

General Description

The MAX3170 is a three-driver/three-receiver multiprotocol transceiver that operates from a +3.3V single supply. The MAX3170, along with the MAX3171/MAX3173 and MAX3172/MAX3174, form a complete softwareselectable data terminal equipment (DTE) or data communications equipment (DCE) interface port that supports the V.28 (RS-232), V.11 (RS-449/V.36, EIA530, EIA530-A, X.21), and V.35 protocols. The MAX3170 transceiver carries the high-speed clock and data signals, while the MAX3171 or MAX3173 carries the control signals. The MAX3170 can be terminated by the MAX3172 or MAX3174 software-selectable resistor termination network or by a discrete termination network.

An internal charge pump and proprietary low-dropout transmitter output stage allow V.11-, V.28-, and V.35compliant operation from a +3.3V single supply. A nocable mode is entered when all mode pins (M0, M1, and M2) are pulled high or left unconnected. In nocable mode, supply current decreases to 1mA and all transmitter and receiver outputs are disabled (high impedance). Short-circuit limiting and thermal shutdown circuitry protect the drivers against excessive power dissipation.

Applications

Data Networking CSU and DSU

PCI Cards

Telecommunications

Data Routers

Features

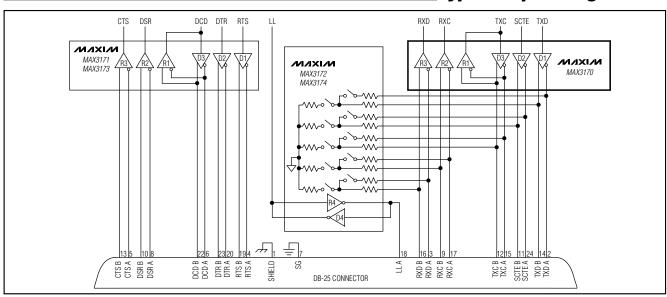
- ♦ Industry's First +3.3V Single-Supply Transceiver
- ♦ 3V/5V Logic-Compatible I/O
- ♦ Certified TBR-1 and TBR-2 Compliant (NET1 and **NET2)—Pending Completion**
- ♦ Supports V.28 (RS-232), V.11 (RS-449/V.36, EIA530, EIA530-A, X.21), and V.35 Protocols
- **♦ Software-Selectable DTE/DCE**
- **♦ Complete DTE/DCE Port with MAX3171/MAX3173** and MAX3172/MAX3174
- ◆ True Fail-Safe Receiver Operation
- ♦ Available in Small 28-Pin SSOP Package
- ♦ 10Mbps Operation (V.11/V.35)
- ♦ Requires Only Four Tiny Surface-Mount **Capacitors**
- **♦** All Transmitter Outputs Are Fault Protected to ±15V to Survive Cable Miswiring

Ordering Information

PART	TEMP. RANGE	PIN-PACKAGE
MAX3170CAI	0°C to +70°C	28 SSOP

Pin Configuration appears at end of data sheet.

Typical Operating Circuit



MIXIM

Maxim Integrated Products 1

ABSOLUTE MAXIMUM RATINGS

(All voltages referenced to GND unless	otherwise noted.)	Transmitter Output
Supply Voltages		T_OUT
V _{CC}	0.3V to +4V	Short-Circuit Dur
V+ (Note 1)	0.3V to +7V	Receiver Inputs
V- (Note 1)	+0.3V to -7V	R_IN
V+ to V- (Note 1)	13V	Continuous Power
Logic Input Voltages		28-Pin SSOP (de
M0, M1, M2, DCE/DTE, T_IN	0.3V to +6V	Operating Tempera
Logic Output Voltages		MAX3170CAI
R_OUT	0.3V to (V _{CC} + 0.3V)	Storage Temperatu
Short-Circuit Duration	Continuous	Lead Temperature

Transmitter Outputs
T_OUT15V to +15V
Short-Circuit Duration60s
Receiver Inputs
R_IN15V to +15V
Continuous Power Dissipation ($T_A = +70^{\circ}C$)
28-Pin SSOP (derate 11.1mW/°C above +70°C)889mW
Operating Temperature Range
MAX3170CAI0°C to +70°C
Storage Temperature Range65°C to +150°C
Lead Temperature (soldering, 10s) +300°C

Note 1: V+ and V- can have maximum magnitudes of 7V, but their absolute difference cannot exceed 13V.

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

 $(V_{CC}=+3.3V\pm5\%, C1=C2=1\mu F, C3=C4=3.3\mu F, T_A=T_{MIN} \text{ to } T_{MAX}, \text{ unless otherwise noted. Typical values are at } T_A=+25^{\circ}C$ and $V_{CC}=+3.3V$.)

PARAMETER	SYMBOL	CONDITIONS		TYP	MAX	UNITS			
DC CHARACTERISTICS									
		V.11 mode	190	250					
		V.11 mode with no load		3	7]			
Supply Current (DCE Mode)		V.35 mode	160	210					
(Digital Inputs = GND or V _{CC})	Icc	V.35 mode with no load		20	40	mA			
(All Outputs Static)		V.28 mode		10	20				
		V.28 mode with no load		4	7				
		No-cable mode		0.8	2				
		V.11 mode, full load		410					
Internal Power Dissipation (DCE Mode)	P_{D}	V.35 mode, full load		510		mW			
·		V.28 mode, full load		15		7			
	V+	V.11 mode	4						
V+ Output Voltage (DCE Mode)		V.35 mode	4.25] ,			
(Full Load)		V.28 mode	5.55						
		No-cable mode 5							
		V.11 mode			-4.1				
V- Output Voltage (DCE Mode)		V.35 mode			-3.7	V			
(Full Load)	V-	V.28 mode		-5.45					
		No-cable mode		-4.25					
Charge-Pump Enable Time		Delay until V+ and V- specifications met		1		ms			
LOGIC INPUTS (M0, M1, M2, DC	E/DTE, T_IN)							
Input High Voltage	VIH		2.0			V			
Input Low Voltage	VIL				0.8	V			
_	I _{IN}	T_IN			±1				
Logic Input Current	lіН	M0, M1, M2, DCE/DTE = V _{CC}			±1	μΑ			
	lıL	M0, M1, M2, DCE/DTE = GND	30	50	100				

ELECTRICAL CHARACTERISTICS (continued)

 $(V_{CC} = +3.3V \pm 5\%, C1 = C2 = 1\mu F, C3 = C4 = 3.3\mu F, T_A = T_{MIN}$ to T_{MAX} , unless otherwise noted. Typical values are at $T_A = +25^{\circ}C$ and $V_{CC} = +3.3V$.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
LOGIC OUTPUTS (R_OUT)			•			
Output High Voltage	VoH	ISOURCE = 1.0mA	V _{CC} - 1.	.0		V
Output Low Voltage	V _{OL}	I _{SINK} = 1.6mA			0.4	V
Rise or Fall Time	t _r , t _f	10% to 90%		15		ns
Output Leakage Current		R_OUT = GND	30	50	100	
(Receiver Output Tristated)		R_OUT = V _{CC}			±1	μΑ
TRANSMITTER OUTPUTS						
Output Leakage Current	IZ	-0.25V < V _{OUT} < +0.25V power-off or no-cable mode			±100	μΑ
Data Bata		V.11/ V.35 mode		10		Mbps
Data Rate		V.28 mode		240		kbps
RECEIVER INPUTS						
Receiver Input Resistance	R _{IN}	-10V < $V_{A,B}$ < +10V, V_{A} or V_{B} grounded (V.11/V.35/no-cable mode)	20	40		kΩ
		-15V < V _A < +15V (V.28 mode)	3	5	7	
Data Rate		V.11/ V.35 mode		10		Mbps
Data nate		V.28 mode		240		kbps
V.11 TRANSMITTER						
Unloaded Differential Output Voltage	V _{ODO}	R = 1.95k Ω , Figure 1	4.0		6.0	V
Loaded Differential Output Voltage	V _{ODL}	$R = 50\Omega$, Figure 1	2.0 0.5 x V _C	DDO		V
Change in Magnitude of Output	ΔV _{OD}	$R = 50\Omega$, Figure 1			0.2	V
Common-Mode Output Voltage	Voc	$R = 50\Omega$, Figure 1			3.0	V
Change in Magnitude of Output Common-Mode Voltage	ΔV _{OC}	R = 50Ω , Figure 1			0.2	V
Short-Circuit Current	I _{SC}	V _{OUT} = GND		60	150	mA
Rise or Fall Time	t _r , t _f	10% to 90%, Figure 2		10	25	ns
Transmitter Input to Output	tphl, tplh	Figure 2		50	80	ns
Data Skew	ItphL - tpLHI	Figure 2		2	10	ns
Output-to-Output Skew	tskew	Figure 2		2		ns
Channel-to-Channel Output				2		ns
V.11 RECEIVER			•			
Differential Threshold Voltage	V _{TH}	-7V < V _{CM} < +7V	-200	-100	-25	mV
Input Hysteresis	ΔV _{TH}	-7V < V _{CM} < +7V		15		mV
Receiver Input to Output	t _{PHL} , t _{PLH}	V _{CM} = 0, Figure 2		60	120	ns
Data Skew	ItphL - tpLHI	V _{CM} = 0, Figure 2		5	16	ns



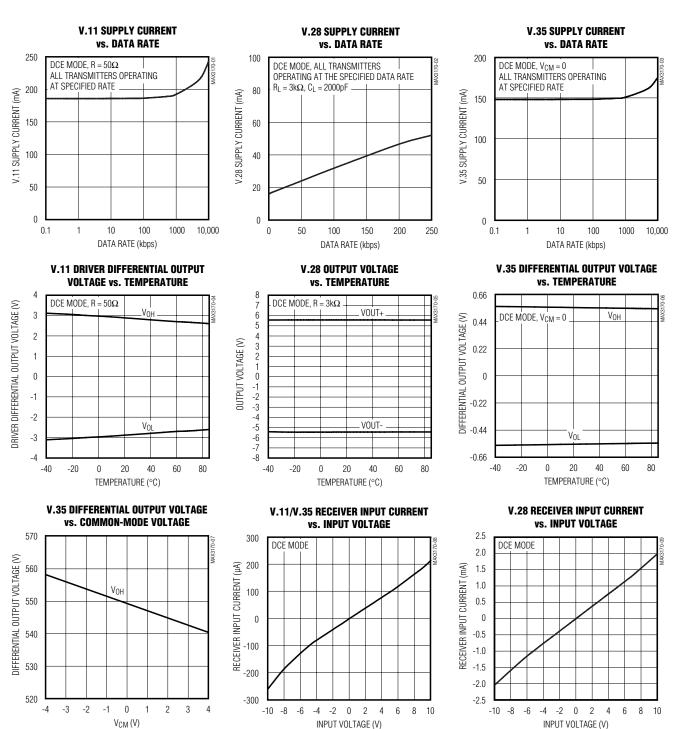
ELECTRICAL CHARACTERISTICS (continued)

 $(V_{CC}=+3.3V\pm5\%, C1=C2=1\mu F, C3=C4=3.3\mu F, T_A=T_{MIN}$ to T_{MAX} , unless otherwise noted. Typical values are at $T_A=+25^{\circ}C$ and $V_{CC}=+3.3V$.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
V.35 TRANSMITTER			•			
Differential Output Voltage		-4V < V _{CM} < +4V, Figure 3	0.44	0.55	0.66	V
Output High Current	Іон	V _A , _B = 0	9	11	13	mA
Output Low Current	loL	V _A , _B = 0	-13	-11	-9	mA
Rise or Fall Time	t _r , t _f	10% to 90%, Figure 3		10		ns
Transmitter Input to Output	tphL, tpLH	Figure 3		50	80	ns
Data Skew	ItphL - tpLHI	Figure 3		5	10	ns
Output-to-Output Skew		Figure 3		2		ns
Channel-to-Channel Output Skew				2		ns
V.35 RECEIVER	•					
Differential Input Voltage	V _{TH}	-4V < V _{CM} < +4V, Figure 3	-200	-100	-25	mV
Input Hysteresis	ΔV_{TH}	-4V < V _{CM} < +4V, Figure 3		15		mV
Receiver Input to Output	tphl, tplh	$V_{CM} = 0$		70	120	ns
Data Skew	ItphL - tpLHI	$V_{CM} = 0$		5	16	ns
V.28 TRANSMITTER						
Output Valtage Cuing	\/-	All transmitters loaded with $R_L = 3k\Omega$	±5.0	±5.4		V
Output Voltage Swing	Vo	No load			±6.5	V
Short-Circuit Current	Isc			±25	±60	mA
Output Class Bata	0.0	$R_L = 3k\Omega$, $C_L = 2500pF$, measured from +3V to -3V or -3V to +3V, Figure 4	4		30	\//
Output Slew Rate	SR	$R_L = 7k\Omega$, $C_L = 150pF$, measured from +3V to -3V or -3V to +3V, Figure 4	6		30	V/μs
Transmitter Input to Output	tphl, tplh	Figure 4		1		μs
Data Skew	ItphL - tpLHI	Figure 4		100		ns
V.28 RECEIVER						
Input Threshold Low	VIL	Figure 5	0.8	1.1		V
Input Threshold High	VIH	Figure 5		1.6	2.0	V
Input Hysteresis	V _H YS			0.5		V
Data Skew	ItphL - tpLHI	Figure 5		100		ns

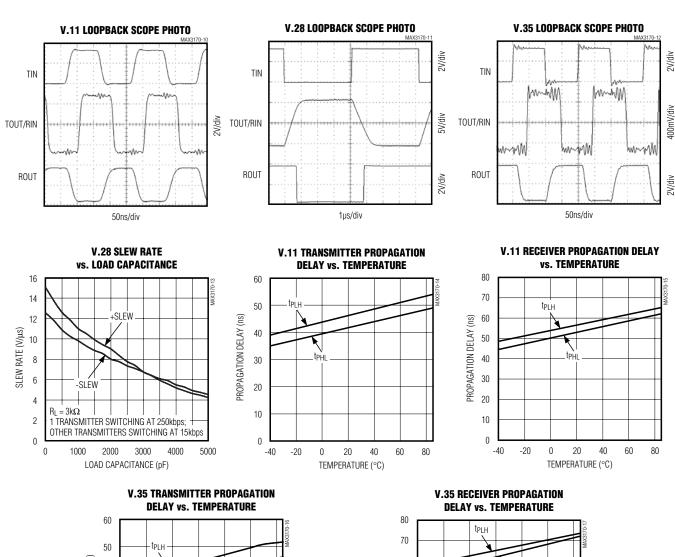
Typical Operating Characteristics

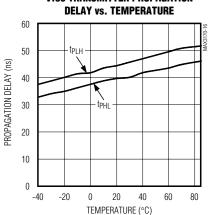
 $(V_{CC} = +3.3V, C1 = C2 = 1.0\mu F, C3 = C4 = 3.3\mu F, T_A = +25^{\circ}C$, unless otherwise noted.)

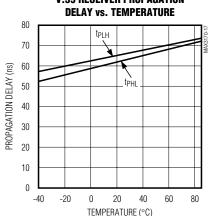


_Typical Operating Characteristics (continued)

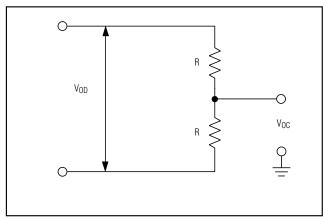
 $(V_{CC} = +3.3V, C1 = C2 = 1.0\mu F, C3 = C4 = 3.3\mu F, T_A = +25^{\circ}C, unless otherwise noted.)$







Test Circuits



100ρF 100ρF 100pF

Figure 1. V.11 DC Test Circuit

Figure 2. V.11 AC Test Circuit

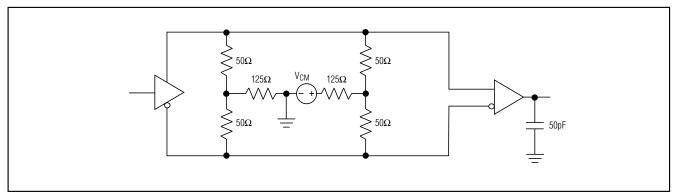


Figure 3. V.35 Transmitter/Receiver Test Circuit

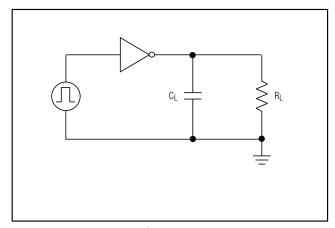


Figure 4. V.28 Driver Test Circuit

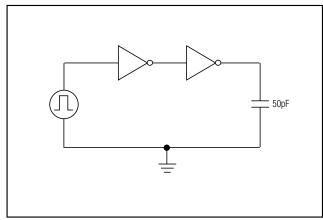


Figure 5. V.28 Receiver Test Circuit

Pin Description

PIN	NAME	FUNCTION
1	V+	Positive Supply Generated by the Charge Pump. Bypass V+ to ground with a 3.3μF ceramic capacitor.
2	C2+	Positive Terminal of the Inverting Charge-Pump Capacitor. Connect C2+ to C2- with a $1\mu\text{F}$ ceramic capacitor.
3	C2-	Negative Terminal of the Inverting Charge-Pump Capacitor. Connect C2+ to C2- with a $1\mu\text{F}$ ceramic capacitor.
4	V-	Negative Supply Generated by the Charge Pump. Bypass V- to ground with a $3.3\mu F$ ceramic capacitor.
5, 6, 7	T_IN	Transmitter CMOS Inputs (T1IN, T2IN, T3IN)
8, 9, 10	R_OUT	Receiver CMOS Outputs (R1OUT, R2OUT, R3OUT)
11, 12, 13	M_	Mode Select Pins (M0, M1, M2). Internally pulled up to V _{CC} . See Table 1 for detailed information.
14	DCE/DTE	DCE/ $\overline{\text{DTE}}$ Mode Select Pin. Logic level high selects DCE interface; logic level low selects DTE interface. Internally pulled up to V _{CC} .
15, 18	R_INB	Noninverting Receiver Inputs (R3INB, R2INB)
16, 17	R_INA	Inverting Receiver Inputs (R3INA, R2INA)
19	T3OUTB/R1INB	Noninverting Transmitter Output/Noninverting Receiver Input
20	T3OUTA/R1INA	Inverting Transmitter Output/Inverting Receiver Input
21, 23	T_OUTB	Noninverting Transmitter Outputs (T2OUTB, T1OUTB)
22, 24	T_OUTA	Inverting Transmitter Outputs (T2OUTA, T1OUTA)
25	C1-	Negative Terminal of the Voltage-Doubler Charge-Pump Capacitor. Connect C1+ to C1- with a $1\mu F$ ceramic capacitor.
26	GND	Ground
27	Vcc	+3.3V Supply Voltage (±5%). Bypass V _{CC} to ground with a 3.3μF capacitor.
28	C1+	Positive Terminal of the Voltage-Doubler Charge-Pump Capacitor. Connect C1+ to C1- with a $1\mu\text{F}$ ceramic capacitor.

Detailed Description

The MAX3170 is a three-driver/three-receiver multiprotocol transceiver that operates from a +3.3V single supply. The MAX3170, along with the MAX3171/MAX3173 and MAX3172/MAX3174, form a complete software-selectable DTE or DCE interface port that supports the V.28 (RS-232), V.11 (RS-449/V.36, EIA530, EIA530-A, and X.21), and V.35 protocols. The MAX3170 transceiver carries the high-speed clock and data signals, while the MAX3171 or MAX3173 carries the control signals.

The MAX3170 can be terminated by the MAX3172 or MAX3174 software-selectable resistor termination network or by a discrete termination network.

The MAX3170 features a 1mA no-cable mode, true fail-safe operation, and thermal shutdown circuitry. Thermal shutdown protects the drivers against excessive power dissipation. When activated, the thermal shutdown circuitry places the driver outputs into a high-impedance state.

Table 1. Mode Selection

PROTOCOL		LOGIC INPUTS			TRANSMITTERS			RECEIVERS		
	M2	M1	МО	DCE/DTE	T1	T2	T3	R1	R2	R3
V.11	0	0	0	0	V.11	V.11	Z	V.11	V.11	V.11
RS-530A	0	0	1	0	V.11	V.11	Z	V.11	V.11	V.11
RS-530	0	1	0	0	V.11	V.11	Z	V.11	V.11	V.11
X.21	0	1	1	0	V.11	V.11	Z	V.11	V.11	V.11
V.35	1	0	0	0	V.35	V.35	Z	V.35	V.35	V.35
RS-449/V.36	1	0	1	0	V.11	V.11	Z	V.11	V.11	V.11
V.28/RS-232	1	1	0	0	V.28	V.28	Z	V.28	V.28	V.28
No cable	1	1	1	0	Z	Z	Z	Z	Z	Z
V.11	0	0	0	1	V.11	V.11	V.11	Z	V.11	V.11
RS-530A	0	0	1	1	V.11	V.11	V.11	Z	V.11	V.11
RS-530	0	1	0	1	V.11	V.11	V.11	Z	V.11	V.11
X.21	0	1	1	1	V.11	V.11	V.11	Z	V.11	V.11
V.35	1	0	0	1	V.35	V.35	V.35	Z	V.35	V.35
RS-449/V.36	1	0	1	1	V.11	V.11	V.11	Z	V.11	V.11
V.28/RS-232	1	1	0	1	V.28	V.28	V.28	Z	V.28	V.28
No cable	1	1	1	1	Z	Z	Z	Z	Z	Z

Z = High impedance

Mode Selection

The state of the mode select pins M0, M1, and M2 determines which serial interface protocol is selected (Table 1). The state of the DCE/ \overline{DTE} input determines whether the transceiver will be configured as a DTE or a DCE serial port. When the DCE/ \overline{DTE} input is logic HIGH, driver T3 is activated and receiver R1 is disabled. When the DCE/ \overline{DTE} input is logic LOW, driver T3 is disabled and receiver R1 is activated. M0, M1, M2, and DCE/ \overline{DTE} are internally pulled up to V_{CC} to ensure a logic HIGH if left unconnected.

The MAX3170's mode can be selected through software control of the M0, M1, M2, and DCE/DTE inputs. Alternatively, the mode can be selected by shorting the appropriate combination of mode control inputs to GND. The inputs left floating will be internally pulled up to VCC (logic HIGH). If the M0, M1, and M2 mode inputs are all unconnected, the MAX3170 will enter nocable mode and the supply current will drop to 1mA.

No-Cable Mode

The MAX3170 will enter no-cable mode when the mode select pins are left unconnected or tied high (M0 = M1 = M2 = 1). In this mode, the multiprotocol drivers and receivers are disabled and the supply current drops to

1mA. The receiver outputs enter a high-impedance state in no-cable mode, which allows these output lines to be shared with other receivers (the receiver outputs have an internal pull-up resistor to pull the outputs HIGH if not driven). Also, in no-cable mode, the transmitter outputs enter a high-impedance state so that these output lines can be shared with other devices.

Dual Charge-Pump Voltage Converter

The MAX3170's internal power supply consists of a regulated dual charge pump that provides positive and negative output voltages from a +3.3V supply. The charge pump operates in discontinuous mode: if the output voltage is less than the regulated voltage, the charge pump is enabled; if the output voltage exceeds the regulated voltage, the charge pump is disabled. Each charge pump requires a flying capacitor (C1, C2) and a reservoir capacitor (C3, C4) to generate the V+ and V- supplies. See Figure 6 for charge-pump connections.

Fail-Safe Receivers

The MAX3170 guarantees a logic-high receiver output when the receiver inputs are shorted or open, or when they are connected to a terminated transmission line with all drivers disabled. This is done by setting the

receiver threshold between -25mV and -200mV in the V.11 and V.35 modes. If the differential receiver input voltage (B - A) is \geq -25mV, R_OUT is logic HIGH. If (B - A) is \leq -200mV, R_OUT is logic LOW. In the case of a terminated bus with all transmitters disabled, the receiver's differential input voltage is pulled to zero by the termination. With the receiver thresholds of the MAX3170, this results in a logic HIGH with a 25mV minimum noise margin.

Applications Information

Capacitor Selection

The capacitors used for the charge pumps, as well as the supply bypassing, should have a low equivalent series resistance (ESR) and low temperature coefficient. Multilayer ceramic capacitors with an X7R dielectric offer the best combination of performance, size, and cost. The flying capacitors (C1, C2) should have a value of $1\mu F$, while the reservoir capacitors (C3, C4) and bypass capacitor (C5) should have a minimum value of $3.3\mu F$ (Figure 6). To reduce the ripple present on the transmitter outputs, capacitors C3, C4, and C5 can be increased. The values of C1 and C2 should not be increased.

Cable Termination

The MAX3172/MAX3174 software-selectable resistor network is designed to be used with the MAX3170. The MAX3172/MAX3174 multiprotocol termination network provides V.11- and V.35-compliant termination, while V.28 receiver termination is internal to the MAX3170. These cable termination networks provide compatibility with V.11, V.28, and V.35 protocols. Using the MAX3172/MAX3174 termination network provides the advantage of not having to build expensive termination networks out of resistors and relays, manually changing termination modules, or building termination networks into custom cables.

Cable-Selectable Mode

A cable-selectable multiprotocol interface is shown in Figure 7. The mode control lines M0, M1, and DCE/DTE are wired to the DB-25 connector. To select the serial interface mode, the appropriate combination of M0, M1, M2, and DCE/DTE are grounded within the cable wiring. The control lines that are not grounded are pulled high by the internal pull-ups on the MAX3170. The serial interface protocol of the MAX3170 (and MAX3171/MAX3173 and MAX3172/MAX3174) is selected based on the cable that is connected to the DB-25 interface.

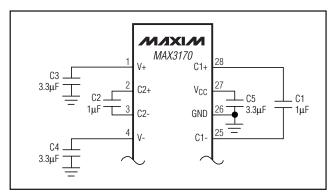


Figure 6. Charge-Pump Connections

V.11 (RS-422) Interface

As shown in Figure 8, the V.11 protocol is a fully balanced differential interface. The V.11 driver generates a minimum of $\pm 2V$ between nodes A and B when a 100Ω (min) resistance is presented at the load. The V.11 receiver is sensitive to $\pm 200 \text{mV}$ differential signals at the receiver inputs, A' and B'. The V.11 receiver rejects common-mode signals developed across the cable (referenced from C to C') of up to $\pm 7V$, allowing for error-free reception in noisy environments. The receiver inputs must comply with the impedance curve shown in Figure 9.

For high-speed data transmission, the V.11 specification recommends terminating the cable at the receiver with a 100Ω minimum resistor. This resistor, although not required, prevents reflections from corrupting transmitted data. In Figure 10, the MAX3172 or MAX3174 is used to terminate the V.11 receiver. Internal to the MAX3172/MAX3174, S1 is closed and S2 is open to present a 100Ω minimum differential resistance. The MAX3170's internal V.28 termination is disabled by opening S3.

V.35 Interface

Figure 11 shows a fully balanced differential standard V.35 interface. The generator and the load must both present a $100\Omega~\pm10\Omega$ differential impedance and a $150\Omega~\pm15\Omega$ common-mode impedance as shown by the resistive T-networks in Figure 11. The V.35 driver generates a current output ($\pm11\text{mA}$ typ) that develops an output voltage of $\pm550\text{mV}$ across the generator and load termination networks. The V.35 receiver is sensitive to $\pm200\text{mV}$ differential signals at the receiver inputs A' and B'. The V.35 receiver rejects common-mode signals developed across the cable (referenced from C to C') of up to $\pm4\text{V}$, allowing for error-free reception in noisy environments.

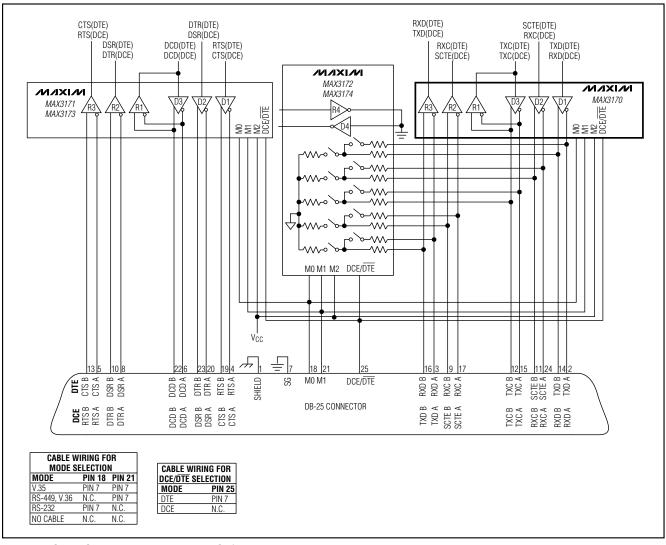


Figure 7. Cable-Selectable Multiprotocol DCE/DTE Port

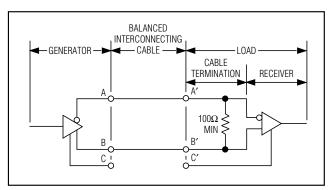
In Figure 12, the MAX3172 or MAX3174 is used to implement the resistive T-network that is needed to properly terminate the V.35 driver and receiver. Internal to the MAX3172/MAX3174, S1 and S2 are closed to connect the T-network resistors to the circuit. The V.28 termination resistor (internal to the MAX3170) is disabled by opening S3 to avoid interference with the T-network impedances.

V.28 Interface

The V.28 interface is an unbalanced single-ended interface (Figure 13). The V.28 driver generates a minimum

of ±5V across the load impedance between A' and C'. The V.28 receiver has a single-ended input and does not reject any common-mode differences between C and C'. The V.28 receiver has input trip points at ±3V. To aid in rejecting system noise, the MAX3170 V.28 receiver has a typical hysteresis of 0.5V.

Figure 14 shows that the MAX3172/MAX3174 termination network is disabled by opening S1 and S2. The MAX3170's internal $5k\Omega$ V.28 termination is enabled by closing S3.



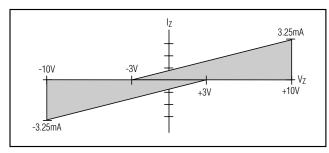


Figure 9. Receiver Input Impedance Curve

Figure 8. Typical V.11 Interface

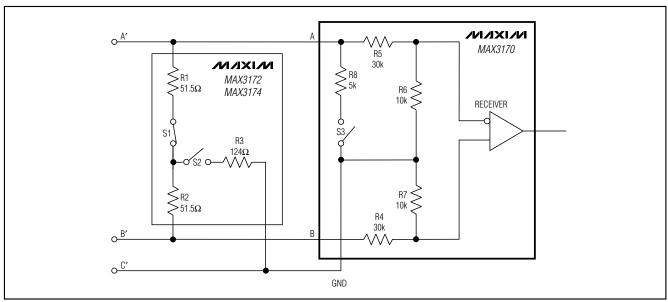


Figure 10. V.11 Termination and Internal Resistance Networks

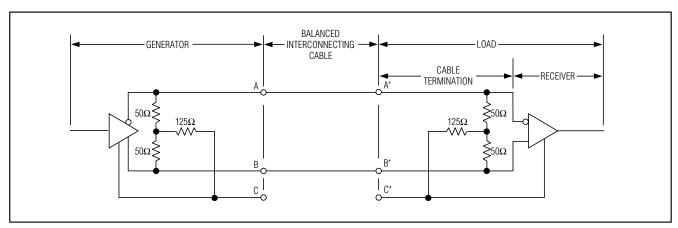


Figure 11. Typical V.35 Interface

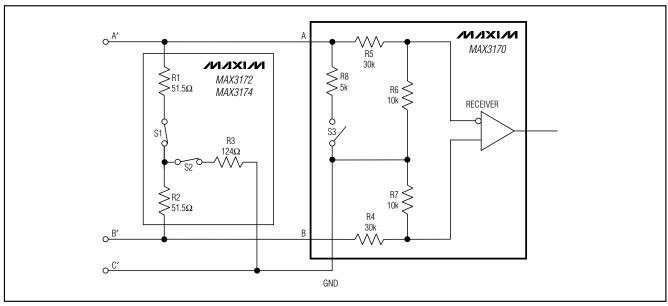


Figure 12. V.35 Termination and Internal Resistance Networks

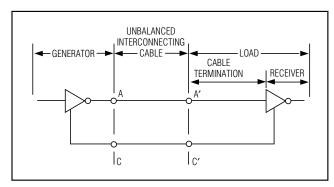


Figure 13. Typical V.28 Interface

DTE vs. DCE Operation

Figure 15 shows a DCE or DTE controller-selectable interface. The DCE/DTE pin (pin 14) switches the port's mode of operation. A logic HIGH selects DCE, which enables D3 on the MAX3170, D3 on the MAX3171/MAX3173, and D4 on the MAX3172/MAX3174. A logic LOW selects DTE, which enables R1 on the MAX3170, R1 on the MAX3171/MAX3173, and R4 on the MAX3172/MAX3174.

This application requires only one DB-25 connector, but separate cables for DCE or DTE signal routing. See Figure 15 for complete signal routing in DCE and DTE modes. For example, the MAX3170's D3 routes the TXC (DCE) signal to pins 12 and 15 in DCE mode, while in DTE mode, the MAX3170's R1 routes pins 12 and 15 to TXC (DTE).

Complete Multiprotocol X.21 Interface

A complete DTE-to-DCE interface operating in X.21 mode is shown in Figure 16. The MAX3170 is used to generate the clock and data signals, and the MAX3171/MAX3173 generate the control signals. The MAX3172/MAX3174 generate local loopback (LL), and are used to terminate the clock and data signals to support the V.11 protocol for cable termination. The control signals do not need external termination.

Compliance Testing

A European Standard EN 45001 test report is available for the MAX3170/MAX3171/MAX3173/MAX3172/MAX3174 chipset. A copy of the test report will be available from Maxim upon completion.

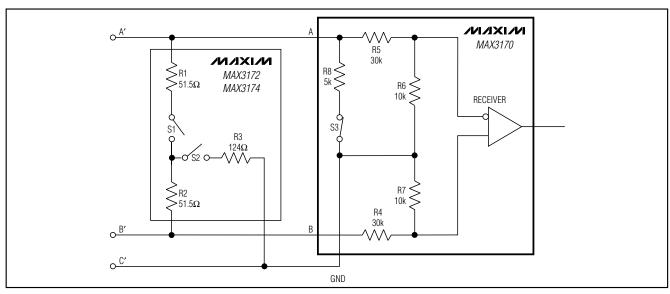


Figure 14. V.28 Termination and Internal Resistance Networks

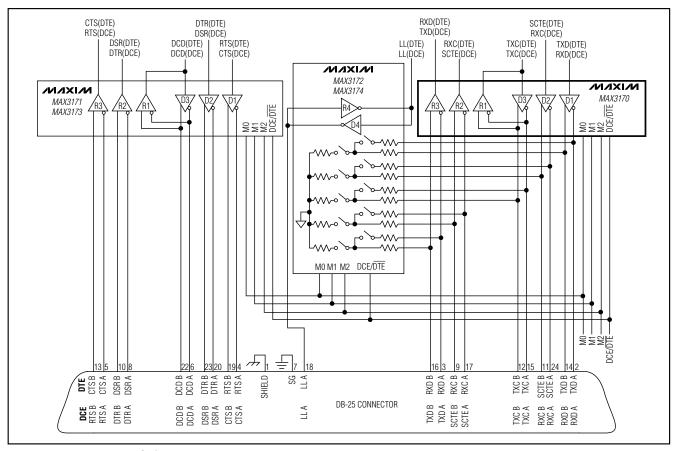


Figure 15. Multiprotocol DCE/DTE Port

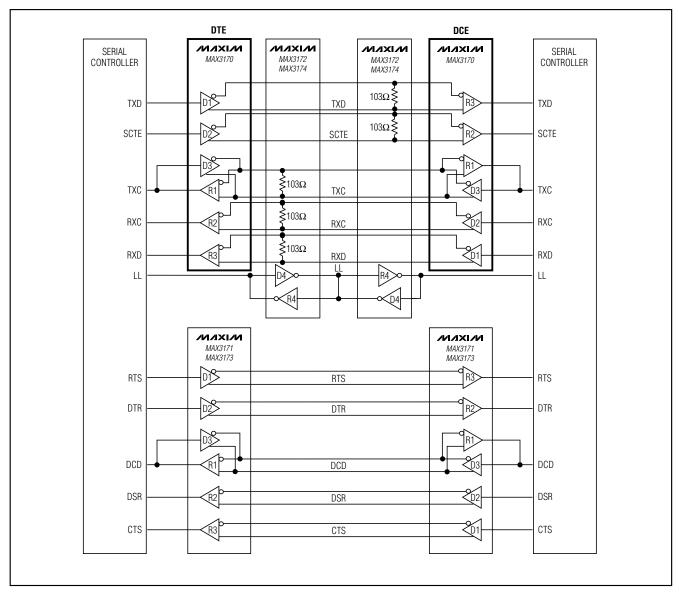
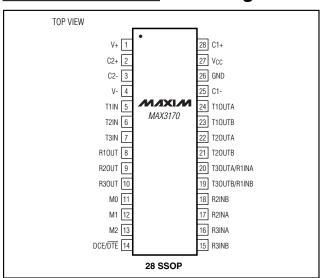


Figure 16. DCE-to-DTE X.21 Interface

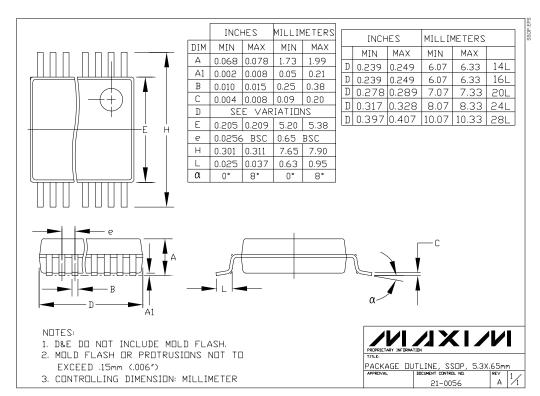
Pin Configuration

Chip Information



TRANSISTOR COUNT: 4058

Package Information



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