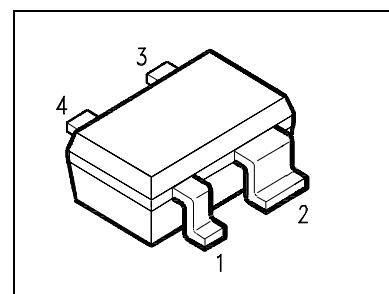


NPN Silicon RF Transistor

- For High Gain Low Noise Amplifiers
- For Oscillators up to 10 GHz
- Noise Figure $F = 1.05$ dB at 1.8 GHz
Outstanding $G_{ms} = 20$ dB at 1.8 GHz
- Transition Frequency $f_T = 25$ GHz
- Gold metalization for high reliability
- SIEGET®25-Line
Siemens Grounded Emitter Transistor-
25 GHz f_T -Line



ESD: Electrostatic discharge sensitive device,
observe handling precautions!

Type	Marking	Ordering Code (8-mm taped)	Pin Configuration				Package ¹⁾
			1	2	3	4	
BFP420	AMs	Q62702-F1591	B	E	C	E	SOT343

Maximum Ratings

Parameter	Symbol		Unit
Collector-emitter voltage	V_{CEO}	4.5	V
Collector-base voltage	V_{CBO}	15	V
Emitter-base voltage	V_{EBO}	1.5	V
Collector current	I_C	35	mA
Base current	I_B	3	mA
Total power dissipation, $T_S \leq 107^\circ\text{C}$ ²⁾³⁾	P_{tot}	160	mW
Junction temperature	T_j	150	°C
Ambient temperature range	T_A	-65...+150°C	°C
Storage temperature range	T_{stg}	-65...+150°C	°C

Thermal Resistance

Junction-soldering point ²⁾	$R_{th JS}$	270	K/W
----------------------------------------	-------------	-----	-----

1) For detailed information see chapter Package

2) T_S is measured on the emitter lead at the soldering point to the pcb.

3) P_{tot} due to Maximum Ratings.

At typical $T_S \leq 80^\circ\text{C}$: $P_{tot} = 250$ mW due to thermal characteristics.

Electrical Characteristicsat $T_A = 25^\circ\text{C}$, unless otherwise specified.

Parameter	Symbol	Value			Unit
		min.	typ.	max.	

DC Characteristics

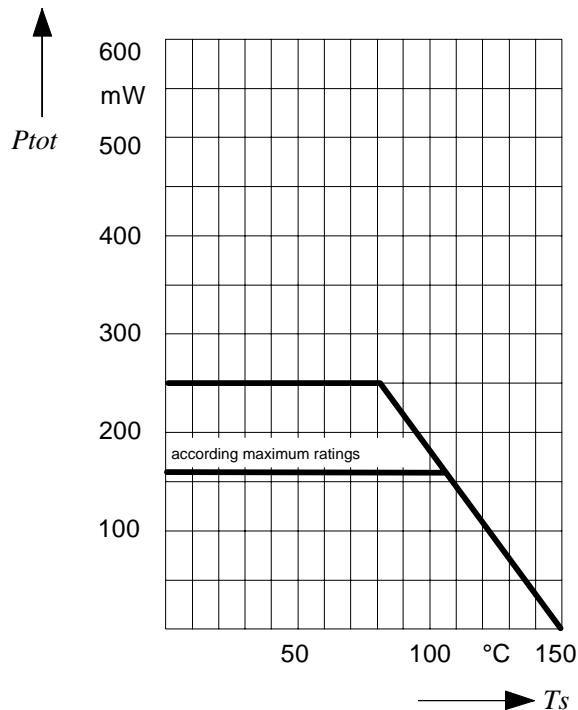
Collector-emitter breakdown voltage $I_C = 1 \text{ mA}$	$V_{(\text{BR})\text{CEO}}$	4.5	5	6.5	V
Collector-cutoff current $V_{CB} = 5 \text{ V}, I_E = 0$	I_{CBO}	-	-	200	nA
Emitter base cutoff current $V_{EB} = 1.5 \text{ V}, I_C = 0$	I_{EBO}	-	-	35	μA
DC current gain $I_C = 20 \text{ mA}, V_{CE} = 4 \text{ V}$	h_{FE}	50	80	150	

AC Characteristics

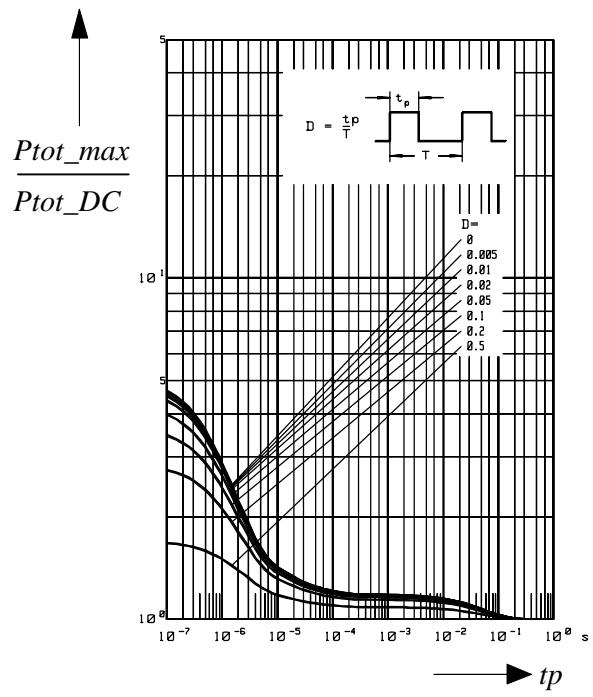
Transition frequency $I_C = 30 \text{ mA}, V_{CE} = 3 \text{ V}, f = 2 \text{ GHz}$	f_T	20	25	-	GHz
Collector-base capacitance $V_{CB} = 2 \text{ V}, V_{BE} = V_{be} = 0, f = 1 \text{ MHz}$	C_{cb}	-	0.15	0.24	pF
Collector-emitter capacitance $V_{CE} = 2 \text{ V}, V_{BE} = V_{be} = 0, f = 1 \text{ MHz}$	C_{ce}	-	0.41	-	pF
Emitter-base capacitance $V_{EB} = 0.5 \text{ V}, V_{CB} = V_{cb} = 0, f = 1 \text{ MHz}$	C_{eb}	-	0.55	-	pF
Noise figure $I_C = 5 \text{ mA}, V_{CE} = 2 \text{ V}, f = 1.8 \text{ GHz}, Z_S = Z_{\text{Sopt}}$	F	-	1.05	1.4	dB
Power gain $I_C = 20 \text{ mA}, V_{CE} = 2 \text{ V}, f = 1.8 \text{ GHz},$ $Z_S = Z_{\text{Sopt}}, Z_L = Z_{\text{Lopt}}$	$G_{ms}^{1)}$	-	20	-	dB
Insertion power gain $I_C = 20 \text{ mA}, V_{CE} = 2 \text{ V}, f = 1.8 \text{ GHz}, Z_S = Z_L = 50\Omega$	$ S_{21} ^2$	14	17	-	dB
Third order intercept point at output $I_C = 20 \text{ mA}, V_{CE} = 2 \text{ V}, f = 1.8 \text{ GHz},$ $Z_S = Z_{\text{Sopt}}, Z_L = Z_{\text{Lopt}}$	IP_3	-	22	-	dBm
1dB Compression point $I_C = 20 \text{ mA}, V_{CE} = 2 \text{ V}, f = 1.8 \text{ GHz},$ $Z_S = Z_{\text{Sopt}}, Z_L = Z_{\text{Lopt}}$	$P_{-1\text{dB}}$	-	12	-	dBm

1) $G_{ms} = \left| \frac{S_{21}}{S_{12}} \right|$

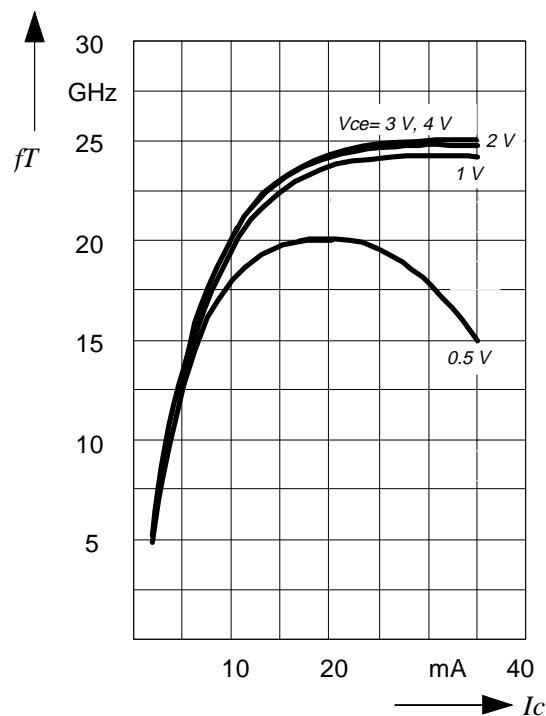
Total Power Dissipation
versus Soldering Point Temperature



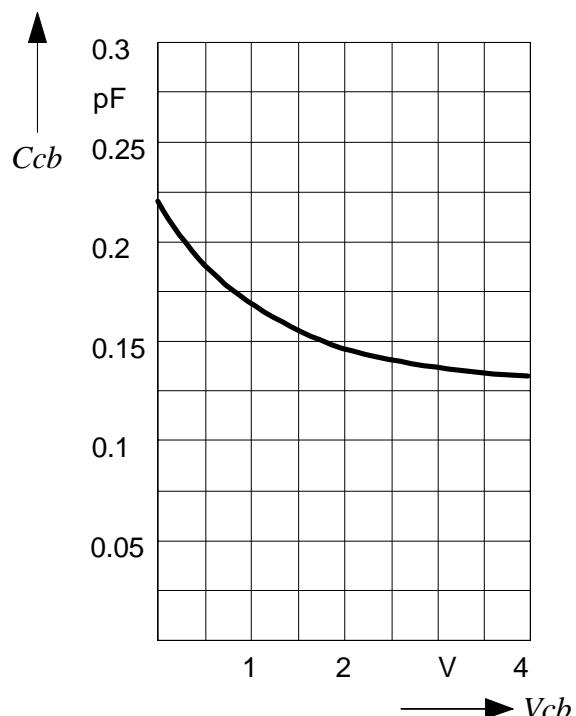
Permissible Pulse Power Dissipation
versus On-Time ($V_{CE0max} = 4.5\text{ V}$)



Transition Frequency
versus Collector Current
 $f = 2\text{ GHz}$



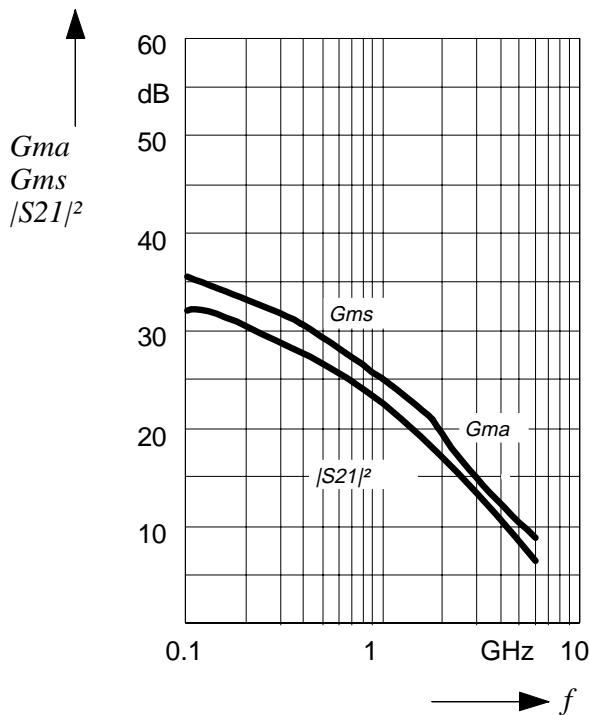
Collector-base Capacitance
versus Collector-base Voltage
 $V_{BE} = 0\text{ V}$, $f = 1\text{ MHz}$



Power Gain

versus Frequency

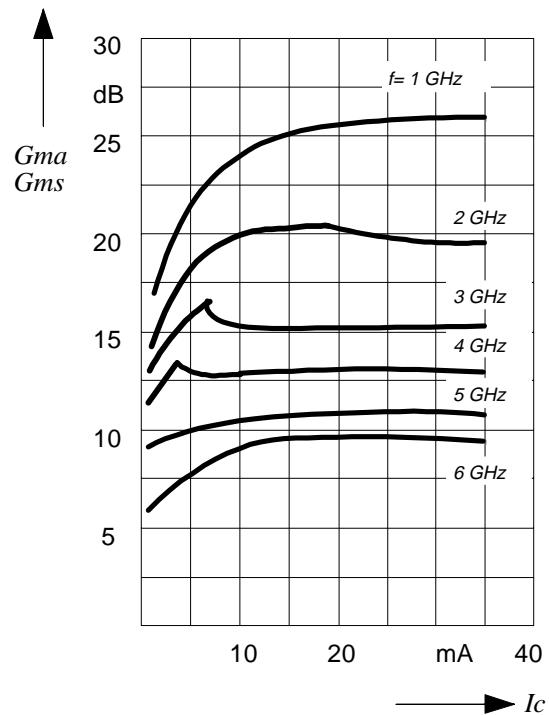
$V_{CE} = 2 \text{ V}$, $I_C = 20 \text{ mA}$



Power Gain

versus Collector Current

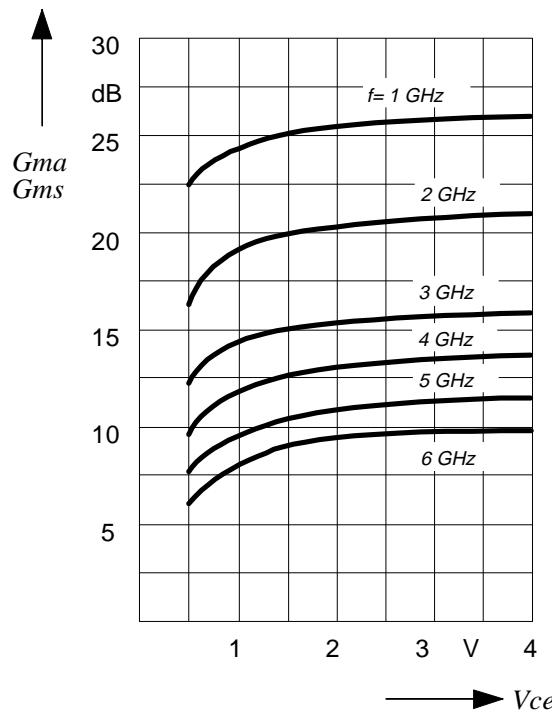
$V_{CE} = 2 \text{ V}$



Power Gain

versus Collector Voltage

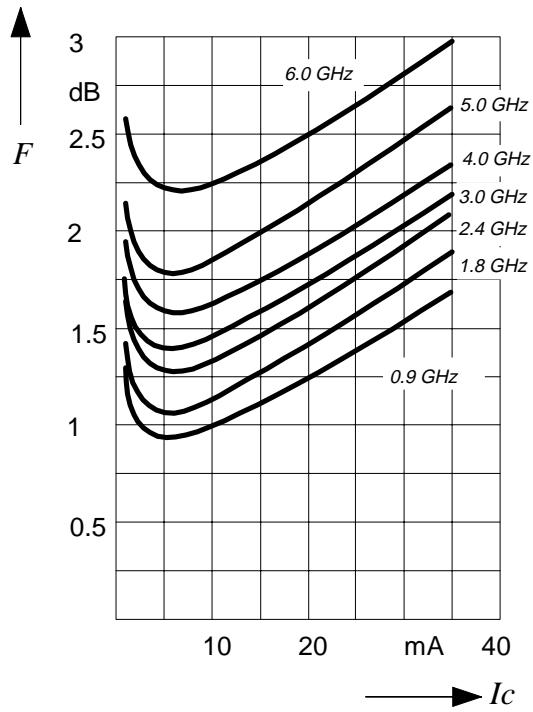
$I_C = 20 \text{ mA}$



Noise Figure

versus Collector Current

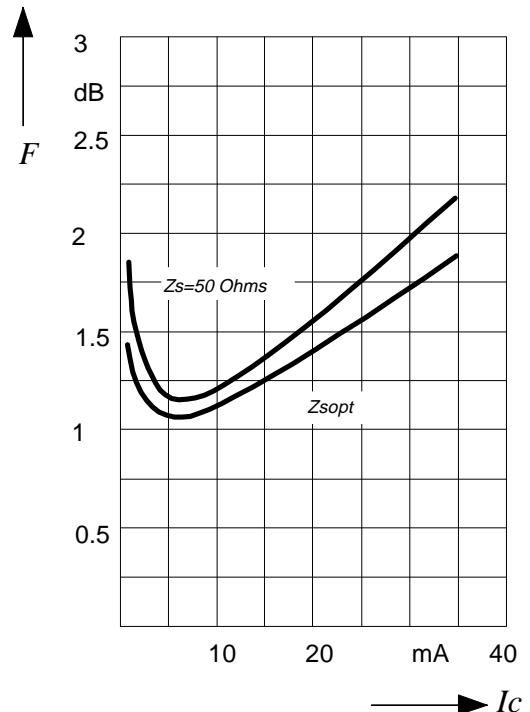
$$V_{CE} = 2 \text{ V}, Z_S = Z_{Sopt}$$



Noise Figure

versus Collector Current

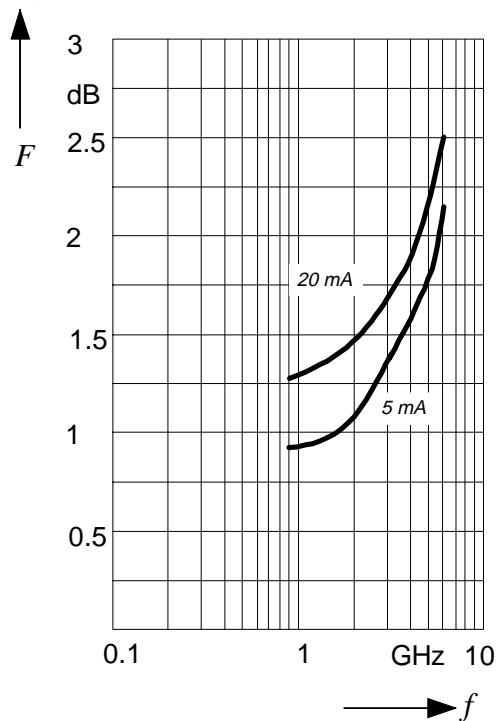
$$V_{CE} = 2 \text{ V}, f = 1.8 \text{ GHz}$$



Noise Figure versus Frequency

$$V_{CE} = 2 \text{ V}, I_C = 5 \text{ mA / 20 mA},$$

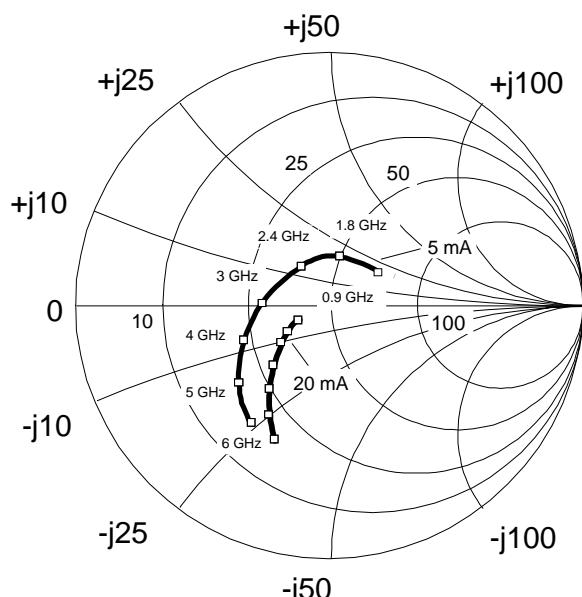
$$Z_S = Z_{Sopt}$$



Source Impedance for min.

Noise Figure versus Frequency

$$V_{CE} = 2 \text{ V}, I_C = 5 \text{ mA / 20 mA}$$



Common Emitter S-Parameters

<i>f</i>	S_{11}		S_{21}		S_{12}		S_{22}	
GHz	MAG	ANG	MAG	ANG	MAG	ANG	MAG	ANG
$V_{CE} = 2 \text{ V}, I_C = 20 \text{ mA}$								
0.01	0.452	-2.3	37.62	178.3	0.0011	94.4	0.956	-0.6
0.1	0.447	-25.1	36.30	164.7	0.0068	82.5	0.941	-12.4
0.5	0.386	-101.1	23.41	121.0	0.0262	61.7	0.632	-47.2
1.0	0.378	-146.2	13.99	96.0	0.0395	57.8	0.395	-63.9
2.0	0.405	173.5	7.18	70.8	0.0664	54.0	0.222	-87.3
3.0	0.446	149.4	4.77	52.6	0.0949	47.1	0.133	-111.3
4.0	0.501	130.0	3.52	36.8	0.1206	38.5	0.133	-158.5
6.0	0.599	104.8	2.27	8.2	0.1646	18.9	0.196	142.0
8.0	0.700	78.5	1.51	-20.8	0.1800	-2.4	0.289	99.3
9.0	0.758	67.6	1.25	-34.4	0.1820	-13.0	0.379	84.1
10.0	0.800	62.0	1.04	-43.5	0.1800	-19.3	0.465	76.6

 $V_{CE} = 2 \text{ V}, I_C = 5 \text{ mA}$

0.01	0.790	-1.0	15.14	179.2	0.0012	83.4	0.988	-0.7
0.1	0.786	-11.6	14.98	171.8	0.0092	84.1	0.982	-6.5
0.5	0.702	-55.7	12.86	140.1	0.0398	62.8	0.857	-29.8
1.0	0.589	-99.1	9.63	112.6	0.0603	46.5	0.647	-48.6
2.0	0.507	-156.0	5.60	79.4	0.0798	34.6	0.401	-70.3
3.0	0.511	168.5	3.84	57.1	0.0957	29.8	0.267	-84.2
4.0	0.549	142.0	2.87	38.5	0.1121	25.1	0.207	-110.5
5.0	0.604	123.9	2.26	22.1	0.1285	19.4	0.150	-137.3
6.0	0.633	110.0	1.86	6.7	0.1442	13.1	0.173	-169.8

Common Emitter Noise Parameters

<i>f</i>	F_{min} 1)	G_a 1)	Γ_{opt}		R_N	r_n	$F_{50\Omega}$ 2)	$ S_{21} ^2$ 2)
GHz	dB	dB	MAG	ANG	Ω	-	dB	dB
$V_{CE} = 2 \text{ V}, I_C = 5 \text{ mA}$								
0.9	0.90	20.5	0.28	41.0	8.7	0.17	1.02	20.3
1.8	1.05	15.2	0.20	82.0	6.7	0.13	1.11	15.8
2.4	1.25	13.0	0.20	124.0	5.5	0.11	1.32	13.5
3.0	1.38	12.1	0.22	-175.0	5.0	0.10	1.48	11.6
4.0	1.55	10.3	0.33	-157.0	5.5	0.11	1.83	9.1
5.0	1.75	8.6	0.45	-142.0	5.0	0.10	2.20	7.0
6.0	2.20	6.4	0.53	-123.0	15.0	0.30	3.30	5.3

1) Input matched for minimum noise figure, output for maximum gain

2) $Z_S = Z_L = 50\Omega$

For more and detailed S- and Noise-parameters please contact your local Siemens distributor or sales office to obtain a SIEMENS Application Notes CD-ROM or see Internet:

<http://www.siemens.de/Semiconductor/products/35/35.htm>

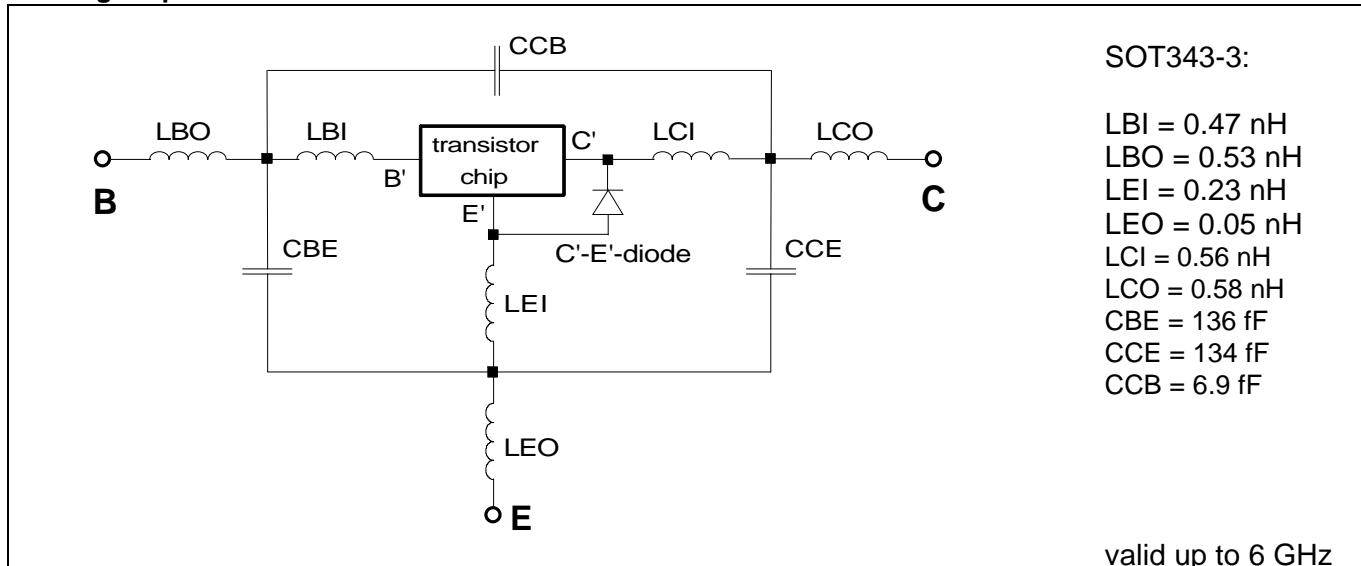
SPICE Parameters:**Transistor Chip Data T502 (Berkeley-SPICE 2G.6 Syntax):**

IS =	0.20045	fA	BF =	72.534	-	NF =	1.2432	-
VAF =	28.383	V	IKF =	0.48731	A	ISE =	19.049	fA
NE =	2.0518	-	BR =	7.8287	-	NR =	1.3325	-
VAR =	19.705	V	IKR =	0.69141	A	ISC =	0.019237	fA
NC =	1.1724	-	RB =	3.4849	OHM	IRB =	0.72983	mA
RBM =	8.5757	OHM	RE =	0.31111	OHM	RC =	0.10105	OHM
CJE =	1.8063	fF	VJE =	0.8051	V	MJE =	0.46576	-
TF =	6.7661	ps	XTF =	0.42199	-	VTF =	0.23794	V
ITF =	1.0	mA	PTF =	0	deg	CJC =	234.53	fF
VJC =	0.81969	V	MJC =	0.30232	-	XCJC =	0.3	-
TR =	2.3249	ns	CJS =	0	fF	VJS =	0.75	V
MJS =	0	-	XTB =	0	-	EG =	1.11	eV
XTI =	3.0	-	FC =	0.73234	-	Tnom =	300	K

C'-E'- Diode Data (Berkeley-SPICE 2G.6 Syntax):

IS =	3.5	fA	N =	1.02	-	RS =	10	Ω
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All parameters are ready to use, no scaling is necessary.

Package Equivalent Circuit:

The SOT343 package has two emitter leads. To avoid high complexity of the package equivalent circuit, both leads are combined in one electrical connection.

Extracted on behalf of SIEMENS Small Signal Semiconductors by:
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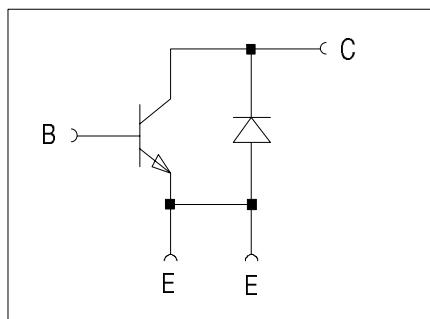
For more examples and ready to use parameters please contact your local Siemens distributor or sales office to obtain a SIEMENS Application Notes CD-ROM or see Internet:
<http://www.siemens.de/Semiconductor/products/35/35.htm>

For non-linear simulation:

- Use transistor chip parameters in Berkeley SPICE 2G.6 syntax for all simulators.
- If you need simulation of the reverse characteristics, add the diode with the C'-E'-diode data between collector and emitter.
- Simulation of the package is not necessary for frequencies < 100 MHz.
For higher frequencies add the wiring of the package equivalent circuit around the non-linear transistor and diode model.

Note:

- This transistor is constructed in a common emitter configuration. This feature causes an additional, reverse biased diode between emitter and collector, which does not effect normal operation.



Transistor Schematic Diagram

The common emitter configuration shows the following advantages:

- Higher gain because of lower emitter inductance.
- Power is dissipated via the grounded emitter leads, because the chip is mounted on the copper emitter leadframe.

Please note, that the broadest lead is the emitter lead.

- The AC-Characteristics are verified by random sampling.