

## PROGRAMMABLE CURRENT SENSING HIGH SIDE SWITCH

### Features

- Load current feedback
- Programmable over current shutdown
- Active clamp
- E.S.D protection
- Input referenced to Vcc
- Reverse battery protection (reverse current operation)

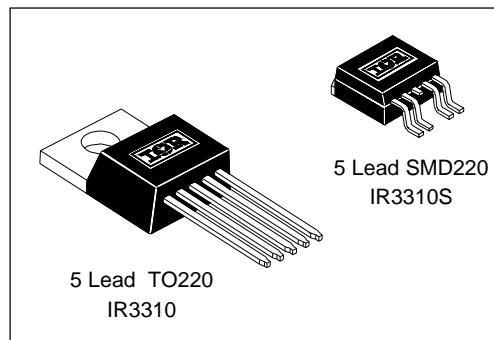
### Description

The IR 3310 is a Fully Protected 4 terminal high side switch. The input signal is referenced to Vcc. When the input voltage Vcc - Vin is higher than the specified Vih threshold, the output power MOSFET is turned-on. When Vcc - Vin is lower than the specified Vil threshold, the output MOSFET is turned-off. A sense current proportional to the current in the power Mosfet is sourced to the ST pin. Over-current shutdown occurs when Vst - Vin > 4 V. Choosing Rst allows to adjust Isd. Either over-current and over-temperature latches off the switch. The device is reset by pulling the input pin high. Other integrated protections ( ESD, reverse battery, active clamp ) make the IR 3310 very rugged and suitable for the automotive environment.

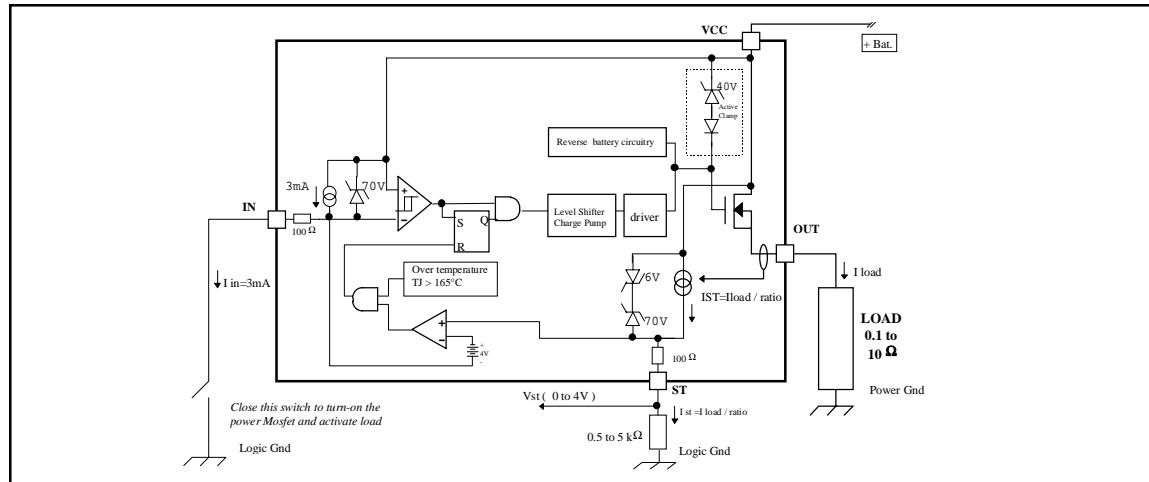
### Product Summary

Rds(on)	5mΩ typ.
Vcc.op.	5.5 to 35V
Current ratio	10 000
Ishutdown	10 to 100A
Active clamp	40V

### Packages



### Typical Connection



# IR3310

International  
**IR** Rectifier

## Absolute Maximum Ratings

Absolute maximum ratings indicates sustained limits beyond which damage to the device may occur. All voltage parameters are referenced to Vcc lead. (TAmbient = 25°C unless otherwise specified).

Symbol	Parameter	Min.	Max.	Units
Vcc - Vin max	Maximum input voltage	-16	50	V
Vcc-Vst max	Maximum status voltage	-16	50	
Vcc - Vout max.	Maximum output voltage	-0.3	37	
Ids cont.	Diode max. permanent current (Rth = 60 °C/W) (1)	—	2.8	A
Ids1 cont	Diode max. permanent current (Rth = 5 °C/W) (1)	—	35	
Ids pulsed	Diode max. pulsed current (1)	—	100	
ESD 1	Electrostatic discharge ( human body model )	—	tbd	kV
ESD 2	Electrostatic discharge ( machine model )	—	tbd	
Pd	Power dissipation ( Rth = 62 °C/W )	—	2	W
TJ max.	Max. storage and junction temperature	-40	150	°C
Min R st	Minimum resistor on the ST pin	0.5	—	kΩ

## Thermal Characteristics

Symbol	Parameter	Typ.	Max.	Units
R <sub>th</sub> free air	Thermal resistance - free air	60	—	°C/W
R <sub>th</sub> std. footprint	Thermal resistance with standard footprint	80	—	°C/W
R <sub>th</sub> 1" footprint	Thermal resistance with 1" footprint	50	—	°C/W
R <sub>th</sub> junc. to case	Thermal resistance junction to case	5	—	°C/W

## Recommended Operating Conditions

These values are given for a quick design. For operation outside these conditions, please consult the application notes.

Symbol	Parameter	Min.	Max.	Units
Vcc - Vin	Continuous input voltage	6	18	V
Vcc-Vst	Continuous status voltage	0	18	
Vcc	Supply to power ground voltage	6	18	
Iout	Continuous output current ( Rth/amb < 5 °C/W, Tj = 125°C )	—	35	A
Iout 85°C amb.	Continuous output current ( Rth/amb < 5 °C/W, Tj = 125°C )	—	8	
Rst	ST resistor to program lsd and scale (2 & 3)	0.5	5	kΩ

## Protection Characteristics

T<sub>j</sub> = 25°C (unless otherwise specified), R<sub>st</sub> = 500 to 5kOhm.

Symbol	Parameter	Typ.	Max.	Units	Test Conditions
Vst - Vin @ lsd	Over-current shutdown threshold	4	—	V	
Tsd	Over-temp. shutdown threshold	165	—	°C	see Fig. 4
Treset	Protection reset time	50	—	μS	see Fig. 4

1) Limited by junction temperature. Pulsed current is also limited by wiring

2) <500 Ohm or shorting ST to gnd may damage the part with lsd around 120A

3) >5000 Ohm or leaving ST open will shutdown the part. No current will flow in the load.

## Static Electrical Characteristics

( $T_j = 25^\circ\text{C}$ ,  $V_{cc} = 14\text{V}$  unless otherwise specified.)

Symbol	Parameter	Min.	Typ.	Max.	Units	Test Conditions
I <sub>q</sub>	Quiescent current	—	7	50	uA	$V_{cc}-V_{in}=0$ , $V_{cc}-V_{out}=12\text{V}$
I <sub>in</sub>	Input current	—	3	—	mA	$V_{cc}-V_{in}=14\text{V}$
V <sub>ih</sub>	High level input threshold voltage (4)	—	5	—	V	
V <sub>il</sub>	Low level input threshold voltage (4)	—	4	—		
V <sub>phys</sub>	Input hysteresis = V <sub>ih</sub> -V <sub>il</sub>	—	1	—		
R <sub>d<sub>1</sub> on</sub>	ON state resistance	—	5	—	mΩ	I <sub>out</sub> =35A, $V_{cc}=14\text{V}$
R <sub>d<sub>2</sub> on</sub>	ON state resistance	—	5.5	—		I <sub>out</sub> =17A, $V_{cc}-V_{in}=6\text{V}$
R <sub>d<sub>3</sub> on</sub>	ON state resistance	—	8	—		I <sub>out</sub> =35A, $T_j = 150^\circ\text{C}$
V <sub>clamp1</sub>	V <sub>cc</sub> to V <sub>out</sub> active clamp voltage	37	40	—	V	I <sub>out</sub> = 10mA
V <sub>clamp2</sub>	V <sub>cc</sub> to V <sub>out</sub> active clamp voltage	—	42	48		I <sub>out</sub> = 35mA
V <sub>sd</sub>	Body diode forward voltage	—	0.85	1	—	I <sub>d</sub> =35A, $V_{cc}-V_{in}=0\text{V}$

## Switching Electrical Characteristics

$V_{cc} = 14\text{V}$ , Resistive Load =  $0.4\Omega$ ,  $T_j = 25^\circ\text{C}$ , (unless otherwise specified).

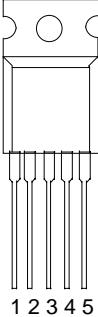
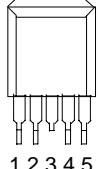
Symbol	Parameter	Min.	Typ.	Max.	Units	Test Conditions
T <sub>on</sub>	Turn-on delay time to $V_{cc}-V_{out}=0.9 V_{cc}$	—	10	—	μs	see figure 2
T <sub>r1</sub>	Rise time to $V_{cc}-V_{out}=5\text{V}$	—	16	—		
T <sub>r2</sub>	Rise time from end of T <sub>r1</sub> to $V_{out} = 10\%$ of $V_{cc}$	—	300	—		
dV/dt (on)	Turn ON dV/dt	—	1.2	—	V/μs	
E <sub>on</sub>	Turn ON energy	—	40	—	mJ	
T <sub>off</sub>	Turn-off delay time	—	tbd	—	μs	see figure 3
T <sub>f</sub>	Fall time to $V_{cc}-V_{out}$ 90% of $V_{cc}$	—	tbd	—		
dV/dt (off)	Turn OFF dV/dt	—	tbd	—	V/μs	
E <sub>off</sub>	Turn OFF energy	—	5	—	mJ	

## Current Sense Characteristics

Symbol	Parameter	Min.	Typ.	Max.	Units	Test Conditions
Ratio	I load / I status	—	10,000	—	—	$T_j = 25^\circ\text{C}$ , $R_{st} = 500\Omega$ , $I = 100\text{A}$
Ratio	Gain variation over temperature range error ToC	-5	—	+5	%	$T_j = 40$ To $+150^\circ\text{C}$
offset	status current when I load = 0	-0.5	0	+0.5	mA	$T_j = 25^\circ\text{C}$ , $R_{st} = 500\Omega$ , $I = 100\text{A}$
offset	Offset variation over temperature range var. ToC	-0.4	0	0.4	mA	$T_j = 40$ To $+150^\circ\text{C}$
Trst	status response time to a small I load step	—	10	—	μs	to get 90% of the I load step

4) Input threshold are measured directly between the input pin and the tab. Any parasitic resistance in common between the load current path and the input signal path can significantly affect the thresholds.

## Lead Assignments

 5 Lead - TO220	 5 Lead - SMD220
<b>IR3310</b>	<b>IR3310S</b>
<b>Part Number</b>	

The following note applies to all curves: 1) they are all typical characteristics. 2) Operation in shaded area is not recommended. 3)  $T_j = 25^\circ\text{C}$ ,  $R_{st}=500\Omega\text{m}$ ,  $V_{bat} = 14\text{V}$  (unless otherwise specified).

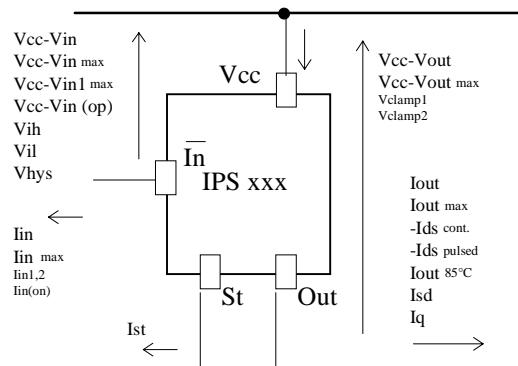


Figure 1 - Voltages and currents definition

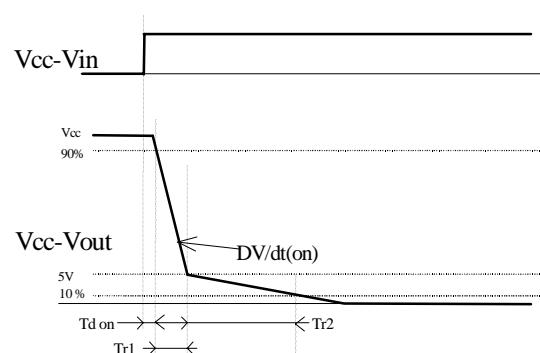


Figure 2 - Switching time definitions (turn-on)

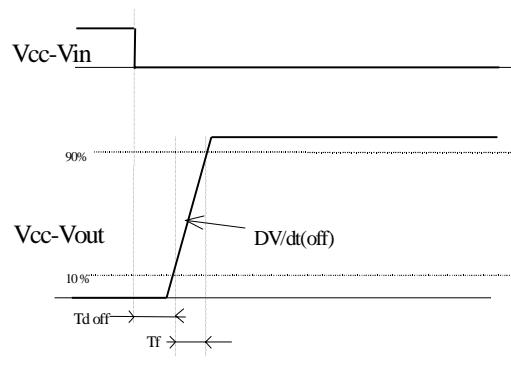


Figure 3 - Switching time definitions (turn-off)

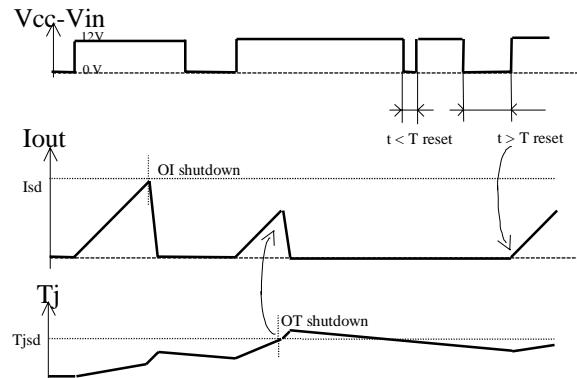


Figure 4 - Protection timing diagram

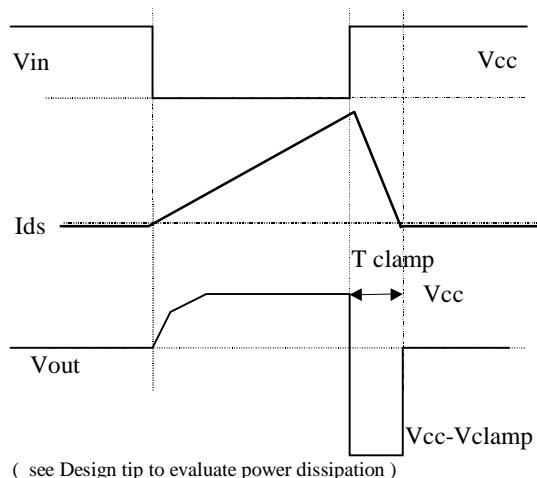


Figure 5 - Active clamp waveform

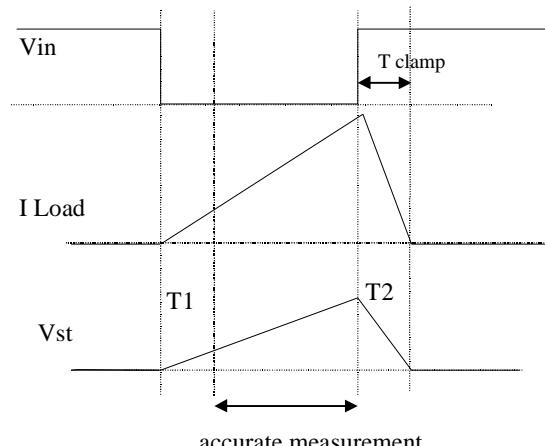


Fig 6 - Current sensing accuracy:  
 Measurement is accurate only when the power Mosfet is fully on (outside  $T_1 = T_{don} + T_{r1} + T_{r2}$ ) and when the part is not in the active clamp (outside  $T_2$ ).

# IR3310

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**IR** Rectifier

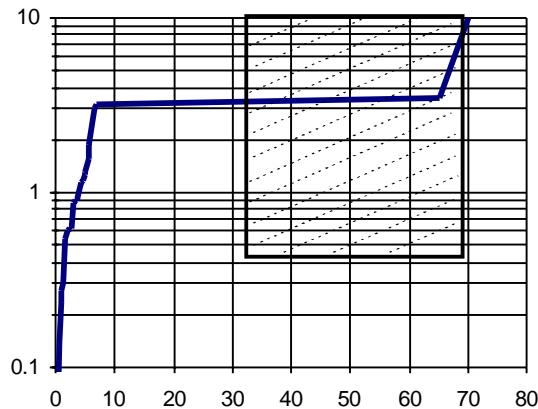


Figure 7 -  $I_{cc}$  (mA) vs  $V_{cc}-V_{in}$  (V)

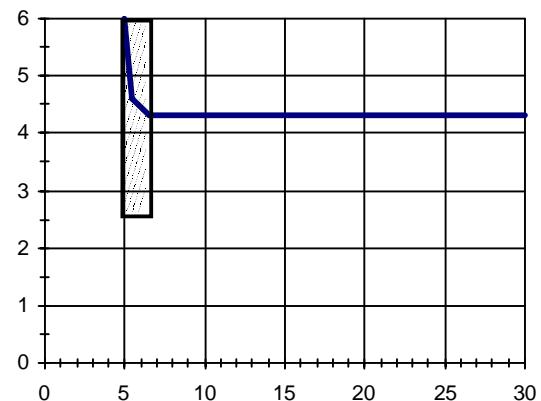


Figure 8 -  $R_{dson}$  ( $m\Omega$ ) vs  $V_{cc}-V_{in}$  (V)



Figure 9 - Normalized  $R_{dson}$  (%) vs  $T_j$  (°C)

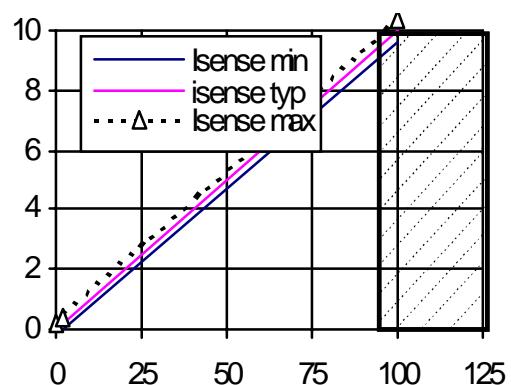


Figure 10 -  $I_{sense}$  (mA) vs  $I_{load}$  (A)

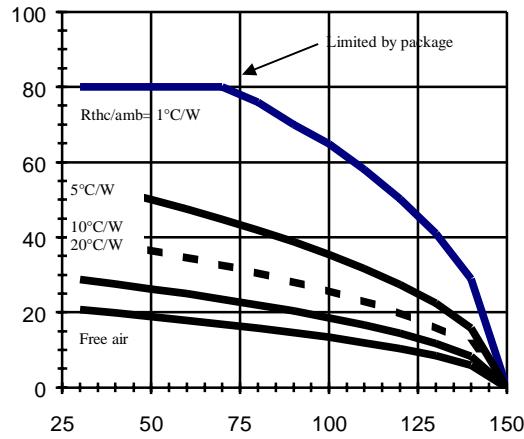


Figure 11 -  $I_{ds}$  (A) vs  $R_{ds}$  ( $\Omega$ )

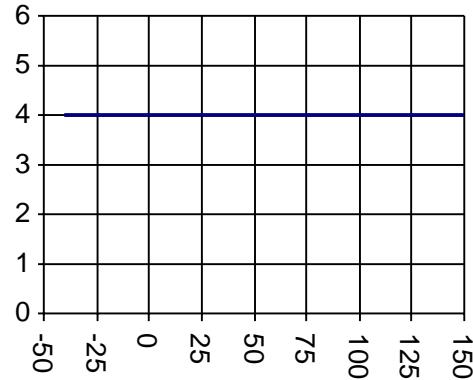


Figure 12 -  $V_{ds} - V_{in}$  (V) vs  $T_j$  ( $^{\circ}$ C)

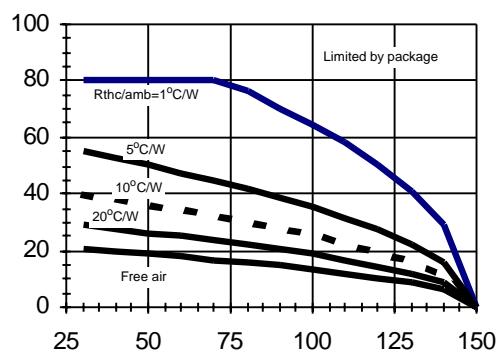


Figure 13 - Max. Cont.  $I_{out}$  (A) vs Temp. ( $^{\circ}$ C)

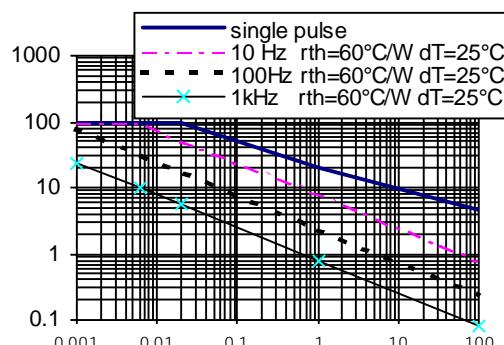


Figure 14 - Max.  $I_{out}$  (A) vs load inductance (uH)

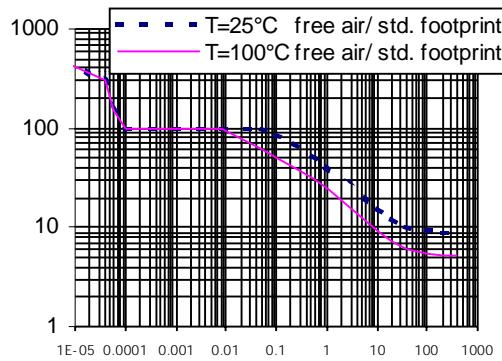


Figure 15 - I out (A) vs Protection resp. Time (s)

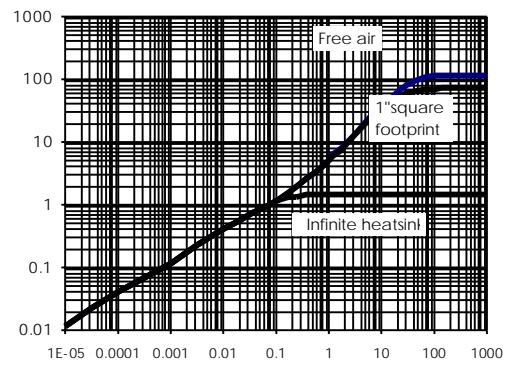


Figure 16 - Transient Rth (°C) vs Time (s)

### How to use the IR 3310 device

- Check max. continuous power dissipation :

Use figure 11 to check that max. continuous load current does not exceed the device capability in worst case ambient temperature.

- Choose  $I_{sd}$  so that it exceeds the maximum transient current with a sufficient margin.  
 $I_{max. load}$  shall not exceed 100A.
- Choose  $R_{st}$  to get  $I_{sd}$  by mean of fig. 10 curve or using the following formula

$$I_{sd} = (V_{st}-V_{in} @ I_{sd}) * \text{gain} / R_{st} = 4V * 10000 / R_{st}$$

This ensures optimum protection and full scale of  $V_{st}$  signal (0V for  $I_{load}=0$  and 4V for  $I_{load} = I_{sd}$ ).

- To reduce power dissipation during reverse battery operation, the inner circuitry takes the potential available on ST pin in order to turn on the power MOSFET. This principle works only if  $R_{st}$  is within the recommended range ( 0.5 to 5 kΩ ) and Rev.bat. voltage > 5V.

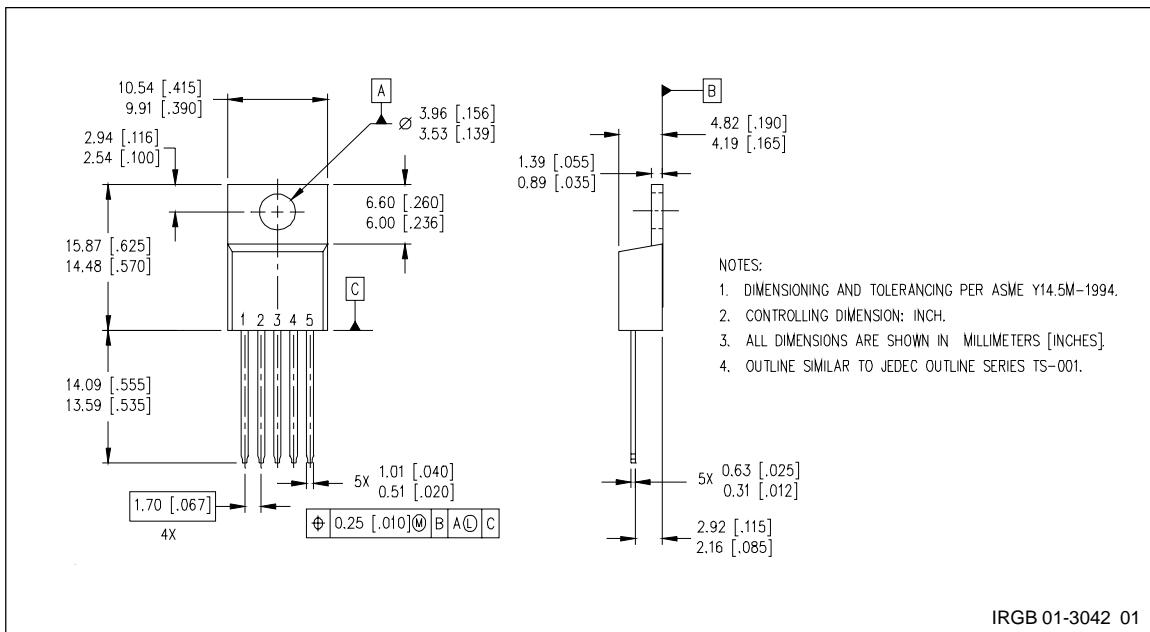
Also, the logic Gnd must not be disconnected from the Power Gnd (due to another reverse battery protection circuitry for example).

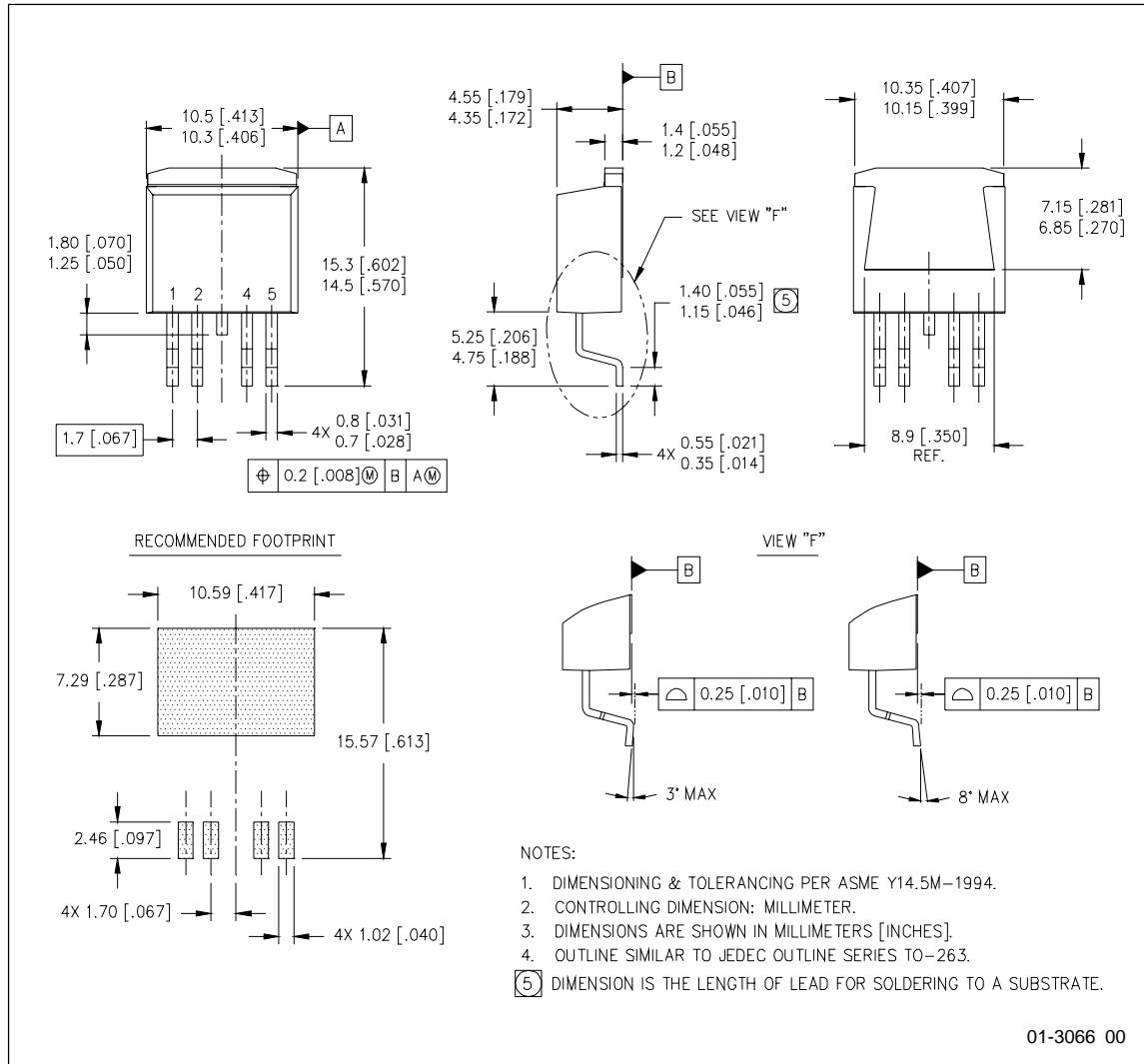
Check that junction temperature does not exceed the max. value (165°C) :

$$( \text{losses multiplied by } r_{th} + \text{max. ambient} ) < 165^{\circ}\text{C}$$

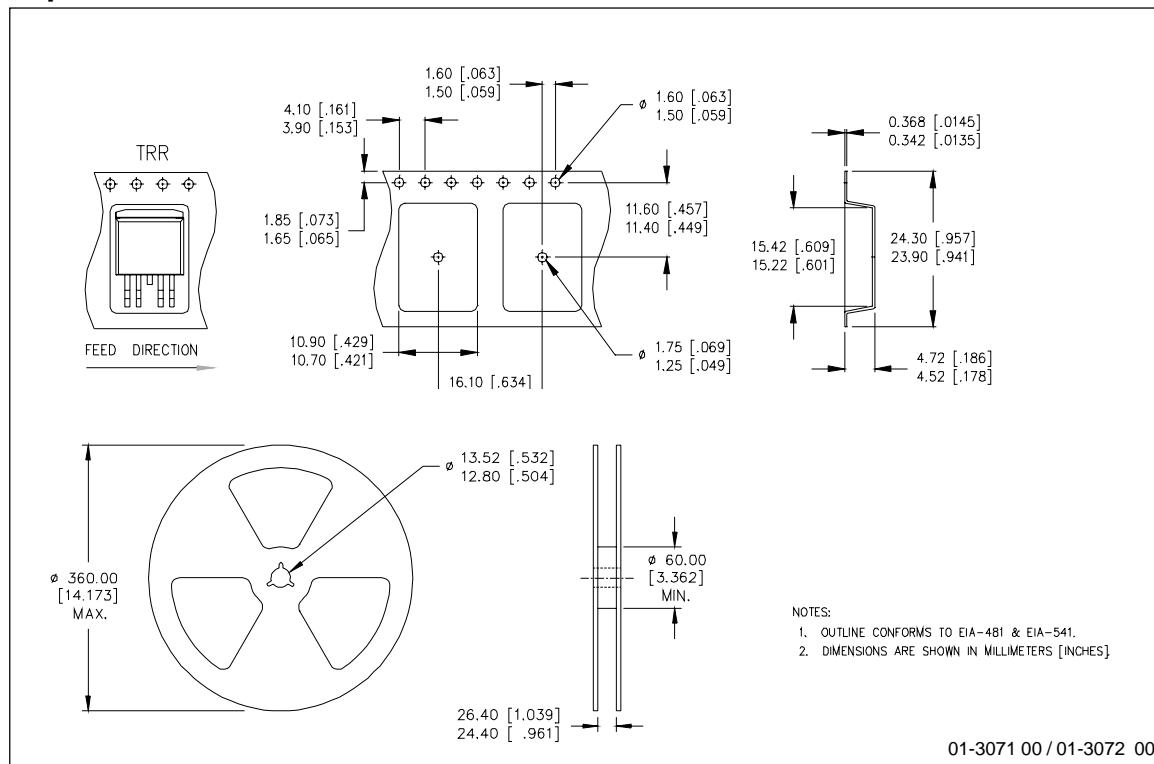
$$\text{Total Losses (w)} = P_{\text{power Mosfet}} + P_{(\text{in})} + P_{(\text{st})}$$

**Case Outline - TO220 (5 lead)**



Case Outline - D<sup>2</sup>PAK (SMD220) - 5 Lead

**Tape & Reel - SMD220 - 5 Lead**



International  
**IR** Rectifier

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*Data and specifications subject to change without notice. 4/16/2000*