

DATA SHEET

TDA5634T

**9 V UHF mixer/oscillator for TV and
VCR tuners**

Product specification
Supersedes data of 1995 Mar 21
File under Integrated Circuits, IC02

1996 Nov 7

9 V UHF mixer/oscillator for TV and VCR tuners

TDA5634T

FEATURES

- Balanced mixer with a common base input
- 4-pin oscillator
- Local oscillator buffer output for external prescaler
- SAW filter preamplifier with a low output impedance to drive the SAW filter directly
- Band gap voltage stabilizer for oscillator stability
- External IF filter between the mixer output and the IF amplifier input.

APPLICATION

- UHF tuners for TV and VCR
- One band tuners.

QUICK REFERENCE DATA

| SYMBOL | PARAMETER | CONDITIONS | MIN. | TYP. | MAX. | UNIT |
|----------|---|-------------------------|------|------|------|------------|
| V_P | supply voltage | | – | 9.0 | – | V |
| I_P | supply current | | – | 35 | – | mA |
| f_{RF} | RF frequency | RF input; note 1 | 430 | – | 860 | MHz |
| G_v | voltage gain | | – | 36 | – | dB |
| NF | noise figure | not corrected for image | – | 9 | – | dB |
| V_o | output voltage causing 1% cross modulation in channel | | – | 121 | – | dB μ V |

Note

1. The limits are related to the tank circuits used in Fig.7 and the intermediate frequency. Frequency bands may be adjusted by the choice of external components.

ORDERING INFORMATION

| TYPE NUMBER | PACKAGE | | |
|-------------|---------|--|----------|
| | NAME | DESCRIPTION | VERSION |
| TDA5634T | SO16 | plastic small outline package; 16 leads; body width 3.9 mm | SOT109-1 |

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BLOCK DIAGRAM

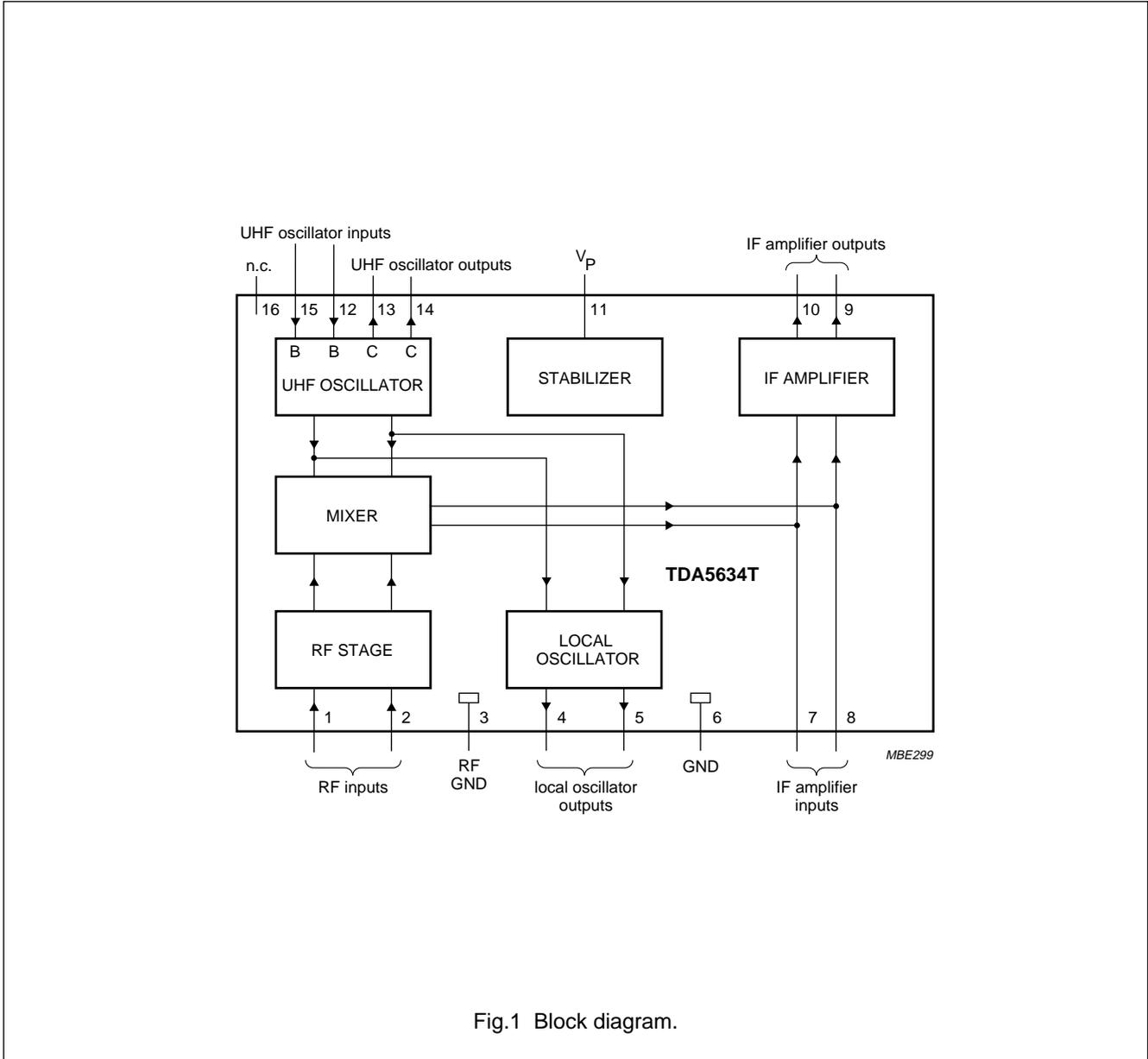


Fig.1 Block diagram.

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PINNING

| SYMBOL | PIN | DESCRIPTION |
|----------------|-----|-------------------------------------|
| CIN1 | 1 | band C input 1 |
| CIN2 | 2 | band C input 2 |
| RFGND | 3 | ground for RF inputs |
| LOOUT1 | 4 | local oscillator amplifier output 1 |
| LOOUT2 | 5 | local oscillator amplifier output 2 |
| GND | 6 | ground (0 V) |
| IFIN1 | 7 | IF amplifier input 1 |
| IFIN2 | 8 | IF amplifier input 2 |
| IFOUT1 | 9 | IF amplifier output 1 |
| IFOUT2 | 10 | IF amplifier output 2 |
| V _P | 11 | supply voltage |
| COSCIB1 | 12 | UHF oscillator input base 1 |
| COSCOC1 | 13 | UHF oscillator output collector 1 |
| COSCOC2 | 14 | UHF oscillator output collector 2 |
| COSCIB2 | 15 | UHF oscillator input base 2 |
| n.c. | 16 | not connected |

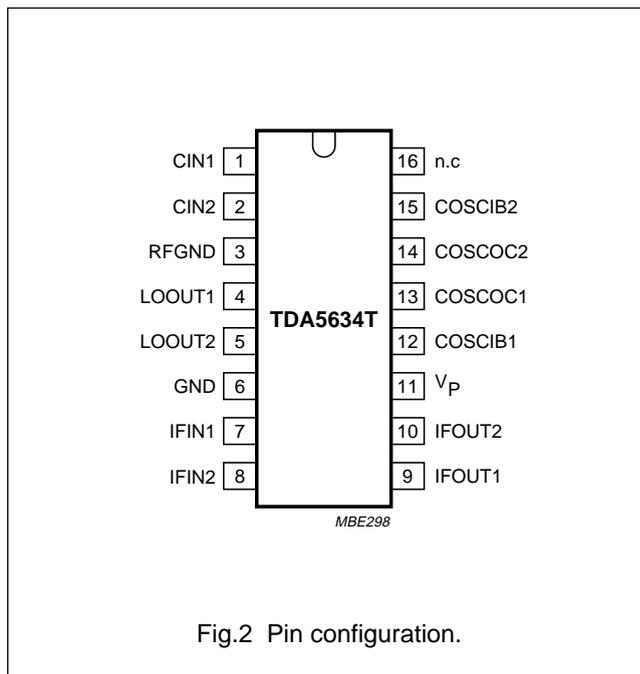


Fig.2 Pin configuration.

LIMITING VALUES

In accordance with the Absolute Maximum Rating System (IEC 134).

| SYMBOL | PARAMETER | MIN. | MAX. | UNIT |
|------------------|---|------|-------|------|
| V _P | supply voltage | -0.3 | +10.5 | V |
| I _O | output current of each pin referenced to ground | - | -10 | mA |
| t _{sc} | maximum short-circuit time (all pins) | - | 10 | s |
| T _{stg} | IC storage temperature | -55 | +150 | °C |
| T _{amb} | operating ambient temperature | -10 | +80 | °C |
| T _j | junction temperature | - | +150 | °C |

HANDLING

Human body model: the IC withstands 2000 V in accordance with the “UZW-BO/FQ-A302”, specification equivalent to the “MIL-STD-883C category B” (2000 V);

R = 1500 Ω, C = 100 pF.

Machine model: the IC withstands 200 V except pin 11 (175 V) in accordance with the “UZW-BO/FQ-B302”, specification (date of issue: Nov 6th, 1990);

R = 0 Ω, C = 200 pF.

THERMAL CHARACTERISTICS

| SYMBOL | PARAMETER | VALUE | UNIT |
|---------------------|---|-------|------|
| R _{th j-a} | thermal resistance from junction to ambient in free air | 120 | K/W |

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CHARACTERISTICS

$V_P = 9\text{ V}$; $T_{\text{amb}} = 25\text{ °C}$; measured in circuit of Fig.7; unless otherwise specified.

| SYMBOL | PARAMETER | CONDITIONS | MIN. | TYP. | MAX. | UNIT |
|---------------------------------------|---|---|------|------|------|------------|
| Supply | | | | | | |
| V_P | supply voltage | | 8.1 | 9.0 | 9.9 | V |
| I_P | supply current | $V_P = 8.1\text{ to }9.9\text{ V}$; $T_{\text{amb}} = -10\text{ to }+80\text{ °C}$ | – | 35 | 45 | mA |
| Mixer (including IF amplifier) | | | | | | |
| f_{RF} | RF frequency | note 1 | 430 | – | 860 | MHz |
| G_V | voltage gain | $f_{RF} = 430\text{ MHz}$; see Fig.3; note 2 | 33 | 36 | 39 | dB |
| | | $f_{RF} = 860\text{ MHz}$; see Fig.3; note 2 | 33 | 36 | 39 | dB |
| NF | noise figure (not corrected for image) | $f_{RF} = 430\text{ MHz}$; see Fig.4 | – | 9 | 11 | dB |
| | | $f_{RF} = 860\text{ MHz}$; see Fig.4 | – | 9 | 11 | dB |
| V_o | output voltage causing 1% cross modulation in channel | $f_{RF} = 430\text{ MHz}$; see Fig.5 | 115 | 118 | – | dB μ V |
| | | $f_{RF} = 860\text{ MHz}$; see Fig.5 | 118 | 121 | – | dB μ V |
| V_i | input voltage causing 10 kHz pulling in channel | $f_{RF} = 860\text{ MHz}$; note 3 | – | 87 | – | dB μ V |
| | input voltage causing $N + 5 - 1\text{ MHz}$ pulling | $f_{RF} = 820\text{ MHz}$; see Fig.6 | 65 | 72 | – | dB μ V |
| Z_i | input impedance ($R_S + jL_S\omega$) | R_S at $f_{RF} = 430\text{ MHz}$; see Fig.8; note 4 | – | 28 | – | Ω |
| | | R_S at $f_{RF} = 860\text{ MHz}$; see Fig.8; note 4 | – | 33 | – | Ω |
| | | L_S at $f_{RF} = 430\text{ to }860\text{ MHz}$; see Fig.8; note 4 | – | 8 | – | nH |
| IF amplifier | | | | | | |
| S_{11} | input reflection coefficient | magnitude; see Fig.10; note 4 | – | –0.6 | – | dB |
| | | phase; see Fig.10; note 4 | – | –2.5 | – | deg |
| S_{12} | reverse transmission coefficient | magnitude; see Fig.11; note 4 | – | –56 | – | dB |
| | | phase; see Fig.11; note 4 | – | 30 | – | deg |
| S_{21} | forward transmission coefficient | magnitude; note 4 | – | –9.5 | – | dB |
| | | phase; note 4 | – | 165 | – | deg |
| S_{22} | output reflection coefficient | magnitude; see Fig.11; note 4 | – | –7 | – | dB |
| | | phase; see Fig.11; note 4 | – | 6 | – | deg |
| Z_o | output impedance ($R_S + jL_S\omega$) | R_S ; see Fig.11; note 4 | – | 100 | – | Ω |
| | | L_S ; see Fig.11; note 4 | – | 32 | – | nH |
| Y_i | input admittance ($G_P + jC_P\omega$) | G_P ; see Fig.10; note 4 | – | 0.8 | – | mS |
| | | C_P ; see Fig.10; note 4 | – | 2.5 | – | pF |

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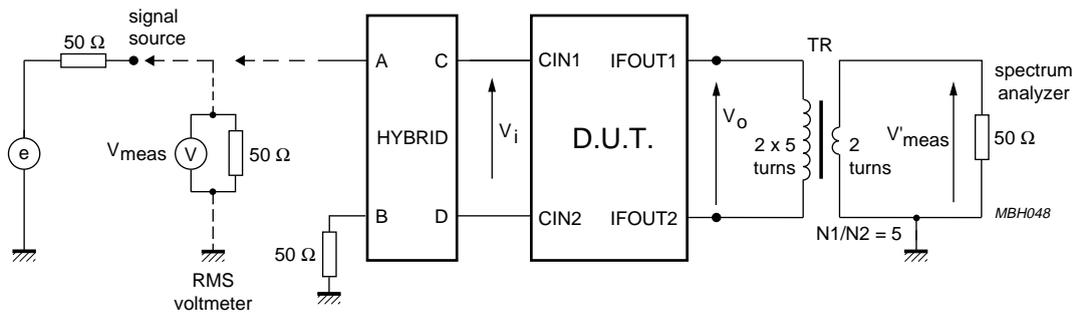
| SYMBOL | PARAMETER | CONDITIONS | MIN. | TYP. | MAX. | UNIT |
|---|---|--|------|------|------|------------------------|
| LO output; $R_L = 100 \Omega$ | | | | | | |
| Y_o | output admittance ($G_P + jC_P\omega$) | $f_{OSC} = 470 \text{ MHz}$; see Fig.9; note 4 | – | 3 | – | mS |
| | | | – | 0.5 | – | pF |
| | | $f_{OSC} = 900 \text{ MHz}$; see Fig.9; note 4 | – | 3.5 | – | mS |
| | | | – | 0.5 | – | pF |
| V_o | output voltage | $R_L = 50 \Omega$; $V_P = 8.1 \text{ to } 9.9 \text{ V}$; $T_{amb} = -10 \text{ to } +80 \text{ }^\circ\text{C}$ | 83 | 91 | 100 | $\text{dB}\mu\text{V}$ |
| SRF | spurious signal on LO output with respect to LO output signal | $R_L = 50 \Omega$; note 5 | – | – | –10 | dBc |
| SHD | LO signal harmonics with respect to LO signal | $R_L = 50 \Omega$ | – | – | –10 | dBc |
| Band C oscillator | | | | | | |
| f_{OSC} | oscillator frequency | $V_t = 0.45 \text{ to } 28 \text{ V}$; $V_P = 8.1 \text{ to } 9.9 \text{ V}$; $T_{amb} = -10 \text{ to } +80 \text{ }^\circ\text{C}$; note 6 | 470 | – | 900 | MHz |
| f_{shift} | frequency shift | $\Delta V_P = 10\%$; note 7 | – | – | 400 | kHz |
| f_{drift} | frequency drift | $\Delta T = 25 \text{ }^\circ\text{C}$ without compensation; NP0 capacitors; note 8 | – | – | 2.5 | MHz |
| | | $\Delta T = 25 \text{ }^\circ\text{C}$ with compensation; note 9 | – | – | 800 | kHz |
| | | 5 s to 15 minutes after switch on; without compensation; note 10 | – | – | 600 | kHz |

Notes

- The RF frequency range is defined by the oscillator frequency range and the intermediate frequency.
- The gain is defined as the transducer gain (measured in Fig.7) plus the voltage transformation ratio of L3 to L2 (10 : 2, 15.4 dB including transformer loss).
- The input level causing 10 kHz frequency detuning at the LO output; $f_{osc} = f_{RF} + 33.4 \text{ MHz}$.
- All S-parameters are referred to a 50Ω system.
- Measured with RF input voltage of $97 \text{ dB}\mu\text{V}$ at $430 \text{ MHz} < f_{RF} < 860 \text{ MHz}$.
- Limits are related to the tank circuits used in Fig.7. Frequency bands may be adjusted by the choice of external components.
- The frequency shift is defined as the change in oscillator frequency when the supply voltage varies from $V_P = 9 \text{ to } 8.1 \text{ V}$ or from $V_P = 9 \text{ to } 9.9 \text{ V}$.
- The frequency drift is defined as the change in oscillator frequency when the ambient temperature varies from $T_{amb} = 25 \text{ }^\circ\text{C}$ to $0 \text{ }^\circ\text{C}$ or from $T_{amb} = 25 \text{ }^\circ\text{C}$ to $50 \text{ }^\circ\text{C}$. Without compensation, the capacitors C11 to C15 are NP0.
- The frequency drift is defined as the change in oscillator frequency when the ambient temperature varies from $T_{amb} = 25 \text{ }^\circ\text{C}$ to $0 \text{ }^\circ\text{C}$ or from $T_{amb} = 25 \text{ }^\circ\text{C}$ to $50 \text{ }^\circ\text{C}$. With compensation, the capacitors C11 to C14 are N750 and C15 is N470.
- Switch on drift is defined as the change in oscillator frequency between 5 s and 15 min after switch on.

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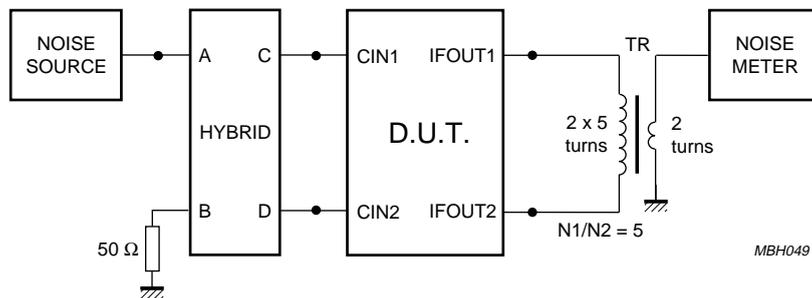
Loss of the hybrid: 1 dB.

$V_i = V_{meas} - \text{loss of the hybrid.}$

$V_o = V'_{meas} + 15.4 \text{ dB (transformer ratio } N1/N2 = 5 \text{ and transformer loss).}$

$G_v = 20 \log \frac{V_o}{V_i} = 20 \log \frac{V'_{meas}}{V_{meas}} + 15.4 \text{ dB} + \text{loss of the hybrid.}$

Fig.3 Gain measurement in band C.



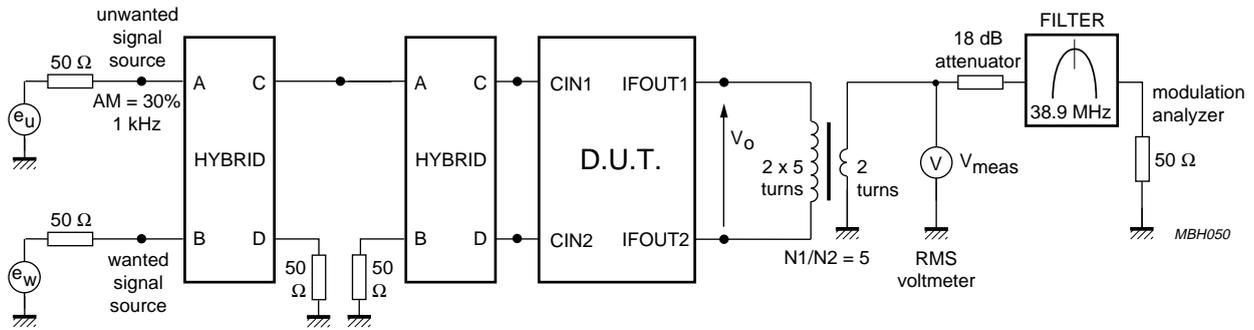
Loss of the hybrid: 1 dB.

$NF = NF_{meas} - \text{loss of the hybrid.}$

Fig.4 Noise figure measurement in band C.

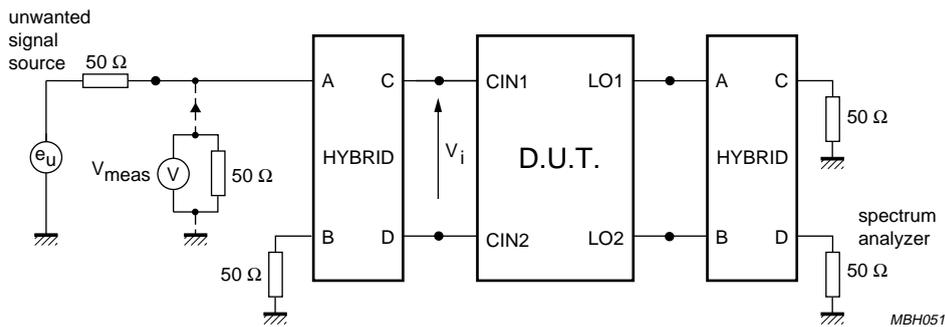
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$V_{meas} = V_o - 15.4 \text{ dB}$ (transformer ratio $N1/N2 = 5$ and transformer loss).
 Wanted output signal at f_{RFW} : $V_{ow} = 108 \text{ dB}\mu\text{V}$, $V_{meas} = 92.6 \text{ dB}\mu\text{V}$.
 Measuring the level of the unwanted output signal V_{ou} causing 0.3 % AM modulation in the wanted output signal.
 $V_{ou} = V_{meas} + 15.4 \text{ dB}$.
 $f_{RFU} = f_{RFW} + 5.5 \text{ MHz}$; $f_{osc} = f_{RF} + 38.9 \text{ MHz}$.
 Filter characteristics: $f_c = 38.9 \text{ MHz}$; $f_{-3\text{dB}} = 1.2 \text{ MHz}$; $f_{-30\text{dB}} = 2.64 \text{ MHz}$.

Fig.5 Cross modulation measurement in band C.

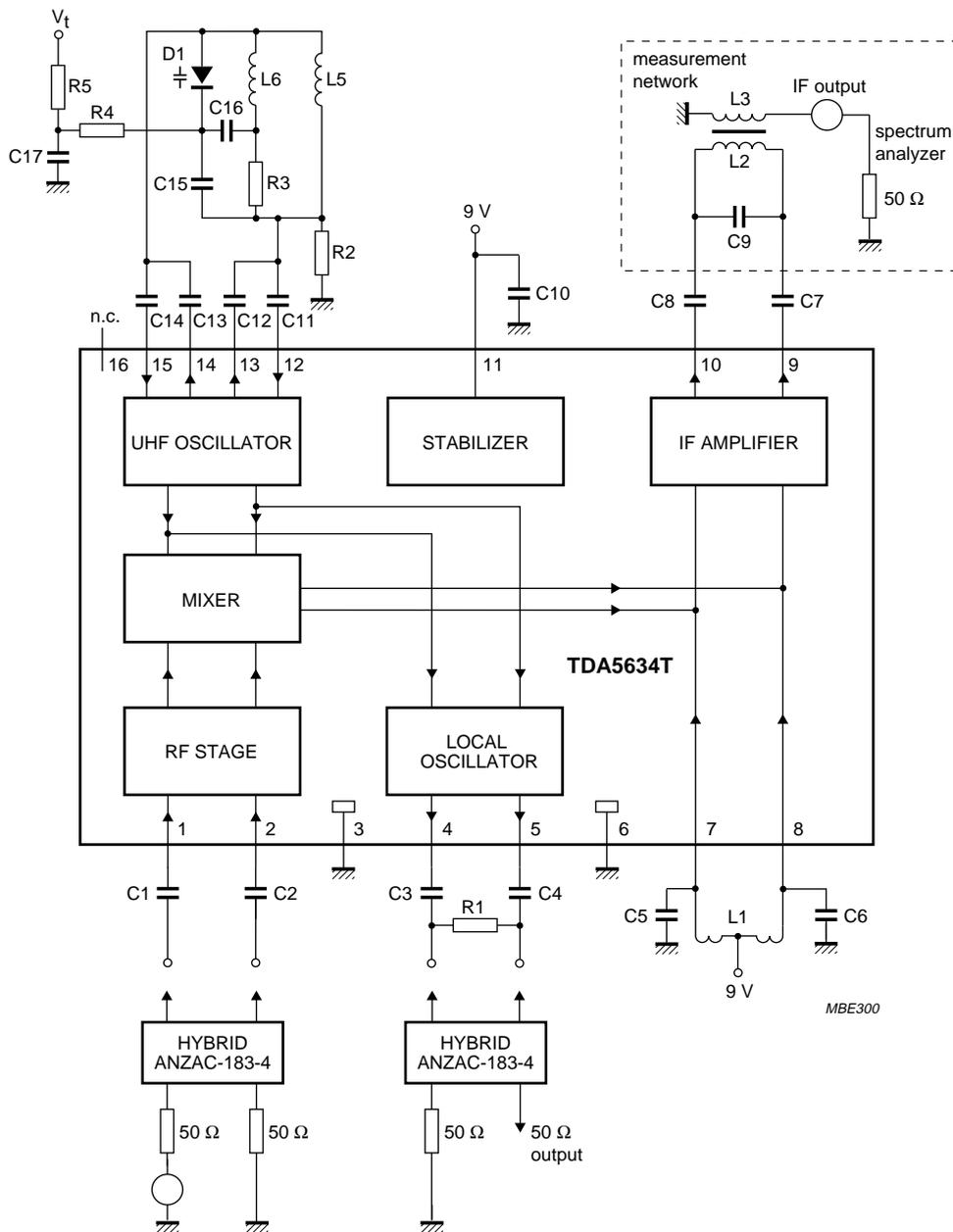


Loss of the hybrid: 1 dB.
 In band C: $f_{RFW} = 781 \text{ MHz}$. These wanted signals are not used during the measurement.
 In band C: $f_{osc} = 819.9 \text{ MHz}$.
 In band C: $f_{RFU} = 820 \text{ MHz} = f_{RFW} + 5 \times 8 \text{ MHz} - 1 \text{ MHz}$.
 Measuring the level of the unwanted signal V_{iu} causing FM sidebands 30 dB below the oscillator carrier at the LO output.
 $V_{iu} = V_{meas} - \text{loss of the hybrid}$.

Fig.6 N + 5 - 1 MHz pulling measurement in band C.

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L2, L3 and C9 are only required for measurement purposes; they are not used in a tuner.

Fig.7 Measurement circuit.

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Component values for measurement circuit

Table 1 Capacitors
(all SMD and NP0 except C11 to C15)

| COMPONENT | VALUE |
|-----------|---|
| C1 | 1 nF |
| C2 | 1 nF |
| C3 | 1 nF |
| C4 | 1 nF |
| C5 | 18 pF |
| C6 | 18 pF |
| C7 | 1 nF |
| C8 | 1 nF |
| C9 | 18 pF |
| C10 | 22 nF |
| C11 | 1 pF (N750) |
| C12 | 1 pF (N750) |
| C13 | 1 pF (N750) |
| C14 | 1 pF (N750) |
| C15 | 6 pF (N470) |
| C16 | 100 pF |
| C17 | 1 μ F (40 V electrolytic capacitor) |

Table 2 Resistors (all SMD)

| COMPONENT | VALUE |
|-----------|----------------|
| R1 | 100 Ω |
| R2 | 22 k Ω |
| R3 | 2.2 k Ω |
| R4 | 22 k Ω |
| R5 | 47 k Ω |

Table 3 Diodes and IC

| COMPONENT | VALUE |
|-----------|----------|
| D1 | BB215 |
| IC | TDA5634T |

Table 4 Coils (wire size 0.4 mm)

| COMPONENT | VALUE |
|-----------|----------------------------|
| L5 | 2.5 turns; diameter 3 mm |
| L6 | 2.5 turns; diameter 2.5 mm |

Table 5 Transformers; note 1

| COMPONENT | VALUE |
|-----------|--------------------|
| L1 | 2 \times 6 turns |
| L2 | 2 \times 5 turns |
| L3 | 2 turns |

Note

- Coil type: TOKO 7kN; material: 113kN; screw core 03-0093; pot core 04-0026.

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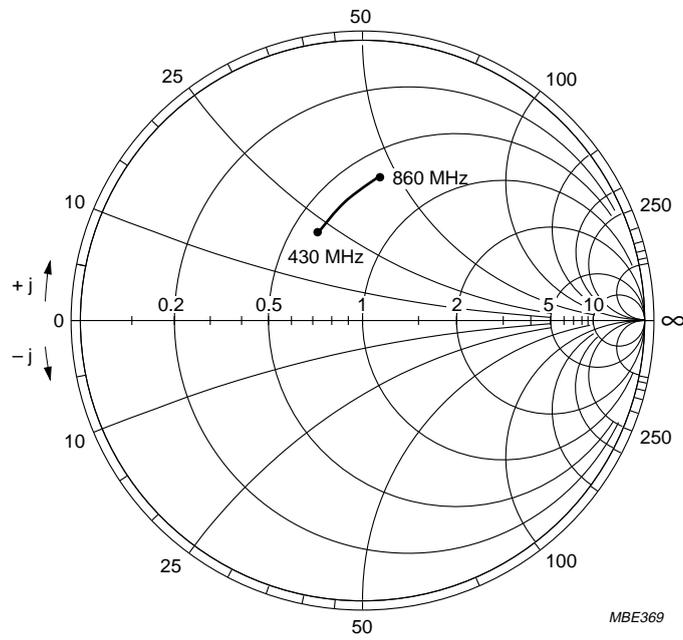


Fig.8 Input impedance (S₁₁) of the mixer input (430 to 860 MHz) (Z chart).

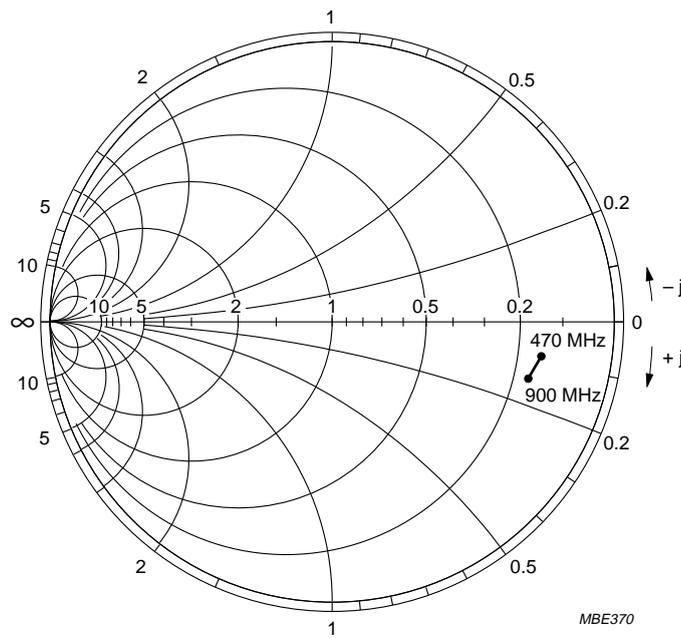


Fig.9 Output admittance (S₂₂) of the LO output (470 to 900 MHz) (Y chart).

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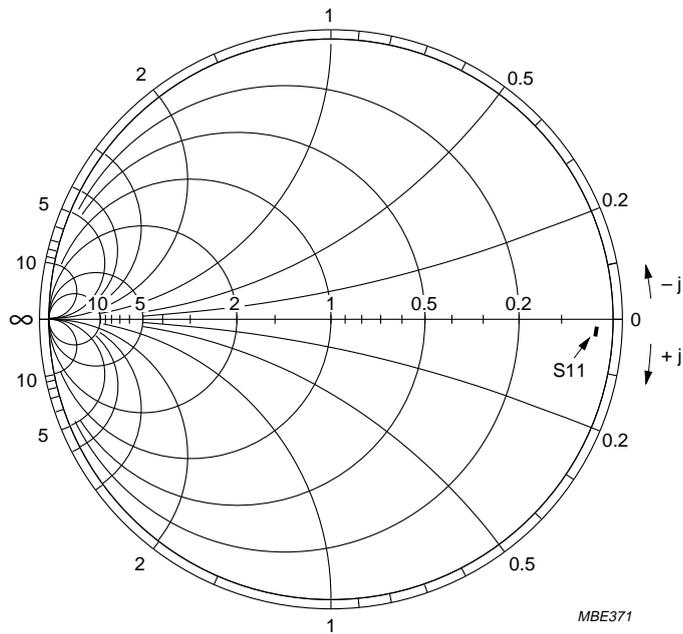


Fig.10 Input admittance (S_{11}) of the IF amplifier (25 to 45 MHz) (Y chart).

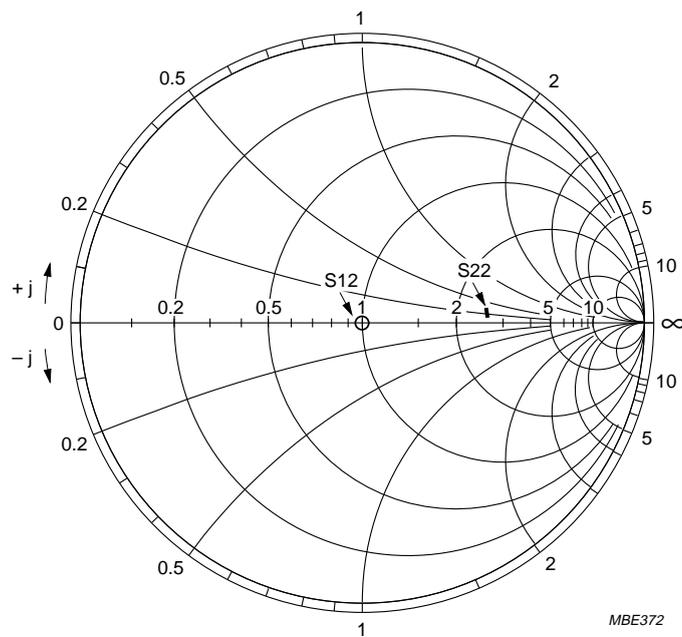
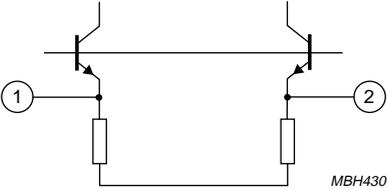
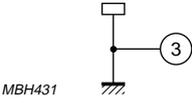
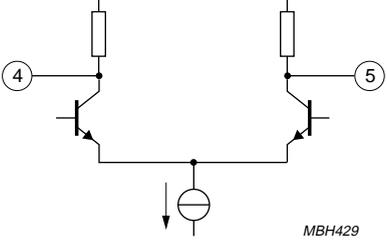
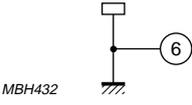
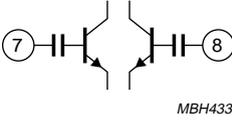
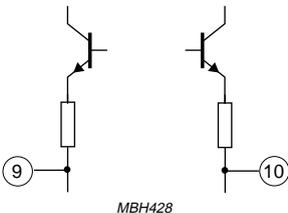


Fig.11 Reverse transmission and output reflection coefficient (S_{12} and S_{22}) of the IF amplifier (25 to 45 MHz) (Z chart).

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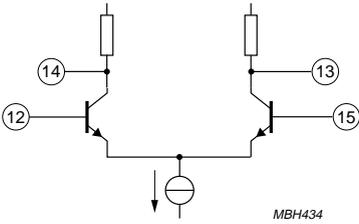
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INTERNAL PIN CONFIGURATION

| SYMBOL | PIN | DESCRIPTION | AVERAGE DC VOLTAGE ⁽¹⁾ IN (V) |
|--------|----------|--|--|
| | TDA5634T | | BAND C |
| CIN1 | 1 |  | 2.2 |
| CIN2 | 2 | | 2.2 |
| RFGND | 3 |  | 0.0 |
| LOOUT1 | 4 |  | 7.3 |
| LOOUT2 | 5 | | 7.3 |
| GND | 6 |  | 0 |
| IFIN1 | 7 |  | 9.0 |
| IFIN2 | 8 | | 9.0 |
| IFOUT1 | 9 |  | 3.8 |
| IFOUT2 | 10 | | 3.8 |

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| SYMBOL | PIN | DESCRIPTION | AVERAGE DC VOLTAGE ⁽¹⁾ IN (V) |
|----------------|----------|--|--|
| | TDA5634T | | BAND C |
| V _P | 11 | supply voltage | 9.0 |
| COSCIB1 | 12 |  | 2.3 |
| COSCOC1 | 13 | | 4.4 |
| COSCOC2 | 14 | | 4.4 |
| COSCIB2 | 15 | | 2.3 |
| n. c. | 16 | not connected | N.R. ⁽²⁾ |

Notes

1. Average DC voltage measured in circuit of Fig.7.
2. N.R. = Not Relevant.

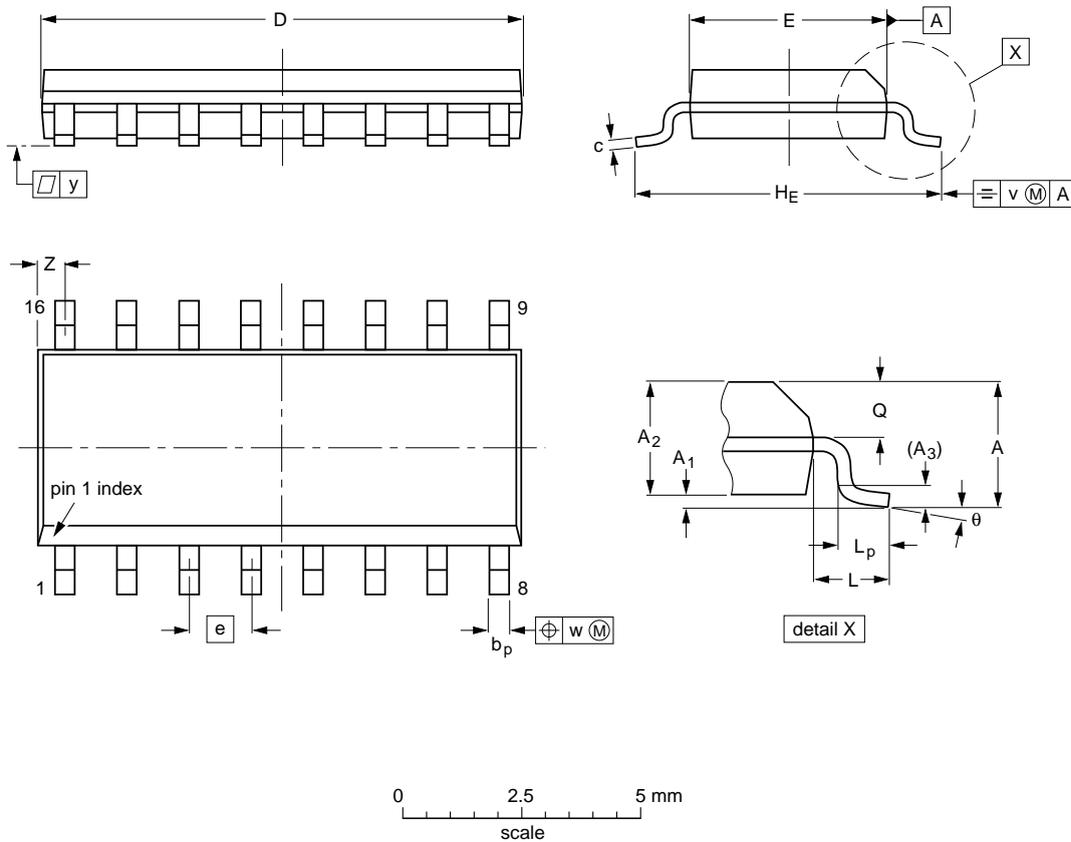
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PACKAGE OUTLINE

SO16: plastic small outline package; 16 leads; body width 3.9 mm

SOT109-1



DIMENSIONS (inch dimensions are derived from the original mm dimensions)

| UNIT | A max. | A ₁ | A ₂ | A ₃ | b _p | c | D ⁽¹⁾ | E ⁽¹⁾ | e | H _E | L | L _p | Q | v | w | y | z ⁽¹⁾ | θ |
|--------|--------|------------------|----------------|----------------|----------------|------------------|------------------|------------------|-------|----------------|-------|----------------|----------------|------|------|-------|------------------|----------|
| mm | 1.75 | 0.25 0.10 | 1.45 1.25 | 0.25 | 0.49 0.36 | 0.25 0.19 | 10.0 9.8 | 4.0 3.8 | 1.27 | 6.2 5.8 | 1.05 | 1.0 0.4 | 0.7 0.6 | 0.25 | 0.25 | 0.1 | 0.7 0.3 | 8° 0° |
| inches | 0.069 | 0.0098 0.0039 | 0.057 0.049 | 0.01 | 0.019 0.014 | 0.0098 0.0075 | 0.39 0.38 | 0.16 0.15 | 0.050 | 0.24 0.23 | 0.041 | 0.039 0.016 | 0.028 0.020 | 0.01 | 0.01 | 0.004 | 0.028 0.012 | |

Note

1. Plastic or metal protrusions of 0.15 mm maximum per side are not included.

| OUTLINE VERSION | REFERENCES | | | | EUROPEAN PROJECTION | ISSUE DATE |
|-----------------|------------|----------|------|--|---------------------|----------------------|
| | IEC | JEDEC | EIAJ | | | |
| SOT109-1 | 076E07S | MS-012AC | | | | 91-08-13 95-01-23 |

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SOLDERING

Introduction

There is no soldering method that is ideal for all IC packages. Wave soldering is often preferred when through-hole and surface mounted components are mixed on one printed-circuit board. However, wave soldering is not always suitable for surface mounted ICs, or for printed-circuits with high population densities. In these situations reflow soldering is often used.

This text gives a very brief insight to a complex technology. A more in-depth account of soldering ICs can be found in our "*IC Package Databook*" (order code 9398 652 90011).

Reflow soldering

Reflow soldering techniques are suitable for all SO packages.

Reflow soldering requires solder paste (a suspension of fine solder particles, flux and binding agent) to be applied to the printed-circuit board by screen printing, stencilling or pressure-syringe dispensing before package placement.

Several techniques exist for reflowing; for example, thermal conduction by heated belt. Dwell times vary between 50 and 300 seconds depending on heating method. Typical reflow temperatures range from 215 to 250 °C.

Preheating is necessary to dry the paste and evaporate the binding agent. Preheating duration: 45 minutes at 45 °C.

Wave soldering

Wave soldering techniques can be used for all SO packages if the following conditions are observed:

- A double-wave (a turbulent wave with high upward pressure followed by a smooth laminar wave) soldering technique should be used.
- The longitudinal axis of the package footprint must be parallel to the solder flow.
- The package footprint must incorporate solder thieves at the downstream end.

During placement and before soldering, the package must be fixed with a droplet of adhesive. The adhesive can be applied by screen printing, pin transfer or syringe dispensing. The package can be soldered after the adhesive is cured.

Maximum permissible solder temperature is 260 °C, and maximum duration of package immersion in solder is 10 seconds, if cooled to less than 150 °C within 6 seconds. Typical dwell time is 4 seconds at 250 °C.

A mildly-activated flux will eliminate the need for removal of corrosive residues in most applications.

Repairing soldered joints

Fix the component by first soldering two diagonally-opposite end leads. Use only a low voltage soldering iron (less than 24 V) applied to the flat part of the lead. Contact time must be limited to 10 seconds at up to 300 °C. When using a dedicated tool, all other leads can be soldered in one operation within 2 to 5 seconds between 270 and 320 °C.

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DEFINITIONS

| | |
|---|---|
| Data sheet status | |
| Objective specification | This data sheet contains target or goal specifications for product development. |
| Preliminary specification | This data sheet contains preliminary data; supplementary data may be published later. |
| Product specification | This data sheet contains final product specifications. |
| Limiting values | |
| Limiting values given are in accordance with the Absolute Maximum Rating System (IEC 134). Stress above one or more of the limiting values may cause permanent damage to the device. These are stress ratings only and operation of the device at these or at any other conditions above those given in the Characteristics sections of the specification is not implied. Exposure to limiting values for extended periods may affect device reliability. | |
| Application information | |
| Where application information is given, it is advisory and does not form part of the specification. | |

LIFE SUPPORT APPLICATIONS

These products are not designed for use in life support appliances, devices, or systems where malfunction of these products can reasonably be expected to result in personal injury. Philips customers using or selling these products for use in such applications do so at their own risk and agree to fully indemnify Philips for any damages resulting from such improper use or sale.