Designing the DP8392 for Longer Cable Applications

The IEEE 802.3 standard is designed for 500 meters of Ethernet cable and 185 meters of Cheapernet (RG58A/U) cable. To extend such segments to 1000 meters of Ethernet cable and 300 meters of Cheapernet cable requires utilization of Transmit mode collision detection. This method is described below.

COLLISION DETECTION SCHEMES

The collision circuitry monitors the coaxial DC level. If the level is more negative than the collision threshold, the collision output is enabled

There are two different collision detection schemes that can be implemented with the CTI; Receive mode, and Transmit mode. The IEEE 802.3 standard allows the use of receive and transmit modes for non-repeater node applications. Repeaters are required to have to receive mode implementation. These different modes are defined as follows:

Receive Mode: Detects a collision between any two stations on the network with certainty at all times.

Transmit Mode: Detects a collision with certainty only when the station is transmitting.

Table I summarizes the receive and transmit mode definitions:

TABLE I

Mode		Re	ceive	9		Tra	nsmi	it
No. of Stations	0	1	2	>2	0	1	2	>2
Transmitting Non-Transmitting	N N	N N	Y Y	Y Y	N N	N N	Y M	Y Y

- Y = Detects Collision
- N = Does Not Detect Collision
- M = Might Detect Collision

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Receive Mode: The Receive mode scheme has a very simple truth table. However, the tight threshold limits make the design of it difficult. The threshold in this case has to be between the maximum DC level of one station ($-1300~{\rm mV}$) and the minimum DC level of two far stations ($-1581~{\rm mV}$). Several factors such as the termination resistor variation, signal skew, and input bias current of non-transmitting nodes contribute to this tight margin. On top of the $-1300~{\rm mV}$ minimum level, the impulse response of the internal low pass filter has to be added. The CTI incorporates a 4-pole Bessel filter in combination with a trimmed on board bandgap reference to provide this mode of collision detection

Transmit Mode: In this case, collision has to be detected only when the station is transmitting. Thus, collision caused by two other nodes may or may not be detected. This feature relaxes the upper limit of the threshold. As a result of this, longer cable segments can be used. With the CTI, a resistor divider can be used at the Collision Detection Sense (CDS) pin to lower the threshold from receive to transmit mode.

COLLISION LEVELS—TRANSMIT MODE

Table II shows the parameter values that are used in calculating the collision levels in transmit mode.

TABLE II. Assumptions and Definitions

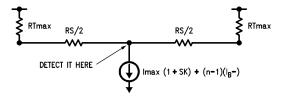
t_T = Temp. Coef. of the Terminator = 0.0001/Deg. ASS	UMPTION
L = Maximum Segment Length = 300m Cheapernet 802.	3
= 1000m Ethernet 802.	3
R_{DC} = Maximum Cable DC Res. at 20°C = 0.0489 Ω /m Cheapernet BELI	DEN
$= 0.0100\Omega/m$ Ethernet BELI	DEN
tc = Temp. Coef. of Copper = 0.004 °C PHY	SICS
Tm = Maximum Cable Temp. = 50°C ASS	UMPTION
SR = Step Response at Max Cable Length = 0.97 Cheapernet NAT	IONAL
= 0.94 Ethernet NAT	IONAL
R_C = Max. Connector Res./Station = 0.0034 Ω Cheapernet MIL:	SPEC
	UMPTION
I_{B+} = Max. Positive Bias Current = 2 μ A 802.	3
$I_{B-} = Max$. Negative Bias Current = 25 μ A 802.	3
I_{max} = Max. DC Drive Current = 45 mA 802.	3
I _{min} = Min. DC Drive Current = 37 mA 802.	3
Ro = Non Transmitting Output Impedance = 100 kΩ 802.	3
N = Max Nodes per Segment = 100 Cheapernet 802.	3
= 100 Ethernet 802.	3
SK = Skew Factor, Effect of Encoder	
Skew on DC Level 802.	3
$= (SKEW \times 4)/100 = 0.02 \text{ for } 0.5 \text{ ns Skew}$	
R _S = Max. DC Loop Res. of a Segment DEF	INITION
R _L = Load Resistance Seen by a Driver DEF	INITION
SEO = Sending End Overshoot = 0.10 Cheapernet ASS	UMPTION
= 0.14 Ethernet ASS	UMPTION

The calculations below explain how the values for the resistor divider in Figure 1 are obtained. First, collision levels V_{max} and V_{min} must be calculated. The V_{max} or "no detect" level is the maximum DC voltage generated by one node. The worst case here occurs when the transmitting node is at the center of a maximum length cable, and the collision is being detected either by itself or by a station right next to it. On the other hand, the V_{min} or "must detect" level is the

minimum DC voltage generated by one minimum transmitting station and another minimum transmitting station at the other end of a maximum length cable.

The filter impulse response is not included in these calculation since it is mutually exclusive with the Sending End Overshoot. If the impulse response is larger than the Sending End Overshoot, the exceeding portion should be added on to the limits.

${\bf Maximum\ Non\ Collision\ Level\ V_{Max}\ (NO\ DETECT)} \color{blue} - {\bf Transmit\ Mode}$



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$$\begin{split} R_{Tmax} &= R_T \times 1.01 \times \left[(Tm-20) \times t_T + 1 \right] \\ R_S &= R_{DC} \times L \times \left[(Tm-20) \times t_c + 1 \right] + N \times RC \\ R_L &= (R_{Tmax} + R_S/2)/2 \end{split}$$

$$\begin{array}{lll} V_{\text{Max}} &= [I_{\text{Max}} \times (1 + \text{SK}) + (\text{N} - 1) \; (I_{\text{B}-})] \times \, \text{R}_{\text{L}} \times \\ &\quad (1 + \text{SEO}) \end{array}$$

CHEAPERNET Cable, 300 Meters, 100 Stations:

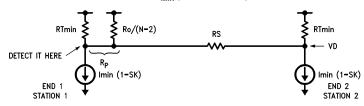
$$\begin{array}{lll} \text{R}_{\text{Tmax}} &= 50 \, \times \, 1.01 \, \times \left[(50 \, - \, 20) \, \times \, 0.0001 \, + \, 1 \right] \\ &= 50.652 \Omega \\ \text{R}_{\text{S}} &= 0.0489 \, \times \, 300 \, \left[(50 \, - \, 20) \, \times \, 0.004 \, + \, 1 \right] \, + \\ &= 100 \, \times \, 0.0034 & = 16.770 \Omega \end{array}$$

$$\begin{array}{lll} {\rm R_L} &= (50.652 + 16.770/2)/2 &= 29.519 \Omega \\ {\rm V_{Max}} &= [45 \times 1.02 + 99 \times 0.025] \times 29.519 \times 1.10 \\ &= 1571 \, {\rm mV} \end{array}$$

ETHERNET Cable, 1000 Meters, 100 Stations:

$$\begin{array}{lll} R_{Tmax} &= 50 \times 1.01 \times [(50-20) \times 0.0001 + 1) \\ &= 50.652 \Omega \\ R_{S} &= 0.01 \times 1000 \, [(50-20) \times 0.004 + 1] + 100 \times \\ &= 0.0001 &= 11.21 \Omega \\ R_{L} &= (50.652 + 11.21/2)/2 &= 28.129 \Omega \\ V_{Max} &= [45 \times 1.02 + 99 \times 0.025] \times 28.129 \times 1.14 \\ &= 1551 \, \text{mV} \end{array}$$

$\label{eq:minimum collision Level V} \mbox{Min (MUST DETECT)} \mbox{--} \mbox{Transmit Mode}$



R_P = Near End Shunt Resistance

 $= [Ro/(N-2)]//R_{Tmin}$

 $R_{T_{\mbox{Min}}} = R_{\mbox{\scriptsize T}} imes 0.99$

VS1(1) = Station 1's DC Voltage at End 1

 $I_{Min} \times (1 - SK) \times [R_P // (R_S + R_{Tmin})]$

VS2(2) = Station 2's DC Voltage at End 2

 $= I_{Min} \times (1 - SK) \times [R_{Tmin} / / (R_S + R_P)]$

 $\begin{array}{l} \text{VS2(1)} = \text{Station 2's DC Voltage at End 1} \\ = \text{VS2(2)} \times \left[\text{R}_{\text{P}}/(\text{R}_{\text{S}} + \text{R}_{\text{P}}) \right] \times \text{SR} \end{array}$

 $V_{Min} = VS1(1) + VS2(1)$

CHEAPERNET Cable, 300 Meters, 100 Stations:

 $R_P = [100k/98]/(50 \times 0.99)$

= 1020//49.5 $= 47.209\Omega$

 $VS1(1) = 37 \times 0.98 \times [47.209//(16.770 + 49.5)]$

= 1000 mV

 $VS2(2) = 37 \times 0.98 \times [49.5//(16.770 + 47.209)]$

= 1012 mV

 $VS2(1) = 1012 \times [47.209/(47.209 + 16.770)] \times 0.97$

= 724 mV

 $V_{Min} = 1000 + 724 = 1724 \text{ mV}$

ETHERNET Cable, 1000 Meters, 100 Stations:

 $R_P = [100k/98]//(50 \times 0.99) = 1020//49.5$

 $= 47.209\Omega$

 $VS1(1) = 37 \times 0.98 \times [47.209//(11.21 + 49.5)]$

 $= 963 \; \text{mV}$

 $VS2(2) = 37 \times 0.98 \times [49.5//(11.21 + 47.209)]$

= 972 mV

 $VS2(1) = 972 \times [47.209/(47.209 + 11.21)] \times 0.94$

= 738 mV

 $V_{Min} = 963 + 738 = 1701 \text{ mV}$

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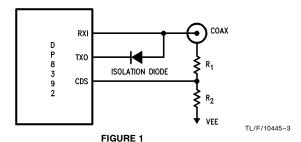
CIRCUIT IMPLEMENTATION

Table III summarizes the design parameters.

TABLE III

Parameter	ETHERNET	CHEAPERNET			
L	1000 Meter	300 Meter			
N	100	30			
V _{Min}	1701 mV	1724 mV			
V _{Max}	1551 mV	1571 mV			
R ₁	125Ω ±1%	150 Ω \pm 1%			
R ₂	10 kΩ ±1%	10 kΩ ±1%			

Circuit implementation is shown in Figure 1



To check the design, subtract the additional offset generated by the resistor divider from these levels (V_{Max} and V_{Min}) and make sure that the internal 8392 collision levels (1450 mV to 1580 mV) are within this window. The supply voltage is assumed to be $9V \pm 5\%$.

1551 mV - 8.55V (125 Ω /(10 k Ω + 125 Ω)) = 1445 mV 1701 mV - 9.45V (125 Ω /10 k Ω + 125 Ω)) = 1584 mV

Cheapernet

 $1571 \text{ mV} - 8.55 \text{V} (150\Omega/(10 \text{ k}\Omega + 150\Omega)) = 1445 \text{ mV}$ $1724 \text{ mV} - 9.45 \text{V} (150\Omega/(10 \text{ k}\Omega + 150\Omega) = 1584 \text{ mV}$ These calculations show that the resistor values are properly selected.

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