

Speech Circuit with Line-Powered Loudspeaker Amplifier

Description

The electronic speech circuit U4050B is a linear integrated circuit for use in telephone sets. It replaces the hybrid transformer, side tone equivalent and ear protection rectifiers.

The circuit is line powered and contains all components necessary for amplification of signals and adaptation to the line. An integrated loudspeaker amplifier allows loudhearing operation.

Features

- Integrated amplifier for loudhearing operation
- Anticlipping for loudspeaker amplifier
- Supply voltages for all functional blocks of a subscriber set
- Adjustable dc characteristics
- Adjustable sending and receiving amplification
- Automatic line loss compensation
- Symmetrical output of earpiece amplifier
- Built in ear protection
- Symmetrical input of microphone amplifier
- Adjustable side tone suppression independent of sending and receiving amplification

- DTMF and MUTE inputs
- Anticlipping in transmit direction
- Squelch
- Integrated transistor for short circuiting the line voltage
- Power down
- Operation possible at line currents above 10 mA

Benefits

- Independent adjustment of transmit gain, receive gain and side tone suppression
- Low number of external components



Block Diagram / Application Circuit

With a squelch function, acoustical feedback during loudhearing can be reduced significantly. The generated

supply voltage is suitable for a wide range of peripheral circuits.

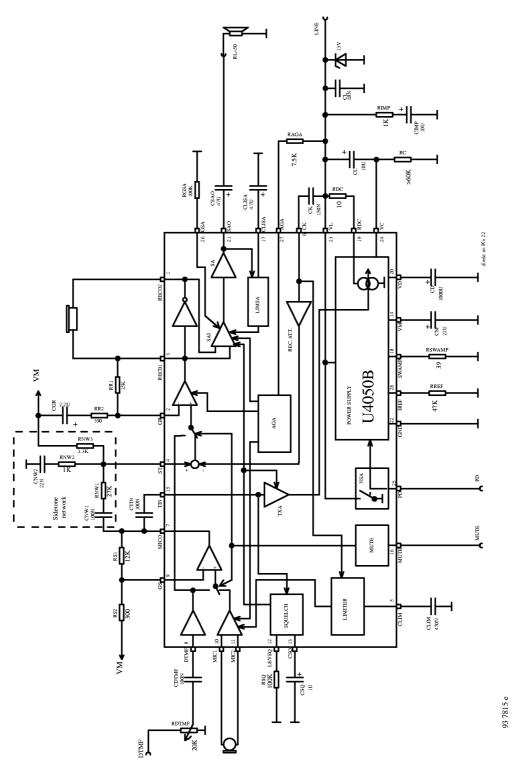


Figure 1. Typical application diagram



Pin Description

Pin	Symbol	Function
1, 3	R _{ECO} 2, R _{ECO} 1	Symmetrical outputs of receiving amplifier
1, 3	R _{ECO} 2, R _{ECO} 1	Symmetrical outputs of receiving amplifier
2	G_R	A resistor connected from this pin to V_M (ac coupled) sets the receiving amplification at the circuit
4	ST	Input of side tone amplifier
5	CLIM	Time constant of anticlipping in transmit patch
6	CK	Input of receiving path
7	MICO	Output of microphone preamplifier
8	DTMF	Input for DTMF signals (ac coupled). In Mute condition a small portion of the signal at this pin is monitored to the receiver output.
9	G_{S}	A resistor from this pin to V _M sets the amplification of microphone and DTMF signals.
10	MIC 1	Inverting input of microphone amplifier
11	MIC 2	Non-inverting input of microphone amplifier
12	LEVSQ	Input for setting the switching level of the squelch circuit
13	C_{SQ}	Time constant of the squelch function
14	V _M	Reference node for microphone, earphone and loudspeaker amplifier. Supply for electret microphone set to $V_{\rm D}/2$.
15	TIN	Input of intermediate transmit stage
16	MUTE	Active high input to switch the circuit into DTMF condition.
17	CLISA	Time constant of anticlipping of speaker amplifier.

Pin	Symbol	Function
18	SWAMP	A resistor connected from this pin to ground converts the excess line current into heat in order to prevent the IC from thermal destruction at high line currents
19	R _{DC}	A small resistor connected from this pin to V_L sets the slope of the characteristic and also affects the line length equalization characteristics and the line current at which the loudspeaker amplifier is switched on.
20	V_{D}	Unregulated supply voltage for peripheral circuits (dialers, microprocessors, etc.). Output current capability and output voltage increase with line current.
21	S _{AO}	Output of loudspeaker amplifier.
22	GND	Reference point for dc and ac output signals
23	$V_{ m L}$	Line voltage
24	$ m V_{C}$	The internal equivalent inductance of the circuit is proportional to the value of the capacitor at this pin. A resistor connected to ground may be used to reduce the line voltage.
25	PD	Active high input for reducing the current consumption of the circuit. Simultaneously $V_{\rm L}$ is shorted by an internal switch.
26	G_{SA}	Current input for setting the gain of the speaker amplifier
27	AGA	Automatic gain adjustment with line current. A resistor connected from this pin to V_L sets the starting point. Maximum gain change is 6 dB.
28	IREF	Internal reference current generation



Absolute Maximum Ratings

Parameters	Symbol	Value	Unit
Line current	${ m I_L}$	140	mA
Line voltage	$V_{ m L}$	15	V
Junction temperature	Tį	150	°C
Ambient temperature	T _{amb}	-25 to +75	°C
Storage temperature	T_{stg}	-55 to +150	°C
Total power dissipation			
$T_{amb} = 60$ °C			
SO28	P_{tot}	750	mW

Thermal Resistance

Parameters	Nymnoi	l Value	Unit
Junction ambient			
SO28	R_{thJA}	120	K/W

Electrical Characteristics

Test conditions unless otherwise specified: f = 1 kHz, 0 dBm = 775 Vrms, I_M = 0.3 mA, I_D = 2 mA, RC = 130 k Ω , T_{amb} = 25°C, R_{GSA} = 560 k Ω , Z_H = Z_M = 68 nF, Pin AGA open

Parameters	Test Conditions / Pins	Symbol	Min.	Тур.	Max.	Unit
DC characteristics	see figure 1					
	$I_L = 2 \text{ mA}$	$V_{ m L}$		1.9		V
DC16 d	$I_L = 15 \text{ mA}$	$V_{ m L}$	4.8	5.2	5.6	V
DC voltage drop over circuit	$I_L = 19 \text{ mA}$	$V_{ m L}$		5.4		V
over encur	$I_L = 30 \text{ mA}$	$V_{ m L}$		6.0		V
	$I_L = 100 \text{ mA}$	$V_{ m L}$		9.5		V
Transmission amplifier	see figure 2 and	8				
Adjustment range of transmit gain	$I_L = 15 \text{ mA}$	G_{S}	40	48	56	dB
Transmitting amplification	$I_L = 15 \text{ mA}$	G_{S}	47.75	48.25	48.75	dB
Frequency response	$I_L \ge 15 \text{ A}, C_L = 4.7 \text{ nF}$ f = 300 to 3400 Hz	$\Delta G_{ m S}$			± 0.5	dB
Gain change with current	Pin AGA open I _L = 15 to 100 mA	$\Delta G_{ m S}$			± 0.5	dB
Gain deviation	$T_{amb} = -10 \text{ to } +60^{\circ}\text{C}$ $I_L = 15 \text{ mA}$	$\Delta G_{ m S}$			± 0.5	dB
CMRR of microphone amplifier		CMRR	60	80		dB
Input resistance of MIC amplifier		R _i	45	60	80	kΩ
Distortion at line	$I_{L} > 15 \text{ mA}$ $V_{L} = 775 \text{ mVrms}$	d_s			2	%
Maximum output voltage	$I_L > 19 \text{ mA d} < 5 \%$ $V_{mic} = 10 \text{ m}$	V _{1max}	1.8	3	4.2	dBm
Noise at line psophometrically weighted	$I_L > 15 \text{ mA}$ $G_S = 48 \text{ dB}$	n _o		-80	-72	dBmp



Parameters	Test Conditions / Pins	Symbol	Min.	Тур.	Max.	Unit
Anticlipping	$V_{\text{mic}} = 20 \text{ mV}$					
attack time	C = 470 nF			0.5		ms
Release time	each 3 dB overdrive			9		ms
Gain at low operating	$I_L = 10 \text{ mA}, I_D = 1 \text{ mA}$					
current	$RC = 68 \text{ k}\Omega$	G_{S}	47		50	dB
	$V_{\text{mic}} = 1 \text{ mV } I_{\text{M}} = 0 \text{ mA}$					
Distortion at low	$I_L = 10 \text{ mA}, I_M = 0 \text{ mA}$					
operating current	$I_D = 1 \text{ mA}, RC = 68 \text{ k}\Omega$	ds			6	%
.	$V_{\text{mic}} = 10 \text{ mV}$	1.0			7	ID.
Line loss compensation	$I_L = 100 \text{ mA}$	$\Delta G_{ m SI}$	-5	-6	-7	dB
7.5	$R_{AGA} = 7.5 \text{ k}\Omega$		60			ID
Mute suppression	$I_L \ge 15 \text{ mA}$	G_{SM}	60			dB
Desciving amplifier	$V_{\text{mute}} = 1.5 \text{ V}$	1 5				
Receiving amplifier	see figure 3 and		0		. 0	.170
Adjustment range of receiving gain	$I_L \ge 15 \text{ mA}$ differential	G_R	-8		+8	dB
		C	-1	-0.5	0	4D
Receiving amplification	I _L = 15 mA differential	G_R	-1	-0.3	0	dB
Amplification of DTMF	$I_F \ge 15 \text{ mA}$	G _{RM}	-15	-12	_9	dB
signal from DTMF IN	1F = 13 mA	ORM	-13	-12	-9	ub
to RECO 1/2	Mute active					
Frequency response	$I_L > 15 \text{ mA}, C_L = 4.7 \text{ nF}$	$\Delta G_{ m RF}$			± 0.5	dB
requency response	f = 300 to 3400 Hz	DOKIT .			0.5	d D
Gain change with current	$I_{L} = 15 \text{ to } 100 \text{ mA}$	$\Delta G_{ m R}$			± 0.5	dB
Gain deviation	$T_{amb} = -10 \text{ to } + 60 ^{\circ}\text{C}$	ΔG_R			± 0.5	dB
	$I_L = 15 \text{ mA}$	- 10				
Ear protection	$I_L \ge 15 \text{ mA}$					
differential	$V_{\rm gen} = 11 V_{\rm rms}$	V_{ep}			2.2	$V_{ m rms}$
Output resistance	each output	Ro			10	Ω
_	against ĜND					
Line loss compensation	$I_L = 100 \text{ mA}$	$\Delta G_{ m RI}$	-5.0	-6.0	-7.0	dB
	$R_{AGA} = 7.5 \text{ k}\Omega$					
Output voltage	$I_L = 15 \text{ mA}, d \le 2 \%$					
Push pull	$Z_H = 68 \text{ nF}$		0.775			
	$Z_{\rm H} = 450 \Omega$		0.6			V _{rms}
Single ended	$Z_{\rm H}$ = 150 Ω		0.3			
Receiving noise	$Z_{\rm H} = 68 \rm nF$					
psophometrically	$G_R = 0 dB$	n _i		-83	70.5	ID.
weighted	$I_{\rm L}$ > 15 mA				-78.5	dBmp
Gain at low operating	$I_L = 10 \text{ mA}$					
current	$I_{D} = 1 \text{ mA}$ $I_{M} = 0 \text{ mA}$	<u></u>	-1.5		+ 0.5	dB
	$V_{gen} = 560 \text{ mV}$	G_R	-1.3		70.3	ub
	$RC = 68 \text{ k}\Omega$					
Distortion at low	$I_{L} = 10 \text{ mA}, I_{D} = 1 \text{ mA}$					
operating current	$V_{gen} = 560 \text{ mV}$					
	$RC = 68 \text{ k}\Omega$	dr			5	%



Parameters	Test Conditions / Pins	Symbol	Min.	Тур.	Max.	Unit
Speaker amplifier	see figure 4					
Minimum line current for operation	No ac signal	I_{Lmin}	10.5		15	mA
Gain from V_L to SAO	$I_{L} \ge 15 \text{ mA}$ $V_{gen} = 10 \text{ mV}$	G_{SA}	27.5	29	30.5	dB
Output power	Load resistance $R_L = 50 \ \Omega$ $d < 5 \ \%$ $V_{gen} = 300 \ mVrms$ $I_L > 15 \ mA$ $I_L = 20 \ mA$	PSA PSA	5	20		mW
Output noise	$I_L > 15 \text{ mA}$ (Input G_{SA} open)	nsa			200	μV
Gain devitation	$I_L = 15 \text{ mA}$ $T_{amb} = -10 \text{ to} + 60^{\circ}\text{C}$	ΔG_{SA}			± 1	dB
Gain change with current	$I_L = 15$ to 100 mA $R_{AGA} = 7.5$ k Ω	ΔG_{SA}			±1.5	dB
Resistor for turning off speaker amplifier	$I_L = 15$ to 100 mA	R _{GSA}	0.8	1.3	2	МΩ
Maximum off-state Output voltage	$I_L = 15 \text{ mA}$ $V_L = 0 \text{ dBm}$ Pin G_{SA} open	V_{SAO}			-50	dBm
Gain change with frequency	I _L = 15 mA f = 300 to 3400 Hz	ΔG_{SA}			± 1	dB
Attack time	20 dB overdrive	t _r		1		ms
Release time		t_{f}		300		ms
Distortion	$I_L = 15 \text{ mA}$ $V_{gen} = 300 \text{ mV}$	d _{SAO}			5	%
DTMF - amplifier	see figure 5	'			•	
Test conditions: $I_D = 2 \text{ mA}$, $I_D = 2 \text{ mA}$	$_{M} = 0.3 \text{ mA}, R_{AGA} = 7.5 \text{ k}\Omega,$	mute active				
Adjustment range of DTMF gain	$I_{L} = 15 \text{ mA}$ $Load = 600 \Omega$	G_{D}	18	26	34	dB
DTMF amplification	$I_L = 15 \text{ mA}$	G_{D}	24.5	26	27	dB
Gain deviation	$I_L = 15 \text{ mA}$ $T_{amb} = -10 \text{ to } 60^{\circ}\text{C}$	G_{D}			± 0.5	dB
Input resistance		R _i	20	25	30	kΩ
Distortion of DTMF signal	$I_L \ge 15 \text{ mA}$ $V_I = 0 \text{ dBm}$	d			2	%
Gain deviation with current	$I_L = 15 \text{ to } 100 \text{ mA}$ $R_{AGA} = 7.5 \text{ k}\Omega$	$\Delta G_{ m D}$			± 0.5	dB



Parameters	Test Conditions / Pins	Symbol	Min.	Тур.	Max.	Unit
Supply voltage	see figure 1	, ,		- ·	· I	I
** *	$0 \text{ mV}; T_{\text{amb}} = -10 \text{ to } 60^{\circ}\text{C}$					
Output voltage	$I_{L} = 15 \text{ mA}$ $I_{D} = 2 \text{ mA}$ $RC = 68 \text{ k}\Omega$	$V_{ m D}$	2.9			V
	$I_{L} = 15 \text{ mA}$ $I_{D} = 2 \text{ mA}$ $RC = 130 \text{ k}\Omega$	V_{D}	3.1			V
	$I_{L} = 100 \text{ mA}$ $I_{D} = 0 \text{ mA}$ $T_{amb} = -10 \text{ to} + 60^{\circ}\text{C}$	V_{D}			6.1	V
Supply voltage for an electret microphone	$\begin{split} I_{M} &= 0.3 \text{ mA} \\ I_{L} &\geq 15 \text{ mA} \\ RC &= 130 \text{ k}\Omega \end{split}$	V_{M}	1.45		3.3	V
Squelch	see figure 6			•		
Attenuation of transmit gain	$I_L \ge 15 \text{ mA}$	$\Delta G_{ m S}$	8	10	12	dB
Attenuation of speaker amplifier	$I_L \ge 15 \text{ mA}$ $R_{GSA} = 18 \text{ to } 560 \text{ k}\Omega$	ΔG_{SA}	7.5	10	12.5	dB
Switching level of squelch	$I_L \ge 15 \text{ mA}$ RSQ = 100 k Ω	V _{mico}	6.5		10	mV
Squelch disable	$I_{\rm L} \ge 15 {\rm mA}$	RSQ	0.5	1	2	ΜΩ
MUTE input	see figure 7					
MUTE input current	$\begin{aligned} & \text{MUTE active} \\ & I_L > 15 \text{ mA} \\ & V_{\text{MUTE}} = V_D \end{aligned}$	I _{MUTE}		20	30	μΑ
MUTE input voltage	Mute inactive I _L > 15 mA	$V_{ m MUTE}$			0,3	V
	Mute active I _L > 15 mA	$V_{ m MUTE}$	1.5		0,3	V
PD input	see figure 7					
PD input current	PD active $I_L > 15 \text{ mA}$ $V_{PD} = V_D$	I_{PD}		20	50	μΑ
Input voltage	PD = active	$V_{ m PD}$	2			V
	PD = inactive	$V_{ m PD}$			0.3	V
Current consumption	$V_D = V_{PD} = 4.5 \text{ V}$ $PD = \text{active}$ $I_L = 15 \text{ mA}$	I _{DPD}		-40	-100	μА
Voltage drop at $V_{\rm L}$	I _L = 15 mA PD = active	$V_{ m L}$		1.5		V
	$I_L=100 \text{ mA}$ PD = active	$ m V_L$		1.7		V

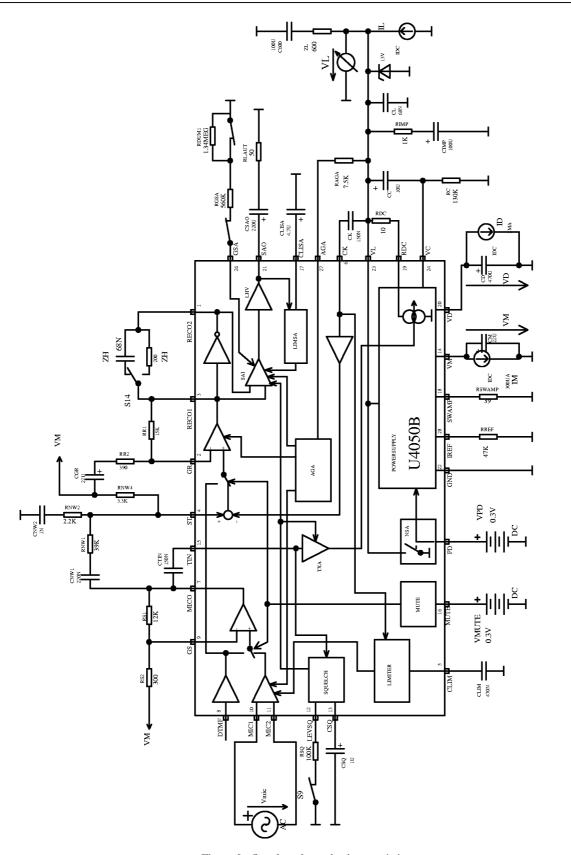


Figure 2. Supply voltage dc characteristics

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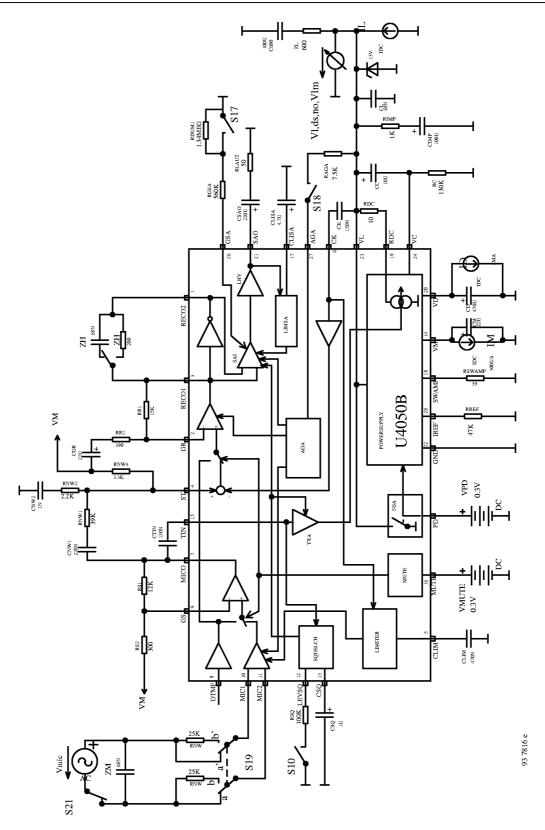


Figure 3. Transmit amplifier

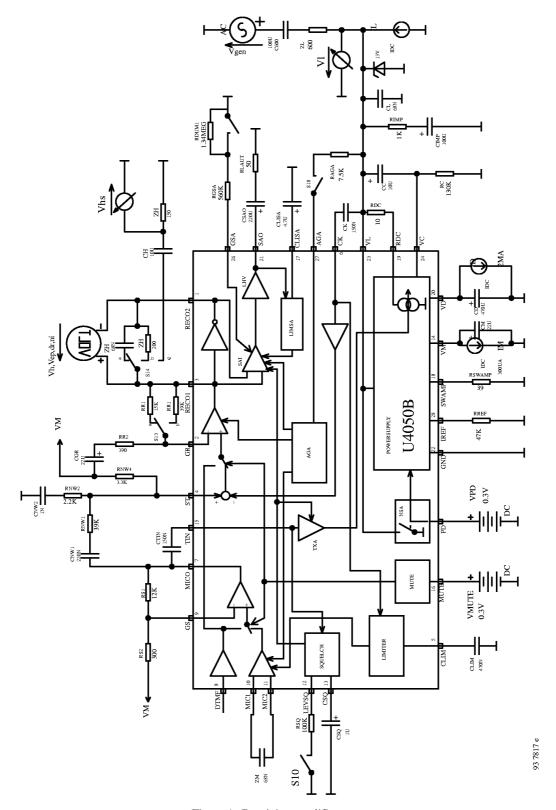


Figure 4. Receiving amplifier

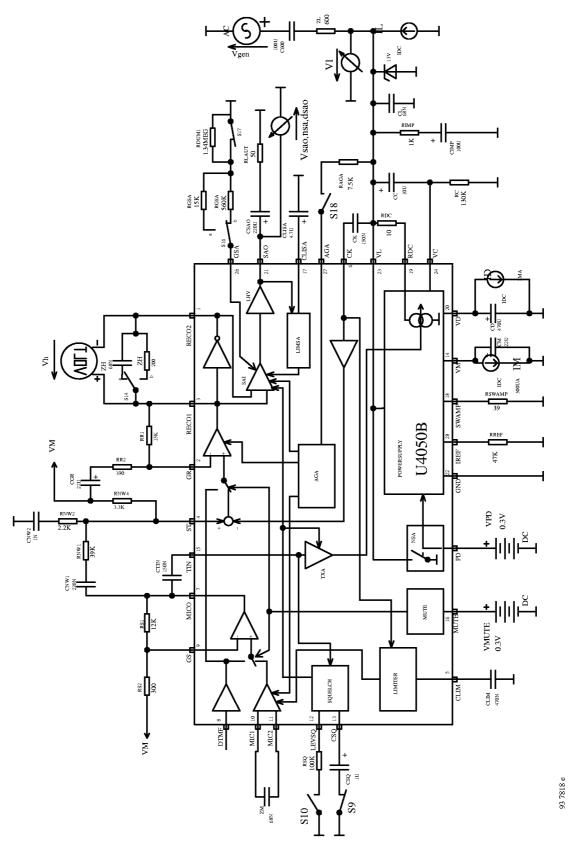


Figure 5. Speaker amplifier

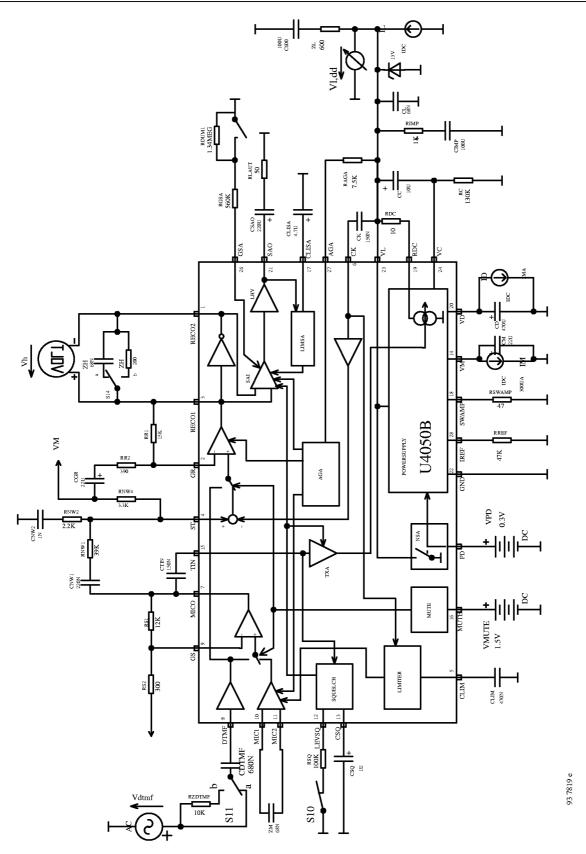


Figure 6. DTMF amplifier

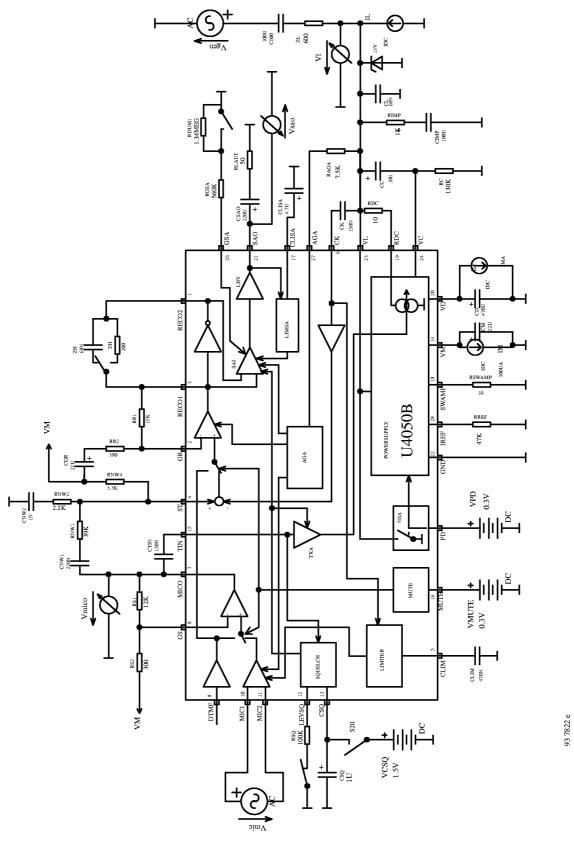


Figure 7. Squelch

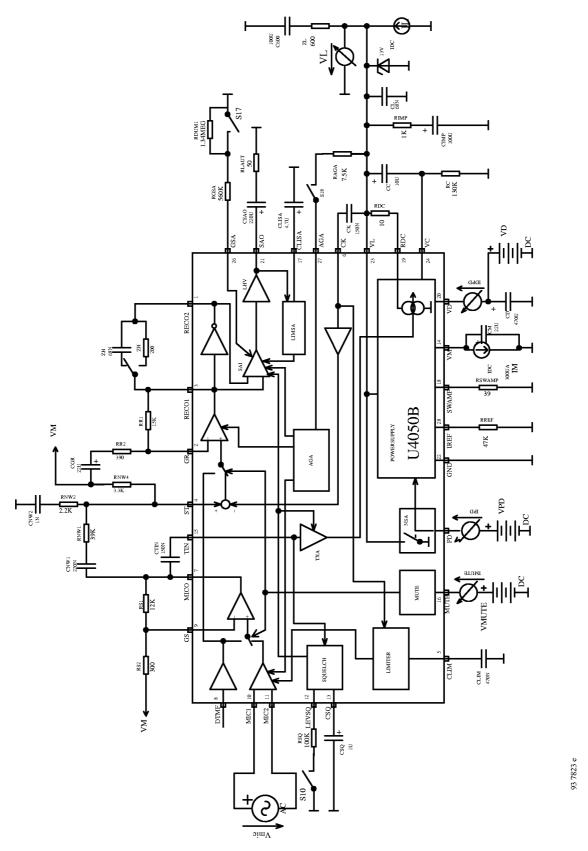


Figure 8. MUTE and PD Test

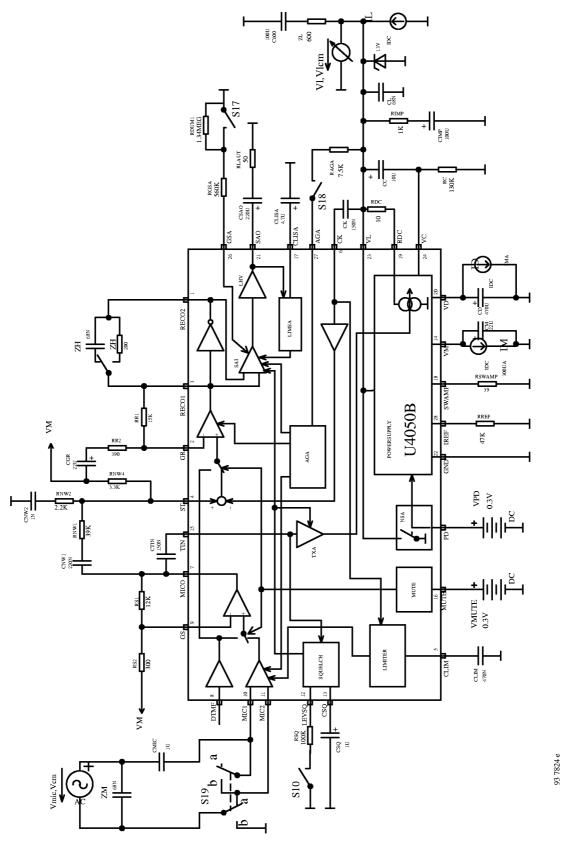


Figure 9. Transmit amplifier (CMRR)



Typical Curves

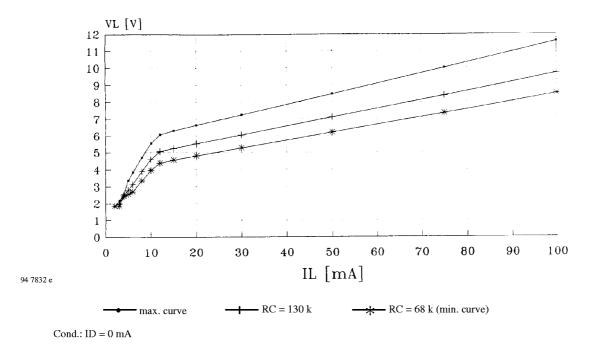


Figure 10. DC characteristics

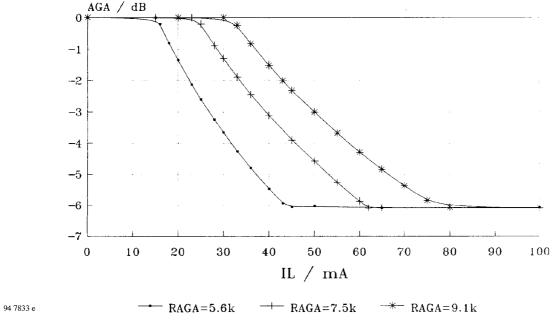
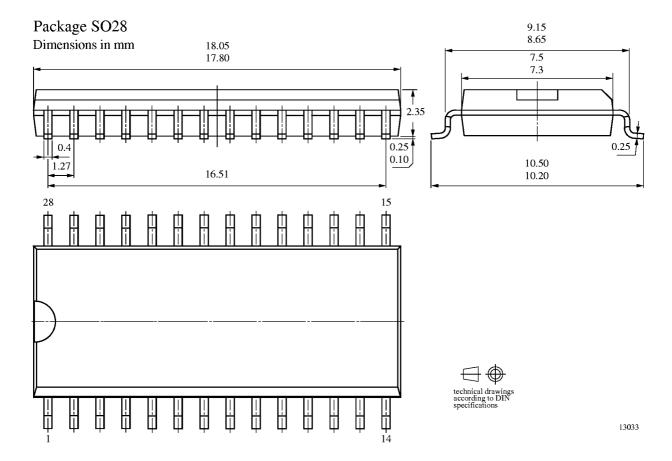


Figure 11. AGA characteristics



Package Information





Ozone Depleting Substances Policy Statement

It is the policy of TEMIC TELEFUNKEN microelectronic GmbH to

- 1. Meet all present and future national and international statutory requirements.
- Regularly and continuously improve the performance of our products, processes, distribution and operating systems with respect to their impact on the health and safety of our employees and the public, as well as their impact on the environment.

It is particular concern to control or eliminate releases of those substances into the atmosphere which are known as ozone depleting substances (ODSs).

The Montreal Protocol (1987) and its London Amendments (1990) intend to severely restrict the use of ODSs and forbid their use within the next ten years. Various national and international initiatives are pressing for an earlier ban on these substances.

TEMIC TELEFUNKEN microelectronic GmbH semiconductor division has been able to use its policy of continuous improvements to eliminate the use of ODSs listed in the following documents.

- 1. Annex A, B and list of transitional substances of the Montreal Protocol and the London Amendments respectively
- 2. Class I and II ozone depleting substances in the Clean Air Act Amendments of 1990 by the Environmental Protection Agency (EPA) in the USA
- 3. Council Decision 88/540/EEC and 91/690/EEC Annex A, B and C (transitional substances) respectively.

TEMIC can certify that our semiconductors are not manufactured with ozone depleting substances and do not contain such substances.

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