TOSHIBA BIPOLAR LINEAR INTEGRATED CIRCUIT SILICON MONOLITHIC

## TA8208H

## 5.8W DUAL AUDIO POWER AMPLIFIER

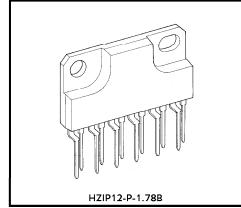
The TA8208H are dual audio power amplifier for consumer applications.

It is designed for high power, low distortion and low

It also contains various kind of protectors.

It is suitable for car-audio power amplifier with high performance.

$$P_{OUT} = 5.8W \text{ (Typ.) / ch}$$
  
( $V_{CC} = 13.2V, f = 1kHz, THD = 10\%, R_L = 4\Omega$ )



Weight: 4.04g (Typ.)

#### **FEATURES**

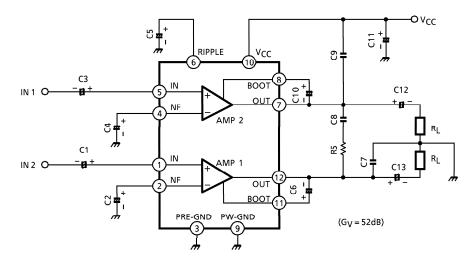
- Low Distortion
  - : THD = 0.06% (Typ.)  $(V_{CC} = 13.2V, f = 1kHz, P_{OUT} = 1W, R_{L} = 4\Omega, G_{V} = 52dB)$
- Low Noise
  - :  $V_{NO} = 0.7 \text{mV}_{rms}$  (Typ.)  $(V_{CC} = 13.2V, R_L = 4\Omega, G_V = 52dB, R_q = 10k\Omega, BW = 20Hz\sim20kHz)$
- **Protector Circuit** 
  - : Thermal Shut Down, Over Voltage Protection. Out-V<sub>CC</sub> short, Out-GND short and Out-Out Short Protection.
- **Operating Supply Voltage Range** 
  - :  $V_{CC (opr.)} = 9 \sim 18V$

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  This product generates heat during normal operation. However, substandard performance or malfunction may cause
- This product generates heat during normal operation. However, substandard performance or malfunction may cause the product and its peripherals to reach abnormally high temperatures.
- The product is often the final stage (the external output stage) of a circuit. Substandard performance or malfunction of the destination device to which the circuit supplies output may cause damage to the circuit or to the product.
- The products described in this document are subject to foreign exchange and foreign trade control laws.

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#### **BLOCK DIAGRAM**



#### **CAUTION AND APPLICATION METHOD**

#### 1. VOLTAGE GAIN ADJUSTMENT

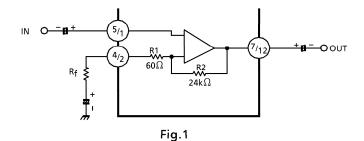
The voltage gain  $G_V$  is determined by R1, R2 and R $_{\rm f}$  in Fig.1.

$$G_V = 20 log \frac{R_f + R1 + R2}{R_f + R1}$$
 (dB)

In case of  $R_f = 0$ 

The voltage gain is set as follows:

$$G_V = 52dB$$
 (Typ.)



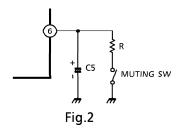
#### 2. MUTING

Audio muting can be accomplished by connecting pin 6 (ripple filter) to GND as shown in Fig.2. Then, the bias circuits are cut off.

However, caution must be exercised to the following items.

- (1) The recovery time from the muting operation is determined by the capacities of the ripple filter capacitor between 6 pin and GND, and of the capacitor for NF.
- (2) As this muting system is operated by the short-circuit of ripple filter: C5, the ripple rejection ratio becomes worse in the muting condition.

At mute ON, some pop noises of breaking the bias are produced.



#### 3. MEASURES AGAINST OSCILLATION

C7, C8, C9: The capacitor of polyester film having small variation of the temperature characteristics is recommended as the capacitor for oscillation prevention.

R5: Increases the oscillation allowance at output clippling of the low range frequency (100Hz or less).

Since the oscillation allowance varies according of the following items, carry out the temperature test to confirm the oscillation allowance.

- (1) Gain to be used (G<sub>V</sub> setting)
- (2) Capacity of capacitor
- (3) Kinds of capacitor
- (4) Layout of printed board

It is recommended to use the capacitor having the capacity exceeding the oscillation stopping value. At using with the voltage gain G<sub>V</sub> lowered or with the high range frequency and the oscillation becomes liable to be produced.

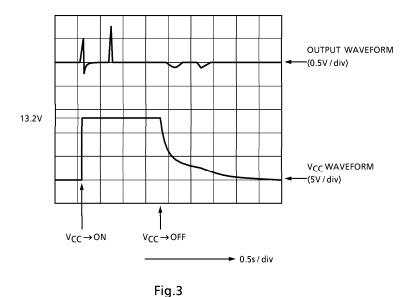
Therefore, Application with  $G_V = 40 dB$  or over is advisable.

Especially for using with the gain lowered in DUAL operation, insert 1000pF between NF (2 pin and 4 pin) and GND.

This capacitor reduces the gain of the high range frequency and is effective for oscillation prevention.

#### 4. OUTPUT WAVEFORM AT POWER SUPPLY ON/OFF (REFERENCE)

Amplifier DC output waveforms in the condition of  $V_{CC}$  = 13.2V,  $R_L$  =  $4\Omega$  at non-signal (input short) are shown in Fig.3.



The popping time at power supply ON/OFF varies according to the rise and fall times. The condition in which the power supply is rapidly and repeatedly made ON/OFF, that is, the condition in which the charge is made on the external capacitor of IC is different from the conditions shown in Fig.3.

## **MAXIMUM RATINGS** (Ta = 25°C)

CHARACTERISTIC	SYMBOL	RATING	UNIT
Peak Supply Voltage (0.2s)	V <sub>CC</sub> (surge)	45	V
DC Supply Voltage	VCC (DC)	25	V
Operating Supply Voltage	V <sub>CC</sub> (opr)	18	V
Output Current (peak)	<sup>I</sup> O (PEAK)	4.5	Α
Power Dissipation	P <sub>D</sub>	25	W
Operating Temperature	T <sub>opr</sub>	<b>-</b> 30∼85	°C
Storage Temperature	T <sub>stg</sub>	<b>-</b> 55∼150	°C

## **ELECTRICAL CHARACTERISTICS**

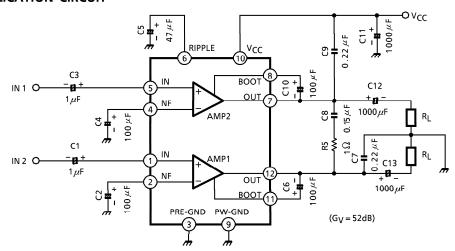
(Unless otherwise specified, V<sub>CC</sub> = 13.2V, R<sub>L</sub> =  $4\Omega$ , R<sub>g</sub> =  $600\Omega$ , f = 1kHz, Ta =  $25^{\circ}$ C)

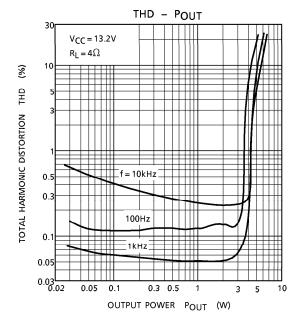
CHARACTERISTIC	SYMBOL	TEST CIR- CUIT	TEST CONDITION	MIN.	TYP.	MAX.	UNIT
Quiescent Current	lccQ		V <sub>IN</sub> = 0	_	80	145	mΑ
Output Power	POUT	<b>—</b>	THD = 10%	5	5.8	_	W
Total Harmonic Distortion	THD	_	P <sub>OUT</sub> = 1W	_	0.06	0.30	%
Voltage Gain	GV	_	$V_{OUT} = 0.775V_{rms}$ (0dBm)	50	52	54	dB
Voltage Gain Ratio	⊿G <sub>V</sub>	_	$V_{OUT} = 0.775V_{rms}$ (0dBm)	- 1	0	1	dB
Output Noise Voltage	V <sub>NO</sub>	_	$R_g = 10k\Omega$ BW = 20Hz~20kHz	_	0.7	1.5	mV <sub>rms</sub>
Ripple Rejection Ratio	R.R.	_	$f_{ripple}$ = 100Hz $V_{ripple}$ = 0.775 $V_{rms}$ (0dBm)	_	- 52	- 40	dB
Cross Talk	C.T.	_	$V_{OUT} = 0.775V_{rms}$ (0dBm)	_	- 57	_	dB
Input Resistance	RIN		f = 1kHz	_	33	_	kΩ

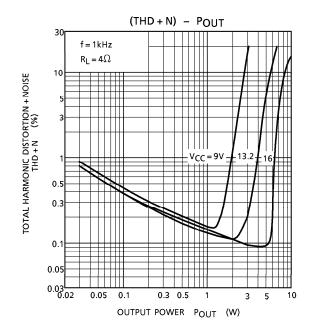
# TYPICAL DC VOLTAGE OF EACH TERMINAL ( $V_{CC} = 13.2V$ , $V_{IN} = 0V$ , $T_{a} = 25$ °C)

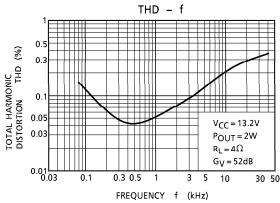
TERMINAL No.	1	2	3	4	5	6	7	8	9	10	11	12
DC Voltage (V)	1.5	1.5	GND	1.5	1.5	6.4	6.4	12.3	GND	۷cc	12.3	6.4

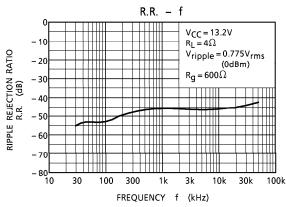
## TEST CIRCUIT / APPLICATION CIRCUIT

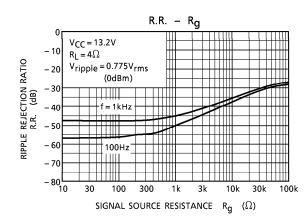


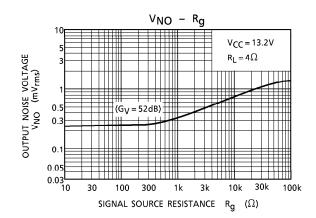


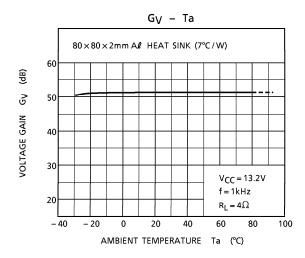


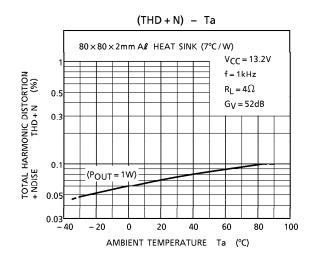


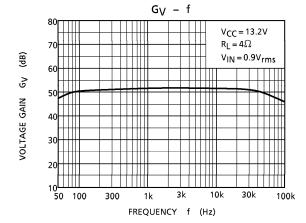


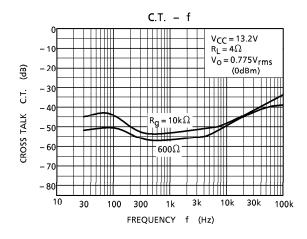


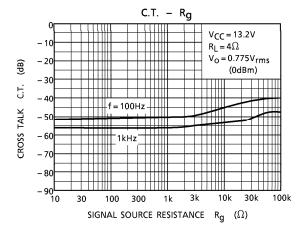


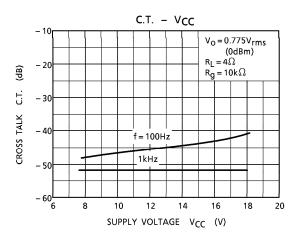


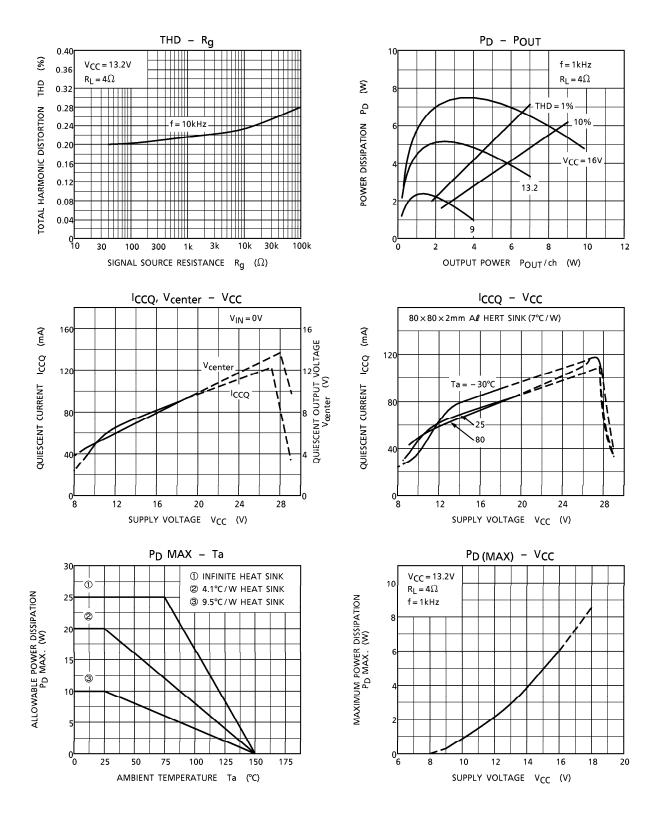








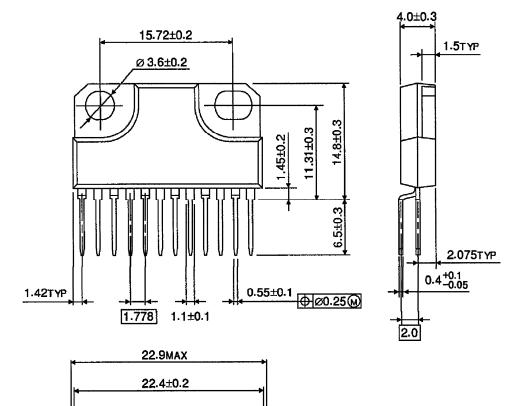




Unit: mm

## **OUTLINE DRAWING**

HZIP12-P-1.78B



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Weight: 4.04g (Typ.)