

# Precision Operational Amplifier

The LM308A operational amplifier provides high input impedance, low input offset and temperature drift, and low noise. These characteristics are made possible by use of a special Super Beta processing technology. This amplifier is particularly useful for applications where high accuracy and low drift performance are essential. In addition high speed performance may be improved by employing feedforward compensation techniques to maximize slew rate without compromising other performance criteria.

The LM308A offers extremely low input offset voltage and drift specifications allowing usage in even the most critical applications without external offset nulling.

- Operation from a Wide Range of Power Supply Voltages
- Low Input Bias and Offset Currents
- Low Input Offset Voltage and Guaranteed Offset Voltage Drift Performance
- High Input Impedance

## **Frequency Compensation Standard Compensation Modified Compensation** Inverting Input • Inverting R1 Noninverting R3 Output Cf Compen B Noninverting Input Compen B Input **Standard Feedforward Feedforward Compensations for** Compensation **Decoupling Load Capacitance** $R_S > 10 \text{ k}$ 100 k 10 k Input • 0.01 μF Output Output 30k

# **LM308A**

# SUPER GAIN OPERATIONAL AMPLIFIER

SEMICONDUCTOR TECHNICAL DATA



N SUFFIX PLASTIC PACKAGE CASE 626



D SUFFIX
PLASTIC PACKAGE
CASE 751
(SO-8)

# 

#### **ORDERING INFORMATION**

Device	Operating Temperature Range	Package
LM308AN LM308AD	$T_A = 0^\circ \text{ to } +70^\circ \text{C}$	Plastic DIP SO–8

### **MAXIMUM RATINGS** ( $T_A = +25$ °C, unless otherwise noted.)

Rating	Symbol	Value	Unit
Power Supply Voltage	VCC, VEE	±18	Vdc
Input Voltage (See Note 1)	VI	±15	V
Input Differential Current ( See Note 2)	l <sub>ID</sub>	±10	mA
Output Short Circuit Duration	tsc	Indefinite	
Operating Ambient Temperature Range	T <sub>A</sub>	0 to +70	°C
Storage Temperature Range	T <sub>stg</sub>	-65 to +150	°C
Junction Temperature	TJ	+150	°C

**NOTES:** 1. For supply voltages less than  $\pm 15$  V, the maximum input voltage is equal to the supply voltage.

# **ELECTRICAL CHARACTERISTICS** (Unless otherwise noted these specifications apply for supply voltages of +5.0 V $\leq$ V<sub>CC</sub> $\leq$ +15 V and -5.0 V $\geq$ V<sub>EE</sub> $\geq$ -15 V, T<sub>A</sub> = +25°C.)

Characteristic	Symbol	Min	Тур	Max	Unit
Input Offset Voltage	VIO	-	0.3	0.5	mV
Input Offset Current	lio	-	0.2	1.0	nA
Input Bias Current	I <sub>IB</sub>	-	1.5	7.0	nA
Input Resistance	rį	10	40	-	MΩ
Power Supply Currents $(V_{CC} = +15 \text{ V}, V_{EE} = -15 \text{ V})$	ICC, IEE	_	±0.3	±0.8	mA
Large Signal Voltage Gain (V <sub>CC</sub> = +15 V, V <sub>EE</sub> = $-15$ V, V <sub>O</sub> = $\pm 10$ V, R <sub>L</sub> $\geq 10$ k $\Omega$ )	A <sub>VOL</sub>	80	300	-	V/mV

#### The following specifications apply over the operating temperature range.

Input Offset Voltage	VIO	-	_	0.73	mV
Input Offset Current	IIO	-	-	1.5	nA
Average Temperature Coefficient of Input Offset Voltage $T_{A} \; (min) \leq T_{A} \leq T_{A} \; (max)$	ΔV <sub>IO</sub> /ΔΤ	_	1.0	5.0	μV/°C
Average Temperature Coefficient of Input Offset Current	ΔΙ <sub>ΙΟ</sub> /ΔΤ	-	2.0	10	pA/°C
Input Bias Current	I <sub>IB</sub>	-	-	10	nA
Large Signal Voltage Gain (V <sub>CC</sub> +15 V, V <sub>EE</sub> = $-15$ V, V <sub>O</sub> = $\pm 10$ V, R <sub>L</sub> $\geq 10$ k $\Omega$ )	AVOL	60	-	-	V/mV
Input Voltage Range (VCC = +15 V, VEE = -15 V)	VICR	±14	-	-	V
Common Mode Rejection $(R_{\mbox{\scriptsize S}} \leq 50 \ \mbox{\scriptsize k}\Omega)$	CMR	96	110	-	dB
Supply Voltage Rejection $(R_{\mbox{\scriptsize S}} \leq 50 \ \mbox{\scriptsize k}\Omega)$	PSR	96	110	_	dB
Output Voltage Range ( $V_{CC}$ = +15 V, $V_{EE}$ = -15 V, $R_L$ = 10 k $\Omega$ )	VOR	±13	±14	_	V

<sup>2.</sup> The inputs are shunted with back—to—back diodes for overvoltage protection. Therefore, excessive current will flow if a differential input voltage in excess of 1.0 V is applied between the inputs, unless some limiting resistance is used.

Figure 1. Input Bias and Input Offset Currents

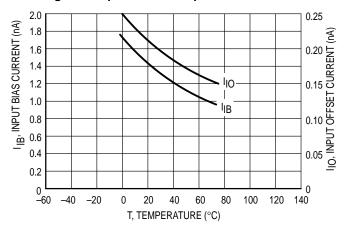


Figure 2. Maximum Equivalent Input Offset Voltage Error versus Input Resistance

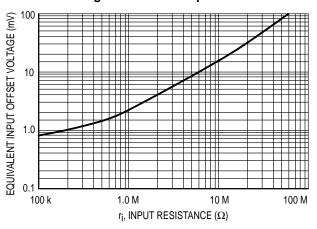


Figure 3. Voltage Gain versus Supply Voltages

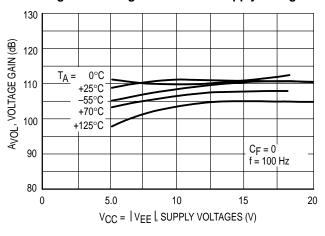


Figure 4. Power Supply Currents versus Power Supply Voltages

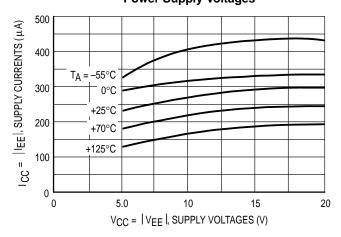


Figure 5. Open Loop Frequency Response

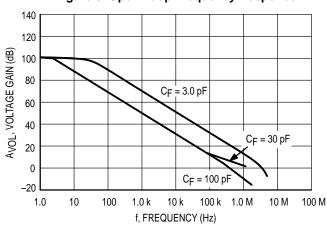
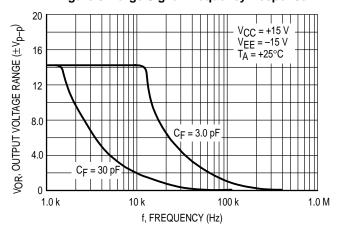


Figure 6. Large Signal Frequency Response

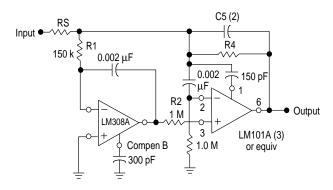


#### SUGGESTED DESIGN APPLICATIONS

#### **INPUT GUARDING**

Special care must be taken in the assembly of printed circuit boards to take full advantage of the low input currents of the LM308A amplifier. Boards must be thoroughly cleaned with alcohol and blown dry with compressed air. After cleaning, the boards should be coated with epoxy or silicone rubber to prevent contamination.

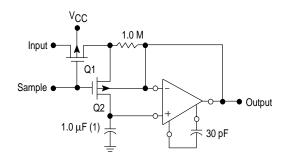
Figure 7. Fast (1) Summing Amplifier with Low Input Current



- Power Bandwidth: 250 kHz Small Signal Bandwidth: 3.5 MHz Slew Rate: 10 V/μs
- (2)  $C5 = \frac{6 \times 10^{-8}}{R1}$
- (3) In addition to increasing speed, the LM101A raises high and low frequency gain, increases output drive capability and eliminates thermal feedback.

Even with properly cleaned and coated boards, leakage currents may cause trouble at +125°C, particularly since the input pins are adjacent to pins that are at supply potentials. This leakage can be significantly reduced by using guarding to lower the voltage difference between the inputs and adjacent metal runs. The guard, which is a conductive ring surrounding the inputs, is connected to a low–impedance point that is at approximately the same voltage as the inputs. Leakage currents from high voltage pins are then absorbed by the guard.

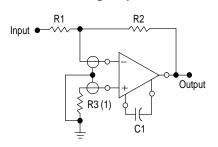
Figure 8. Sample and Hold



(1) Teflon, Polyethylene or Polycarbonate Dielectric Capacitor

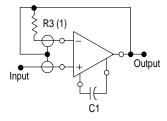
Figure 9. Connection of Input Guards

#### **Inverting Amplifier**

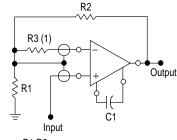


(1) Used to compensate for large source resistances.

#### **Follower**

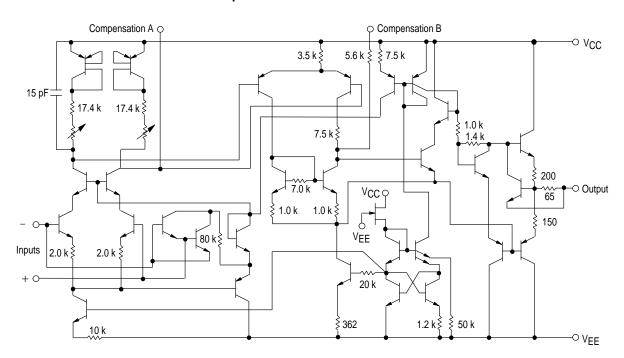


#### Noninverting Amplifier

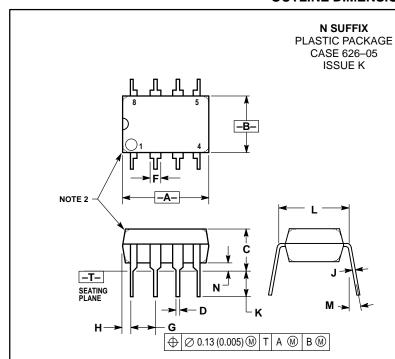


Note:  $\frac{R1 R2}{R1 + R2}$  must be an impedance.

### **Representative Circuit Schematic**

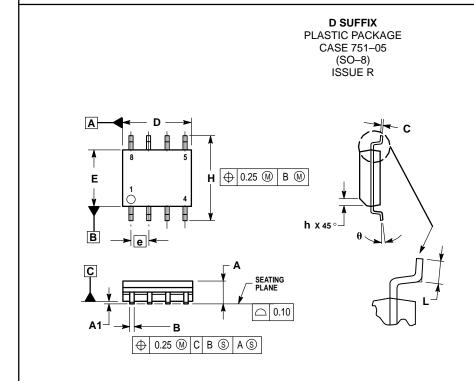


### **OUTLINE DIMENSIONS**



- NOTES:
  1. DIMENSION L TO CENTER OF LEAD WHEN FORMED PARALLEL.
  2. PACKAGE CONTOUR OPTIONAL (ROUND OR SQUARE CORNERS).
  3. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982.

	MILLIN	IETERS	INCHES			
DIM	MIN	MAX	MIN	MAX		
Α	9.40	10.16	0.370	0.400		
В	6.10	6.60	0.240	0.260		
С	3.94	4.45	0.155	0.175		
D	0.38	0.51	0.015	0.020		
F	1.02	1.78	0.040	0.070		
G	2.54	BSC	0.100	BSC		
Н	0.76	1.27	0.030	0.050		
J	0.20	0.30	0.008	0.012		
K	2.92	3.43	0.115	0.135		
L	7.62 BSC		0.300			
M		10°		10°		
N	0.76	1.01	0.030	0.040		



- NOTES:
  1. DIMENSIONING AND TOLERANCING PER ASME Y14.5M, 1994.
  2. DIMENSIONS ARE IN MILLIMETERS.
  3. DIMENSION D AND E DO NOT INCLUDE MOLD PROTECTION OF AND EDO NOT INCLUDE MOLD PROTECTION OF AND EDO NOT INCLUDE MOLD PROT PROTRUSION.

  4. MAXIMUM MOLD PROTRUSION 0.15 PER SIDE.
- DIMENSION B DOES NOT INCLUDE MOLD PROTRUSION. ALLOWABLE DAMBAR PROTRUSION. SHALL BE 0.127 TOTAL IN EXCESS OF THE B DIMENSION AT MAXIMUM MATERIAL.

	MILLIMETERS			
DIM	MIN	MAX		
Α	1.35	1.75		
A1	0.10	0.25		
В	0.35	0.49		
C	0.18	0.25		
D	4.80	5.00		
Е	3.80	4.00		
е	1.27	1.27 BSC		
Η	5.80	6.20		
h	0.25	0.50		
L	0.40	1.25		
θ	0 °	7 °		

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LM308A/D