TIMER interrupt is generated by the timer when either a timer1 overflow or a input capture or output compare has occurred as described in **Section 9**. The interrupt flags and enable bits for the Timer1 interrupts are located in the Timer1 Control & Status Register (TSR) located at \$0012, \$0013. The I-bit in the CCR must be clear in order for the TIMER1 interrupt to be enabled. Either of these three interrupts will vector to the same interrupt service routine located at the address specified by the contents of memory locations \$1FF6 and \$1FF7.

4.4.5 USB Interrupt (USB)

The USB interrupt is generated by the USB module as described in **Section 10**. The interrupt enable bits for the USB interrupt are located at bit3-bit2 of UIR0 REG and bit3-bit2 of UIR1 REG. Also Once the device goes into Suspend Mode, any bus activities will cause the USB to generate an interrupt to CPU to come out from the Suspend mode. The I-bit in the CCR must be clear in order for the USB interrupt to be enabled. Either of these two interrupts will vector to the same interrupt service routine located at the address specified by the contents of memory locations \$1FF8 and \$1FF9.

4.4.6 MFT Interrupt (MFT)

The MFT interrupt is generated by the MFT module as described in **Section 8**. These interrupts will vector to the same interrupt service routine located at the address specified by the contents of memory locations \$1FF4 and \$1FF5.

SECTION 5 RESETS

The MCU can be reset in five ways:

- by an active low input to the $\overline{\text{RESET}}$ pin,
- by initial power-on reset,
- by an USB reset,
- by an illegal address access, and
- by a low voltage reset function.

The $\overline{\text{RESET}}$ pin is an I/O pin as shown in **Figure 5-1**. The internal steering diode for discharge and pull-up device are not shown here. All the peripheral modules which drive external pins will be reset by the synchronous reset signal (RST) coming from a latch, which is synchronized to the internal bus clock and set by any of the five reset sources.

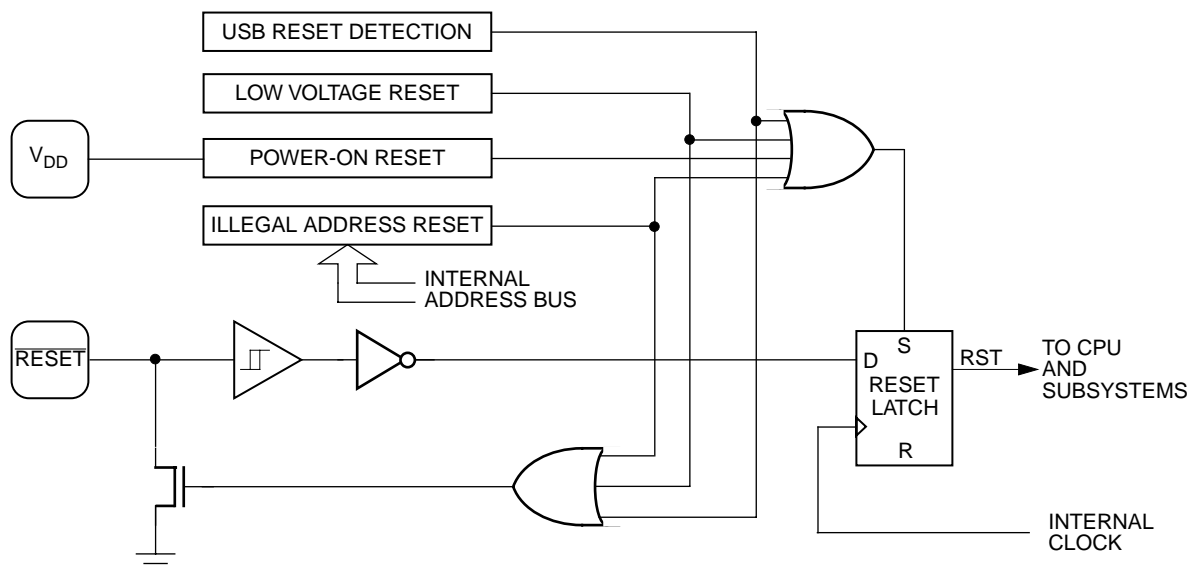


Figure 5-1. Reset Block Diagram

5.1 EXTERNAL RESET ($\overline{\text{RESET}}$)

The $\overline{\text{RESET}}$ pin is the only external source of a reset. This pin is connected to a Schmitt trigger input gate to provide an upper and lower threshold voltage separated by a minimum amount of hysteresis. This external reset occurs whenever the $\overline{\text{RESET}}$ pin is pulled below the lower threshold and remains in reset until the $\overline{\text{RESET}}$ pin rises above the upper threshold. This active low input will generate the RST signal and reset the CPU and peripherals. This pin is also an output pin whenever the LVR triggers an internal reset. Termination of the external RESET input or the USB reset or LVR are the only reset sources that can alter the operating mode of the MCU.

NOTE

Activation of the RST signal is generally referred to as reset of the device, unless otherwise specified.

5.2 INTERNAL RESETS

The four internally generated resets are the initial power-on reset function, the USB reset flag reset, the illegal address detector reset and the low voltage reset (LVR). Termination of the external RESET input or LVR or USB reset or ILADR are the reset sources that can alter the operating mode of the MCU. The other internal resets will not have any effect on the mode of operation when their reset state ends.

5.2.1 Power-On Reset (POR)

The internal POR is generated on power-up to allow the clock oscillator to stabilize. The POR is strictly for power turn-on conditions and is not able to detect a drop in the power supply voltage (brown-out). There is an oscillator stabilization delay of 128 or 4064 internal processor bus clock cycles (PH2) for Ceramic Resonator or Crystal after the oscillator becomes active.

The POR will generate the RST signal which will reset the CPU. If any other reset function is active at the end of this 128 or 4064 cycle delay, the RST signal will remain in the reset condition until the other reset condition(s) end.

5.2.2 USB Reset

The USB reset is generated by a detection on the USB bus reset signal. For MC68HC705JB2, seeing a single-end zero on its upstream port for 4 to 8 bit times will set RSTF bit in UIR0 register. The detections will also generate the RST signal to reset the CPU and other peripherals in the MCU.

5.2.3 Illegal Address Reset (ILADR)

The internal ILADR reset is generated when an instruction opcode fetch occurs from an address which is not implemented in the RAM (\$0080 - \$00FF) nor ROM (\$1600-\$1FFF). The ILADR will generate the RST signal which will reset the CPU and other peripherals. If any other reset function is active at the end of the ILADR reset signal, the RST signal will remain in the reset condition until the other reset condition(s) end. Notice that ILADR also forces the $\overline{\text{RESET}}$ pin low

5.2.4 Low Voltage Reset (LVR)

The internal LVR reset is generated when V_{DD} falls below the specified LVR trigger value V_{LVR} for at least one t_{CYC} . In typical applications, the power supply decoupling circuit will eliminate negative-going voltage glitches of less than one t_{CYC} . This reset will hold the MCU in the reset state until V_{DD} rises above V_{LVR} . Whenever V_{DD} is above V_{LVR} and below 4.2V, the MCU is guaranteed to operate although not within specification. The output from the LVR is connected directly to the internal reset circuitry and also forces the $\overline{\text{RESET}}$ pin low. The internal reset will be removed once the power supply voltage rises above V_{LVR} , at which time a normal power-on-reset sequence occurs. LVR function will still be active during Stop or Suspend mode.

SECTION 6

LOW POWER MODES

The MC68HC705JB2 has two low-power operating modes: STOP mode and WAIT mode.

The STOP and WAIT instructions provide two modes that reduce the power required for the MCU by stopping various internal clocks and/or the oscillator. The flow of the STOP, and WAIT modes are shown in **Figure 6-1**.

6.1 STOP MODE

Execution of the STOP instruction in this mode places the MCU in its lowest power consumption mode. In the STOP Mode the internal oscillator is turned off, halting all internal processing.

Execution of the STOP instruction automatically clears the I-bit in the Condition Code Register and sets the IRQE enable bit in the IRQ Control/Status Register so that the IRQ external interrupt is enabled. All other registers, including the other bits in the TCSR, and memory remain unaltered. All input/output lines remain unchanged.

The MCU can be brought out of the STOP Mode by an IRQ external interrupt or a USB coming out from Suspend Mode Interrupt (Bus activity detection) or an externally generated RESET, USB Reset or an LVR reset. When exiting the STOP Mode the internal oscillator will resume after a 128 or 4064 internal processor clock cycle oscillator stabilization delay.

6.2 WAIT MODE

The WAIT instruction places the MCU in a low-power mode, which consumes more power than the STOP Mode. In the WAIT Mode the internal processor clock is halted, suspending all processor and internal bus activity. Execution of the WAIT instruction automatically clears the I-bit in the Condition Code Register and sets the IRQE enable bit in the IRQ Control/Status Register so that the IRQ external interrupt is enabled. All other registers, memory, and input/output lines remain in their previous states.

The WAIT Mode may be exited when an external IRQ or a USB or Timer1 or MFT interrupt, an LVR reset or an external RESET occurs.

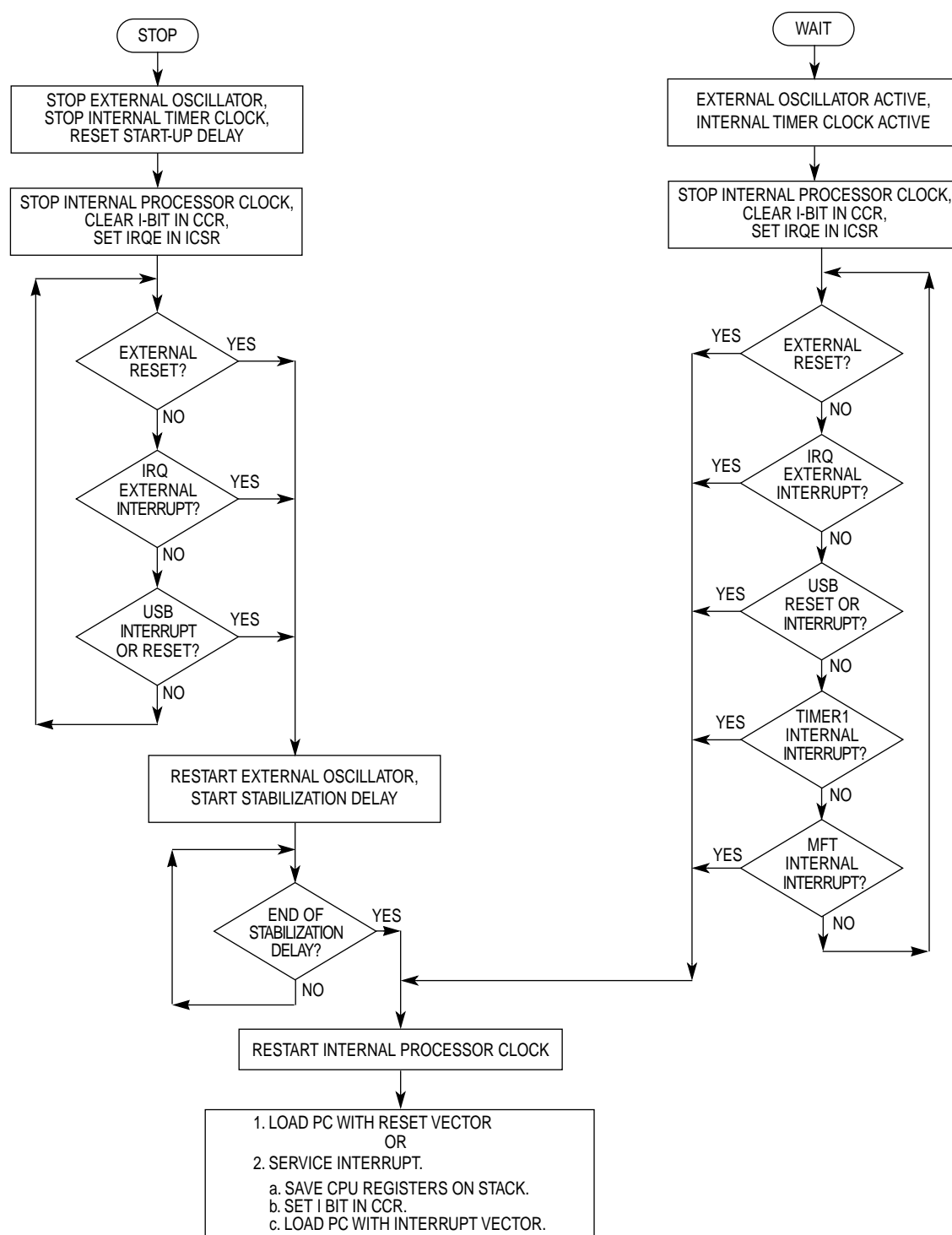


Figure 6-1. STOP/WAIT Flowchart

6.3 DATA-RETENTION MODE

The contents of RAM and CPU registers are retained at supply voltages as low as 2.0Vdc. This is called the data-retention mode where the data is held, but the device is not guaranteed to operate. The $\overline{\text{RESET}}$ pin must be held low during data-retention mode.

NOTE

The voltage threshold of the LVR is higher than the Data-Retention Mode minimum voltage, therefore the Data-Retention mode will not be available if the LVR function is enabled in the mask option.

SECTION 7

INPUT/OUTPUT PORTS

In the Normal Operating Mode, there are 11 bidirectional I/O lines arranged as one 8-bit I/O port (Port A), and one 3-bit I/O port (Port B). Each port line can be programmed as either input or output, under software control, by the data direction registers (DDR's). Also, if enabled by a mask option, all Port A and Port B I/O pins may have individual software programmable pulldown or pullup devices. PA4 to PA7 and PB1 & PB2 pins have the additional properties of sinking higher current.

PA0 to PA3 may function as additional IRQ interrupt input sources (mask option).

PB1 and PB2 have open drain output drivers, with optional slow falling-edge output transitions. The transition delay is 170ns (typical), with a bus rate of 3MHz and a loading of 50pF.

7.1 PORT A

Port A is an 8-bit bidirectional port which shares four of its pins with the IRQ interrupt system as shown in **Figure 7-1**. Each Port A pin is controlled by the corresponding bits in a data direction register, a data register, and a pulldown register. The Port A Data Register is located at address \$0000. The Port A Data Direction Register (DDRA) is located at address \$0004. The Port A Pulldown Register (PDURA) is located at address \$0010. Reset clears the DDRA and the PDURA. The Port A Data Register is unaffected by reset.

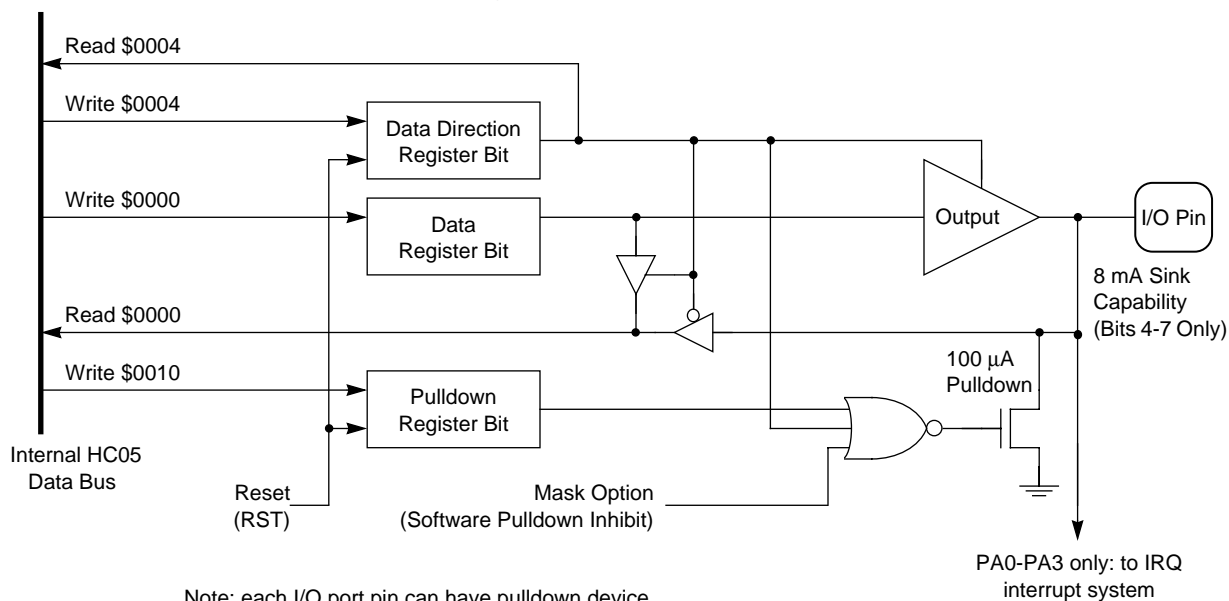


Figure 7-1. Port A I/O Circuitry

7.1.1 Port A Data Register

Each Port A I/O pin has a corresponding bit in the Port A Data Register. When a Port A pin is programmed as an output the state of the corresponding data register bit determines the state of the output pin. When a Port A pin is programmed as an input, any read of the Port A Data Register will return the logic state of the corresponding I/O pin. The Port A data register is unaffected by reset.

7.1.2 Port A Data Direction Register

Each Port A I/O pin may be programmed as an input by clearing the corresponding bit in the DDRA, or programmed as an output by setting the corresponding bit in the DDRA. The DDRA can be accessed at address \$0004. The DDRA is cleared by reset.

7.1.3 Port A Pulldown Register

All Port A I/O pins may have software programmable pulldown devices enabled by a mask option. If the pulldown/up mask option is selected, the pulldown is activated whenever the corresponding bit in the PDURA is clear. If the corresponding bit in the PDURA bit is set or the mask option for pulldown is not chosen, the pulldown will be disabled. A pulldown on an I/O pin is activated only if the I/O pin is programmed as an input.

The PDURA is a **write-only** register. Any reads of location \$0010 will return undefined results. Since reset clears both the DDRA and the PDURA, all pins will initialize as inputs with the pulldown active (if enabled by mask option).

7.1.4 Port A Drive Capability

The outputs of the PA4, PA5, PA6 and PA7 are capable of sinking 8 mA (typical) of current to V_{SS} .

7.1.5 Port A I/O Pin Interrupts

The inputs to PA0, PA1, PA2, PA3 may be connected to the IRQ input of the CPU if enabled by a mask option. PA0 to PA4 also has a Schmitt trigger circuit implemented as part of its input circuitry.

When connected as an alternate source of an IRQ interrupt, PA0-3 input pins will behave the same as the \overline{IRQ} pin itself, except that their active state is a logical one or a rising edge. The \overline{IRQ} pin has an active state that is a logical zero or a falling edge.

If the mask option for edge-and-level trigger sensitivity interrupts are chosen, the presence of a logic one or occurrence of a rising edge on any one of the lower four Port A pins will cause an IRQ interrupt request. If the edge-only sensitivity is selected, the occurrence of a rising edge on any one of the lower four Port A pins will cause an IRQ interrupt request. As long as any one of the lower four Port A IRQ inputs remains at a logic one level, the other of the lower four Port A IRQ inputs are effectively ignored.

NOTE

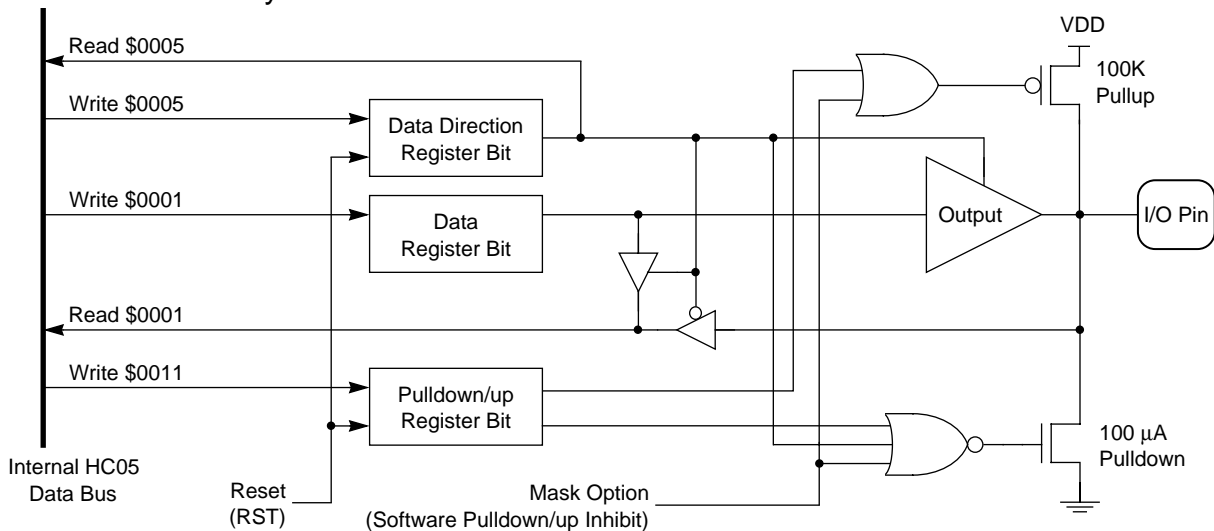
The BIH and BIL instructions will only apply to the level on the \overline{IRQ} pin itself, and not to the internal IRQ input to the CPU. Therefore BIH and BIL cannot be used to test the state of the lower four Port A input pins as a group.

7.2 PORT B

Port B is a 3-bit bidirectional port which functions as shown in **Figure 7-2**. Each Port B pin is controlled by the corresponding bits in a data direction register, a data register, and a pulldown/up register. The Port B Data Register is located at address \$0001. The Port B Data Direction Register (DDRB) is located at address \$0005. The Port B Pulldown/up Register (PDURB) is located at address \$0011. Reset clears the DDRB and the PDURB. The Port B Data Register is unaffected by reset.

PB0 is a standard push-pull I/O pin with pulldown option; and is shared with TCAP.

PB1 and PB2 are of open-drain type, with pullup option, each capable of sinking 25mA (typical) current at V_{OL} 0.5V max. These two pins may be connected together to constitute a single pin capable of sinking 50mA (typical). In this case, both PB1 and PB2 data bits will have to be written with the same value at the same write cycle.



Note: Each I/O port pin can have either pullup or pulldown device, but not both.
PB1 and PB2 output drivers are of open-drain type

Figure 7-2. Port B I/O Circuitry

7.2.1 Port B Data Register

All Port B I/O pins have a corresponding bit in the Port B Data Register. When a Port B pin is programmed as an output the state of the corresponding data register bit determines the state of the output pin. When a Port B pin is programmed as an input, any read of the Port B Data Register will return the logic state of the corresponding I/O pin. The Port B data register is unaffected by reset. Unused bits will always read as logic zeros, and any write to these bits will be ignored. The Port B data register is unaffected by reset.

7.2.2 Port B Data Direction Register

Port B I/O pins may be programmed as an input by clearing the corresponding bit in the DDRB, or programmed as an output by setting the corresponding bit in the DDRB. The DDRB can be accessed at address \$0005. Unused bits will always read as logic zeros, and any write to these bits will be ignored. The DDRB is cleared by reset.

If configured as output pins, PB1 and PB2 have slow output falling-edge transition feature. The slow transition feature is controlled by the SLOWE bit of DDRB. SLOWE bit, if set and if the pin is configured as an output pin, enables the slow falling-edge output transition feature of PB1 and PB2.

7.2.3 Slow Output Falling-Edge Transition

		BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
DDRB	R	SLOWE					DDRB2	DDRB1	DDRB0
\$0005	W								
reset:		0	0	0	0	0	0	0	0

Figure 7-3. Port B Data Direction Register

SLOWE - Slow Transition Enable

The slow transition feature is controlled by the SLOWE bit of DDRB (Port B Data Direction Register). Default value of SLOWE bit is clear on reset.

- 1 = Enables the slow falling-edge output transition feature on both PB1 and PB2, if the pin is configured as an output pin. PB2 falling edge transition is a sharp falling edge transition delayed by $t_{CYC}/2$ after the write cycle to PB2 data register. PB1 is a true slow transition I/O line.
- 0 = Disables slow falling-edge output transition feature on both PB1 and PB2.

7.2.4 Port B Pulldown/Pullup Register

All Port B I/O pins may have software programmable pulldown/pullup devices enabled by a mask option. If the pulldown/pullup mask option is selected, the pulldown/pullup is activated whenever the corresponding bit in the PDURB is clear. A pulldown on an I/O pin is activated only if the I/O pin is programmed as an input; whereas a pullup device on an I/O pin is always activated whenever enabled, regardless of port direction.

The PDURB is a write-only register. Any reads of location \$0011 will return undefined results. Since reset clears both the DDRB and the PDURB, all pins will initialize as inputs with the pulldown devices active and pullup devices active (if chosen via mask option).

Typical value of PB1 and PB2 pullup is 100K Ω (typical).

7.3 I/O PORT PROGRAMMING

All I/O pins can be programmed as inputs or outputs, with or without pulldown/pullup devices.

7.3.1 Pin Data Direction

The direction of a pin is determined by the state of its corresponding bit in the associated port Data Direction Register (DDR). A pin is configured as an output if its corresponding DDR bit is set to a logic one. A pin is configured as an input if its corresponding DDR bit is cleared to a logic zero.

The data direction bits DDRB0 thru DDRB2 and DDRA0 thru DDRA7 are read/write bits which can be manipulated with read-modify-write instructions. At power-on or reset, all DDR's are cleared which configures all port pins as inputs. If the pulldown/up mask option is chosen, all pins will initially power-up with their software programmable Pulldown/ups enabled.

7.3.2 Output Pin

When an I/O pin is programmed as an output pin, the state of the corresponding data register bit will determine the state of the pin. The state of the data register bits can be altered by writing to address \$0000 for Port A and address \$0001 for Port B. Reads of the corresponding data register bit at address \$0000 or \$0001 will return the state of the data register bit (not the state of the I/O pin itself). Therefore bit manipulation is possible on all pins programmed as outputs.

If the corresponding bit in the pulldown/up register is clear (and the pulldown/up mask option is chosen), only output pins with pullups have an activated pullup device connected to the pin. For those pins with pulldowns and configured as output pins, the pulldowns will be inactivated regardless of the state of the corresponding pulldown/up register bit. Since the pulldown/up register bits are write-only, bit manipulation should not be used on these register bits.

7.3.3 Input Pin

When an I/O pin is programmed as an input pin, the state of the pin can be determined by reading the corresponding data register bit. Any writes to the corresponding data register bit for an input pin will be ignored in the sense that the written value will not be reflected on the pin, rather it is only reflected in the port data register. Please refer to **Table 7-1** and **Table 7-2** for details.

If the corresponding bit in the pulldown/up register is clear (and the pulldown/up mask option is chosen) the input pin will also have an activated pulldown/up device. Since the pulldown/up register bits are write-only, bit manipulation should not be used on these register bits.

7.3.4 I/O Pin Transitions

A "glitch" can be generated on an I/O pin when changing it from an input to an output unless the data register is first preconditioned to the desired state before changing the corresponding DDR bit from a zero to a one.

If pulldowns are enabled by mask option, a floating input can be avoided by clearing the pulldown/pullup register bit before changing the corresponding DDR from a one to a zero. This will insure that the pulldown device will be activated before the I/O pin changes from a driven output to a pulled low/high input.

7.3.5 I/O Pin Truth Tables

Every pin on Port A and Port B may be programmed as an input or an output under software control as shown in **Table 7-1** and **Table 7-2**. All port I/O pins may also have software programmable pulldown/pullup devices if selected by the appropriate mask option.

Table 7-1. Port A I/O Pin Functions

DDRA	I/O Pin Mode	Accesses to PDURA at \$0010		Accesses to DDRA @ \$0004	Accesses to Data Register @ \$0000	
		Read	Write	Read/Write	Read	Write
0	IN, Hi-Z	U	PDURA0-7	DDRA0-7	I/O Pin PA0-7	*
1	OUT	U	PDURA0-7	DDRA0-7		PA0-7

NOTE: U is undefined.

* Does not affect input, but stored to data register

Table 7-2. Port B I/O Pin Functions

DDRB	I/O Pin Mode	Accesses to PDURB at \$0011		Accesses to DDRB @ \$0005	Accesses to Data Register @ \$0001	
		Read	Write	Read/Write	Read	Write
0	IN, Hi-Z	U	PDURB0-2	DDRB0-2	I/O Pin PB0-2	*
1	OUT	U	PDURB0-2	DDRB0-2		PB0-2

NOTE: U is undefined.

* Does not affect input, but stored to data register

SECTION 8

MULTI-FUNCTION TIMER

The MC68HC705JB2 core timer is a multi-function ripple counter. The features include Timer Over Flow (TOF) and Power-On Reset (POR).

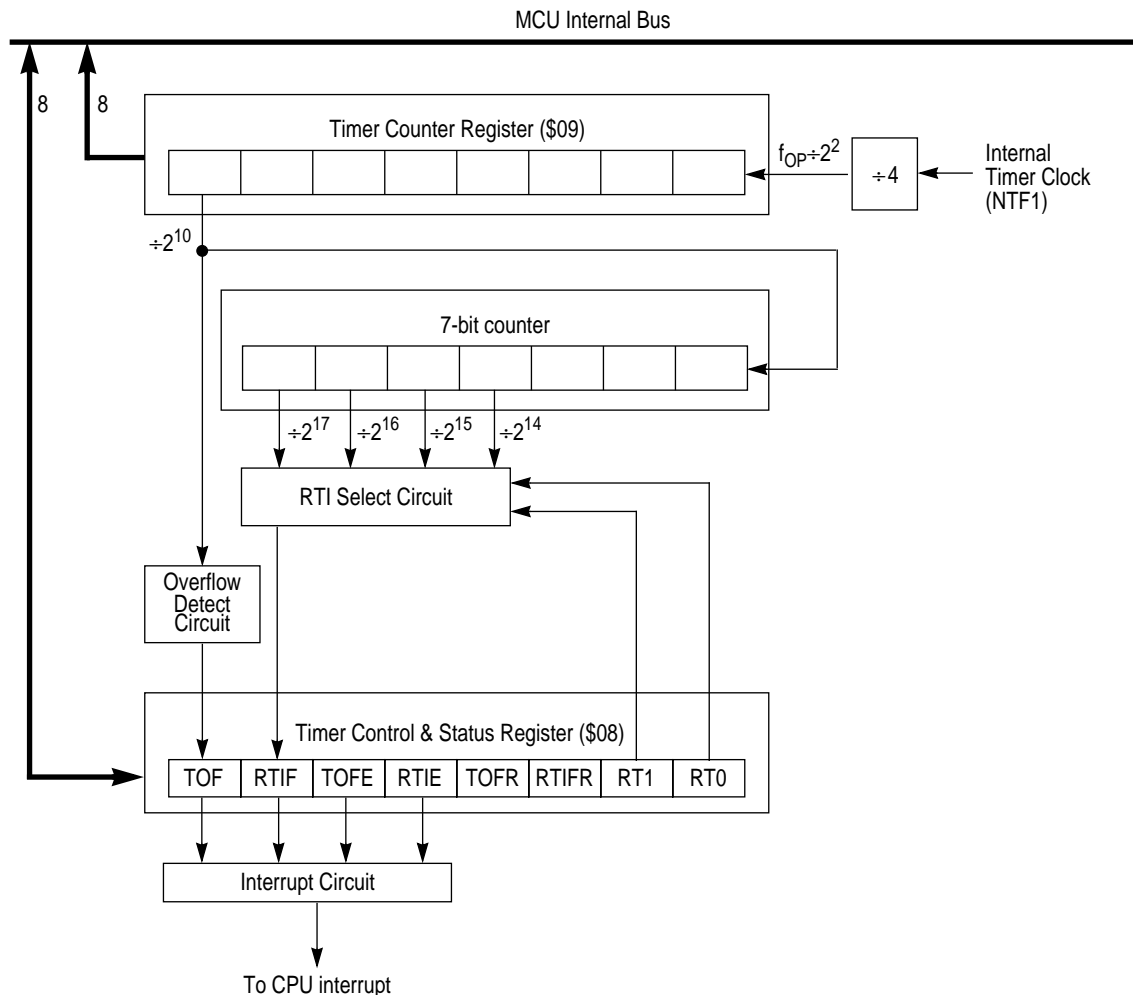


Figure 8-1. Multi-Function Timer Block Diagram

As shown in **Figure 8-1**, the Timer is driven by the timer clock, NTF1, divided by four ($\div 4$). NTF1 has the same phase and frequency as the processor bus clock, PH2, but is not stopped by the WAIT Modes. This signal drives an 8-bit ripple counter. The value of this 8-bit ripple counter can be read by the CPU at any time by accessing the Timer Counter Register (TCNT) at address \$09. A timer overflow

function is implemented on the last stage of this counter, giving a possible interrupt at the rate of $f_{op}/1024$. Two additional stages produce the POR function at $f_{op}/4064$. The Timer Counter Bypass circuitry (available only in Expanded Test Mode) is at this point in the timer chain. This circuit is followed by two more stages, with the resulting clock ($f_{op}/16384$) driving the Real Time Interrupt circuit. The RTI circuit consists of three divider stages with a 1 of 4 selector. The RTI rate selector bits, and the RTI and TOF enable bits and flags are located in the Timer Control and Status Register at location \$08.

The Real Time Interrupt circuit consists of a three stage divider and a 1 of 4 selector. The clock frequency that drives the RTI circuit is $f_{op}/2^{14}$ (or $f_{op}/16384$) with three additional divider stages giving a maximum interrupt period of $f_{op}/2^{17}$ (or $f_{op}/131072$).

The power-on cycle clears the entire counter chain and begins clocking the counter. After 128 or 4064 cycles, the power-on reset circuit is released which again clears the counter chain and allows the device to come out of reset. At this point, if \overline{RESET} is not asserted, the timer will start counting up from zero and normal device operation will begin. If \overline{RESET} is asserted at any time during operation the counter chain will be cleared.

8.1 TIMER REGISTERS

The 15-stage Multi-function Timer contains two registers: a Timer Counter Register and a Timer Control/Status Register.

8.1.1 Timer Counter Register (TCNT) \$09

The Timer Counter Register is a read-only register which contains the current value of the 8-bit ripple counter at the beginning of the timer chain. This counter is clocked at f_{op} divided by 4 and can be used for various functions including a software input capture. Extended time periods can be attained using the TOF function to increment a temporary RAM storage location thereby simulating a 16-bit (or more) counter. The value of each bit of the TCNT is shown in **Figure 8-2**. This register is cleared by reset.

		BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
TCNT	R	TMR7	TMR6	TMR5	TMR4	TMR3	TMR2	TMR1	TMR0
\$0009	W								
reset:		0	0	0	0	0	0	0	0

Figure 8-2. Timer Counter Register

8.1.2 Timer Control/Status Register (TCSR) \$08

The TCSR contains the timer interrupt flag bits, the timer interrupt enable bits, and the real time interrupt rate select bits. Bit 2 and bit 3 are write-only bits which will read as logical zeros. **Figure 8-3** shows the value of each bit in the TCSR following reset.

	BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
TCSR	R	TOF	RTIF	TOFE	RTIE	0	0	RT1
\$0008	W					TOFR	RTIFR	
reset:	0	0	0	0	0	0	0	1

Figure 8-3. Timer Control/Status Register (TCSR)

TOF - Timer Overflow Flag

The TOF is a read-only flag bit.

- 1 = Set when the 8-bit ripple counter rolls over from \$FF to \$00. A TIMER Interrupt request will be generated if TOFE is also set.
- 0 = Reset by writing a logical one to the TOF acknowledge bit, TOFR. Writing to the TOF flag bit has no effect on its value. This bit is cleared by reset.

RTIF - Real Time Interrupt Flag

The RTIF is a read-only flag bit.

- 1 = Set when the output of the chosen (1 of 4 selections) Real Time Interrupt stage goes active. A TIMER Interrupt request will be generated if RTIE is also set.
- 0 = Reset by writing a logical one to the RTIF acknowledge bit, RTIFR. Writing to the RTIF flag bit has no effect on its value. This bit is cleared by reset.

TOFE - Timer Overflow Enable

The TOFE is an enable bit that allows generation of a TIMER Interrupt upon overflow of the Timer Counter Register.

- 1 = When set, the TIMER Interrupt is generated when the TOF flag bit is set.
- 0 = When cleared, no TIMER interrupt caused by TOF bit set will be generated. This bit is cleared by reset.

RTIE - Real Time Interrupt Enable

The RTIE is an enable bit that allows generation of a TIMER Interrupt by the RTIF bit.

- 1 = When set, the TIMER Interrupt is generated when the RTIF flag bit is set.
- 0 = When cleared, no TIMER interrupt caused by RTIF bit set will be generated. This bit is cleared by reset.

TOFR - Timer Overflow Acknowledge

The TOFR is an acknowledge bit that resets the TOF flag bit. This bit is unaffected by reset. Reading the TOFR will always return a logical zero.

- 1 = Clears the TOF flag bit.
- 0 = Does not clear the TOF flag bit.

RTIFR - Real Time Interrupt Acknowledge

The RTIFR is an acknowledge bit that resets the RTIF flag bit. This bit is unaffected by reset. Reading the RTIFR will always return a logical zero.

- 1 = Clears the RTIF flag bit.
- 0 = Does not clear the RTIF flag bit.

RT1, RT0 - Real Time Interrupt period select bits

These two bits select one of the four real time interrupt periods.

		Bus Frequency, $f_{BUS}=f_{OP}=3.0\text{ MHz}$	
RT1	RT0	Divide Ratio	RTI Rate
0	0	2^{14}	5.46ms
0	1	2^{15}	10.92ms
1	0	2^{16}	21.85ms
1	1	2^{17}	43.69ms

8.2 OPERATION DURING STOP MODE

When STOP is exited by an external interrupt or an LVR reset or an external $\overline{\text{RESET}}$, the internal oscillator will resume, followed by a 128 or 4064 internal processor oscillator stabilization delay.

SECTION 9

PROGRAMMABLE TIMER

This 16-bit Programmable Timer (Timer1) has an Input Capture function and an Output Compare function. **Figure 9-1** shows a block diagram of the 16-bit programmable timer.

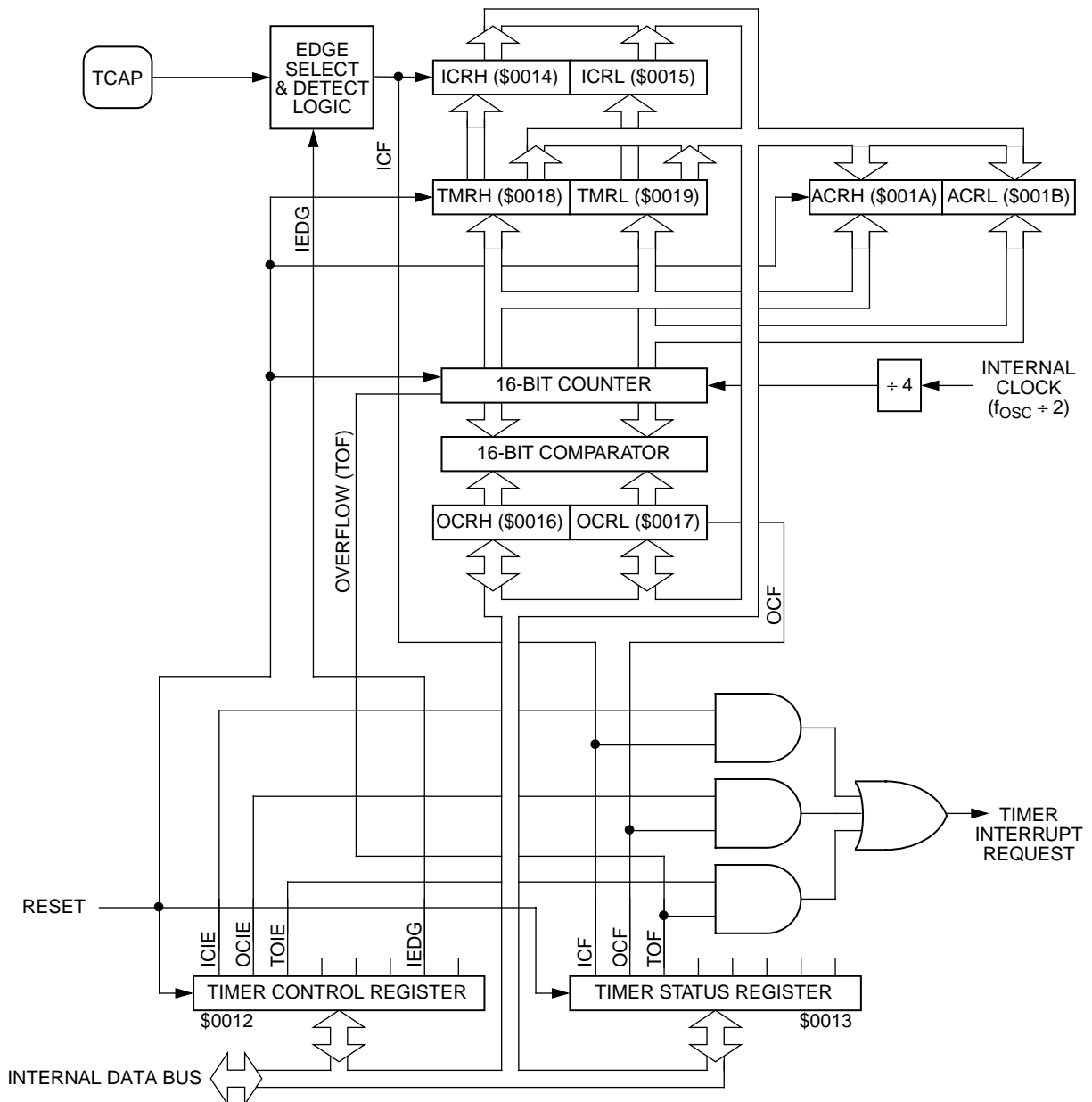


Figure 9-1. Programmable Timer Block Diagram

The basis of the 16-bit Timer is a 16-bit free-running counter which increases in count with each internal bus clock cycle. The counter is the timing reference for the input capture and output compare functions. The input capture and output compare functions provide a means to latch the times at which external events occur, to measure input waveforms, and to generate output waveforms and timing delays. Software can read the value in the 16-bit free-running counter at any time without affect the counter sequence.

Because of the 16-bit timer architecture, the I/O registers for the input capture and output compare functions are pairs of 8-bit registers. Each register pair contains the high and low byte of that function. Generally, accessing the low byte of a specific timer function allows full control of that function; however, an access of the high byte inhibits that specific timer function until the low byte is also accessed.

Because the counter is 16 bits long and preceded by a fixed divide-by-four prescaler, the counter rolls over every 262,144 internal clock cycles. Timer resolution with a 4MHz crystal oscillator is 2 microsecond/count.

The interrupt capability, the input capture edge, and the output compare state are controlled by the timer control register (TCR) located at \$0012 and the status of the interrupt flags can be read from the timer status register (TSR) located at \$0013.

9.1 TIMER REGISTERS (TMRH, TMRL)

The functional block diagram of the 16-bit free-running timer counter and timer registers is shown in **Figure 9-2**. The timer registers include a transparent buffer latch on the LSB of the 16-bit timer counter.

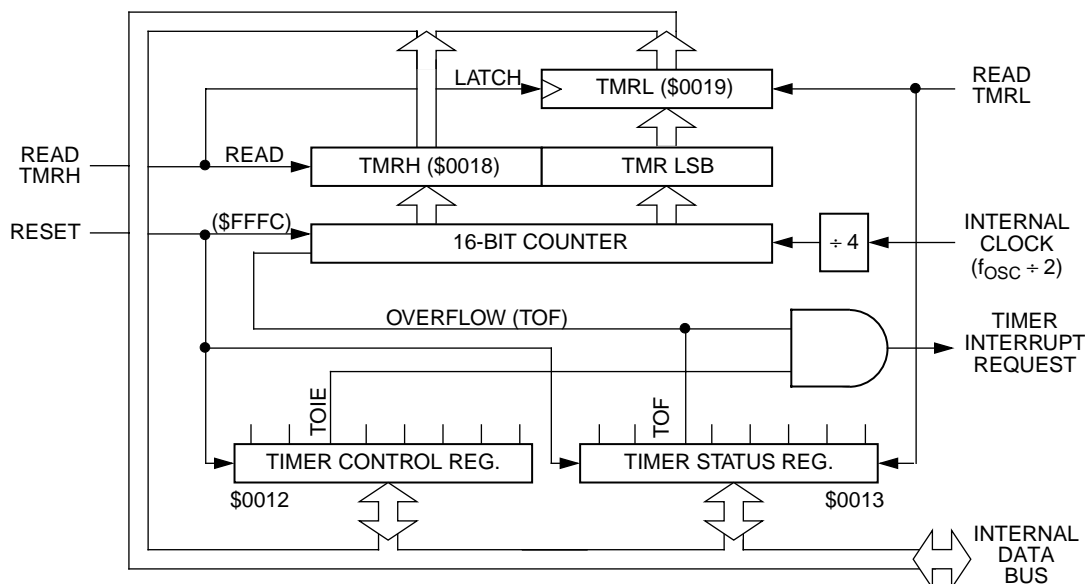


Figure 9-2. Programmable Timer Counter Block Diagram

The timer registers (TMRH, TMRL) shown in **Figure 9-3** are read-only locations which contain the current high and low bytes of the 16-bit free-running counter. Writing to the timer registers has no effect. Reset of the device presets the timer counter to \$FFFC.

		BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
TMRH	R	TMRH7	TMRH6	TMRH5	TMRH4	TMRH3	TMRH2	TMRH1	TMRH0
\$0018	W								
reset:		1	1	1	1	1	1	1	1

TMRL	R	TMRL7	TMRL6	TMRL5	TMRL4	TMRL3	TMRL2	TMRL1	TMRL0
\$0019	W								
reset:		1	1	1	1	1	1	0	0

Figure 9-3. Programmable Timer Counter Registers (TMRH, TMRL)

The TMRL latch is a transparent read of the LSB until the a read of the TMRH takes place. A read of the TMRH latches the LSB into the TMRL location until the TMRL is again read. The latched value remains fixed even if multiple reads of the TMRH take place before the next read of the TMRL. Therefore, when reading the MSB of the timer at TMRH the LSB of the timer at TMRL must also be read to complete the read sequence.

During power-on-reset (POR), the counter is initialized to \$FFFC and begins counting after the oscillator start-up delay. Because the counter is sixteen bits and preceded by a fixed divide-by-four prescaler, the value in the counter repeats every 262, 144 internal bus clock cycles (524, 288 oscillator cycles).

When the free-running counter rolls over from \$FFFF to \$0000, the timer overflow flag bit (TOF) is set in the TSR. When the TOF is set, it can generate an interrupt if the timer overflow interrupt enable bit (TOIE) is also set in the TCR. The TOF flag bit can only be reset by reading the TMRL after reading the TSR.

Other than clearing any possible TOF flags, reading the TMRH and TMRL in any order or any number of times does not have any effect on the 16-bit free-running counter.

NOTE

To prevent interrupts from occurring between readings of the TMRH and TMRL, set the I bit in the condition code register (CCR) before reading TMRH and clear the I bit after reading TMRL.

9.2 ALTERNATE COUNTER REGISTERS (ACRH, ACRL)

The functional block diagram of the 16-bit free-running timer counter and alternate

counter registers is shown in **Figure 9-4**. The alternate counter registers behave the same as the timer registers, except that any reads of the alternate counter will not have any effect on the TOF flag bit and Timer interrupts. The alternate counter registers include a transparent buffer latch on the LSB of the 16-bit timer counter.

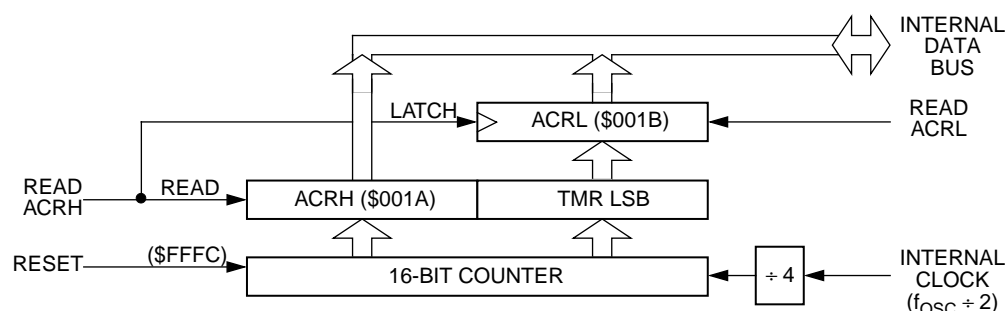


Figure 9-4. Alternate Counter Block Diagram

The alternate counter registers (ACRH, ACRL) shown in **Figure 9-5** are read-only locations which contain the current high and low bytes of the 16-bit free-running counter. Writing to the alternate counter registers has no effect. Reset of the device presets the timer counter to \$FFFC.

		BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
ACRH	R	ACRH7	ACRH6	ACRH5	ACRH4	ACRH3	ACRH2	ACRH1	ACRH0
\$001A	W								
reset:		1	1	1	1	1	1	1	1

		BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
ACRL	R	ACRL7	ACRL6	ACRL5	ACRL4	ACRL3	ACRL2	ACRL1	ACRL0
\$001B	W								
reset:		1	1	1	1	1	1	0	0

Figure 9-5. Alternate Counter Registers (ACRH, ACRL)

The ACRL latch is a transparent read of the LSB until the a read of the ACRH takes place. A read of the ACRH latches the LSB into the ACRL location until the ACRL is again read. The latched value remains fixed even if multiple reads of the ACRH take place before the next read of the ACRL. Therefore, when reading the MSB of the timer at ACRH the LSB of the timer at ACRL must also be read to complete the read sequence.

During power-on-reset (POR), the counter is initialized to \$FFFC and begins counting after the oscillator start-up delay. Because the counter is sixteen bits and preceded by a fixed divide-by-four prescaler, the value in the counter repeats every 262,144 internal bus clock cycles (524,288 oscillator cycles).

Reading the ACRH and ACRL in any order or any number of times does not have any effect on the 16-bit free-running counter or the TOF flag bit.

NOTE

To prevent interrupts from occurring between readings of the ACRH and ACRL, set the I bit in the condition code register (CCR) before reading ACRH and clear the I bit after reading ACRL.

9.3 INPUT CAPTURE REGISTERS

The input capture function is a technique whereby an external signal (connected to PB0/TCAP pin) is used to trigger the 16-bit timer counter. In this way it is possible to relate the timing of an external signal to the internal counter value, and hence to elapsed time.

When the input capture circuitry detects an active edge on the TCAP pin, it latches the contents of the free-running timer counter registers into the input capture registers as shown in **Figure 9-6**.

Latching values into the input capture registers at successive edges of the same polarity measures the period of the selected input signal. Latching the counter values at successive edges of opposite polarity measures the pulse width of the signal.

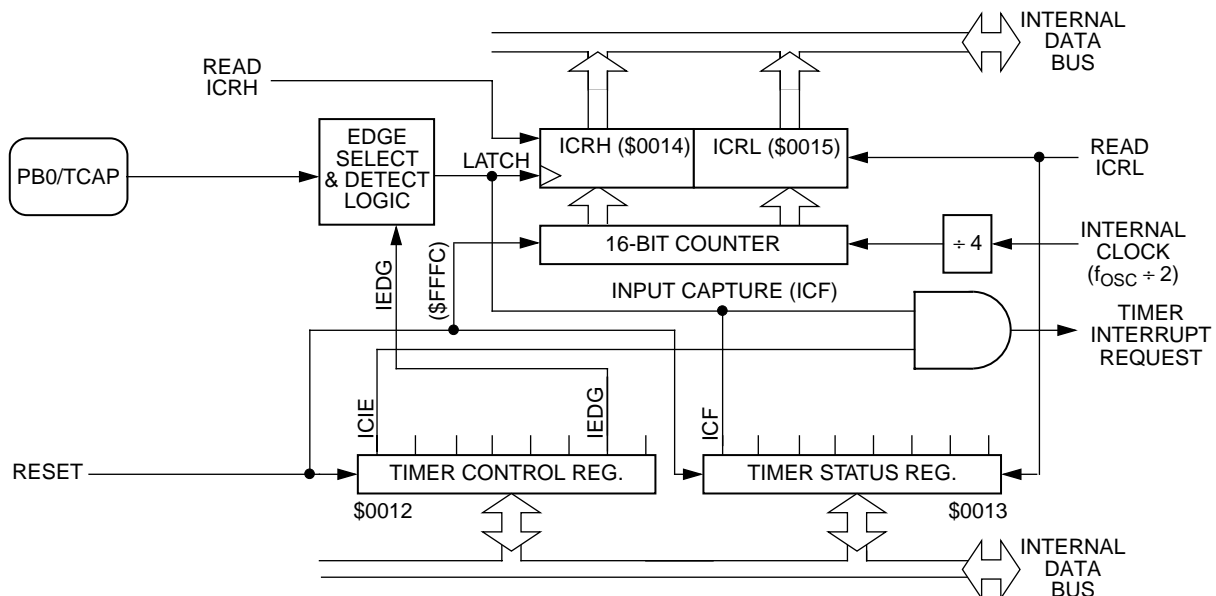


Figure 9-6. Timer Input Capture Block Diagram

The input capture registers are made up of two 8-bit read-only registers (ICRH, ICRL) as shown in **Figure 9-7**. The input capture edge detector contains a Schmitt

trigger to improve noise immunity. The edge that triggers the counter transfer is defined by the input edge bit (IEDG) in the TCR. Reset does not affect the contents of the input capture registers.

The result obtained by an input capture will be one count higher than the value of the free-running timer counter preceding the external transition. This delay is required for internal synchronization. Resolution is affected by the prescaler, allowing the free-running timer counter to increment once every four internal clock cycles (eight oscillator clock cycles).

		BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
ICRH	R	ICRH7	ICRH6	ICRH5	ICRH4	ICRH3	ICRH2	ICRH1	ICRH0
\$0014	W								
reset:		U	U	U	U	U	U	U	U

		ICRL7	ICRL6	ICRL5	ICRL4	ICRL3	ICRL2	ICRL1	ICRL0
ICRL	R								
\$0015	W								
reset:		U	U	U	U	U	U	U	U

U = UNAFFECTED BY RESET

Figure 9-7. Input Capture Registers (ICRH, ICRL)

Reading the ICRH inhibits further captures until the ICRL is also read. Reading the ICRL after reading the timer status register (TSR) clears the ICF flag bit. does not inhibit transfer of the free-running counter. There is no conflict between reading the ICRL and transfers from the free-running timer counters. The input capture registers always contain the free-running timer counter value which corresponds to the most recent input capture.

NOTE

To prevent interrupts from occurring between readings of the ICRH and ICRL, set the I bit in the condition code register (CCR) before reading ICRH and clear the I bit after reading ICRL.

9.4 OUTPUT COMPARE REGISTERS

The Output Compare function is a means of generating an interrupt when the 16-bit timer counter reaches a selected value as shown in **Figure 9-8**. Software writes the selected value into the output compare registers. On every fourth internal clock cycle (every eight oscillator clock cycle) the output compare circuitry compares the value of the free-running timer counter to the value written in the output compare registers. When a match occurs, the output compare interrupt flag, OCF is set. A timer interrupt request to the CPU is generated if the output compare interrupt enable is set, i.e. OCIE=1.

Software can use the output compare register to measure time periods, to generate timing delays, or to generate a pulse of specific duration or a pulse train of specific frequency and duty cycle.

Writing to the OCRH before writing to the OCRL inhibits timer compares until the OCRL is written. Reading or writing to the OCRL after reading the TSR will clear the output compare flag bit (OCF).

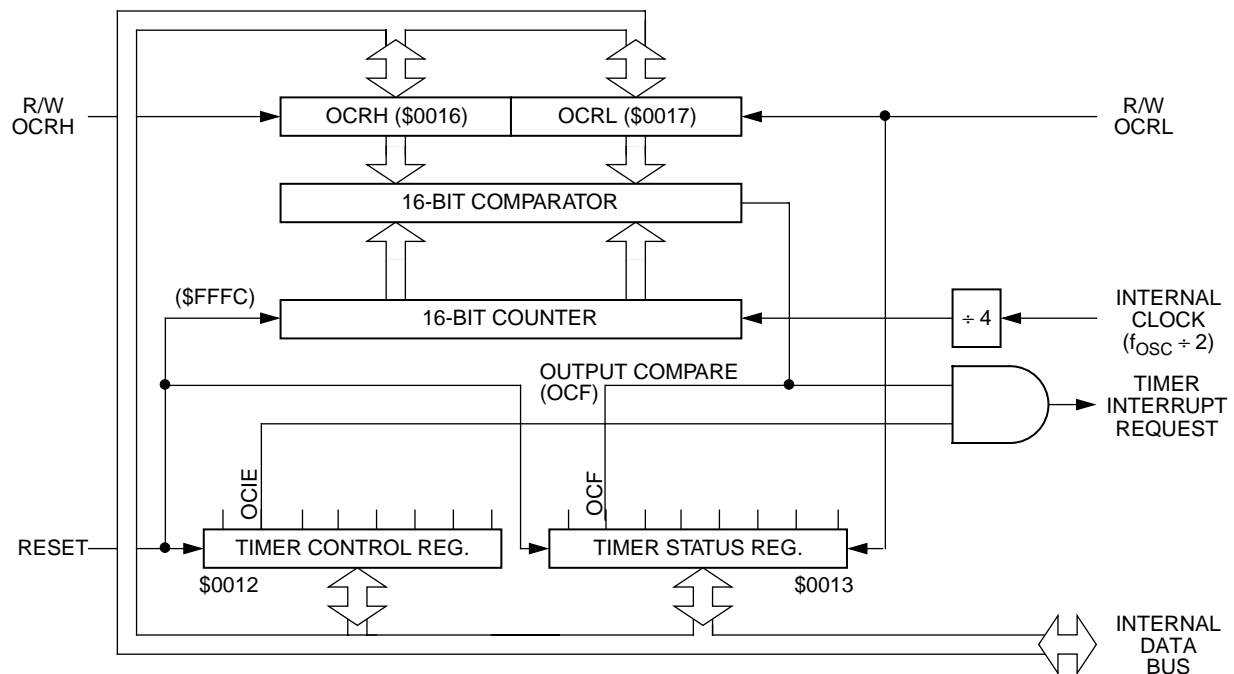


Figure 9-8. Timer Output Compare Block Diagram

		BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
OCRH	R	OCRH7	OCRH6	OCRH5	OCRH4	OCRH3	OCRH2	OCRH1	OCRH0
\$0016	W								
reset:		U	U	U	U	U	U	U	U
U = UNAFFECTED BY RESET									
OCRL	R	OCRL7	OCRL6	OCRL5	OCRL4	OCRL3	OCRL2	OCRL1	OCRL0
\$0017	W								
reset:		U	U	U	U	U	U	U	U
U = UNAFFECTED BY RESET									

Figure 9-9. Output Compare Registers (OCRH, OCRL)

To prevent OCF from being set between the time it is read and the time the output compare registers are updated, use the following procedure:

1. Disable interrupts by setting the I bit in the condition code register.
2. Write to the OCRH. Compares are now inhibited until OCRL is written.
3. Read the TSR to arm the OCF for clearing.
4. Enable the output compare registers by writing to the OCRL. This also clears the OCF flag bit in the TSR.
5. Enable interrupts by clearing the I bit in the condition code register.

A software example of this procedure is shown below.

```

9B          SEI          DISABLE INTERRUPTS
...         ...         .....
...         ...         .....
B7    16    STA    OCRH    INHIBIT OUTPUT COMPARE
B6    13    LDA    TSR     ARM OCF FLAG FOR CLEARING
BF    17    STX    OCRL    READY FOR NEXT COMPARE, OCF CLEARED
...         ...         .....
...         ...         .....
9A          CLI          ENABLE INTERRUPTS

```

9.5 TIMER CONTROL REGISTER (TCR)

The timer control register is shown in **Figure 9-10** performs the following functions:

- Enables input capture interrupts
- Enables output compare interrupts
- Enables timer overflow interrupts
- Control the active edge polarity of the TCAP signal on pin PB0/TCAP

Reset clears all the bits in the TCR with the exception of the IEDG bit which is unaffected.

		BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
TCR	R	ICIE	OCIE	TOIE	0	0	0	IEDG	0
\$0012	W								
reset:		0	0	0	0	0	0	Unaffected	0

Figure 9-10. Timer Control Register (TCR)

ICIE - INPUT CAPTURE INTERRUPT ENABLE

This read/write bit enables interrupts caused by an active signal on the PB0/TCAP pin. Reset clears the ICIE bit.

- 1 = Input capture interrupts enabled.
 0 = Input capture interrupts disabled.

OCIE - OUTPUT COMPARE INTERRUPT ENABLE

This read/write bit enables interrupts caused by a successful compare between the timer counter and the output compare registers. Reset clears the OCIE bit.

- 1 = Output compare interrupts enabled.
- 0 = Output compare interrupts disabled.

TOIE - TIMER OVERFLOW INTERRUPT ENABLE

This read/write bit enables interrupts caused by a timer overflow. Reset clears the TOIE bit.

- 1 = Timer overflow interrupts enabled.
- 0 = Timer overflow interrupts disabled.

IEDG - INPUT CAPTURE EDGE SELECT

The state of this read/write bit determines whether a positive or negative transition on the TCAP pin triggers a transfer of the contents of the timer register to the input capture register. Reset has no effect on the IEDG bit.

- 1 = Positive edge (low to high transition) triggers input capture.
- 0 = Negative edge (high to low transition) triggers input capture.

9.6 TIMER STATUS REGISTER (TSR)

The timer status register (TSR) shown in **Figure 9-11** contains flags for the following events:

- An active signal on the PB0/TCAP pin, transferring the contents of the timer registers to the input capture registers.
- A match between the 16-bit counter and the output compare registers
- An overflow of the timer registers from \$FFFF to \$0000.

Writing to any of the bits in the TSR has no effect. Reset does not change the state of any of the flag bits in the TSR.

		BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0
TSR	R	ICF	OCF	TOF	0	0	0	0	0
\$0013	W								
reset:		U	U	U	0	0	0	0	0

U = UNAFFECTED BY RESET

Figure 9-11. Timer Status Registers (TSR)

ICF - INPUT CAPTURE FLAG

The ICF bit is automatically set when an edge of the selected polarity occurs on the PB0/TCAP pin. Clear the ICF bit by reading the timer status register with the ICF set, and then reading the low byte (ICRL, \$0015) of the input capture registers. Reset has no effect on ICF.

OCF - OUTPUT COMPARE FLAG

The OCF bit is automatically set when the value of the timer registers matches the contents of the output compare registers. Clear the OCF bit by reading the timer status register with the OCF set, and then accessing the low byte (OCRL, \$0017) of the output compare registers. Reset has no effect on OCF.

TOF - TIMER OVERFLOW FLAG

The TOF bit is automatically set when the 16-bit timer counter rolls over from \$FFFF to \$0000. Clear the TOF bit by reading the timer status register with the TOF set, and then accessing the low byte (TMRL, \$0019) of the timer registers. Reset has no effect on TOF.

9.7 TIMER OPERATION DURING WAIT MODE

During WAIT mode the 16-bit timer continues to operate normally and may generate an interrupt to trigger the MCU out of the WAIT mode.

9.8 TIMER OPERATION DURING STOP MODE

When the MCU enters the STOP mode the free-running counter stops counting (the internal processor clock is stopped). It remains at that particular count value until the STOP mode is exited by applying a low signal to the $\overline{\text{IRQ}}$ pin, at which time the counter resumes from its stopped value as if nothing had happened. If STOP mode is exited via an external reset (logic low applied to the $\overline{\text{RESET}}$ pin) the counter is forced to \$FFFC.

If a valid input capture edge occurs at the PB0/TCAP pin during the STOP mode the input capture detect circuitry will be armed. This action does not set any flags or "wake up" the MCU, but when the MCU does "wake up" there will be an active input capture flag (and data) from the first valid edge. If the STOP mode is exited by an external reset, no input capture flag or data will be present even if a valid input capture edge was detected during the STOP mode.

SECTION 10

UNIVERSAL SERIAL BUS MODULE

This USB Module is designed for USB application in LS products. With minimized software effort, it can fully comply with USB LS device specification. See USB specification version 1.0 for the detail description of USB.

10.1 FEATURES

- Integrated 3.3 Volt Regulator with 3.3V Output Pin
- Integrated USB transceiver supporting Low Speed functions
- USB Data Control Logic
 - Packet decoding/generation
 - CRC generation and checking
 - NRZI encoding/decoding
 - Bit-stuffing
- USB reset support
- Control Endpoint 0 and Interrupt Endpoints 1 and 2
- Two 8-byte transmit buffers
- One 8-byte receive buffer
- Suspend and resume operations
- Remote Wake-up support
- USB generated interrupts
 - transaction interrupt driven
 - Resume interrupt
 - End of Packet interrupt
 - STALL, NAK, and ACK handshake generation

10.2 OVERVIEW

This section provides an overview of the Universal Serial Bus (USB) module in the MC68HC705JB2. This USB module is designed to serve as a low-speed (LS) USB device per the Universal Serial Bus Specification Rev 1.0. Three types of USB data transfers are supported: control, interrupt, and bulk (transmit only). Endpoint 0 functions as a receive/transmit control endpoint. Endpoints 1 and 2 can function as interrupt or bulk, but only in the transmit direction.

A block diagram of the USB module is shown **Figure 10-1**. The USB module manages communications between the host and the USB function. The module is partitioned into four functional blocks. These blocks consist of a 3.3 volt regulator, a dual function transceiver, the USB control logic, and the endpoint registers. The blocks are further detailed in **Section 10.4**.

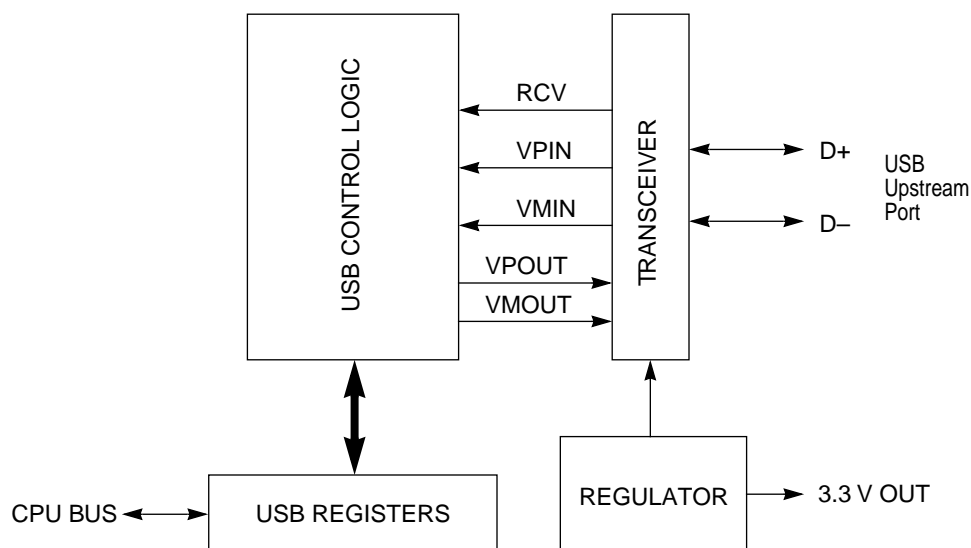


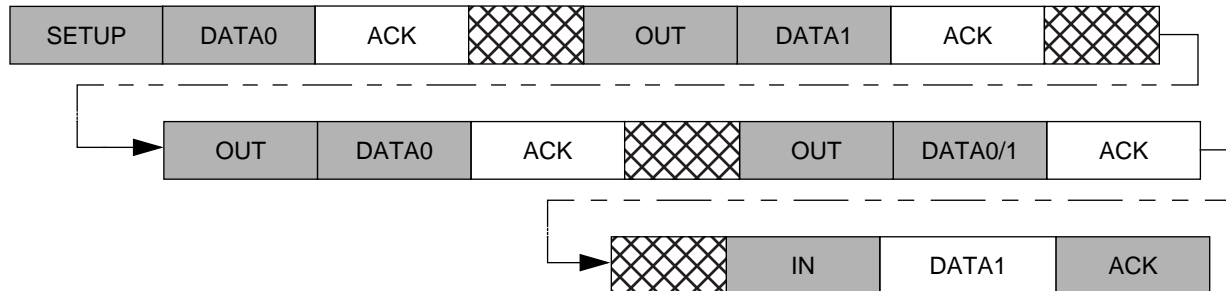
Figure 10-1. USB Block Diagram

10.2.1 USB Protocol

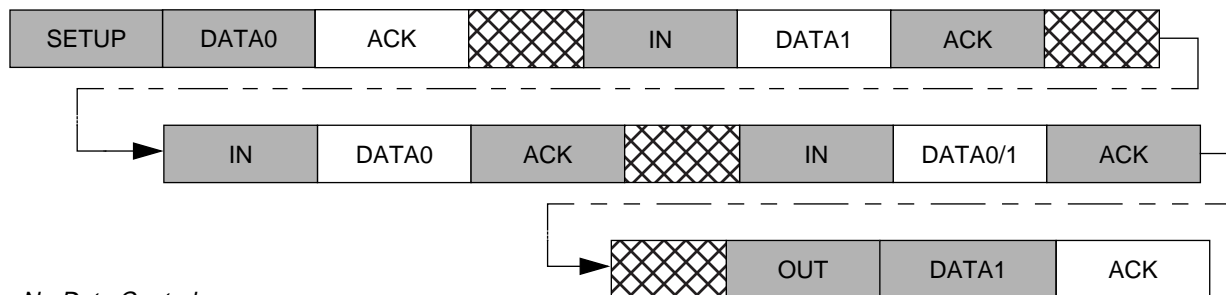
Figure 10-2 shows the various transaction types supported by the MC68HC705JB2 USB module. The transactions are portrayed as error free. The effect of errors in the data flow are discussed later.

ENDPOINT 0 TRANSACTIONS:

Control Write



Control Read

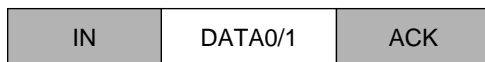


No-Data Control

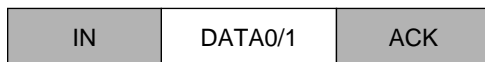


ENDPOINTS 1 & 2 TRANSACTIONS:

Interrupt



Bulk Transmit



KEY:

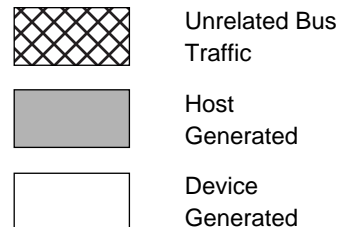
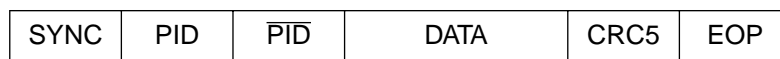
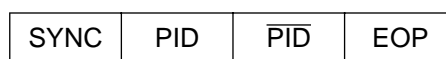


Figure 10-2. Supported Transaction Types per Endpoint

Each USB transaction is comprised of a series of packets. The MC68HC705JB2 USB module supports the packet types shown in **Figure 10-3**. Token packets are generated by the USB host and decoded by the USB device. Data and Handshake packets are both decoded and generated by the USB device depending on the type of transaction.

Token Packet:*IN**OUT**SETUP***Data Packet:***DATA0**DATA1*

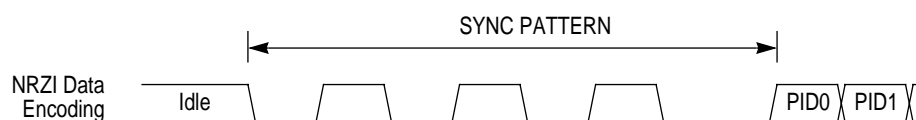
0 - 8 bytes

Handshake Packet:*ACK**NAK**STALL***Figure 10-3. Supported USB Packet Types**

The following sections will give some detail on each segment used to form a complete USB transaction.

10.2.1.1 Sync Pattern

The NRZI (See **Section 10.4.4.1**) bit pattern shown in **Figure 10-4** is used as a synchronization pattern and is prefixed to each packet. This pattern is equivalent to a data pattern of seven 0's followed by a 1 (0x80).

**Figure 10-4. Sync Pattern**

The start of a packet (SOP) is signaled by the originating port by driving the D+ and D- lines from the idle state (also referred to as the "J" state) to the opposite logic level (also referred to as the "K" state). This switch in levels represents the first bit of the Sync field. **Figure 10-5** shows the data signaling and voltage levels for the start of packet and the sync pattern.

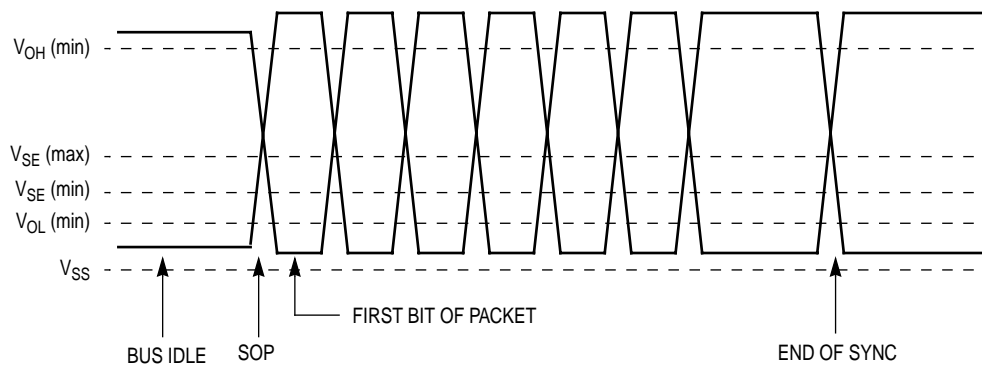


Figure 10-5. SOP, Sync Signaling and Voltage Levels

10.2.1.2 Packet Identifier Field

The Packet Identifier field is an eight bit number comprised of the four bit packet identification (PID) and its complement. The field follows the sync pattern and determines the direction and type of transaction on the bus. **Table 10-1** shows the PID values for the supported packet types.

Table 10-1. Supported Packet Identifiers

PID Value	PID Type
%1001	IN Token
%0001	OUT Token
%1101	SETUP Token
%0011	DATA0 Packet
%1011	DATA1 Packet
%0010	ACK Handshake
%1010	NAK Handshake
%1110	STALL Handshake

10.2.1.3 Address Field (ADDR)

The Address field is a seven bit number that is used to select a particular USB device. This field is compared to the lower seven bits of the UADDR register to determine if a given transaction is targeting the MC68HC705JB2 USB device.

10.2.1.4 Endpoint Field (ENDP)

The Endpoint field is a four bit number that is used to select a particular endpoint within a USB device. For the MC68HC705JB2, this will be a binary number between zero and two inclusive. Any other value will cause the transaction to be ignored.

10.2.1.5 Cyclic Redundancy Check (CRC)

Cyclic Redundancy Checks are used to verify the address and data stream of a USB transaction. This field is five bits wide for token packets and sixteen bits wide for data packets. CRCs are generated in the transmitter and sent on the USB data lines after both the endpoint field and the data field. **Figure 10-6** shows how the five bit CRC value is calculated from the data stream and verified for the address and endpoint fields of a token packet. **Figure 10-7** shows how the sixteen bit CRC value is calculated and either transmitted or verified for the data packet of a given transaction.

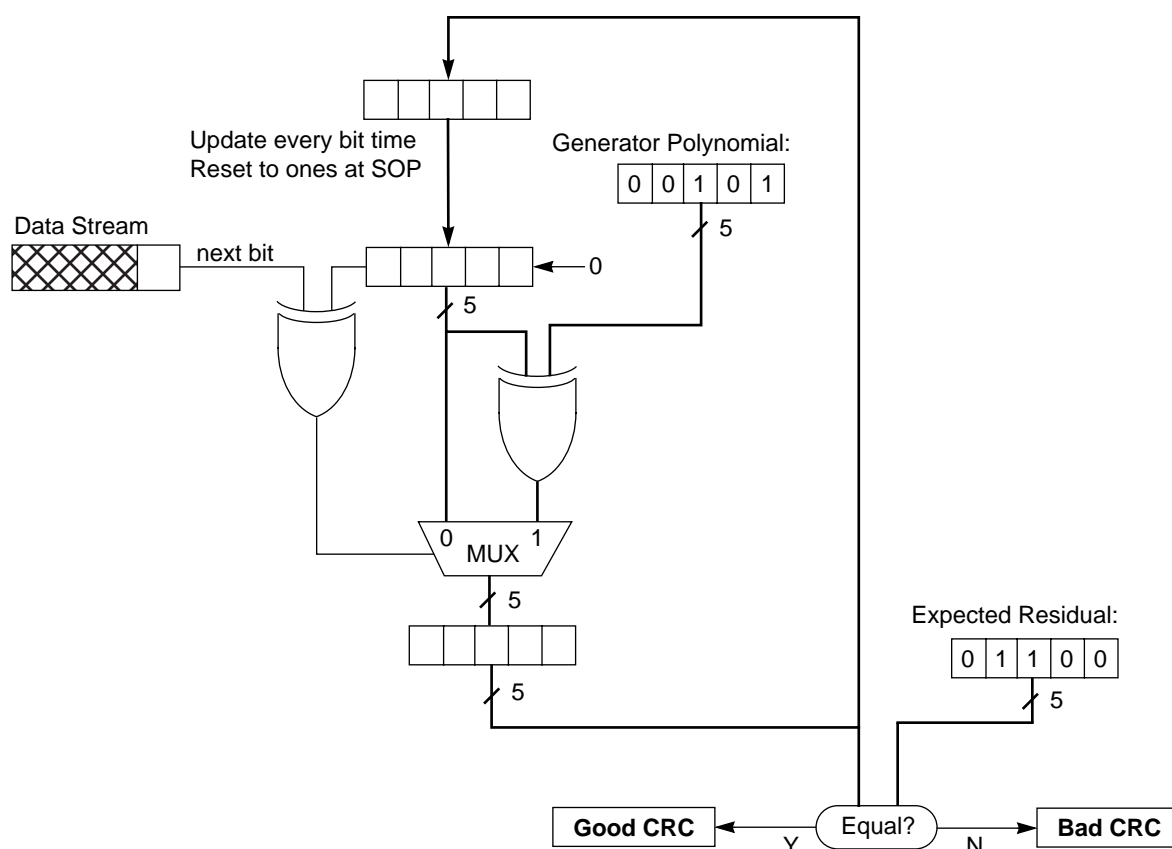


Figure 10-6. CRC Block Diagram for Address and Endpoint Fields

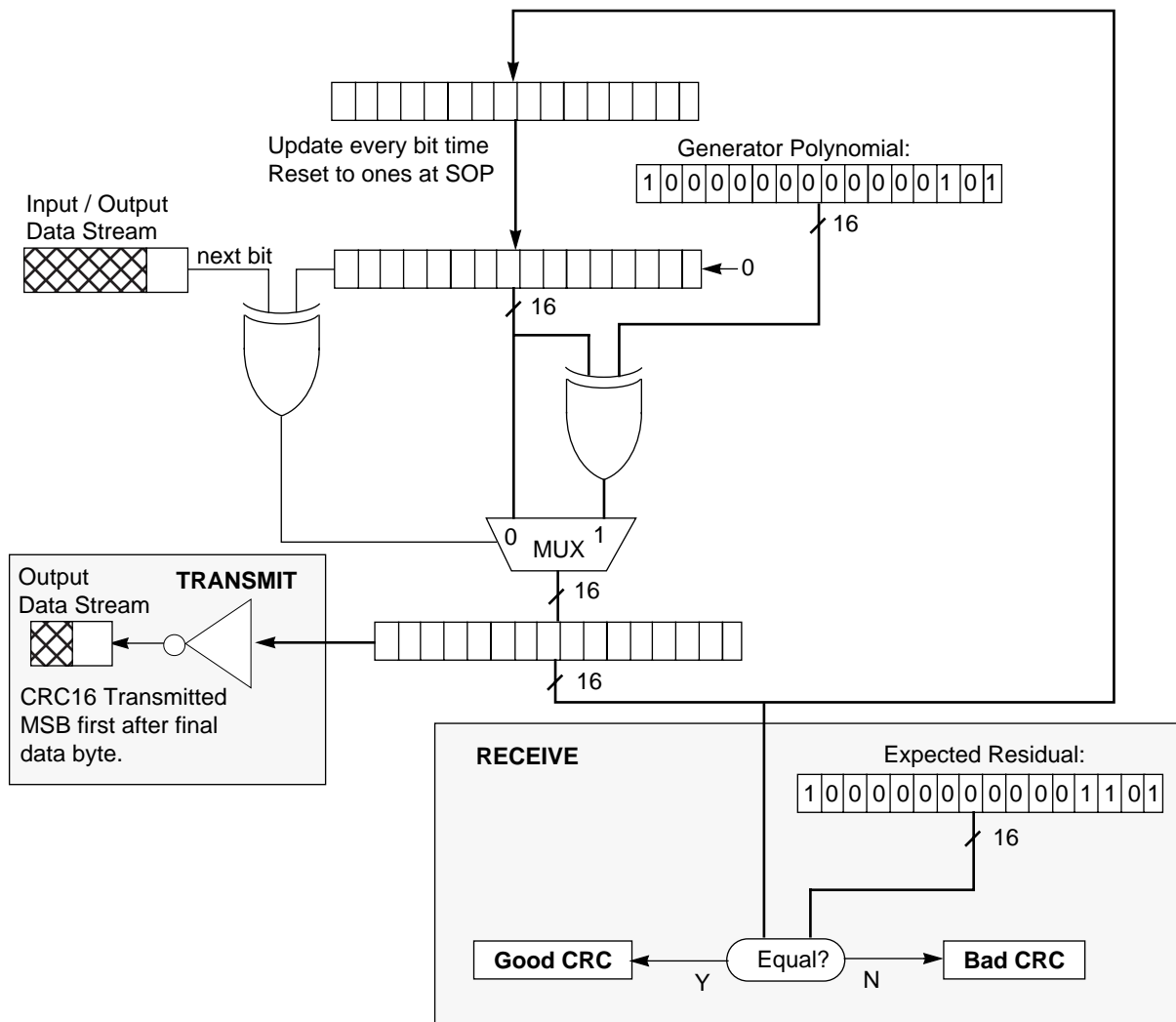


Figure 10-7. CRC Block Diagram for Data Packets

10.2.1.6 End Of Packet (EOP)

The single-ended 0 (SE0) state is used to signal an end of packet (EOP). The single-ended 0 state is indicated by both D+ and D- being below 0.8 V. EOP will be signaled by driving D+ and D- to the single-ended 0 state for two bit times followed by driving the lines to the idle state for one bit time. The transition from the single-ended 0 to the idle state defines the end of the packet. The idle state is asserted for one bit time and then both the D+ and D- output drivers are placed in their high-impedance state. The bus termination resistors hold the bus in the idle state. **Figure 10-8** shows the data signaling and voltage levels for an end of packet transaction.

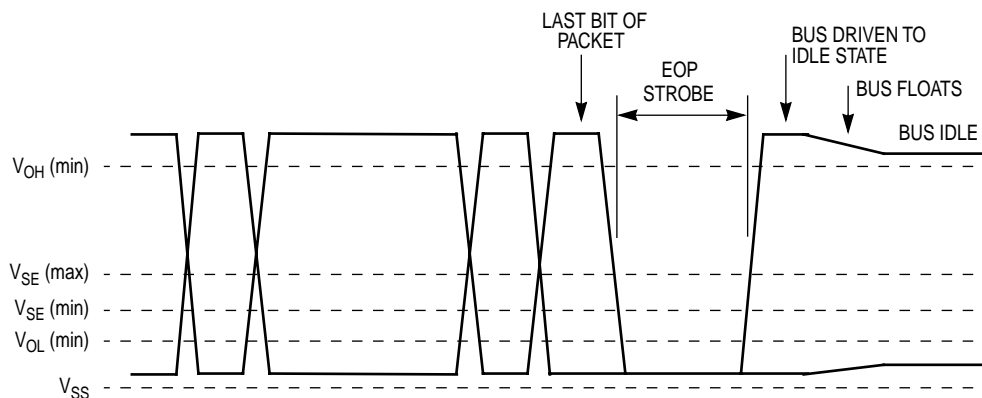


Figure 10-8. EOP Transaction Voltage Levels

The width of the SE0 in the EOP is about two bit times. The EOP width is measured with the same capacitive load used for maximum rise and fall times and is measured at the same level as the differential signal crossover points of the data lines.

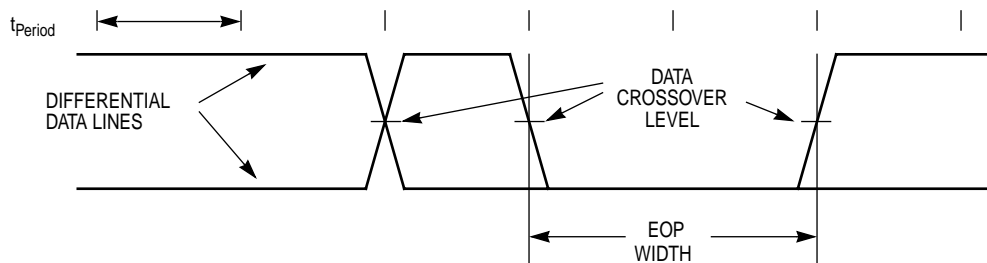


Figure 10-9. EOP Width Timing

10.2.2 Reset Signaling

A reset is signaled on the bus by the presence of an extended SE0 at the USB data pins of a device. The reset signaling is specified to be present for a minimum of 10 ms. An active device (powered and not in the suspend state) seeing a single-ended zero on its USB data inputs for more than 2.5 μ s may treat that signal as a reset, but must have interpreted the signaling as a reset within 5.5 μ s. For a Low speed device, an SE0 condition between 4 and 8 low speed bit times represents a valid USB reset.

A USB sourced reset will hold the MC68HC705JB2 in reset for the duration of the reset on the USB bus. The RSTF bit in the USB interrupt register 0 (UIR0) will be set after the internal reset is removed (See **Section 10.5.2** for more detail).

After a reset is removed, the device will be in the attached, but not yet addressed or configured state (refer to Section 9.1 of the USB specification). The device must be able to accept a device address via a SET_ADDRESS command (refer to section 9.4 of the USB specification) no later than 10 ms after the reset is removed.

Reset can wake a device from the suspended mode. A device may take up to 10ms to wake up from the suspended state.

10.2.3 Suspend

The MC68HC705JB2 supports suspend mode for low power. Suspend mode should be entered when the USB data lines are in the idle state for more than 3.0 ms. Entry into Suspend mode is controlled by the SUSPND bit in the USB Interrupt Register. Any low speed bus activity should keep the device out of the suspend state. Low speed devices are kept awake by periodic low speed EOP signals from the host. This is referred to as Low speed keep alive (refer to Section 11.2.5.1 of the USB specification).

Firmware should monitor the EOPF flag and enter suspend mode by setting the SUSPND bit if an EOP is not detected for 3 ms.

Per the USB specification, the MC68HC705JB2 is required to draw less than 500 μ A from the V_{DD} supply when in the suspend state. This includes the current supplied by the voltage regulator to the 15 K Ω to ground termination resistors placed at the host end of the USB bus. This low current requirement means that firmware is responsible for entering STOP mode once the USB module has been placed in the suspend state.

10.2.4 Resume After Suspend

The MC68HC705JB2 can be activated from the suspend state by normal bus activity, a USB reset signal, or by a forced resume driven from the MC68HC705JB2.

10.2.4.1 Host Initiated Resume

The host signals resume by initiating resume signalling ("K" state) for at least 20 ms followed by a standard low speed EOP signal. This 20 ms ensures that all devices in the USB network are awakened.

After resuming the bus, the host must begin sending bus traffic within 3 ms to prevent the device from re-entering suspend mode.

10.2.4.2 USB Reset Signalling

Reset can wake a device from the suspended mode. A device may take up to 10 ms to wake up from the suspended state.

10.2.4.3 Remote Wake-up

The MC68HC705JB2 also supports the remote wake-up feature. The firmware has the ability to exit suspend mode by signaling a resume state to the upstream Host or Hub. A non-idle state ("K" state) on the USB data lines is accomplished by asserting the FRESUM bit in the UCR1 register.

When using the remote wake-up capability, the firmware must wait for at least 5 ms after the bus is in the idle state before sending the remote wake-up resume signaling. This allows the upstream devices to get into their suspend state and prepare for propagating resume signaling. The FRESUM bit should be asserted to cause the resume state on the USB data lines for at least 10ms, but not more than 15ms. Note that the resume signalling is controlled by the FRESUM bit and meeting the timing specifications is dependent on the firmware. When FRESUM is cleared by firmware, the data lines will return to their high impedance state. Refer to **Section 10.5.5** for more information about how the Force Resume (FRESUM) bit can be used to initiate the remote wake-up feature.

10.2.5 Low Speed Device

Externally, low speed devices are configured by the position of a pull-up resistor on the USB D- pin of the MC68HC705JB2. Low speed devices are terminated as shown in **Figure 10-10** with the pull-up on the D- line.

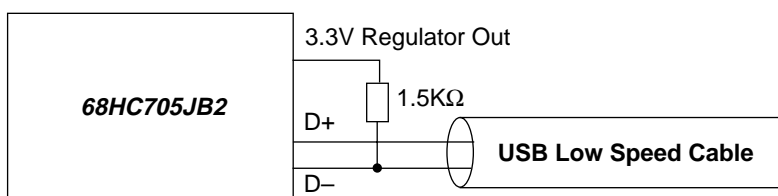


Figure 10-10. External Low Speed Device Configuration

For low speed transmissions, the transmitter's EOP width must be between 1.25μs and 1.50μs. These ranges include timing variations due to differential buffer delay and rise/fall time mismatches and to noise and other random effects. A low speed receiver must accept a 670 ns wide SE0 followed by a J transition as a valid EOP. An SE0 narrower than 330 ns or an SE0 not followed by a J transition must be rejected as an EOP. An EOP between 330ns and 670ns may be rejected or accepted as above. Any SE0 that is 2.5μs or wider is automatically a reset.

10.3 CLOCK REQUIREMENTS

The low speed data rate is nominally 1.5 Mbs. The OSCXCLK signal driven by the oscillator circuits is the clock source for the USB module and requires that a 6 MHz oscillator circuit be connected to the OSC1 and OSC2 pins. The permitted frequency tolerance for low speed functions is approximately $\pm 1.5\%$ (15000 ppm). This tolerance includes inaccuracies from all sources: initial frequency accuracy, crystal capacitive loading, supply voltage on the oscillator, temperature, and aging. The jitter in the low speed data rate must be less than 10 ns. This tolerance allows the use of resonators in low cost, low speed devices.

10.4 HARDWARE DESCRIPTION

The USB module as previously shown in **Figure 10-1** contains four functional blocks: a 3.3 volt Regulator, a LS USB transceiver, the USB control logic, and the USB registers. The following will detail the function of the regulator, transceiver and control logic. See **Section 10.5** for the register discussion.

10.4.1 Voltage Regulator

The USB data lines are required by the USB Specification to have a maximum output voltage between 2.8V and 3.6V. The data lines are also required to have an external $1.5K\Omega$ pullup resistor connected between a data line and a voltage source between 3.0V and 3.6V. Since the power provided by the USB cable is specified to be between 4.4V and 5.0V, an on-chip regulator is used to drop the voltage to the appropriate level for sourcing the USB transceiver and external pullup resistor. An output pin driven by the regulator voltage is provided to source the $1.5K\Omega$ external resistor. **Figure 10-11** shows the worst case electrical connection for the voltage regulator.

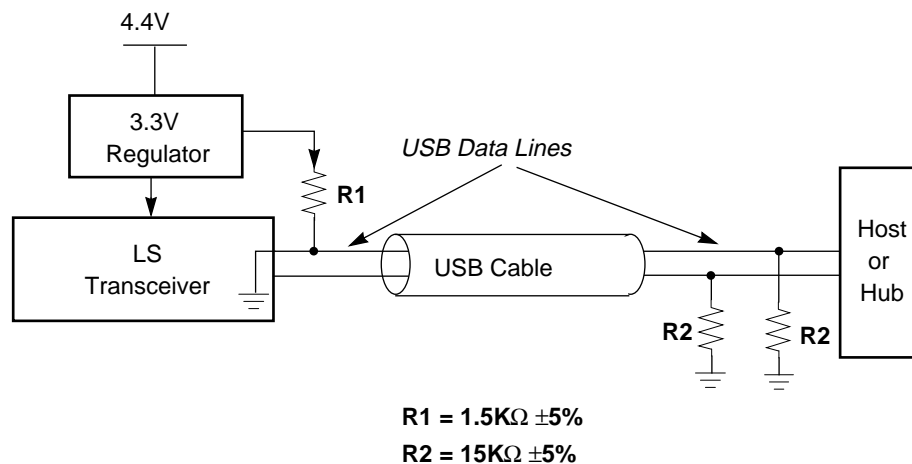


Figure 10-11. Regulator Electrical Connections

10.4.2 USB Transceiver

The USB transceiver provides the physical interface to the USB D+ and D- data lines. The transceiver is composed of two parts: an output drive circuit and a differential receiver.

10.4.2.1 Output Driver Characteristics

The USB transceiver uses a differential output driver to drive the USB data signal onto the USB cable. The static output swing of the driver in its low state is below the V_{OL} of 0.3 V with a 1.5 k Ω load to 3.6 V and in its high state is above the V_{OH} of 2.8 V with a 15 k Ω load to ground. The output swings between the differential high and low state are well balanced to minimize signal skew. Slew rate control on the driver is used to minimize the radiated noise and cross talk. The driver's outputs support three-state operation to achieve bi-directional half duplex operation. The driver can tolerate a voltage on the signal pins of -0.5 V to 3.8 V with respect to local ground reference without damage.

10.4.2.2 Low Speed (1.5 Mbs) Driver Characteristics

The rise and fall time of the signals on this cable are greater than 75 ns to keep RFI emissions under FCC class B limits, and less than 300 ns to limit timing delays and signaling skews and distortions. The driver reaches the specified static signal levels with smooth rise and fall times, and minimal reflections and ringing when driving the cable. This driver is used only on network segments between low speed devices and the ports to which they are connected.

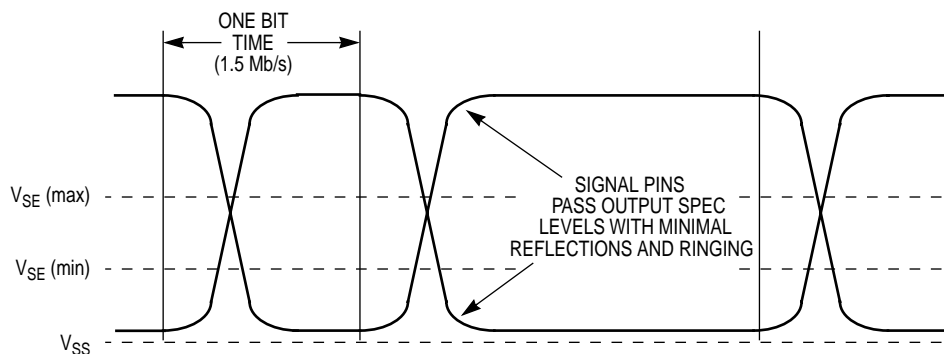


Figure 10-12. Low Speed Driver Signal Waveforms

10.4.3 Receiver Characteristics

USB data transmission is done with differential signals. A differential input receiver is used to accept the USB data signal. A differential 1 on the bus is represented by D+ being at least 200 mV more positive than D- as seen at the receiver, and a differential 0 is represented by D- being at least 200 mV more positive than D+ as seen at the receiver. The signal cross over point must be between 1.3V and 2.0V.

The receiver features an input sensitivity of 200 mV when both differential data inputs are in the range of 0.8 V to 2.5 V with respect to the local ground reference. This is called the common mode input voltage range. Proper data reception is also achieved when the differential data lines are outside the common mode range, as shown in **Figure 10-13**. The receiver can tolerate static input voltages between -0.5V to 3.8 V with respect to its local ground reference without damage. In addition to the differential receiver, there is a single-ended receiver (schmitt trigger) for each of the two data lines.

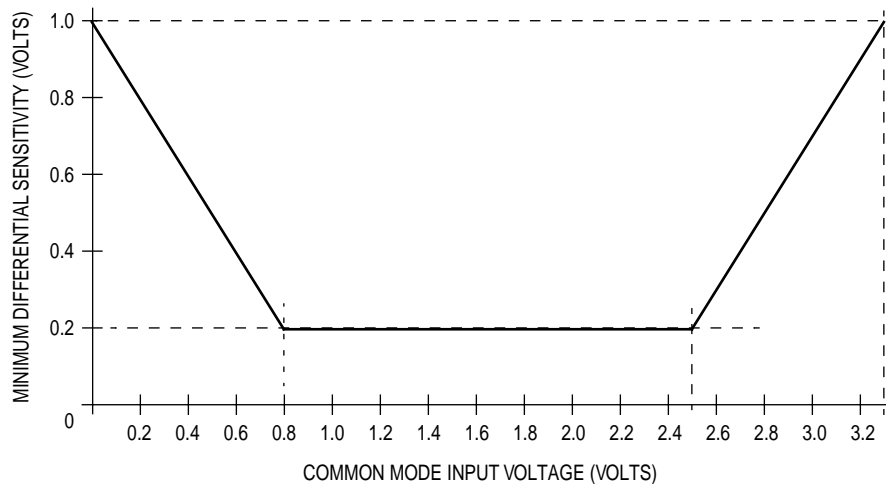


Figure 10-13. Differential Input Sensitivity Over Entire Common Mode Range

10.4.3.1 Receiver Data Jitter

The data receivers for all types of devices must be able to properly decode the differential data in the presence of jitter. The more of the bit cell that any data edge can occupy and still be decoded, the more reliable the data transfer will be. Data receivers are required to decode differential data transitions that occur in a window plus and minus a nominal quarter bit cell from the nominal (centered) data edge position.

Jitter will be caused by the delay mismatches and by mismatches in the source and destination data rates (frequencies). The receive data jitter budget for low speed is given in the electrical section of the this specification. The specification includes the consecutive (next) and paired transition values for each source of jitter.

10.4.3.2 Data Source Jitter

The source of data can have some variation (jitter) in the timing of edges of the data transmitted. The time between any set of data transitions is $N * T_{PERIOD} \pm$ jitter time, where 'N' is the number of bits between the transitions and T_{PERIOD} is defined as the actual period of the data rate. The data jitter is measured with the same capacitive load used for maximum rise and fall times and is measured at the crossover points of the data lines as shown in **Figure 10-14**.

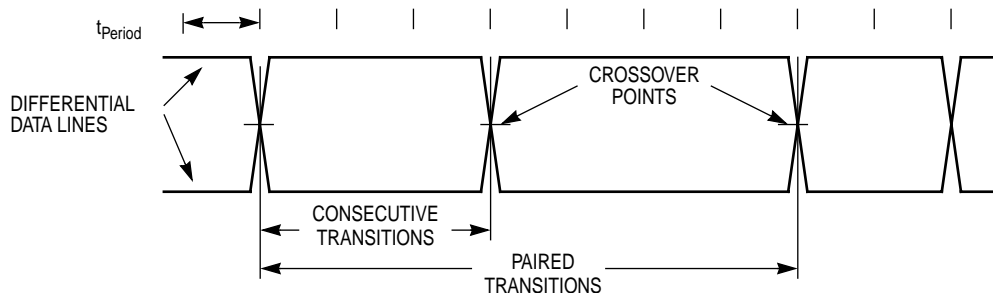


Figure 10-14. Data Jitter

For low speed transmissions, the jitter time for any consecutive differential data transitions must be within ± 25 ns and within ± 10 ns for any set of paired differential data transitions. These jitter numbers include timing variations due to differential buffer delay, rise/fall time mismatches, internal clock source jitter, and to noise and other random effects.

10.4.3.3 Data Signal Rise and Fall Time

The output rise time and fall time are measured between 10% and 90% of the signal. Edge transition time for the rising and falling edges of low speed signals is 75 ns (minimum) into a capacitive load (C_L) of 50 pF and 300 ns (maximum) into a capacitive load of 350 pF. The rising and falling edges should be smooth transitional (monotonic) when driving the cable to avoid excessive EMI.

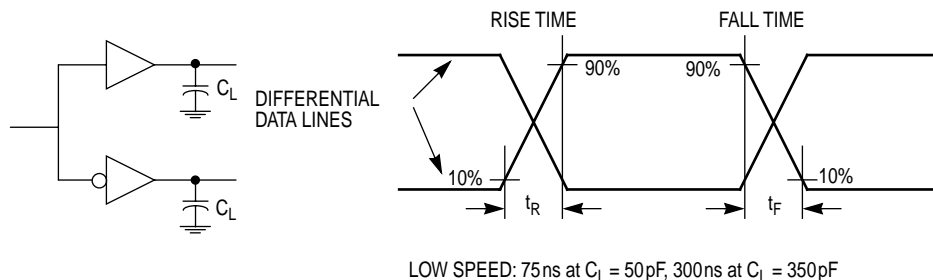


Figure 10-15. Data Signal Rise and Fall Time

10.4.4 USB Control Logic

The USB control logic manages data movement between the CPU and the transceiver. The control logic handles both transmit and receive operations on the USB. It contains the logic used to manipulate the transceiver and the endpoint registers. The logic contains byte count buffers for transmit operations that load the active transmit endpoints byte count and use this to determine the number of bytes to transfer. This same buffer is used for receive transactions to count the number of bytes received and, upon the end of the transaction, transfer that number to the receive endpoints byte count register.

When transmitting, the control logic handles parallel to serial conversion, CRC generation, NRZI encoding, and bit stuffing.

When Receiving, the control logic handles Sync detection, packet identification, end of packet detection, bit (un)stuffing, NRZI decoding, CRC validation, and serial to parallel conversion. Errors detected by the control logic include bad CRC, time-out while waiting for EOP, and bit stuffing violations.

10.4.4.1 Data Encoding/Decoding

The USB employs NRZI data encoding when transmitting packets. In NRZI encoding, a 1 is represented by no change in level and a 0 is represented by a change in level. **Figure 10-16** shows a data stream and the NRZI equivalent and **Figure 10-17** is a flow diagram for NRZI. The high level represents the J state on the data lines in this and subsequent figures showing NRZI encoding. A string of zeros causes the NRZI data to toggle each bit time. A string of ones causes long periods with no transitions in the data.

10.4.4.2 Bit Stuffing

In order to ensure adequate signal transitions, bit stuffing is employed by the transmitting device when sending a packet on the USB (see **Figure 10-18** and **Figure 10-19**). A 0 is inserted after every six consecutive 1's in the data stream before the data is NRZI encoded to force a transition in the NRZI data stream. This gives the receiver logic a data transition at least once every seven bit times to guarantee the data and clock lock. The receiver must decode the NRZI data, recognize the stuffed bits, and discard them. Bit stuffing is enabled beginning with the Sync Pattern and throughout the entire transmission. The data "one" that ends the Sync Pattern is counted as the first one in a sequence. Bit stuffing is always enforced, without exception. If required by the bit stuffing rules, a zero bit will be inserted even if it is the last bit before the end-of-packet (EOP) signal.

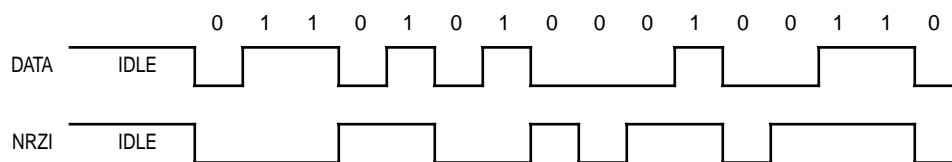


Figure 10-16. NRZI Data Encoding

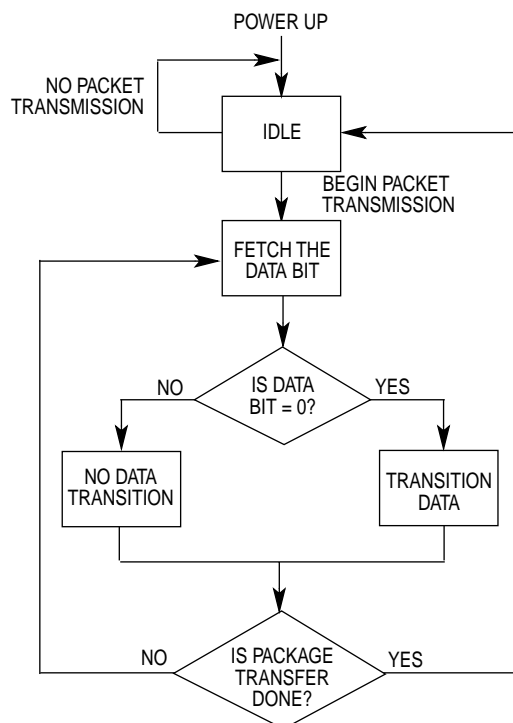


Figure 10-17. Flow Diagram for NRZI

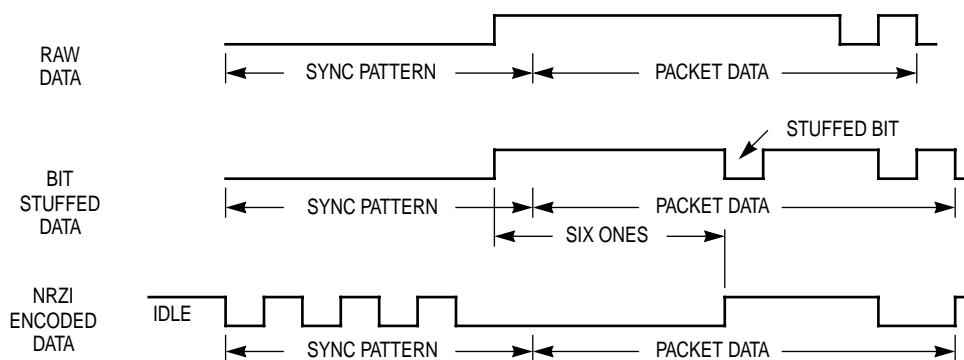
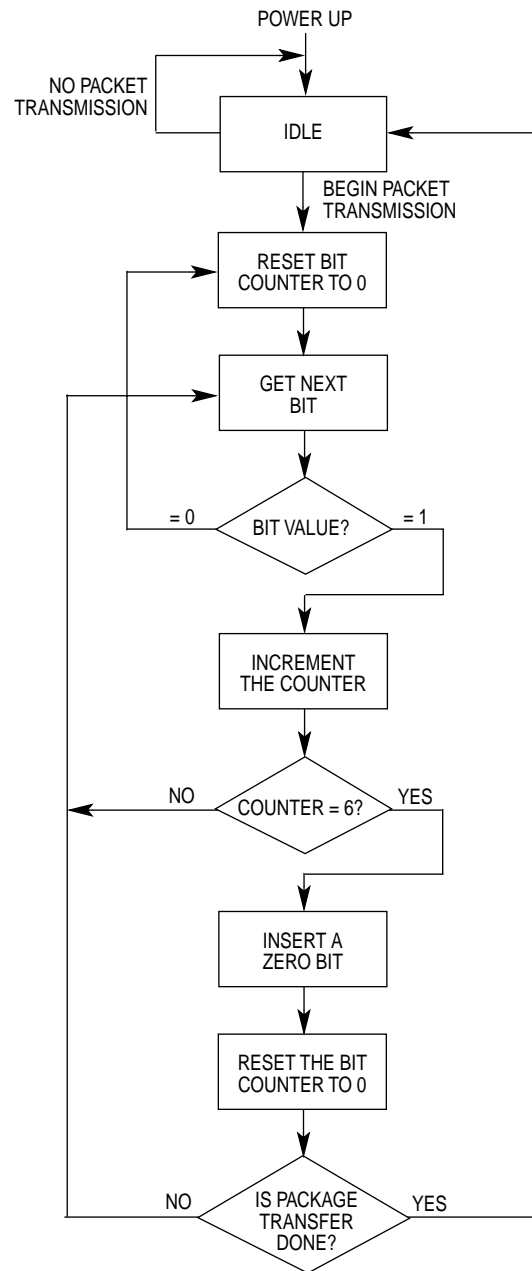


Figure 10-18. Bit Stuffing


**Figure 10-19. Flow Diagram for Bit Stuffing**

10.5 I/O REGISTER DESCRIPTION

The USB Endpoint registers are comprised of a set of control/status registers and twenty-four data registers that provide storage for the buffering of data between the USB and the CPU. These registers are shown in **Table 10-2**.

Table 10-2. Register Summary

Register Name	Bit 7	6	5	4	3	2	1	Bit 0	Addr
USB Control Register 2 (UCR2)		0	TX1ST	0	ENABLE2	ENABLE1	STALL2	STALL1	\$0037
		TX1STR							
USB Address Register (UADDR)	USBEN	UADD6	UADD5	UADD4	UADD3	UADD2	UADD1	UADD0	\$0038
USB Interrupt Register 0 (UIR0)	TXD0F	RXD0F	RSTF	SUSPND	TXD0IE	RXD0IE	0	0	\$0039
							TXD0FR	RXD0FR	
USB Interrupt Register 1 (UIR1)	TXD1F	EOPF	RESUMF	0	TXD1IE	EOPIE	0	0	\$003A
				RESUMFR			TXD1FR	EOPFR	
USB Control Register 0 (UCR0)	T0SEQ	STALL0	TX0E	RX0E	TP0SIZ3	TP0SIZ2	TP0SIZ1	TP0SIZ0	\$003B
USB Control Register 1 (UCR1)	T1SEQ	ENDADD	TX1E	FRESUM	TP1SIZ3	TP1SIZ2	TP1SIZ1	TP1SIZ0	\$003C
USB Status Register (USR)	RSEQ	SETUP	0	0	RPSIZ3	RPSIZ2	RPSIZ1	RPSIZ0	\$003D
USB Endpoint 0 Data Register 0 (UE0D0)	UE0RD7	UE0RD6	UE0RD5	UE0RD4	UE0RD3	UE0RD2	UE0RD1	UE0RD0	\$0020
	UE0TD7	UE0TD6	UE0TD5	UE0TD4	UE0TD3	UE0TD2	UE0TD1	UE0TD0	
↓									
USB Endpoint 0 Data Register 7 (UE0D7)	UE0RD7	UE0RD6	UE0RD5	UE0RD4	UE0RD3	UE0RD2	UE0RD1	UE0RD0	\$0027
	UE0TD7	UE0TD6	UE0TD5	UE0TD4	UE0TD3	UE0TD2	UE0TD1	UE0TD0	
USB Endpoint 1/2 Data Register 0 (UE1D0)									\$0028
	UE1TD7	UE1TD6	UE1TD5	UE1TD4	UE1TD3	UE1TD2	UE1TD1	UE1TD0	
↓									
USB Endpoint 1/2 Data Register 7 (UE1D7)									\$002F
	UE1TD7	UE1TD6	UE1TD5	UE1TD4	UE1TD3	UE1TD2	UE1TD1	UE1TD0	

 = Unimplemented

10.5.1 USB Address Register (UADDR)

		Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
UADDR	R	USBEN	UADD6	UADD5	UADD4	UADD3	UADD2	UADD1	UADD0
\$0038	W								
reset ⇒		0	0	0	0	0	0	0	0

Figure 10-20. USB Address Register (UADDR)

USBEN — USB Module Enable

This read/write bit enables and disables the USB module and the USB pins. When USBEN is clear, the USB module will not respond to any tokens. Reset clears this bit.

1 = USB function enabled

0 = USB function disabled

UADD6-UADD0 — USB Function Address

These bits specify the USB address of the device. Reset clears these bits.

10.5.2 USB Interrupt Register 0 (UIR0)

		Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
UIR0	R	TXD0F	RXD0F	RSTF	SUSPND	TXD0IE	RXD0IE	0	0
\$0039	W							TXD0FR	RXD0FR
reset ⇒		0	0	0	0	0	0	0	0

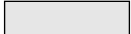
 = Unimplemented

Figure 10-21. USB Interrupt Register 0 (UIR0)

TXD0F — Endpoint 0 Data Transmit Flag

This read only bit is set after the data stored in Endpoint 0 transmit buffers has been sent and an ACK handshake packet from the host is received. Once the next set of data is ready in the transmit buffers, software must clear this flag by writing a logic 1 to the TXD0FR bit. To enable the next data packet transmission, TX0E must also be set. If TXD0F bit is not cleared, a NAK handshake will be returned in the next IN transaction.

Reset clears this bit. Writing a logic 0 to TXD0F has no effect.

1 = Transmit on Endpoint 0 has occurred

0 = Transmit on Endpoint 0 has not occurred

RXD0F — Endpoint 0 Data Receive Flag

This read only bit is set after the USB module has received a data packet and responded with an ACK handshake packet. Software must clear this flag by writing a logic 1 to the RXD0FR bit after all of the received data has been read. Software must also set RX0E bit to one to enable the next data packet reception. If RXD0F bit is not cleared, a NAK handshake will be returned in the next OUT transaction.

Reset clears this bit. Writing a logic 0 to RXD0F has no effect.

- 1 = Receive on Endpoint 0 has occurred
 0 = Receive on Endpoint 0 has not occurred

RSTF — USB Reset Flag

This read only bit is set when a valid reset signal state is detected on the D+ and D- lines. This reset detection will also generate an internal reset signal to reset the CPU and other peripherals including the USB module. This bit is cleared by writing a logic 1 to the RSTFR bit in the UCR2 register. This bit is cleared by a POR reset.

SUSPND — USB Suspend Flag

To save power, this read/write bit should be set by the software if a 3ms constant idle state is detected on USB bus. Setting this bit stops the clock to the USB and causes the USB module to enter Suspend mode. Unnecessary analog circuitry will be powered down. Software must clear this bit after the Resume flag (RESUMF) is set while this Resume interrupt flag is serviced.

TXD0IE — Endpoint 0 Transmit Interrupt Enable

This read/write bit enables the Transmit Endpoint 0 to generate a USB interrupt when the TXD0F bit becomes set.

- 1 = USB interrupts enabled for Transmit Endpoint 0
 0 = USB interrupts disabled for Transmit Endpoint 0

RXD0IE — Endpoint 0 Receive Interrupt Enable

This read/write bit enables the Transmit Endpoint 0 to generate a USB interrupt when the RXD0F bit becomes set.

- 1 = USB interrupts enabled for Receive Endpoint 0
 0 = USB interrupts disabled for Receive Endpoint 0

TXD0FR — Endpoint 0 Transmit Flag Reset

Writing a logic 1 to this write only bit will clear the TXD0F bit if it is set. Writing a logic 0 to TXD0FR has no effect. Reset clears this bit.

RXD0FR — Endpoint 0 Receive Flag Reset

Writing a logic 1 to this write only bit will clear the RXD0F bit if it is set. Writing a logic 0 to RXD0FR has no effect. Reset clears this bit.

10.5.3 USB Interrupt Register 1 (UIR1)

		Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
UIR1	R	TXD1F	EOPF	RESUMF	0	TXD1IE	EOPIE	0	0
\$003A	W				RESUMFR			TXD1FR	EOPFR
reset ⇒		0	0	0	0	0	0	0	0


 = Unimplemented

Figure 10-22. USB Interrupt Register 1 (UIR1)

TXD1F — Endpoint 1/Endpoint 2 Data Transmit Flag

This read only bit is shared by Endpoint 1 and Endpoint 2. It is set after the data stored in the shared Endpoint 1/Endpoint 2 transmit buffer has been sent and an ACK handshake packet from the host is received. Once the next set of data is ready in the transmit buffers, software must clear this flag by writing a logic 1 to the TXD1FR bit. To enable the next data packet transmission, TX1E must also be set. If TXD1F bit is not cleared, a NAK handshake will be returned in the next IN transaction.

Reset clears this bit. Writing a logic 0 to TXD1F has no effect.

- 1 = Transmit on Endpoint 1 or Endpoint 2 has occurred
- 0 = Transmit on Endpoint 1 or Endpoint 2 has not occurred

EOPF — End of Packet Detect Flag

This read only bit is set when a valid End-of-Packet sequence is detected on the D+ and D- lines. Software must clear this flag by writing a logic 1 to the EOPFR bit.

Reset clears this bit. Writing a logic 0 to EOPF has no effect.

- 1 = End-of-Packet sequence has been detected
- 0 = End-of-Packet sequence has not been detected

RESUMF — Resume Flag

This read only bit is set when USB bus activity is detected while the SUSPND bit is set. Software must clear this flag by writing a logic 1 to the RESUMFR bit.

Reset clears this bit. Writing a logic 0 to RESUMF has no effect.

- 1 = USB bus activity has been detected
- 0 = No USB bus activity has been detected

RESUMFR — Resume Flag Reset

Writing a logic 1 to this write only bit will clear the RESUMF bit if it is set. Writing a logic 0 to RESUMFR has no effect. Reset clears this bit.

TXD1IE — Endpoint 1/Endpoint 2 Transmit Interrupt Enable

This read/write bit enables the USB to generate an interrupt when the shared Transmit Endpoint 1/Endpoint 2 interrupt flag (TXD1F) bit becomes set. Reset clears this bit.

- 1 = USB interrupts enabled for Transmit Endpoints 1 and 2
- 0 = USB interrupts disabled for Transmit Endpoints 1 and 2

EOPIE — End of Packet Detect Interrupt Enable

This read/write bit enables the USB to generate an interrupt when the EOPF bit becomes set. Reset clears this bit.

- 1 = USB interrupts enabled for Transmit Endpoints 1 and 2
- 0 = USB interrupts disabled for Transmit Endpoint 1 and 2

TXD1FR — Endpoint 1/Endpoint 2 Transmit Flag Reset

Writing a logic 1 to this write only bit will clear the TXD1F bit if it is set. Writing a logic 0 to TXD1FR has no effect. Reset clears this bit.

EOPFR — End of Packet Flag Reset

Writing a logic 1 to this write only bit will clear the EOPF bit if it is set. Writing a logic 0 to the EOPFR has no effect. Reset clears this bit.

10.5.4 USB Control Register 0 (UCR0)

		Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
UCR0	R	T0SEQ	STALL0	TX0E	RX0E	TP0SIZ3	TP0SIZ2	TP0SIZ1	TP0SIZ0
\$003B	W								
reset ⇒		0	0	0	0	0	0	0	0

Figure 10-23. USB Control Register 0 (UCR0)**T0SEQ — Endpoint 0 Transmit Sequence Bit**

This read/write bit determines which type of data packet (DATA0 or DATA1) will be sent during the next IN transaction. Toggling of this bit must be controlled by software. Reset clears this bit.

- 1 = DATA1 Token active for next Endpoint 0 transmit
- 0 = DATA0 Token active for next Endpoint 0 transmit

STALL0 — Endpoint 0 Force Stall Bit

This read/write bit causes Endpoint 0 to return a STALL handshake when polled by either an IN or OUT token by the USB Host Controller. The USB hardware clears this bit when a SETUP token is received. Reset clears this bit.

- 1 = Send STALL handshake
- 0 = Default

TX0E — Endpoint 0 Transmit Enable

This read/write bit enables a transmit to occur when the USB Host controller sends an IN token to Endpoint 0. Software should set this bit when data is ready to be transmitted. It must be cleared by software when no more Endpoint 0 data needs to be transmitted.

If this bit is 0 or the TXD0F is set, the USB will respond with a NAK handshake to any Endpoint 0 IN tokens. Reset clears this bit.

- 1 = Data is ready to be sent.
- 0 = Data is not ready. Respond with NAK.

RX0E — Endpoint 0 Receive Enable

This read/write bit enables a receive to occur when the USB Host controller sends an OUT token to Endpoint 0. Software should set this bit when data is ready to be received. It must be cleared by software when data cannot be received.

If this bit is 0 or the RXD0F is set, the USB will respond with a NAK handshake to any Endpoint 0 OUT tokens. Reset clears this bit.

- 1 = Data is ready to be received.
- 0 = Not ready for data. Respond with NAK.

TP0SIZ3-TP0SIZ0 — Endpoint 0 Transmit Data Packet Size

These read/write bits store the number of transmit data bytes for the next IN token request for Endpoint 0. These bits are cleared by reset.

10.5.5 USB Control Register 1 (UCR1)

		Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
UCR1 \$003C	R	T1SEQ	ENDADD	TX1E	FRESUM	TP1SZ3	TP1SZ2	TP1SZ1	TP1SZ0
	W								
reset ⇒		0	0	0	0	0	0	0	0

Figure 10-24. USB Control Register 1 (UCR1)**T1SEQ — Endpoint1/Endpoint 2 Transmit Sequence Bit**

This read/write bit determines which type of data packet (DATA0 or DATA1) will be sent during the next IN transaction directed to Endpoint 1 or Endpoint 2. Toggling of this bit must be controlled by software. Reset clears this bit.

- 1 = DATA1 Token active for next Endpoint 1/Endpoint 2 transmit
- 0 = DATA0 Token active for next Endpoint 1/Endpoint 2 transmit

ENDADD — Endpoint Address Select

This read/write bit specifies whether the data inside the registers UE1D0-UE1D7 are used for Endpoint 1 or Endpoint 2. If all the conditions for a successful Endpoint 2 USB response to a hosts IN token are satisfied (TXD1F=0, TX1E=1, STALL2=0, and ENABLE2=1) except that the ENDADD bit is configured for Endpoint 1, the USB responds with a NAK handshake packet.

- 1 = The data buffers are used for Endpoint 2
- 0 = The data buffers are used for Endpoint 1

TX1E — Endpoint 1/Endpoint 2 Transmit Enable

This read/write bit enables a transmit to occur when the USB Host controller sends an IN token to Endpoint 1 or Endpoint 2. The appropriate endpoint enable bit, ENABLE1 or ENABLE2 bit in the UCR2 register, should also be set. Software should set the TX1E bit when data is ready to be transmitted. It must be cleared by software when no more data needs to be transmitted.

If this bit is 0 or the TXD1F is set, the USB will respond with a NAK handshake to any Endpoint 1 or Endpoint 2 directed IN tokens. Reset clears this bit.

- 1 = Data is ready to be sent.
- 0 = Data is not ready. Respond with NAK.

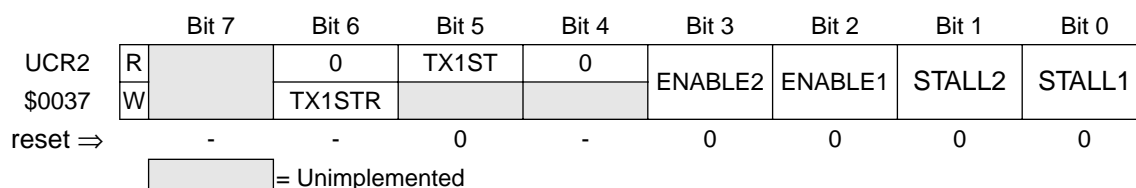
FRESUM — Force Resume

This read/write bit forces a resume state ("K" or non-idle state) onto the USB data lines to initiate a remote wake-up. Software should control the timing of the forced resume to be between 10ms and 15 ms. Setting this bit will not cause the RESUMF bit to set.

- 1 = Force data lines to "K" state
- 0 = Default

TP1SIZ3-TP1SIZ0 — Endpoint 1/Endpoint 2 Transmit Data Packet Size

These read/write bits store the number of transmit data bytes for the next IN token request for Endpoint 1 or Endpoint 2. These bits are cleared by reset.

10.5.6 USB Control Register 2 (UCR2)**Figure 10-25. USB Control Register 2 (UCR2)****TX1STR — Clear Transmit First Flag**

Writing a logic 1 to this write-only bit will clear the TX1ST bit if it is set. Writing a logic 0 to the TX1STR has no effect. Reset clears this bit.

TX1ST — Transmit First Flag

This read-only bit is set if the Endpoint 0 Data Transmit Flag (TXD0F) is set when the USB control logic is setting the Endpoint 0 Data Receive Flag (RXD0F). That is, this bit will be set if an Endpoint 0 Transmit Flag is still set at the end of an Endpoint 0 reception. This bit lets the firmware know that the Endpoint 0 transmission happened before the Endpoint 0 reception. Reset clears this bit.

- 1 = IN transaction occurred before SETUP/OUT.
- 0 = IN transaction occurred after SETUP/OUT.

ENABLE2 — Endpoint 2 Enable

This read/write bit enables Endpoint 2 and allows the USB to respond to IN packets addressed to Endpoint 2. Reset clears this bit.

- 1 = Endpoint 2 is enabled and can respond to an IN token.
- 0 = Endpoint 2 is disabled

ENABLE1 — Endpoint 1 Enable

This read/write bit enables Endpoint 1 and allows the USB to respond to IN packets addressed to Endpoint 1. Reset clears this bit.

- 1 = Endpoint 1 is enabled and can respond to an IN token.
- 0 = Endpoint 1 is disabled

STALL2 — Endpoint 2 Force Stall Bit

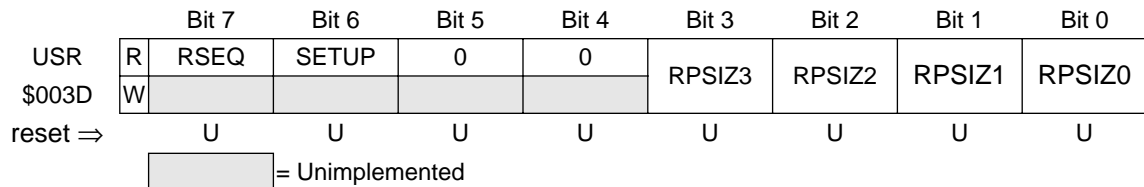
This read/write bit causes Endpoint 2 to return a STALL handshake when polled by either an IN or OUT token by the USB Host Controller. Reset clears this bit.

- 1 = Send STALL handshake.
- 0 = Default

STALL1 — Endpoint 1 Force Stall Bit

This read/write bit causes Endpoint 1 to return a STALL handshake when polled by either an IN or OUT token by the USB Host Controller. Reset clears this bit.

- 1 = Send STALL handshake
- 0 = Default

10.5.7 USB Status Register (USR)**Figure 10-26. USB Status Register (USR)****RSEQ — Endpoint 0 Receive Sequence Bit**

This read only bit indicates the type of data packet last received for Endpoint 0 (DATA0 or DATA1).

- 1 = DATA1 Token received in last Endpoint 0 receive
- 0 = DATA0 Token received in last Endpoint 0 receive

SETUP — SETUP Token Detect Bit

This read only bit indicates that a valid SETUP token has been received.

- 1 = Last token received for Endpoint 0 was a SETUP token
- 0 = Last token received for Endpoint 0 was not a SETUP token

RPSIZ3-RPSIZ0 — Endpoint 0 Receive Data Packet Size

These read only bits store the number of data bytes received for the last OUT or SETUP transaction for Endpoint 0. These bits are not affected by reset.

10.5.8 USB Endpoint 0 Data Registers (UE0D0-UE0D7)

		Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
UE0D0	R	UE0RD7	UE0RD6	UE0RD5	UE0RD4	UE0RD3	UE0RD2	UE0RD1	UE0RD0
\$0020	W	UE0TD7	UE0TD6	UE0TD5	UE0TD4	UE0TD3	UE0TD2	UE0TD1	UE0TD0
		:	:	:	:	:	:	:	:
to		:	:	:	:	:	:	:	:
UE0D7	R	UE0RD7	UE0RD6	UE0RD5	UE0RD4	UE0RD3	UE0RD2	UE0RD1	UE0RD0
\$0027	W	UE0TD7	UE0TD6	UE0TD5	UE0TD4	UE0TD3	UE0TD2	UE0TD1	UE0TD0
reset ⇒		X	X	X	X	X	X	X	X

Figure 10-27. USB Endpoint 0 Data Register (UE0D0-UE0D7)**UE0RD7 - UE0RD0 — Endpoint 0 Receive Data Buffer**

These read only bits are serially loaded with OUT token or SETUP token data received over the USB's D+ and D- pins.

UE0TD7 - UE0TD0 — Endpoint 0 Transmit Data Buffer

These write only buffers are loaded by software with data to be sent on the USB bus on the next IN token directed at Endpoint 0.

10.5.9 USB Endpoint 1/Endpoint 2 Data Registers (UE1D0-UE1D7)

		Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
UE1D0	R								
\$0028	W	UE1TD7	UE1TD6	UE1TD5	UE1TD4	UE1TD3	UE1TD2	UE1TD1	UE1TD0
		:	:	:	:	:	:	:	:
to		:	:	:	:	:	:	:	:
UE1D7	R								
\$002F	W	UE1TD7	UE1TD6	UE1TD5	UE1TD4	UE1TD3	UE1TD2	UE1TD1	UE1TD0
reset ⇒		X	X	X	X	X	X	X	X

Figure 10-28. USB Endpoint 1/Endpoint2 Data Registers (UE1D0-UE1D7)**UE1TD7 - UE1TD0 — Endpoint 1/ Endpoint 2 Transmit Data Buffer**

These write only buffers are loaded by software with data to be sent on the USB bus on the next IN token directed at Endpoint 1 or Endpoint 2. These buffers are shared by Endpoints 1 and 2 and depend on proper configuration of the ENDADD bit.

10.6 USB INTERRUPTS

The USB module is capable of generating interrupts and causing the CPU to execute the USB interrupt service routine. There are three types of USB interrupts:

- End of Transaction interrupts signify a completed transaction (receive or transmit)
- Resume interrupts signify that the USB bus is reactivated after having been suspended
- End of Packet interrupts signify that a low speed end of packet signal was detected

All USB interrupts share the same interrupt vector. Firmware is responsible for determining which interrupt is active.

10.6.1 USB End of Transaction Interrupt

There are three possible end of transaction interrupts: Endpoint 0 Receive, Endpoint 0 Transmit, and a shared Endpoint 1 or Endpoint 2 Transmit. End of transaction interrupts occur as detailed in the following sections.

10.6.1.1 Receive Control Endpoint 0

For a Control OUT transaction directed at Endpoint 0, the USB module will generate an interrupt by setting the RXD0F flag in the UIR0 register. The conditions necessary for the interrupt to occur are shown in the flowchart of **Figure 10-29**.

SETUP transactions cannot be stalled by the USB function. A SETUP received by a control endpoint will clear the STALL0 bit if it is set. The conditions for receiving a SETUP interrupt are shown in **Figure 10-30**.

10.6.1.2 Transmit Control Endpoint 0

For a Control IN transaction directed at Endpoint 0, the USB module will generate an interrupt by setting the TXD0F flag in the UIR0 register. The conditions necessary for the interrupt to occur are shown in the flowchart of **Figure 10-31**.

10.6.1.3 Transmit Endpoint 1 and Transmit Endpoint 2

Transmit Endpoints 1 & 2 share their interrupt flag. For an IN transaction directed at Endpoint 1 or 2, the USB module will generate an interrupt by setting the TXD1F flag in the UIR1 register. The conditions necessary for the interrupt to occur are shown in the flowchart of **Figure 10-32**.

10.6.2 Resume Interrupt

The USB module will generate a USB interrupt if low speed bus activity is detected after entering the suspend state. A transition of the USB data lines to the non-idle state ("K" state) while in the suspend mode will set the RESUMF flag in the UIR1 register. There is no interrupt enable bit for this interrupt source and an interrupt will be executed if the I bit in the CCR is cleared. A resume interrupt can only occur while the MC68HC705JB2 is in the suspend mode.

10.6.3 End of Packet Interrupt

The USB module can generate a USB interrupt upon detection of an end of packet signal (a single ended 0) for low speed devices. Upon detection of an SE0 sequence, the USB module sets the EOPF bit and will generate an interrupt if the EOPIE bit in the UIR1 register is set.

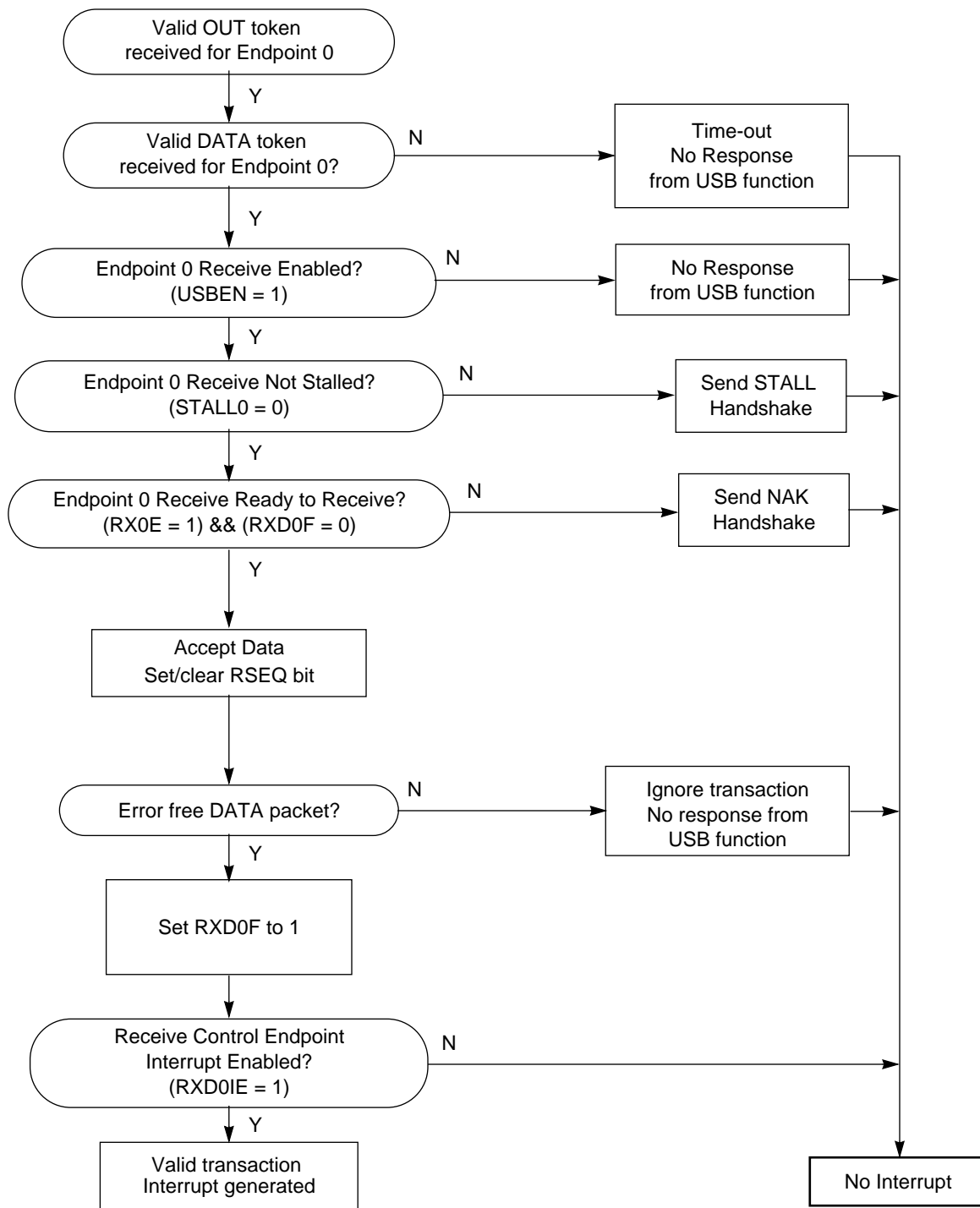


Figure 10-29. OUT Token Data Flow for Receive Endpoint 0

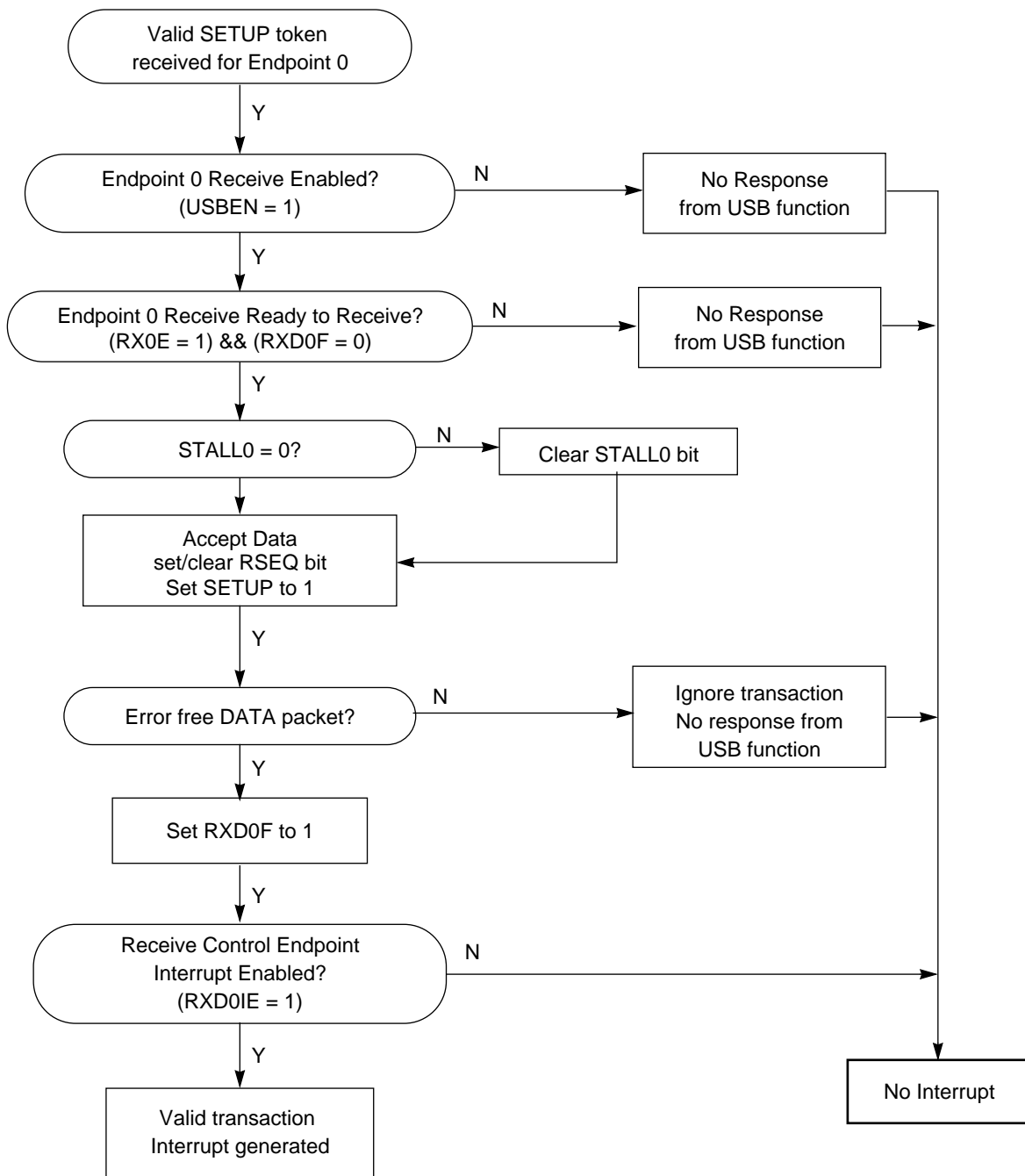


Figure 10-30. SETUP Token Data Flow for Receive Endpoint 0

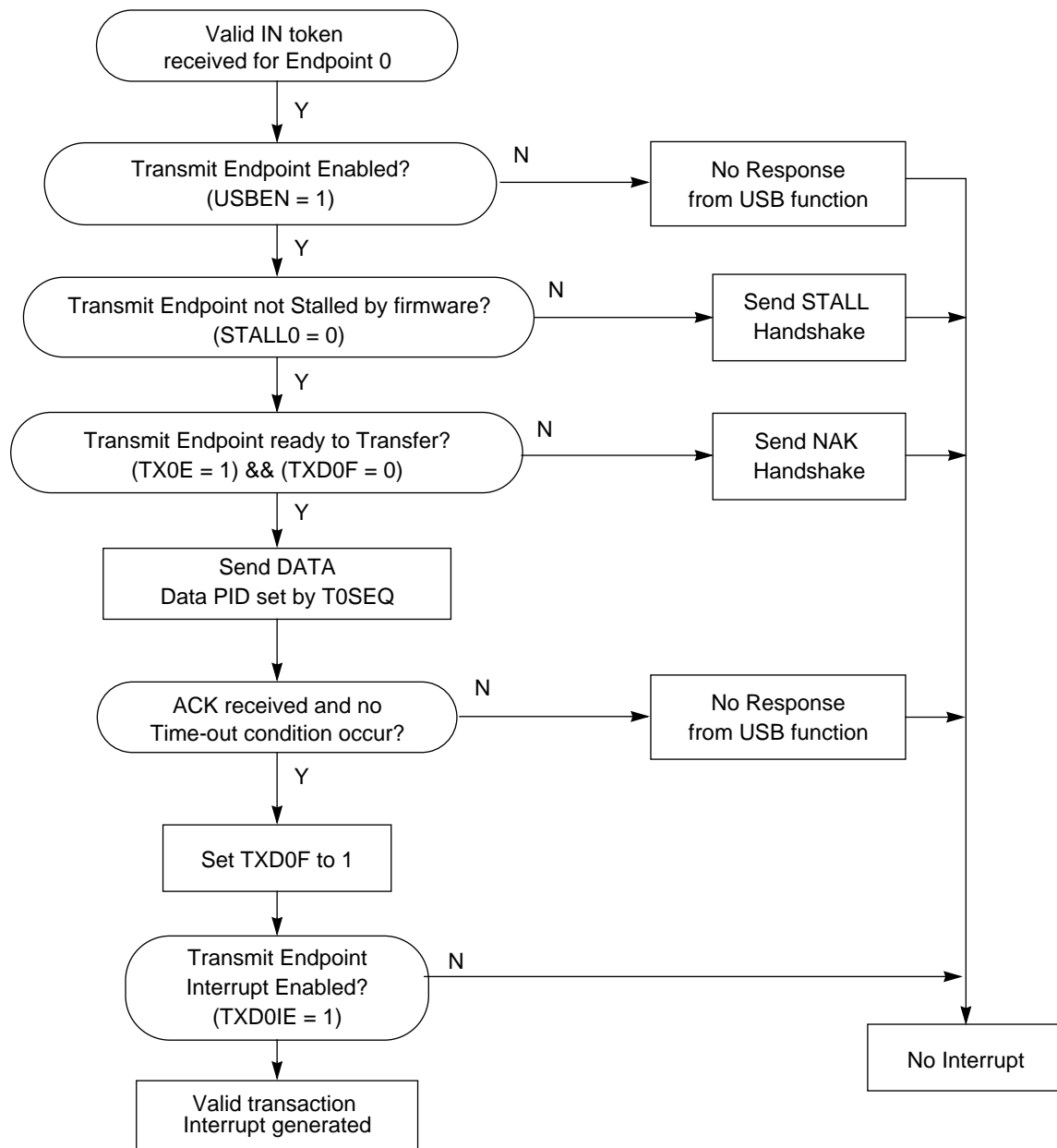


Figure 10-31. IN Token Data Flow for Transmit Endpoint 0

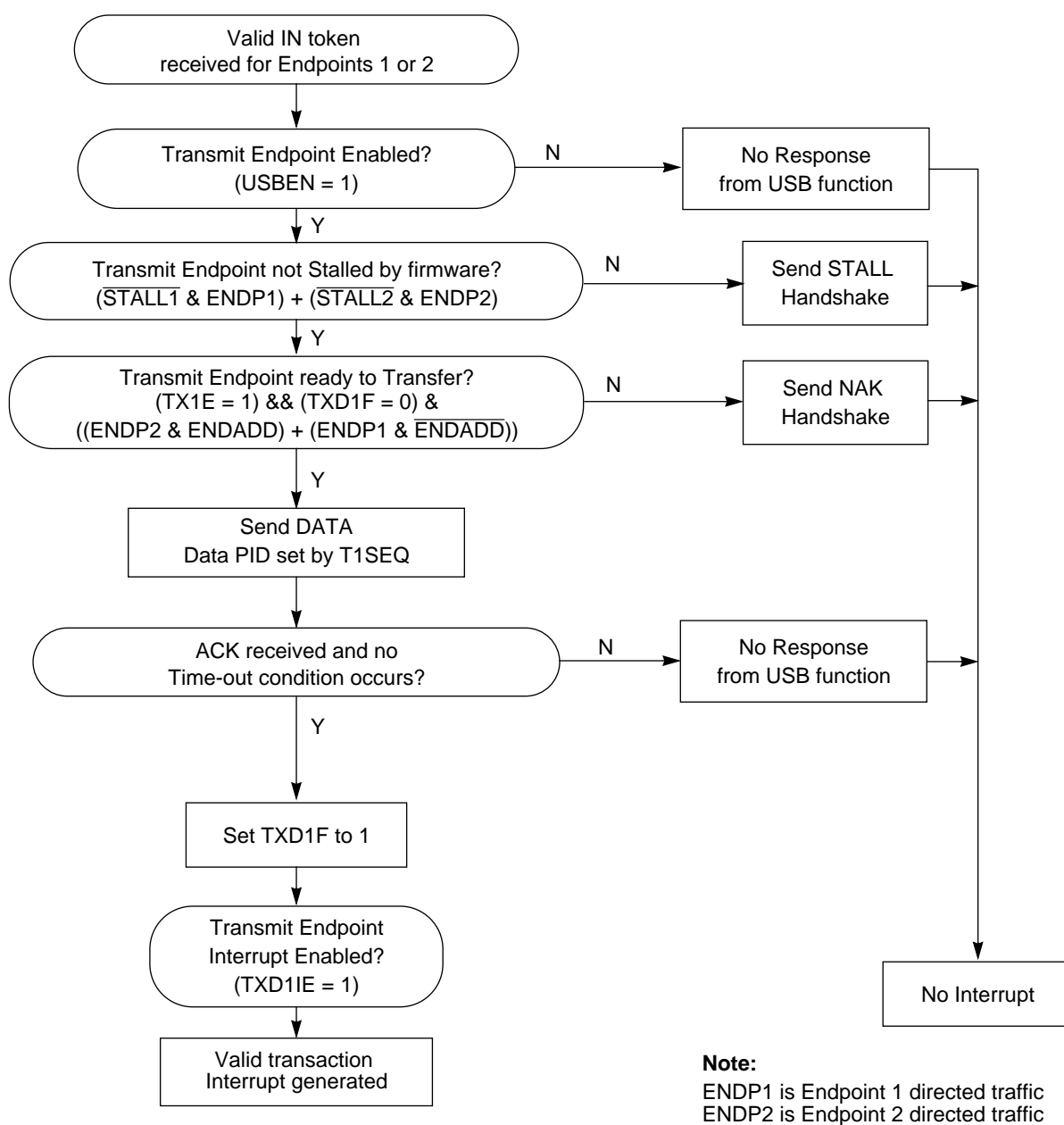


Figure 10-32. IN Token Data Flow for Transmit Endpoint 1/ Endpoint 2

SECTION 11

EPROM

This section describes erasable programmable read-only memory (EPROM) programming.

11.1 EPROM

The on-chip user EPROM consists of 2048 bytes of EPROM from \$1600 to \$1DFF and 16 bytes of user vectors from \$1FF0 to \$1FFF.

The bootloader ROM and vectors are located from \$1E00 to \$1FEF.

12 of the user vectors, \$1FF4-\$1FFF, are dedicated to reset and interrupt vectors. The four remaining locations, \$1FF0-\$1FF3, are reserved for test functions. The Mask Option Register is located at \$01FF.


11.2 BOOTLOADER

This program (contained in an on-chip boot ROM) handles copying of user code from an external EPROM into the on-chip EPROM. The bootloader function does not have to be done from an EPROM, but can be done from a host.

11.2.1 Bootloader Mode

Bootloader mode is entered upon the rising edge of $\overline{\text{RESET}}$ if the $\overline{\text{IRQ}}/\text{V}_{\text{PP}}$ pin is at V_{TST} and the PB0 pin is at logic zero. The bootloader performs one programming pass at 1 ms per byte then does a verify pass.

Table 11-1. Operation Mode Condition After Reset

RESET Pin	IRQ/VPP	PB0/TCAP	MODE
	V_{SS} to V_{DD}	V_{SS} to V_{DD}	Single-Chip (Normal)
	V_{TST}	V_{SS}	Bootloader

$$V_{\text{TST}} = 2 \times V_{\text{DD}}$$

The user code must be a one-to-one correspondence with the internal EPROM addresses.

11.3 EPROM PROGRAMMING

Programming the on-chip EPROM is achieved by using the Program Control Register located at address \$3E.

Please contact Motorola for programming board availability.

11.3.1 EPROM Program Control Register (PCR)

This register is provided for programming the on-chip EPROM.

PCR \$003E		bit-7	bit-6	bit-5	bit-4	bit-3	bit-2	bit-1	bit-0
	Read	RESERVED					MORON	ELAT	PGM
	Write								
	Reset	0	0	0	0	0	0	0	0

MORON – Mask Option Register ON

- 0 = Disable programming to Mask Option Register (\$01FF)
- 1 = Enable programming to Mask Option Register (\$01FF)

ELAT – EPROM LATch control

- 0 = EPROM address and data bus configured for normal reads
- 1 = EPROM address and data bus configured for programming (writes to EPROM cause address and data to be latched). EPROM is in programming mode and cannot be read if ELAT is 1. This bit should not be set when no programming voltage is applied to the V_{pp} pin.

PGM – EPROM ProGraM command

- 0 = Programming power is switched OFF from EPROM array.
- 1 = Programming power is switched ON to EPROM array. If ELAT \neq 1, then PGM = 0.

11.3.2 Programming Sequence

The EPROM programming sequence is:

1. Set the ELAT bit
2. Write the data to the address to be programmed
3. Set the PGM bit
4. Delay for a time t_{PGMR}
5. Clear the PGM bit
6. Clear the ELAT bit

The last two steps must be performed with separate CPU writes.

CAUTION

It is important to remember that an external programming voltage must be applied to the V_{PP} pin while programming, but it should be equal to V_{DD} during normal operations.

Figure 11-1 shows the flow required to successfully program the EPROM.

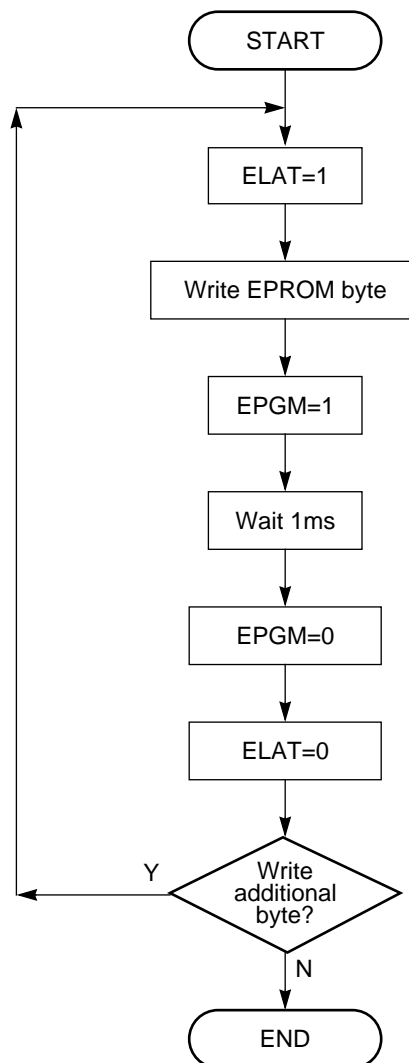


Figure 11-1. EPROM Programming Sequence

11.4 MASK OPTION REGISTER (MOR), \$01FF

The Mask Option Register (MOR) contains programmable EPROM bits to control mask options. In order to program this register the MORON bit in PCR need to be set to "1" before doing the EPROM programming process.

	bit-7	bit-6	bit-5	bit-4	bit-3	bit-2	bit-1	bit-0
MOR \$01FF	Read			IRQTRIG	PULLREN	PAINTEN	OSCDLY	LVREN
	Write							
	Erased	0	0	1	1	1	1	1
	Reset	Unaffected						

IRQTRIG – IRQ, PA0-PA3 Interrupt Options

- 1 = Edge-trigger only
- 0 = Edge-and-level-triggered

PULLREN – Port A and B Pullup/Pulldown Options

- 1 = Connected
- 0 = Disconnected

PAINTEN – PA0-PA3 External Interrupt Options

- 1 = Disabled
- 0 = Enabled

OSCDLY – Oscillator Delay Option

- 1 = 128 internal clock cycles
- 0 = 4064 internal clock cycles

LVREN – LVR Option

- 1 = Enabled
- 0 = Disabled

SECTION 12

INSTRUCTION SET

This section describes the addressing modes and instruction types.

12.1 ADDRESSING MODES

The CPU uses eight addressing modes for flexibility in accessing data. The addressing modes define the manner in which the CPU finds the data required to execute an instruction. The eight addressing modes are the following:

- Inherent
- Immediate
- Direct
- Extended
- Indexed, No Offset
- Indexed, 8-Bit Offset
- Indexed, 16-Bit Offset
- Relative

12.1.1 Inherent

Inherent instructions are those that have no operand, such as return from interrupt (RTI) and stop (STOP). Some of the inherent instructions act on data in the CPU registers, such as set carry flag (SEC) and increment accumulator (INCA). Inherent instructions require no memory address and are one byte long.

12.1.2 Immediate

Immediate instructions are those that contain a value to be used in an operation with the value in the accumulator or index register. Immediate instructions require no memory address and are two bytes long. The opcode is the first byte, and the immediate data value is the second byte.

12.1.3 Direct

Direct instructions can access any of the first 256 memory addresses with two bytes. The first byte is the opcode, and the second is the low byte of the operand address. In direct addressing, the CPU automatically uses \$00 as the high byte of the operand address. BRSET and BRCLR are three-byte instructions that use direct addressing to access the operand and relative addressing to specify a branch destination.

12.1.4 Extended

Extended instructions use only three bytes to access any address in memory. The first byte is the opcode; the second and third bytes are the high and low bytes of the operand address.

When using the Motorola assembler, the programmer does not need to specify whether an instruction is direct or extended. The assembler automatically selects the shortest form of the instruction.

12.1.5 Indexed, No Offset

Indexed instructions with no offset are one-byte instructions that can access data with variable addresses within the first 256 memory locations. The index register contains the low byte of the conditional address of the operand. The CPU automatically uses \$00 as the high byte, so these instructions can address locations \$0000–\$00FF.

Indexed, no offset instructions are often used to move a pointer through a table or to hold the address of a frequently used RAM or I/O location.

12.1.6 Indexed, 8-Bit Offset

Indexed, 8-bit offset instructions are two-byte instructions that can access data with variable addresses within the first 511 memory locations. The CPU adds the unsigned byte in the index register to the unsigned byte following the opcode. The sum is the conditional address of the operand. These instructions can access locations \$0000–\$01FE.

Indexed 8-bit offset instructions are useful for selecting the *k*th element in an *n*-element table. The table can begin anywhere within the first 256 memory locations and could extend as far as location 510 (\$01FE). The *k* value is typically in the index register, and the address of the beginning of the table is in the byte following the opcode.

12.1.7 Indexed, 16-Bit Offset

Indexed, 16-bit offset instructions are three-byte instructions that can access data with variable addresses at any location in memory. The CPU adds the unsigned byte in the index register to the two unsigned bytes following the opcode. The sum is the conditional address of the operand. The first byte after the opcode is the high byte of the 16-bit offset; the second byte is the low byte of the offset. These instructions can address any location in memory.

Indexed, 16-bit offset instructions are useful for selecting the *k*th element in an *n*-element table anywhere in memory.

As with direct and extended addressing, the Motorola assembler determines the shortest form of indexed addressing.

12.1.8 Relative

Relative addressing is only for branch instructions. If the branch condition is true, the CPU finds the conditional branch destination by adding the signed byte following the opcode to the contents of the program counter. If the branch condition is not true, the CPU goes to the next instruction. The offset is a signed, two's complement byte that gives a branching range of -128 to $+127$ bytes from the address of the next location after the branch instruction.

When using the Motorola assembler, the programmer does not need to calculate the offset, because the assembler determines the proper offset and verifies that it is within the span of the branch.

12.1.9 Instruction Types

The MCU instructions fall into the following five categories:

- Register/Memory Instructions
- Read-Modify-Write Instructions
- Jump/Branch Instructions
- Bit Manipulation Instructions
- Control Instructions

12.1.10 Register/Memory Instructions

Most of these instructions use two operands. One operand is in either the accumulator or the index register. The CPU finds the other operand in memory. **Table 12-1** lists the register/memory instructions.

Table 12-1. Register/Memory Instructions

Instruction	Mnemonic
Add Memory Byte and Carry Bit to Accumulator	ADC
Add Memory Byte to Accumulator	ADD
AND Memory Byte with Accumulator	AND
Bit Test Accumulator	BIT
Compare Accumulator	CMP
Compare Index Register with Memory Byte	CPX
EXCLUSIVE OR Accumulator with Memory Byte	EOR
Load Accumulator with Memory Byte	LDA
Load Index Register with Memory Byte	LDX
Multiply	MUL
OR Accumulator with Memory Byte	ORA
Subtract Memory Byte and Carry Bit from Accumulator	SBC
Store Accumulator in Memory	STA
Store Index Register in Memory	STX
Subtract Memory Byte from Accumulator	SUB

12.1.11 Read-Modify-Write Instructions

These instructions read a memory location or a register, modify its contents, and write the modified value back to the memory location or to the register. The test for negative or zero instruction (TST) is an exception to the read-modify-write sequence because it does not write a replacement value. **Table 12-2** lists the read-modify-write instructions.

Table 12-2. Read-Modify-Write Instructions

Instruction	Mnemonic
Arithmetic Shift Left	ASL
Arithmetic Shift Right	ASR
Clear Bit in Memory	BCLR
Set Bit in Memory	BSET
Clear	CLR
Complement (One's Complement)	COM
Decrement	DEC
Increment	INC
Logical Shift Left	LSL
Logical Shift Right	LSR
Negate (Two's Complement)	NEG
Rotate Left through Carry Bit	ROL
Rotate Right through Carry Bit	ROR
Test for Negative or Zero	TST

12.1.12 Jump/Branch Instructions

Jump instructions allow the CPU to interrupt the normal sequence of the program counter. The unconditional jump instruction (JMP) and the jump to subroutine instruction (JSR) have no register operand. Branch instructions allow the CPU to interrupt the normal sequence of the program counter when a test condition is met. If the test condition is not met, the branch is not performed. All branch instructions use relative addressing.

Bit test and branch instructions cause a branch based on the state of any readable bit in the first 256 memory locations. These three-byte instructions use a combination of direct addressing and relative addressing. The direct address of the byte to be tested is in the byte following the opcode. The third byte is the signed offset byte. The CPU finds the conditional branch destination by adding the

third byte to the program counter if the specified bit tests true. The bit to be tested and its condition (set or clear) is part of the opcode. The span of branching is from –128 to +127 from the address of the next location after the branch instruction. The CPU also transfers the tested bit to the carry/borrow bit of the condition code register. **Table 12-3** lists the jump and branch instructions.

Table 12-3. Jump and Branch Instructions

Instruction	Mnemonic
Branch if Carry Bit Clear	BCC
Branch if Carry Bit Set	BCS
Branch if Equal	BEQ
Branch if Half-Carry Bit Clear	BHCC
Branch if Half-Carry Bit Set	BHCS
Branch if Higher	BHI
Branch if Higher or Same	BHS
Branch if $\overline{\text{IRQ}}$ Pin High	BIH
Branch if $\overline{\text{IRQ}}$ Pin Low	BIL
Branch if Lower	BLO
Branch if Lower or Same	BLS
Branch if Interrupt Mask Clear	BMC
Branch if Minus	BMI
Branch if Interrupt Mask Set	BMS
Branch if Not Equal	BNE
Branch if Plus	BPL
Branch Always	BRA
Branch if Bit Clear	BRCLR
Branch Never	BRN
Branch if Bit Set	BRSET
Branch to Subroutine	BSR
Unconditional Jump	JMP
Jump to Subroutine	JSR

12.1.13 Bit Manipulation Instructions

The CPU can set or clear any writable bit in the first 256 bytes of memory. Port registers, port data direction registers, timer registers, and on-chip RAM locations are in the first 256 bytes of memory. The CPU can also test and branch based on the state of any bit in any of the first 256 memory locations. Bit manipulation instructions use direct addressing. **Table 12-4** lists these instructions.

Table 12-4. Bit Manipulation Instructions

Instruction	Mnemonic
Clear Bit	BCLR
Branch if Bit Clear	BRCLR
Branch if Bit Set	BRSET
Set Bit	BSET

12.1.14 Control Instructions

These register reference instructions control CPU operation during program execution. Control instructions, listed in **Table 12-5**, use inherent addressing.

Table 12-5. Control Instructions

Instruction	Mnemonic
Clear Carry Bit	CLC
Clear Interrupt Mask	CLI
No Operation	NOP
Reset Stack Pointer	RSP
Return from Interrupt	RTI
Return from Subroutine	RTS
Set Carry Bit	SEC
Set Interrupt Mask	SEI
Stop Oscillator and Enable \overline{IRQ} Pin	STOP
Software Interrupt	SWI
Transfer Accumulator to Index Register	TAX
Transfer Index Register to Accumulator	TXA
Stop CPU Clock and Enable Interrupts	WAIT

12.1.15 Instruction Set Summary

Table 12-6 is an alphabetical list of all M68HC05 instructions and shows the effect of each instruction on the condition code register.

Table 12-6. Instruction Set Summary

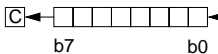
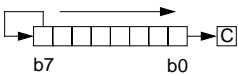
Source Form	Operation	Description	Effect on CCR					Address Mode	Opcode	Operand	Cycles
			H	I	N	Z	C				
ADC #opr ADC opr ADC opr ADC opr,X ADC opr,X ADC ,X	Add with Carry	$A \leftarrow (A) + (M) + (C)$	↑	—	↑	↑	↑	IMM DIR EXT IX2 IX1 IX	A9 B9 C9 D9 E9 F9	ii dd hh ll ee ff ff	2 3 4 5 4 3
ADD #opr ADD opr ADD opr ADD opr,X ADD opr,X ADD ,X	Add without Carry	$A \leftarrow (A) + (M)$	↑	—	↑	↑	↑	IMM DIR EXT IX2 IX1 IX	AB BB CB DB EB FB	ii dd hh ll ee ff ff	2 3 4 5 4 3
AND #opr AND opr AND opr AND opr,X AND opr,X AND ,X	Logical AND	$A \leftarrow (A) \wedge (M)$	—	—	↑	↑	—	IMM DIR EXT IX2 IX1 IX	A4 B4 C4 D4 E4 F4	ii dd hh ll ee ff ff	2 3 4 5 4 3
ASL opr ASLA ASLX ASL opr,X ASL ,X	Arithmetic Shift Left (Same as LSL)		—	—	↑	↑	↑	↑	38 48 58 68 78	dd ff	5 3 3 6 5
ASR opr ASRA ASRX ASR opr,X ASR ,X	Arithmetic Shift Right		—	—	↑	↑	↑	DIR INH INH IX1 IX	37 47 57 67 77	dd ff	5 3 3 6 5
BCC rel	Branch if Carry Bit Clear	$PC \leftarrow (PC) + 2 + rel ? C = 0$	—	—	—	—	—	REL	24	rr	3
BCLR n opr	Clear Bit n	$M_n \leftarrow 0$	—	—	—	—	—	DIR (b0) DIR (b1) DIR (b2) DIR (b3) DIR (b4) DIR (b5) DIR (b6) DIR (b7)	11 13 15 17 19 1B 1D 1F	dd dd dd dd dd dd dd dd	5 5 5 5 5 5 5 5
BCS rel	Branch if Carry Bit Set (Same as BLO)	$PC \leftarrow (PC) + 2 + rel ? C = 1$	—	—	—	—	—	REL	25	rr	3
BEQ rel	Branch if Equal	$PC \leftarrow (PC) + 2 + rel ? Z = 1$	—	—	—	—	—	REL	27	rr	3

Table 12-6. Instruction Set Summary (Continued)

Source Form	Operation	Description	Effect on CCR					Address Mode	Opcode	Operand	Cycles
			H	I	N	Z	C				
BHCC <i>rel</i>	Branch if Half-Carry Bit Clear	$PC \leftarrow (PC) + 2 + rel ? H = 0$	—	—	—	—	—	REL	28	rr	3
BHCS <i>rel</i>	Branch if Half-Carry Bit Set	$PC \leftarrow (PC) + 2 + rel ? H = 1$	—	—	—	—	—	REL	29	rr	3
BHI <i>rel</i>	Branch if Higher	$PC \leftarrow (PC) + 2 + rel ? C \vee Z = 0$	—	—	—	—	—	REL	22	rr	3
BHS <i>rel</i>	Branch if Higher or Same	$PC \leftarrow (PC) + 2 + rel ? C = 0$	—	—	—	—	—	REL	24	rr	3
BIH <i>rel</i>	Branch if \overline{IRQ} Pin High	$PC \leftarrow (PC) + 2 + rel ? \overline{IRQ} = 1$	—	—	—	—	—	REL	2F	rr	3
BIL <i>rel</i>	Branch if \overline{IRQ} Pin Low	$PC \leftarrow (PC) + 2 + rel ? \overline{IRQ} = 0$	—	—	—	—	—	REL	2E	rr	3
BIT # <i>opr</i> BIT <i>opr</i> BIT <i>opr</i> BIT <i>opr</i> ,X BIT <i>opr</i> ,X BIT ,X	Bit Test Accumulator with Memory Byte	(A) \wedge (M)	—	—	↑	↑	—	IMM DIR EXT IX2 IX1 IX	A5 B5 C5 D5 E5 F5	ii dd hh ll ee ff ff p	2 3 4 5 4 3
BLO <i>rel</i>	Branch if Lower (Same as BCS)	$PC \leftarrow (PC) + 2 + rel ? C = 1$	—	—	—	—	—	REL	25	rr	3
BLS <i>rel</i>	Branch if Lower or Same	$PC \leftarrow (PC) + 2 + rel ? C \vee Z = 1$	—	—	—	—	—	REL	23	rr	3
BMC <i>rel</i>	Branch if Interrupt Mask Clear	$PC \leftarrow (PC) + 2 + rel ? I = 0$	—	—	—	—	—	REL	2C	rr	3
BMI <i>rel</i>	Branch if Minus	$PC \leftarrow (PC) + 2 + rel ? N = 1$	—	—	—	—	—	REL	2B	rr	3
BMS <i>rel</i>	Branch if Interrupt Mask Set	$PC \leftarrow (PC) + 2 + rel ? I = 1$	—	—	—	—	—	REL	2D	rr	3
BNE <i>rel</i>	Branch if Not Equal	$PC \leftarrow (PC) + 2 + rel ? Z = 0$	—	—	—	—	—	REL	26	rr	3
BPL <i>rel</i>	Branch if Plus	$PC \leftarrow (PC) + 2 + rel ? N = 0$	—	—	—	—	—	REL	2A	rr	3
BRA <i>rel</i>	Branch Always	$PC \leftarrow (PC) + 2 + rel ? 1 = 1$	—	—	—	—	—	REL	20	rr	3
BRCLR <i>n opr rel</i>	Branch if bit <i>n</i> clear	$PC \leftarrow (PC) + 2 + rel ? Mn = 0$	—	—	—	—	↑	DIR (b0) DIR (b1) DIR (b2) DIR (b3) DIR (b4) DIR (b5) DIR (b6) DIR (b7)	01 03 05 07 09 0B 0D 0F	dd rr dd rr dd rr dd rr dd rr dd rr dd rr dd rr	5 5 5 5 5 5 5 5
BRSET <i>n opr rel</i>	Branch if Bit <i>n</i> Set	$PC \leftarrow (PC) + 2 + rel ? Mn = 1$	—	—	—	—	↑	DIR (b0) DIR (b1) DIR (b2) DIR (b3) DIR (b4) DIR (b5) DIR (b6) DIR (b7)	00 02 04 06 08 0A 0C 0E	dd rr dd rr dd rr dd rr dd rr dd rr dd rr dd rr	5 5 5 5 5 5 5 5
BRN <i>rel</i>	Branch Never	$PC \leftarrow (PC) + 2 + rel ? 1 = 0$	—	—	—	—	—	REL	21	rr	3

Table 12-6. Instruction Set Summary (Continued)

Source Form	Operation	Description	Effect on CCR					Address Mode	Opcode	Operand	Cycles
			H	I	N	Z	C				
BSET <i>n opr</i>	Set Bit <i>n</i>	$M_n \leftarrow 1$	—	—	—	—	—	DIR (b0) DIR (b1) DIR (b2) DIR (b3) DIR (b4) DIR (b5) DIR (b6) DIR (b7)	10 12 14 16 18 1A 1C 1E	dd dd dd dd dd dd dd dd	5 5 5 5 5 5 5 5
BSR <i>rel</i>	Branch to Subroutine	$PC \leftarrow (PC) + 2$; push (PCL) $SP \leftarrow (SP) - 1$; push (PCH) $SP \leftarrow (SP) - 1$ $PC \leftarrow (PC) + rel$	—	—	—	—	—	REL	AD	rr	6
CLC	Clear Carry Bit	$C \leftarrow 0$	—	—	—	—	0	INH	98		2
CLI	Clear Interrupt Mask	$I \leftarrow 0$	—	0	—	—	—	INH	9A		2
CLR <i>opr</i> CLRA CLR X CLR <i>opr,X</i> CLR ,X	Clear Byte	$M \leftarrow \$00$ $A \leftarrow \$00$ $X \leftarrow \$00$ $M \leftarrow \$00$ $M \leftarrow \$00$	—	—	0	1	—	DIR INH INH IX1 IX	3F 4F 5F 6F 7F	dd ff	5 3 3 6 5
CMP # <i>opr</i> CMP <i>opr</i> CMP <i>opr</i> CMP <i>opr,X</i> CMP <i>opr,X</i> CMP ,X	Compare Accumulator with Memory Byte	$(A) - (M)$	—	—	↑	↑	↑	IMM DIR EXT IX2 IX1 IX	A1 B1 C1 D1 E1 F1	ii dd hh ll ee ff ff	2 3 4 5 4 3
COM <i>opr</i> COMA COM X COM <i>opr,X</i> COM ,X	Complement Byte (One's Complement)	$M \leftarrow (\overline{M}) = \$FF - (M)$ $A \leftarrow (\overline{A}) = \$FF - (M)$ $X \leftarrow (\overline{X}) = \$FF - (M)$ $M \leftarrow (\overline{M}) = \$FF - (M)$ $M \leftarrow (\overline{M}) = \$FF - (M)$	—	—	↑	↑	1	DIR INH INH IX1 IX	33 43 53 63 73	dd ff	5 3 3 6 5
CPX # <i>opr</i> CPX <i>opr</i> CPX <i>opr</i> CPX <i>opr,X</i> CPX <i>opr,X</i> CPX ,X	Compare Index Register with Memory Byte	$(X) - (M)$	—	—	↑	↑	↑	IMM DIR EXT IX2 IX1 IX	A3 B3 C3 D3 E3 F3	ii dd hh ll ee ff ff	2 3 4 5 4 3
DEC <i>opr</i> DECA DEC X DEC <i>opr,X</i> DEC ,X	Decrement Byte	$M \leftarrow (M) - 1$ $A \leftarrow (A) - 1$ $X \leftarrow (X) - 1$ $M \leftarrow (M) - 1$ $M \leftarrow (M) - 1$	—	—	↑	↑	—	DIR INH INH IX1 IX	3A 4A 5A 6A 7A	dd ff	5 3 3 6 5
EOR # <i>opr</i> EOR <i>opr</i> EOR <i>opr</i> EOR <i>opr,X</i> EOR <i>opr,X</i> EOR ,X	EXCLUSIVE OR Accumulator with Memory Byte	$A \leftarrow (A) \oplus (M)$	—	—	↑	↑	—	IMM DIR EXT IX2 IX1 IX	A8 B8 C8 D8 E8 F8	ii dd hh ll ee ff ff	2 3 4 5 4 3

Table 12-6. Instruction Set Summary (Continued)

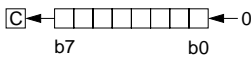
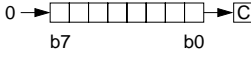
Source Form	Operation	Description	Effect on CCR					Address Mode	Opcode	Operand	Cycles
			H	I	N	Z	C				
INC <i>opr</i> INCA INCX INC <i>opr</i> ,X INC ,X	Increment Byte	$M \leftarrow (M) + 1$ $A \leftarrow (A) + 1$ $X \leftarrow (X) + 1$ $M \leftarrow (M) + 1$ $M \leftarrow (M) + 1$	—	—	↑	↑	—	DIR INH INH IX1 IX	3C 4C 5C 6C 7C	dd ff	5 3 3 6 5
JMP <i>opr</i> JMP <i>opr</i> JMP <i>opr</i> ,X JMP <i>opr</i> ,X JMP ,X	Unconditional Jump	$PC \leftarrow \text{Jump Address}$	—	—	—	—	—	DIR EXT IX2 IX1 IX	BC CC DC EC FC	dd hh ll ee ff ff	2 3 4 3 2
JSR <i>opr</i> JSR <i>opr</i> JSR <i>opr</i> ,X JSR <i>opr</i> ,X JSR ,X	Jump to Subroutine	$PC \leftarrow (PC) + n$ ($n = 1, 2, \text{ or } 3$) Push (PCL); $SP \leftarrow (SP) - 1$ Push (PCH); $SP \leftarrow (SP) - 1$ $PC \leftarrow \text{Conditional Address}$	—	—	—	—	—	DIR EXT IX2 IX1 IX	BD CD DD ED FD	dd hh ll ee ff ff	5 6 7 6 5
LDA # <i>opr</i> LDA <i>opr</i> LDA <i>opr</i> LDA <i>opr</i> ,X LDA <i>opr</i> ,X LDA ,X	Load Accumulator with Memory Byte	$A \leftarrow (M)$	—	—	↑	↑	—	IMM DIR EXT IX2 IX1 IX	A6 B6 C6 D6 E6 F6	ii dd hh ll ee ff ff	2 3 4 5 4 3
LDX # <i>opr</i> LDX <i>opr</i> LDX <i>opr</i> LDX <i>opr</i> ,X LDX <i>opr</i> ,X LDX ,X	Load Index Register with Memory Byte	$X \leftarrow (M)$	—	—	↑	↑	—	IMM DIR EXT IX2 IX1 IX	AE BE CE DE EE FE	ii dd hh ll ee ff ff	2 3 4 5 4 3
LSL <i>opr</i> LSLA LSLX LSL <i>opr</i> ,X LSL ,X	Logical Shift Left (Same as ASL)		—	—	↑	↑	↑	DIR INH INH IX1 IX	38 48 58 68 78	dd ff	5 3 3 6 5
LSR <i>opr</i> LSRA LSRX LSR <i>opr</i> ,X LSR ,X	Logical Shift Right		—	—	0	↑	↑	DIR INH INH IX1 IX	34 44 54 64 74	dd ff	5 3 3 6 5
MUL	Unsigned Multiply	$X : A \leftarrow (X) \times (A)$	0	—	—	—	0	INH	42		11
NEG <i>opr</i> NEGA NEGX NEG <i>opr</i> ,X NEG ,X	Negate Byte (Two's Complement)	$M \leftarrow -(M) = \$00 - (M)$ $A \leftarrow -(A) = \$00 - (A)$ $X \leftarrow -(X) = \$00 - (X)$ $M \leftarrow -(M) = \$00 - (M)$ $M \leftarrow -(M) = \$00 - (M)$	—	—	↑	↑	↑	DIR INH INH IX1 IX	30 40 50 60 70	ii ff	5 3 3 6 5
NOP	No Operation		—	—	—	—	—	INH	9D		2

Table 12-6. Instruction Set Summary (Continued)

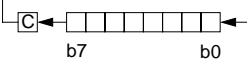
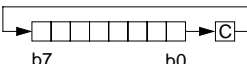
Source Form	Operation	Description	Effect on CCR					Address Mode	Opcode	Operand	Cycles
			H	I	N	Z	C				
ORA #opr ORA opr ORA opr ORA opr,X ORA opr,X ORA ,X	Logical OR Accumulator with Memory	$A \leftarrow (A) \vee (M)$	—	—	↑	↑	—	IMM DIR EXT IX2 IX1 IX	AA BA CA DA EA FA	ii dd hh ll ee ff ff	2 3 4 5 4 3
ROL opr ROLA ROLX ROL opr,X ROL ,X	Rotate Byte Left through Carry Bit		—	—	↑	↑	↑	DIR INH INH IX1 IX	39 49 59 69 79	dd ff	5 3 3 6 5
ROR opr RORA RORX ROR opr,X ROR ,X	Rotate Byte Right through Carry Bit		—	—	↑	↑	↑	DIR INH INH IX1 IX	36 46 56 66 76	dd ff	5 3 3 6 5
RSP	Reset Stack Pointer	$SP \leftarrow \$00FF$	—	—	—	—	—	INH	9C		2
RTI	Return from Interrupt	$SP \leftarrow (SP) + 1$; Pull (CCR) $SP \leftarrow (SP) + 1$; Pull (A) $SP \leftarrow (SP) + 1$; Pull (X) $SP \leftarrow (SP) + 1$; Pull (PCH) $SP \leftarrow (SP) + 1$; Pull (PCL)	↑	↑	↑	↑	↑	INH	80		9
RTS	Return from Subroutine	$SP \leftarrow (SP) + 1$; Pull (PCH) $SP \leftarrow (SP) + 1$; Pull (PCL)	—	—	—	—	—	INH	81		6
SBC #opr SBC opr SBC opr SBC opr,X SBC opr,X SBC ,X	Subtract Memory Byte and Carry Bit from Accumulator	$A \leftarrow (A) - (M) - (C)$	—	—	↑	↑	↑	IMM DIR EXT IX2 IX1 IX	A2 B2 C2 D2 E2 F2	ii dd hh ll ee ff ff	2 3 4 5 4 3
SEC	Set Carry Bit	$C \leftarrow 1$	—	—	—	—	1	INH	99		2
SEI	Set Interrupt Mask	$I \leftarrow 1$	—	1	—	—	—	INH	9B		2
STA opr STA opr STA opr,X STA opr,X STA ,X	Store Accumulator in Memory	$M \leftarrow (A)$	—	—	↑	↑	—	DIR EXT IX2 IX1 IX	B7 C7 D7 E7 F7	dd hh ll ee ff ff	4 5 6 5 4
STOP	Stop Oscillator and Enable \overline{IRQ} Pin		—	0	—	—	—	INH	8E		2
STX opr STX opr STX opr,X STX opr,X STX ,X	Store Index Register In Memory	$M \leftarrow (X)$	—	—	↑	↑	—	DIR EXT IX2 IX1 IX	BF CF DF EF FF	dd hh ll ee ff ff	4 5 6 5 4

Table 12-6. Instruction Set Summary (Continued)

Source Form	Operation	Description	Effect on CCR					Address Mode	Opcode	Operand	Cycles
			H	I	N	Z	C				
SUB # <i>opr</i> SUB <i>opr</i> SUB <i>opr</i> SUB <i>opr</i> ,X SUB <i>opr</i> ,X SUB ,X	Subtract Memory Byte from Accumulator	$A \leftarrow (A) - (M)$	—	—	↓	↓	↓	IMM DIR EXT IX2 IX1 IX	A0 B0 C0 D0 E0 F0	ii dd hh ll ee ff ff	2 3 4 5 4 3
SWI	Software Interrupt	PC ← (PC) + 1; Push (PCL) SP ← (SP) - 1; Push (PCH) SP ← (SP) - 1; Push (X) SP ← (SP) - 1; Push (A) SP ← (SP) - 1; Push (CCR) SP ← (SP) - 1; I ← 1 PCH ← Interrupt Vector High Byte PCL ← Interrupt Vector Low Byte	—	1	—	—	—	INH	83		10
TAX	Transfer Accumulator to Index Register	$X \leftarrow (A)$	—	—	—	—	—	INH	97		2
TST <i>opr</i> TSTA TSTX TST <i>opr</i> ,X TST ,X	Test Memory Byte for Negative or Zero	$(M) - \$00$	—	—	↓	↓	—	DIR INH INH IX1 IX	3D 4D 5D 6D 7D	dd ff	4 3 3 5 4
TXA	Transfer Index Register to Accumulator	$A \leftarrow (X)$	—	—	—	—	—	INH	9F		2
WAIT	Stop CPU Clock and Enable Interrupts		—	0	—	—	—	INH	8F		2

A Accumulator
 C Carry/borrow flag
 CCR Condition code register
 dd Direct address of operand
 dd rr Direct address of operand and relative offset of branch instruction
 DIR Direct addressing mode
 ee ff High and low bytes of offset in indexed, 16-bit offset addressing
 EXT Extended addressing mode
 ff Offset byte in indexed, 8-bit offset addressing
 H Half-carry flag
 hh ll High and low bytes of operand address in extended addressing
 I Interrupt mask
 ii Immediate operand byte
 IMM Immediate addressing mode
 INH Inherent addressing mode
 IX Indexed, no offset addressing mode
 IX1 Indexed, 8-bit offset addressing mode
 IX2 Indexed, 16-bit offset addressing mode
 M Memory location
 N Negative flag
 n Any bit

opr Operand (one or two bytes)
 PC Program counter
 PCH Program counter high byte
 PCL Program counter low byte
 REL Relative addressing mode
rel Relative program counter offset byte
 rr Relative program counter offset byte
 SP Stack pointer
 X Index register
 Z Zero flag
 # Immediate value
 ^ Logical AND
 v Logical OR
 ⊕ Logical EXCLUSIVE OR
 () Contents of
 -() Negation (two's complement)
 ← Loaded with
 ? If
 : Concatenated with
 ↓ Set or cleared
 — Not affected

Table 12-7. Opcode Map

Bit Manipulation			Branch			Read-Modify-Write			Control			Register/Memory							
DIR	DIR	REL	DIR	INH	INH	IX1	IX	INH	INH	INH	IMM	DIR	EXT	IX2	IX1	IX			
MSB LSB	0	1	2	3	4	5	6	7	8	9	A	B	C	D	E	F	MSB LSB		
0	BRSET0 DIR 2	BSET0 DIR 2	BRA REL 3	NEG DIR 1	NEGA INH 1	NEG INH 2	NEG IX1 1	NEG IX 5	RTI INH 9		SUB IMM 2	SUB DIR 3	SUB EXT 4	SUB IX2 5	SUB IX1 4	SUB IX 3	0		
1	BRCLR0 DIR 2	BCLR0 DIR 2	BRN REL 3						RTS INH 6		CMP IMM 2	CMP DIR 3	CMP EXT 4	CMP IX2 5	CMP IX1 4	CMP IX 3	1		
2	BRSET1 DIR 2	BSET1 DIR 2	BHI REL 3	MUL ¹¹							SBC IMM 2	SBC DIR 3	SBC EXT 4	SBC IX2 5	SBC IX1 4	SBC IX 3	2		
3	BRCLR1 DIR 2	BCLR1 DIR 2	BLS REL 3	COM DIR 1	COMA INH 1	COMX INH 3	COM IX1 1	COM IX 5	SWI INH 10		CPX IMM 2	CPX DIR 3	CPX EXT 4	CPX IX2 5	CPX IX1 4	CPX IX 3	3		
4	BRSET2 DIR 2	BSET2 DIR 2	BCC REL 3	LSR DIR 1	LSRA INH 1	LSRX INH 3	LSR IX1 1	LSR IX 5			AND IMM 2	AND DIR 3	AND EXT 4	AND IX2 5	AND IX1 4	AND IX 3	4		
5	BRCLR2 DIR 2	BCLR2 DIR 2	BCH/BLO REL 3								BIT IMM 2	BIT DIR 3	BIT EXT 4	BIT IX2 5	BIT IX1 4	BIT IX 3	5		
6	BRSET3 DIR 2	BSET3 DIR 2	BNE REL 3	ROR DIR 1	RORA INH 1	RORX INH 3	ROR IX1 1	ROR IX 5			LDA IMM 2	LDA DIR 3	LDA EXT 4	LDA IX2 5	LDA IX1 4	LDA IX 3	6		
7	BRCLR3 DIR 2	BCLR3 DIR 2	BEQ REL 3	ASR DIR 1	ASRA INH 1	ASRX INH 3	ASR IX1 1	ASR IX 5	TAX INH 2			STA DIR 3	STA EXT 4	STA IX2 5	STA IX1 4	STA IX 3	7		
8	BRSET4 DIR 2	BSET4 DIR 2	BHCC REL 3	ASL/ASL DIR 1	ASLA/ASLA INH 1	ASLX/ASLX INH 3	ASL/SL IX1 1	ASL/SL IX 5		CLC INH 2	EOR IMM 2	EOR DIR 3	EOR EXT 4	EOR IX2 5	EOR IX1 4	EOR IX 3	8		
9	BRCLR4 DIR 2	BCLR4 DIR 2	BHCS REL 3	ROL DIR 1	ROLA INH 1	ROLX INH 3	ROL IX1 1	ROL IX 5		SEC INH 2	ADC IMM 2	ADC DIR 3	ADC EXT 4	ADC IX2 5	ADC IX1 4	ADC IX 3	9		
A	BRSET5 DIR 2	BSET5 DIR 2	BPL REL 3	DEC DIR 1	DECA INH 1	DECX INH 3	DEC IX1 1	DEC IX 5		CLI INH 2	ORA IMM 2	ORA DIR 3	ORA EXT 4	ORA IX2 5	ORA IX1 4	ORA IX 3	A		
B	BRCLR5 DIR 2	BCLR5 DIR 2	BMI REL 3							SEI INH 2	ADD IMM 2	ADD DIR 3	ADD EXT 4	ADD IX2 5	ADD IX1 4	ADD IX 3	B		
C	BRSET6 DIR 2	BSET6 DIR 2	BMC REL 3	INC DIR 1	INCA INH 1	INCX INH 3	INC IX1 1	INC IX 5		RSP INH 2		JMP DIR 3	JMP EXT 3	JMP IX2 2	JMP IX1 1	JMP IX 2	C		
D	BRCLR6 DIR 2	BCLR6 DIR 2	BMS REL 3	TST DIR 1	TSTA INH 1	TSTX INH 3	TST IX1 1	TST IX 4		NOP INH 2	BSR REL 2	JSR DIR 3	JSR EXT 3	JSR IX2 2	JSR IX1 1	JSR IX 2	D		
E	BRSET7 DIR 2	BSET7 DIR 2	BIL REL 3						STOP INH 2		LDX IMM 2	LDX DIR 3	LDX EXT 3	LDX IX2 2	LDX IX1 1	LDX IX 2	E		
F	BRCLR7 DIR 2	BCLR7 DIR 2	BIH REL 3	CLR DIR 1	CLRA INH 1	CLR INH 2	CLR IX1 1	CLR IX 5	WAIT INH 2	TXA INH 2		STX DIR 3	STX EXT 3	STX IX2 2	STX IX1 1	STX IX 2	F		

INH = Inherent
 IMM = Immediate
 DIR = Direct
 EXT = Extended

REL = Relative
 IX = Indexed, No Offset
 IX1 = Indexed, 8-Bit Offset
 IX2 = Indexed, 16-Bit Offset

MSB of Opcode in Hexadecimal

MSB of Opcode in Hexadecimal
 Number of Cycles
 Opcode Mnemonic
 Number of Bytes/Addressing Mode

SECTION 13 ELECTRICAL SPECIFICATIONS

13.1 MAXIMUM RATINGS

Table 13-1. Maximum Ratings

(Voltages referenced to V_{SS})

Rating	Symbol	Value	Unit
Supply Voltage	V_{DD}	-0.3 to +7.0	V
Input Voltage	V_{IN}	$V_{SS}-0.3$ to $V_{DD}+0.3$	V
EPROM Programming Voltage	V_{PP}	15	V
Current Drain Per Pin Excluding PB1, PB2, V_{DD} and V_{SS}	I	-25	mA
Operating Temperature Range (Standard) (Extended)	T_A	T_L to T_H 0 to +70 -40 to +85	°C
Storage Temperature Range	T_{STG}	-65 to +150	°C

This device contains circuitry to protect the inputs against damage due to high static voltages or electric fields; however, it is advised that normal precautions be taken to avoid application of any voltage higher than maximum-rated voltages to this high-impedance circuit. For proper operation, it is recommended that V_{IN} and V_{OUT} be constrained to the range $V_{SS} \leq (V_{IN} \text{ or } V_{OUT}) \leq V_{DD}$. Reliability of operation is enhanced if unused inputs are connected to an appropriate logic voltage level (e.g., either V_{SS} or V_{DD}).

13.2 THERMAL CHARACTERISTICS

Table 13-2. Thermal Characteristics

Characteristic	Symbol	Value	Unit
Thermal Resistance			
20-Pin PDIP	θ_{JA}	68	°C/W
20-Pin SOIC	θ_{JA}	91	°C/W

13.3 DC ELECTRICAL CHARACTERISTICS**Table 13-3. DC Electrical Characteristics**(V_{DD} = 4.2V to 5.5V, V_{SS} = 0 Vdc, T_A = 0°C to +70°C, unless otherwise noted)

Characteristic	Symbol	Min	Typ	Max	Unit
Output Voltage I _{Load} = 10.0 μA	V _{OL} V _{OH}	— V _{DD} -0.1	— —	0.1 —	V
Output High Voltage (I _{Load} = -0.8 mA) PA0-5, PB0	V _{OH}	V _{DD} -0.8	—	—	V
Output Low Voltage (I _{Load} = 1.6 mA) PA0-3, PB0 (I _{Load} = 8.0 mA) PA4-7 (I _{Load} = 25.0 mA) PB1, PB2 (note 8)	V _{OL}	— — —	— — —	0.4 0.4 0.5	V
Input High Voltage PA0-7, PB0-2, $\overline{\text{IRQ}}$, $\overline{\text{RESET}}$, OSC1	V _{IH}	0.7×V _{DD}	—	V _{DD}	V
Input Low Voltage PA0-7, PB0-2, $\overline{\text{IRQ}}$, $\overline{\text{RESET}}$, OSC1	V _{IL}	V _{SS}	—	0.2×V _{DD}	V
Supply Current (see Notes) Run (USB active) Run (USB suspended) Wait (USB active) Wait (USB suspended) Stop 25°C 0°C to +70°C	I _{DD}	—	10 2.3 8.8 1.0 200 300	12 4.0 10 2.0 250 400	mA mA mA mA μA μA
I/O Ports Hi-Z Leakage Current PA0-7, PB0-2 (without individual pulldown/up activated)	I _Z	—	—	±10	μA
Input Pulldown Current PA0-7, PB0 (with individual pulldown activated)	I _{IL}	50	100	200	μA
Input Current $\overline{\text{RESET}}$, $\overline{\text{IRQ}}$, OSC1	I _{in}	—	—	±1	μA
Capacitance Ports (as Input or Output) $\overline{\text{RESET}}$, $\overline{\text{IRQ}}$, OSC1, OSC2	C _{out} C _{in}	— —	— —	12 8	pF pF
Crystal/Ceramic Resonator Oscillator Mode Internal Resistor OSC1 to OSC2	R _{OSC}	1.0	2.0	3.0	MΩ
Pullup Resistor PB1, PB2	R _{PULL-UP}	25	100	200	KΩ
LVR Inhibit (see note 9)	V _{LVRI}	3.0	3.3	3.6	V
LVR Recover (see note 9)	V _{LVRR}	3.1	3.5	3.7	V

NOTES:

1. All values shown reflect average measurements.
2. Typical values at midpoint of voltage range, 25°C only.
3. Wait I_{DD} : Only timer system (MFT) active.
4. Run (Operating) I_{DD} , Wait I_{DD} : Measured using external square wave clock source to OSC1 ($f_{OSC} = 6.0$ MHz), all inputs 0.2 VDC from rail; no DC loads, less than 50pF on all outputs, $C_L = 20$ pF on OSC2.
5. Wait, Stop I_{DD} : All ports configured as inputs, $V_{IL} = 0.2$ VDC, $V_{IH} = V_{DD} - 0.2$ VDC.
6. Stop I_{DD} measured with OSC1 = V_{SS} .
7. Wait I_{DD} is affected linearly by the OSC2 capacitance.
8. $T_A = 0^\circ\text{C}$ to $+40^\circ\text{C}$.
9. These are preliminary specifications.

13.4 USB DC ELECTRICAL CHARACTERISTICS**Table 13-4. USB DC Electrical Characteristics**

($V_{DD} = 4.2\text{V}$ to 5.5V , $V_{SS} = 0$ Vdc, $T_A = 0^\circ\text{C}$ to $+70^\circ\text{C}$, unless otherwise noted)

Characteristic	Symbol	Conditions	Min	Typ	Max	Unit
Hi-Z State Data Line Leakage	I_{LO}	$0\text{V} < V_{in} < 3.3\text{V}$	-10		+10	μA
Differential Input Sensitivity	V_{DI}	$ (D+) - (D-) $, and Figure 12-1	0.2			V
Differential Common Mode Range	V_{CM}	Includes V_{DI} range	0.8		2.5	V
Single Ended Receiver Threshold	V_{SE}		0.8		2.0	V
Static Output Low	V_{OL}	R_L of 1.5k to 3.6V			0.3	V
Static Output High	V_{OH}	R_L of 15k to GND	2.8		3.6	V
3.3V External Reference Pin	$V_{3.3}$	$I_L = 200\mu\text{A}$	3.0	3.3	3.6	V

13.5 USB LOW SPEED SOURCE ELECTRICAL CHARACTERISTICS**Table 13-5. USB Low Speed Source Electrical Characteristics**

Parameter	Symbol	Conditions (Notes 1,2,3)	Min	Typ	Max	Unit
Transition time: Rise Time	T_R	Notes 4, 5, 8 $C_L=50\text{pF}$	75			ns
Fall Time	T_F	$C_L=350\text{pF}$ $C_L=50\text{pF}$ $C_L=350\text{pF}$	75		300 300	ns ns ns
Rise/Fall Time Matching	T_{RFM}	T_R/T_F	80		120	%
Output Signal Crossover Voltage	V_{CRS}		1.3		2.0	V
Low Speed Data Rate	T_{DRATE}	1.5Mbs $\pm 1.5\%$	1.4775 676.8	1.500 666.0	1.5225 656.8	Mbs ns
Source Differential Driver Jitter To Next Transition For Paired Transitions	T_{UDJ1} T_{UDJ2}	$C_L=350\text{pF}$ Notes 6,7 and Figure 12-2	-25 -10		25 10	ns ns
Receiver Data Jitter Tolerance To Next Transition For Paired Transitions	T_{DJR1} T_{DJR2}	$C_L=350\text{pF}$ Notes 7 and Figure 12-4	-75 -45		75 45	ns ns
Source EOP Width	$TEOPT$	Note 7 and Figure 12-3	1.25		1.50	μs
Differential to EOP Transition Skew	$TDEOP$	Note 7 and Figure 12-3	-40		100	ns
Receiver EOP Width Must Reject as EOP Must Accept	T_{EOPR1} T_{EOPR2}	Note 7 and Figure 12-3	330 675			ns ns

NOTES:

1. All voltages measured from local ground, unless otherwise specified.
2. All timings use a capacitive load of 50pF, unless otherwise specified.
3. Low speed timings have a 1.5k pull-up to 2.8V on the D- data line.
4. Measured from 10% to 90% of the data signal.
5. The rising and falling edges should be smooth transitional (monotonic).
6. Timing differences between the differential data signals.
7. Measured at crossover point of differential data signals.
8. Capacitive loading includes 50pF of tester capacitance.

13.6 CONTROL TIMING

Table 13-6. Control Timing

($V_{DD} = 4.2V$ to $5.5V$, $V_{SS} = 0$ Vdc, $T_A = 0^\circ C$ to $+70^\circ C$, unless otherwise noted)

Characteristic	Symbol	Min	Max	Units
Frequency of Operation Crystal Oscillator Option External Clock Source	f_{OSC} f_{OSC}	— DC	6 6	MHz MHz
Internal Operating Frequency Crystal Oscillator ($f_{OSC} \div 2$) External Clock ($f_{OSC} \div 2$)	f_{OP} f_{OP}	— DC	3 3	MHz MHz
Cycle Time ($1/f_{OP}$)	t_{CYC}	330	—	ns
RESET Pulse Width Low	t_{RL}	1.5	—	t_{CYC}
IRQ Interrupt Pulse Width Low (Edge-Triggered)	t_{ILIH}	0.5	—	t_{CYC}
IRQ Interrupt Pulse Period	t_{ILIL}	note 1	—	t_{CYC}
PA0 to PA3, Interrupt Pulse Width High (Edge-Triggered)	t_{IHIL}	0.5	—	t_{CYC}
PA0 to PA3, Interrupt Pulse Period	t_{IHIH}	note 1	—	t_{CYC}
OSC1 Pulse Width	t_{OH} , t_{OL}	—	—	ns
Output High to Low Transition Period on PB1 (note 3)	t_{SLOW}	—	—	ns

1. The minimum period t_{ILIL} or t_{IHIH} should not be less than the number of cycles it takes to execute the interrupt service routine plus $19 t_{CYC}$.
2. Effects of processing, temperature, and supply voltage (excluding tolerances of external R and C)
3. t_{slow} is a parameter dependent on f_{OSC} and loading. Typical value of t_{slow} is TENTATIVELY set at 170 ns with minimal value of 130ns and maximal value of 185ns under the SIMULATION conditions that f_{OSC} is 6.0 MHz and slow output transition feature is enabled. Capacitive loadings of 50pF on PB1-PB2, are assumed. Actual transition time will be specified to replace the TBDs when enough characterization has been done on various wafers from different lots. The values listed here represent data off simulation runs under the specified conditions. Under no circumstances should they be treated as the final specification.

13.7 EPROM PROGRAMMING SPECIFICATIONS

Table 13-7. EPROM Programming Electrical Characteristics

($V_{DD} = 4.2V$ to $5.5V$, $V_{SS} = 0$ Vdc, $T_A = 0^\circ C$ to $+70^\circ C$, unless otherwise noted)

Characteristic	Symbol	Min	Typ	Max	Unit
Programming Voltage IRQ/ V_{PP}	V_{PP}	10	12	15	V
Programming Current IRQ/ V_{PP}	I_{PP}	—	3	—	mA
Programming Time per byte	t_{EPGM}	1	4	—	ms

SECTION 14

MECHANICAL SPECIFICATIONS

This section provides the mechanical dimensions for the two available packages for MC68HC705JB2: the 20-Pin PDIP and 20-Pin SOIC.

14.1 20-PIN PLASTIC DUAL-IN-LINE PACKAGE (PDIP)

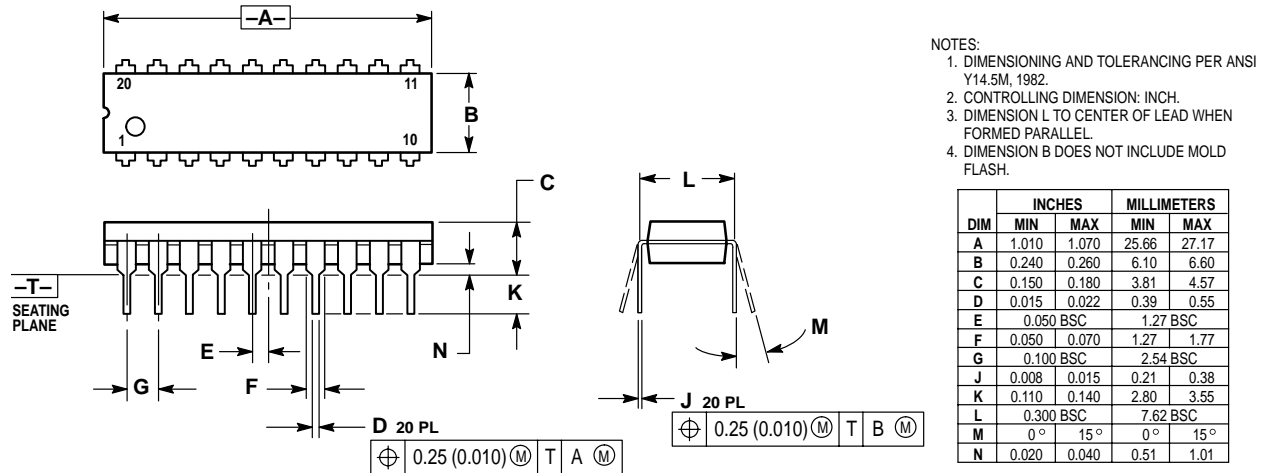


Figure 14-1. 20-Pin PDIP Mechanical Dimensions

14.2 20-PIN SURFACE-MOUNT SMALL OUTLINE PACKAGE (SOIC)

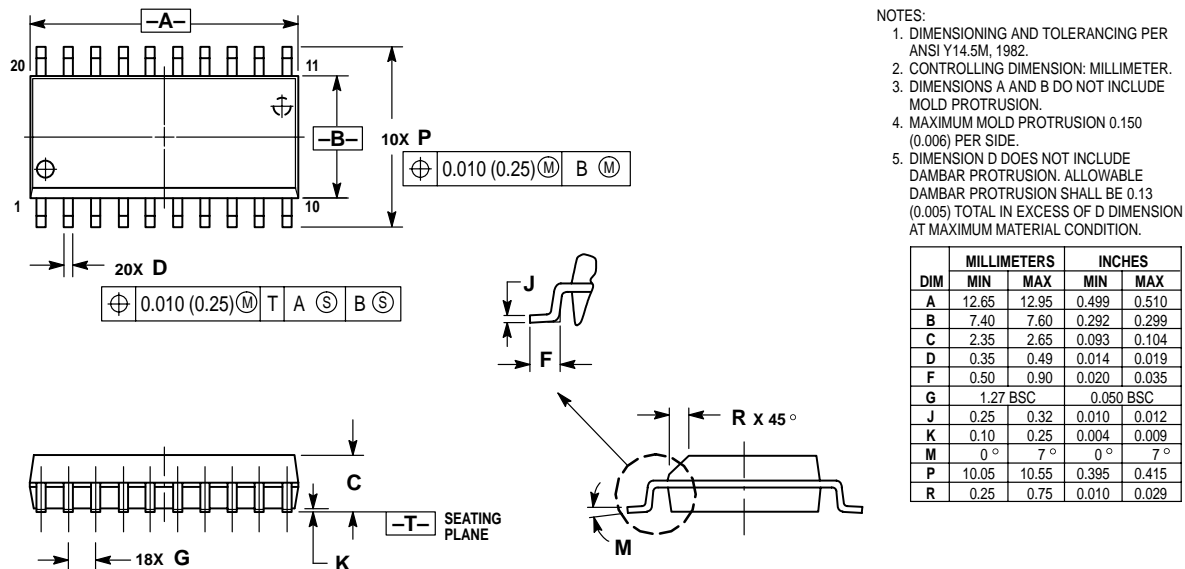



Figure 14-2. 20-Pin SOIC Mechanical Dimensions

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