

# DATA SHEET

## **TDA8776A** 10-bit, 1000 Msps Digital-to-Analog Converter (DAC)

Product specification  
Supersedes data of 1995 May 11  
File under Integrated Circuits, IC02

1996 Jun 04

# 10-bit, 1000 Msps Digital-to-Analog Converter (DAC)

TDA8776A

**FEATURES**

- 10-bit resolution
- Conversion rate up to 1000 MHz
- 10K/100K ECL input levels
- Internal reference voltage generator
- No deglitching circuit required
- Internal input register
- Power dissipation only 925 mW (typical)
- Internal 50 Ω output load (connected to the analog ground)
- Very few external components required.

**APPLICATIONS**

- High-speed digital-to-analog conversion for:
- High resolution video and graphics
  - Direct digital synthesis (DDS)
  - Telecommunication
  - High-speed modems.

**GENERAL DESCRIPTION**

The TDA8776A is a 10-bit Digital-to-Analog Converter (DAC) for high resolution video and other high frequency applications. It converts the digital input signal into an analog output voltage at a maximum conversion rate of 1000 Msps. No external reference voltage is required and all digital inputs are 10K/100K-ECL compatible.

**QUICK REFERENCE DATA**

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
$V_{EEA}$	analog supply voltage		-5.46	-5.20	-4.94	V
$V_{EED}$	digital supply voltage		-5.46	-5.20	-4.94	V
$V_{EEI}$	input stages digital supply voltage	note 1	-5.46	-5.20	-4.94	V
$I_{EEA}$	analog supply current	note 1	-	108	145	mA
$I_{EED}$	digital supply current	note 1	-	60	85	mA
$I_{EEI}$	input stages digital supply current	note 1	-	10	15	mA
$V_{OUT} - V_{\overline{OUT}}$	full-scale analog output voltage (peak-to-peak value)	notes 1 and 2; $Z_L = 50 \Omega$	1.7	2.0	2.5	V
INL	DC integral non-linearity	note 3	-	$\pm 0.3$	$\pm 0.5$	LSB
DNL	DC differential non-linearity	note 3	-	$\pm 0.2$	$\pm 0.45$	LSB
$f_{clk(max)}$	maximum clock frequency		1000	-	-	MHz
$t_{S1}$	settling time (differential)	10% to 90% full scale; Fig.9	-	0.5	-	ns
$P_{tot}$	total power dissipation		-	925	-	mW

**Notes**

1. D0 to D9 connected to either HIGH or LOW level, CLK is HIGH and  $\overline{CLK}$  is LOW.
2. The analog output voltages ( $V_{OUT}$  and  $V_{\overline{OUT}}$ ) are negative with respect to AGND (see Table 1). The external output resistance between AGND and each of these outputs is typically 50 Ω.
3. A warm-up time is necessary to reach optimal performances.

**ORDERING INFORMATION**

TYPE NUMBER	PACKAGE		
	NAME	DESCRIPTION	VERSION
TDA8776AK	PLCC28	plastic leaded chip carrier; 28 leads	SOT261-2

# 10-bit, 1000 Msps Digital-to-Analog Converter (DAC)

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## BLOCK DIAGRAM

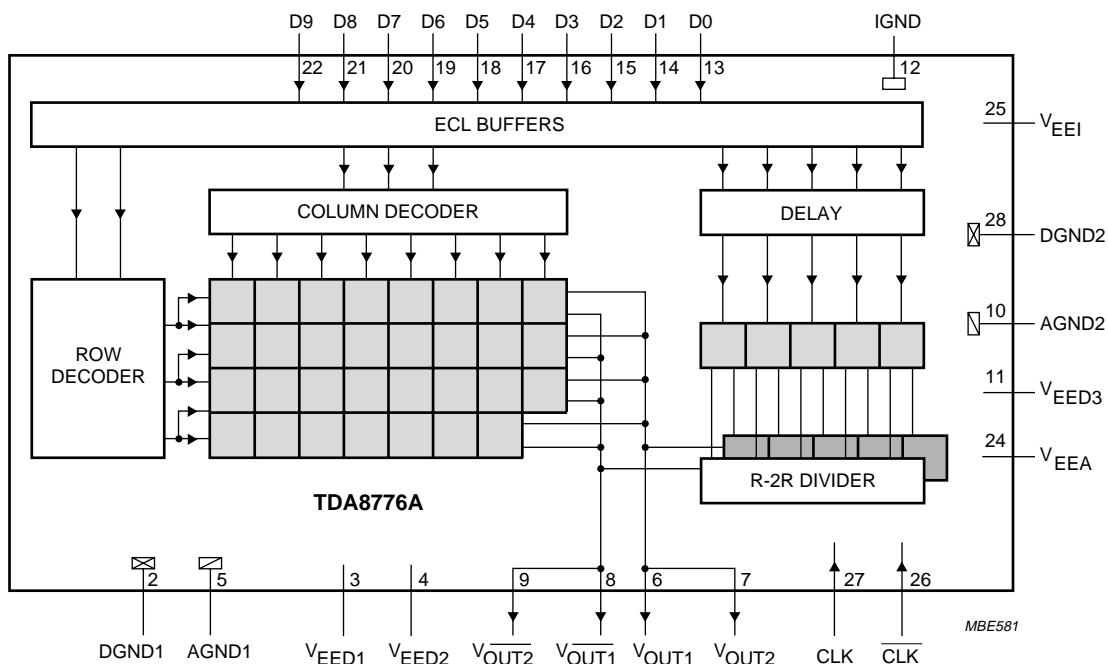


Fig.1 Block diagram.

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## PINNING

SYMBOL	PIN	DESCRIPTION
n.c.	1	not connected
DGND1	2	digital ground 1
V <sub>EED1</sub>	3	digital supply voltage 1 (-5.2 V)
V <sub>EED2</sub>	4	digital supply voltage 2 (-5.2 V)
AGND1	5	analog ground 1
V <sub>OUT1</sub>	6	analog voltage output 1
V <sub>OUT2</sub>	7	analog voltage output 2
V <sub>OUT1</sub>	8	complementary analog voltage output 1
V <sub>OUT2</sub>	9	complementary analog voltage output 2
AGND2	10	analog ground 2
V <sub>EED3</sub>	11	digital supply voltage 3 (-5.2 V)
IGND	12	input ground for ECL input buffers
D0	13	data input; bit 0 (LSB)
D1	14	data input; bit 1

SYMBOL	PIN	DESCRIPTION
D2	15	data input; bit 2
D3	16	data input; bit 3
D4	17	data input; bit 4
D5	18	data input; bit 5
D6	19	data input; bit 6
D7	20	data input; bit 7
D8	21	data input; bit 8
D9	22	data input; bit 9 (MSB)
n.c.	23	not connected
V <sub>EEA</sub>	24	analog supply voltage (-5.2 V)
V <sub>EEI</sub>	25	input supply voltage for ECL input buffers (-5.2 V)
CLK	26	complementary clock input
CLK	27	clock input
DGND2	28	digital ground 2

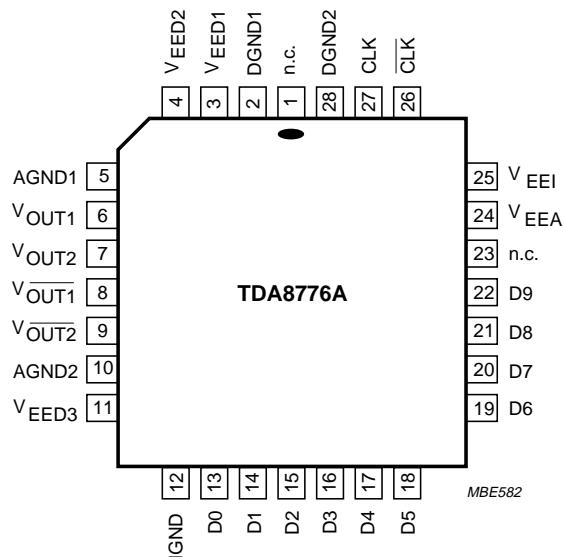


Fig.2 Pin configuration.

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**LIMITING VALUES**

In accordance with the Absolute Maximum Rating System (IEC 134).

SYMBOL	PARAMETER	CONDITIONS	MIN.	MAX.	UNIT
$V_{EEA}$	analog supply voltage		-7.0	$\pm 0.3$	V
$V_{EED}$	digital supply voltage		-7.0	$\pm 0.3$	V
$V_{EEI}$	input stages digital supply voltage		-7.0	$\pm 0.3$	V
$V_{EEA} - V_{EED}$	supply voltage differential		-0.5	+0.5	V
$AGND - DGND$	ground voltage differential		-0.1	+0.1	V
$V_I$	input voltage		$V_{EEI}$	$\pm 0.3$	V
$I_{OUT}/I_{\bar{OUT}}$	total output current	$Z_L = 50 \Omega$	-5	+50	mA
$T_{stg}$	storage temperature		-55	+150	°C
$T_{amb}$	operating ambient temperature		0	70	°C
$T_j$	junction temperature		-	150	°C

**HANDLING**

Inputs and outputs are protected against electrostatic discharges in normal handling. However, to be totally safe, it is desirable to take normal precautions appropriate to handling integrated circuits.

**THERMAL CHARACTERISTICS**

SYMBOL	PARAMETER	VALUE	UNIT
$R_{th\ j-a}$	thermal resistance from junction to ambient in free air	55 (typ.)	K/W

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## CHARACTERISTICS

$V_{EEA} = V_24$  to  $V_5$  and  $V_{10} = -5.46$  to  $-4.94$  V;  $V_{EED} = V_3, V_4$  and  $V_{11}$  to  $V_2$  and  $V_{28} = -5.46$  to  $-4.94$  V;  
 $V_{EEI} = V_{25}$  to  $V_{12} = -5.46$  to  $-4.94$  V;  $V_{EED}$  and  $V_{EEI}$  shorted together;  $T_{amb} = 0$  to  $+70$  °C; AGND, DGND and IGND  
shorted together;  $V_{OUT} - V_{OUT} = 2$  V (p-p);  $Z_L = 50 \Omega$ ; unless otherwise specified (typical values measured at  
 $V_{EEA} = V_{EED} = -5.2$  V and  $T_{amb} = 25$  °C).

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
<b>Supply</b>						
$V_{EEA}$	analog supply voltage		-5.46	-5.20	-4.94	V
$V_{EED}$	digital supply voltage		-5.46	-5.20	-4.94	V
$V_{EEI}$	input stages digital supply voltage	note 1	-5.46	-5.20	-4.94	V
$I_{EEA}$	analog supply current	note 1	-	108	145	mA
$I_{EED}$	digital supply current	note 1	-	60	85	mA
$I_{EEI}$	input stages digital supply current	note 1	-	10	15	mA
AGND – DGND	ground voltage differential		-0.1	-	+0.1	V
<b>Inputs</b>						
DIGITAL INPUTS (D9 TO D0) AND CLOCK INPUTS (CLK AND $\overline{CLK}$ )						
$V_{IL}$	LOW level input voltage		-1.9	-1.8	-1.6	V
$V_{IH}$	HIGH level input voltage		-1.2	-0.9	-0.8	V
$I_{IL}$	LOW level input current	$V_I = -1.8$ V	-	-	10	$\mu A$
$I_{IH}$	HIGH level input current	$V_I = -0.9$ V	-	-	20	$\mu A$
$f_{clk(max)}$	maximum clock frequency		1000	-	-	MHz
<b>Outputs (referenced to AGND); notes 1 and 2</b>						
$V_{OUT} - V_{\overline{OUT}(p-p)}$	full-scale analog output voltage (peak-to-peak value)	$Z_L = 50 \Omega$	1.7	2.0	2.5	V
$Z_O$	output impedance		-	50	-	$\Omega$
<b>Transfer function</b>						
INL	DC integral non-linearity	note 3	-	$\pm 0.3$	$\pm 0.5$	LSB
DNL	DC differential non-linearity	note 3	-	$\pm 0.2$	$\pm 0.45$	LSB
<b>Spurious free dynamic range (<math>f_{clk} = 1000</math> MHz); <math>V_{EEA} = V_{EED} = 5.2</math> V; <math>T_{amb} = 25</math> °C; note 4; see Fig.3</b>						
SFDR	spurious free dynamic range $f_{OUT} = 10$ MHz $f_{OUT} = 50$ MHz $f_{OUT} = 100$ MHz $f_{OUT} = 200$ MHz		-65	-69	-	dB
			-	-60	-	dB
			-52	-57	-	dB
			-	-46	-	dB

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SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
<b>Switching characteristics (<math>f_{clk} = 1000</math> MHz); notes 5 and 6; see Figs 8 and 9</b>						
$t_{SU;DAT}$	data set-up time		—	400	500	ps
$t_{HD;DAT}$	data hold time		100	150	—	ps
$t_{PD}$	propagation delay time		—	0.8	0.9	ns
$t_{S1}$	settling time	10% to 90% full scale	—	0.5	—	ns
$t_{S2}$	settling time	change to $\pm 1$ LSB	—	2.0	—	ns
$t_d$	input to 50% output delay time		—	1.4	1.5	ns
<b>Output transients; glitches (<math>f_{clk} = 1000</math> MHz); note 7; see Fig.10</b>						
$E_g$	differential glitch energy from code transition 511 to 512		—	1	2	pV.s

## Notes

1. D0 to D9 connected to either HIGH or LOW level, CLK is HIGH and  $\overline{CLK}$  is LOW.
2. The analog output voltages ( $V_{OUT}$  and  $V_{\overline{OUT}}$ ) are negative with respect to AGND (see Table 1). The external output resistance between AGND and each of these outputs is typically  $50 \Omega$ .
3. Due to on-chip regulator behaviour a warm-up time is necessary to reach optimal performances; a typical time is 1 minute.
4. Devices with higher SFDR (min.) can be delivered on special request.
5. The worst case characteristics are obtained at the transition from input code 0 to 1023 and if an external load impedance greater than  $50 \Omega$  is connected between  $V_{OUT}$  or  $V_{\overline{OUT}}$  and AGND in parallel with the external  $50 \Omega$  load. The specified values have been measured directly on a  $50 \Omega$  load between  $V_{OUT}$  and AGND. No further load impedance between  $V_{\overline{OUT}}$  and AGND has been applied. All input data is latched at the falling edge of the clock.
6. The data set-up ( $t_{SU;DAT}$ ) is the minimum period preceding the falling edge of the clock that the input data must be stable in order to be correctly registered. A negative set-up time indicates that the data may be initiated after the falling edge of the clock and still be recognized. The data hold time ( $t_{HD;DAT}$ ) is the minimum period following the falling edge of the clock that the input data must be stable in order to be correctly registered. A negative hold time indicates that the data may be released prior to the falling edge of the clock and still be recognized.
7. The definition of glitch energy and the measurement set-up are shown in Fig.10. The glitch energy is measured at the input transition between code 511 to 512.

**Table 1** Input coding and DAC output voltages (typical values; referenced to AGND regardless of the offset voltage)

CODE	BINARY INPUT DATA										DAC OUTPUT VOLTAGES (V) $Z_L = 50 \Omega$	
	D9	D8	D7	D6	D5	D4	D3	D2	D1	D0	$V_{OUT}$	$V_{\overline{OUT}}$
0	0	0	0	0	0	0	0	0	0	0	0	-1.0
1	0	0	0	0	0	0	0	0	0	1	-0.0010	-0.9990
.	.	.	.	.	.	.	.	.	.	.	.	.
512	1	0	0	0	0	0	0	0	0	0	-0.5	-0.5
.	.	.	.	.	.	.	.	.	.	.	.	.
1022	1	1	1	1	1	1	1	1	1	0	-0.9990	-0.0010
1023	1	1	1	1	1	1	1	1	1	1	-1.0	0

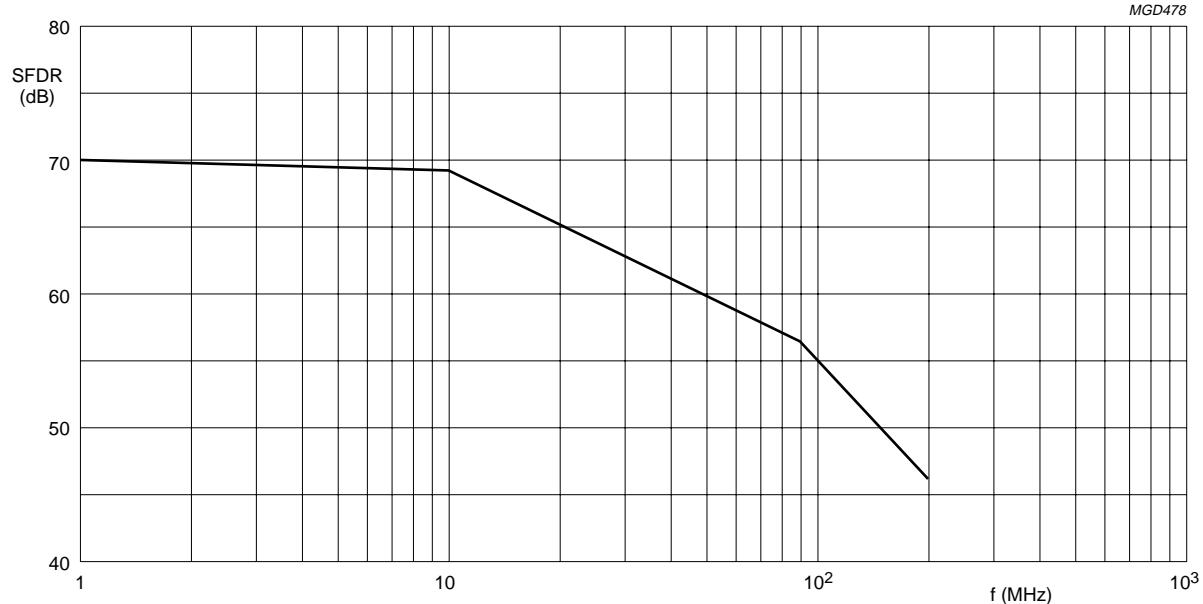
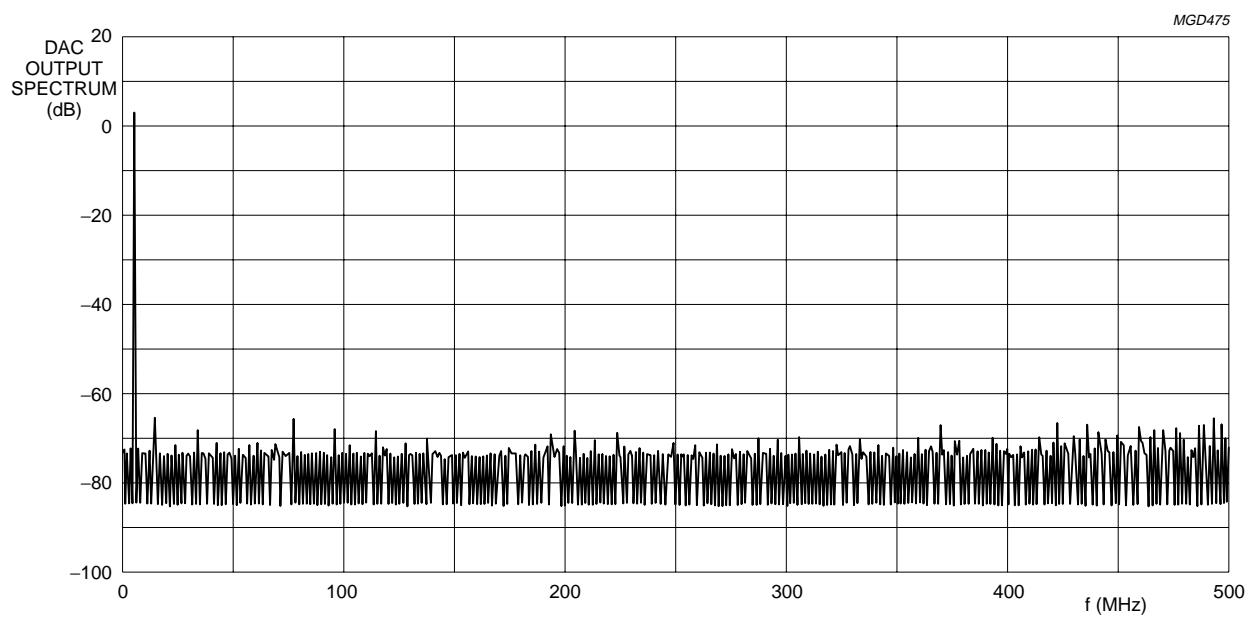
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Fig.3 Typical spurious free dynamic range (SFDR) as a function of output frequency.

Fig.4 Typical output spectrum;  $f_{\text{clk}} = 1000$  MHz;  $f_{\text{OUT}} = 10$  MHz.

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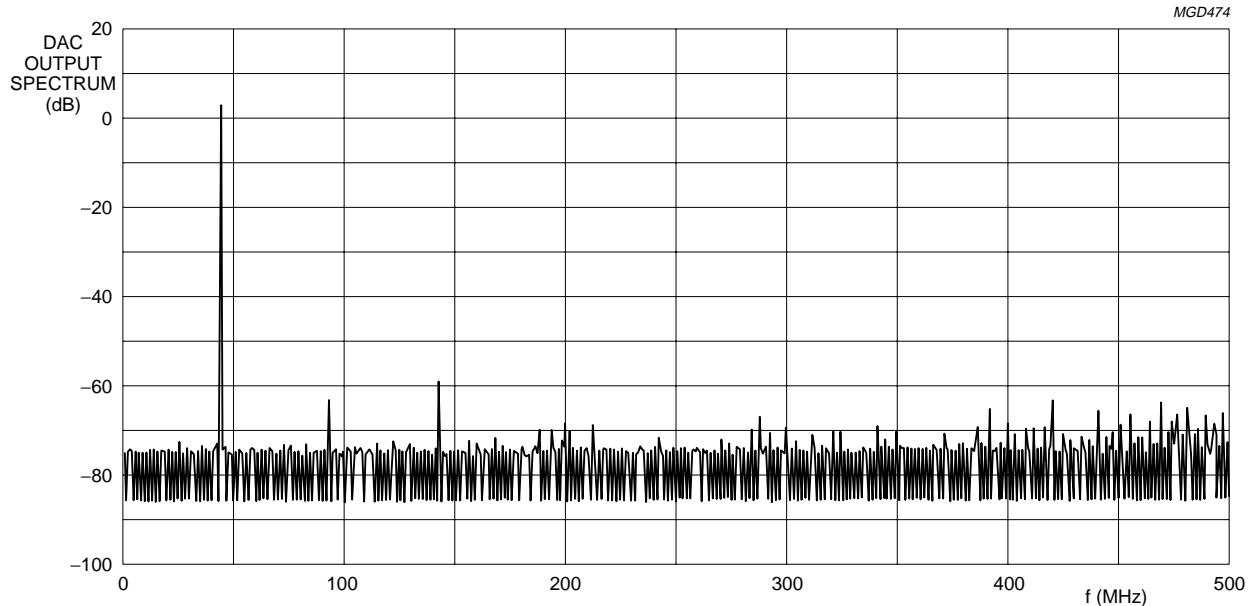


Fig.5 Typical output spectrum;  $f_{\text{clk}} = 1000 \text{ MHz}$ ;  $f_{\text{OUT}} = 50 \text{ MHz}$ .

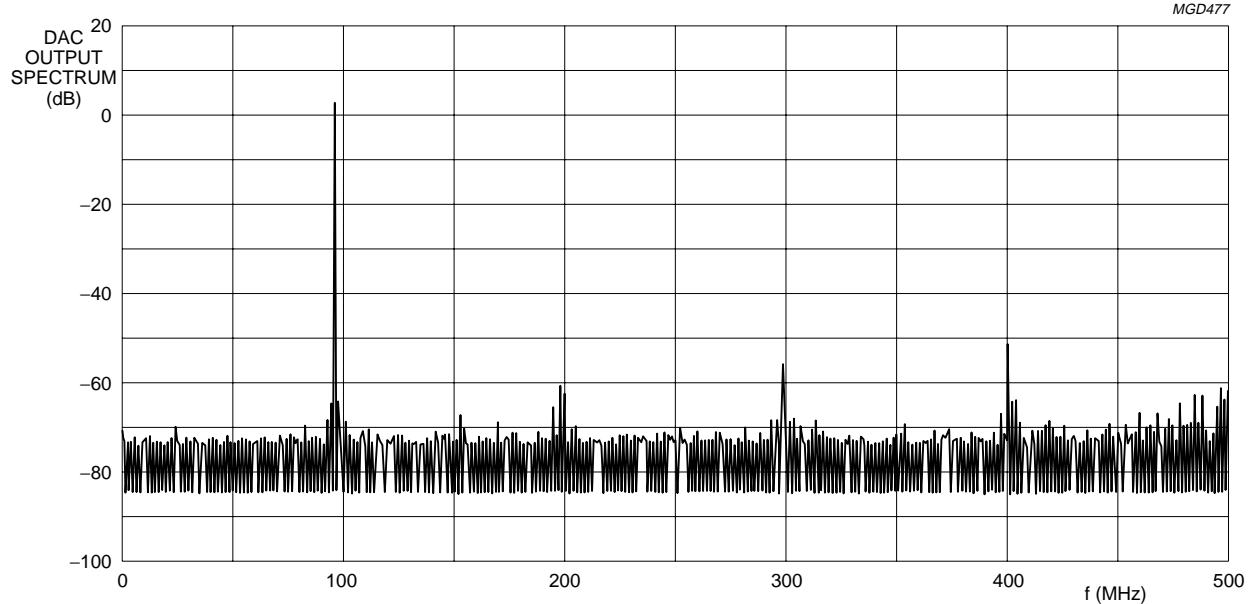


Fig.6 Typical output spectrum;  $f_{\text{clk}} = 1000 \text{ MHz}$ ;  $f_{\text{OUT}} = 100 \text{ MHz}$ .

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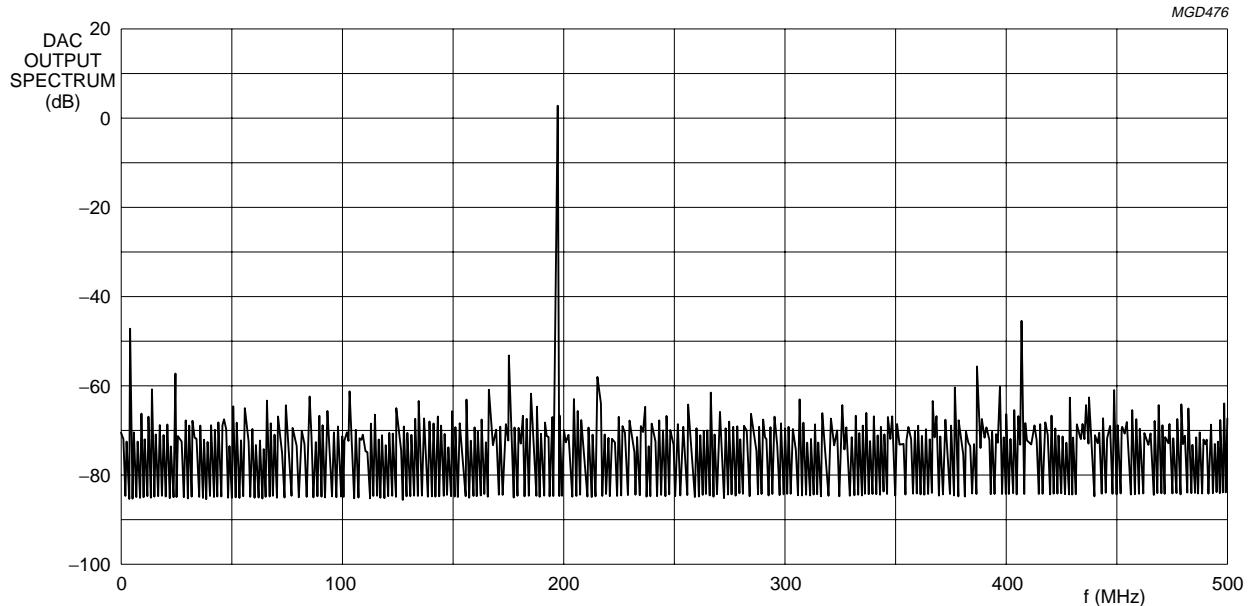
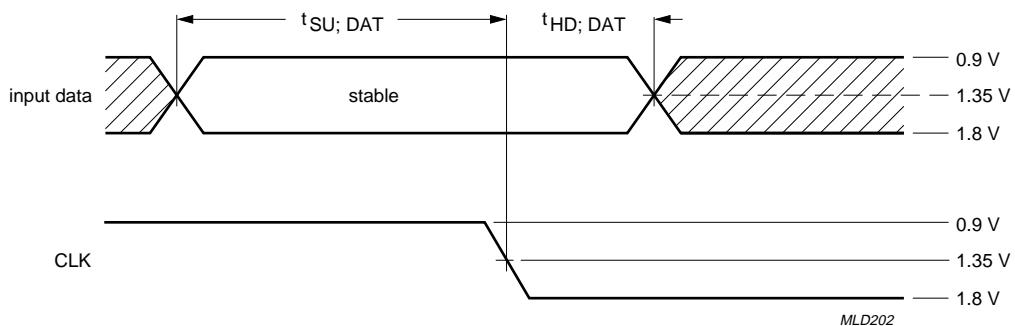


Fig.7 Typical output spectrum;  $f_{\text{clk}} = 1000 \text{ MHz}$ ;  $f_{\text{OUT}} = 200 \text{ MHz}$ .



The shaded areas indicate when the input data may change and be correctly registered. Data input update must be completed within  $t_{\text{bf}}$  ns after the falling edge of the clock ( $t_{\text{SU}; \text{DAT}}$  is negative;  $t_{\text{bf}}$  ns). Data must be held at least  $t_{\text{bf}}$  ns after the falling edge ( $t_{\text{HD}; \text{DAT}} = t_{\text{bf}}$  ns).

Fig.8 Data set-up and hold times.

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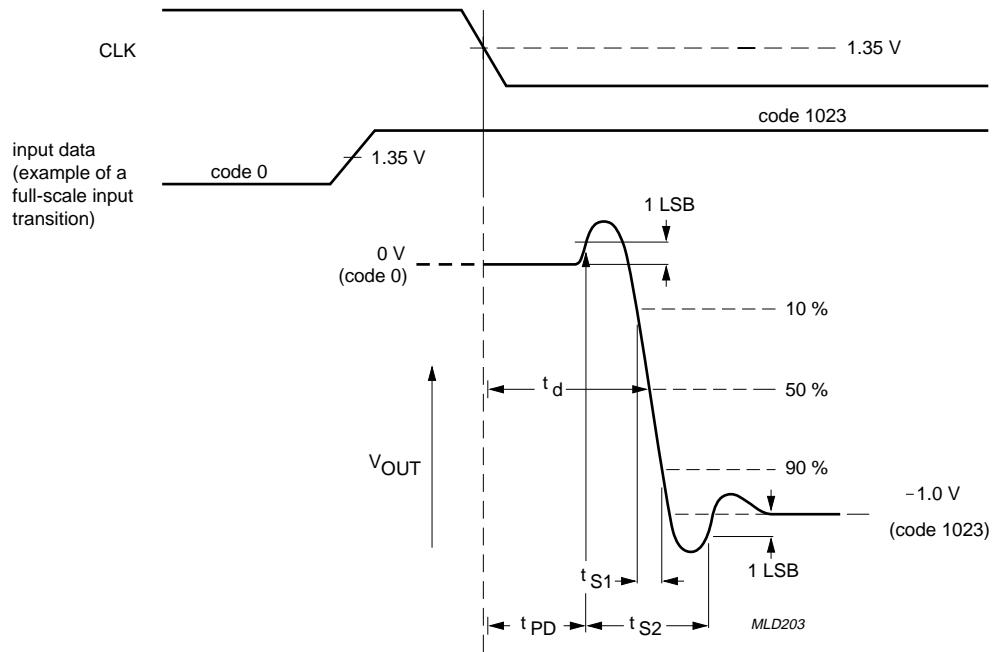
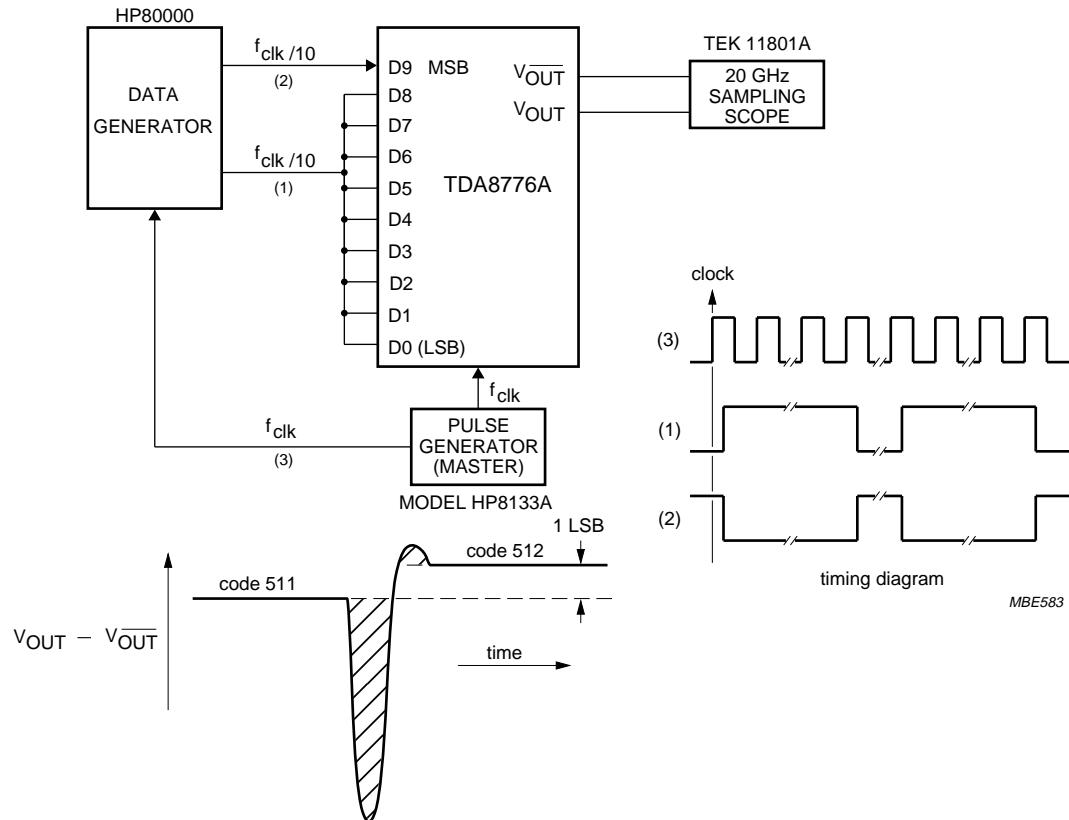


Fig.9 Switching characteristics.

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The value of the glitch energy is the sum of the shaded areas measured in pV.s.

Fig.10 Glitch energy measurement.

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## INTERNAL PIN CONFIGURATIONS

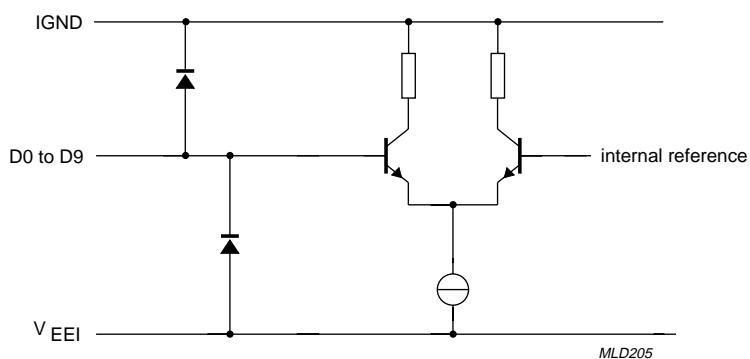


Fig.11 D9 to D0.

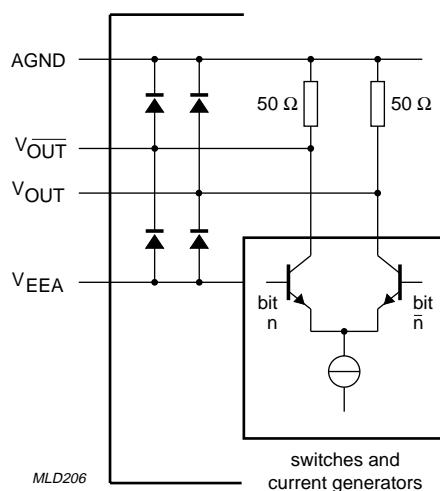


Fig.12 Analog outputs.

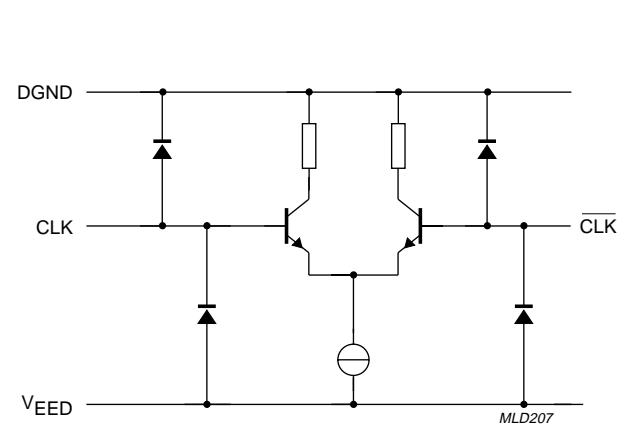
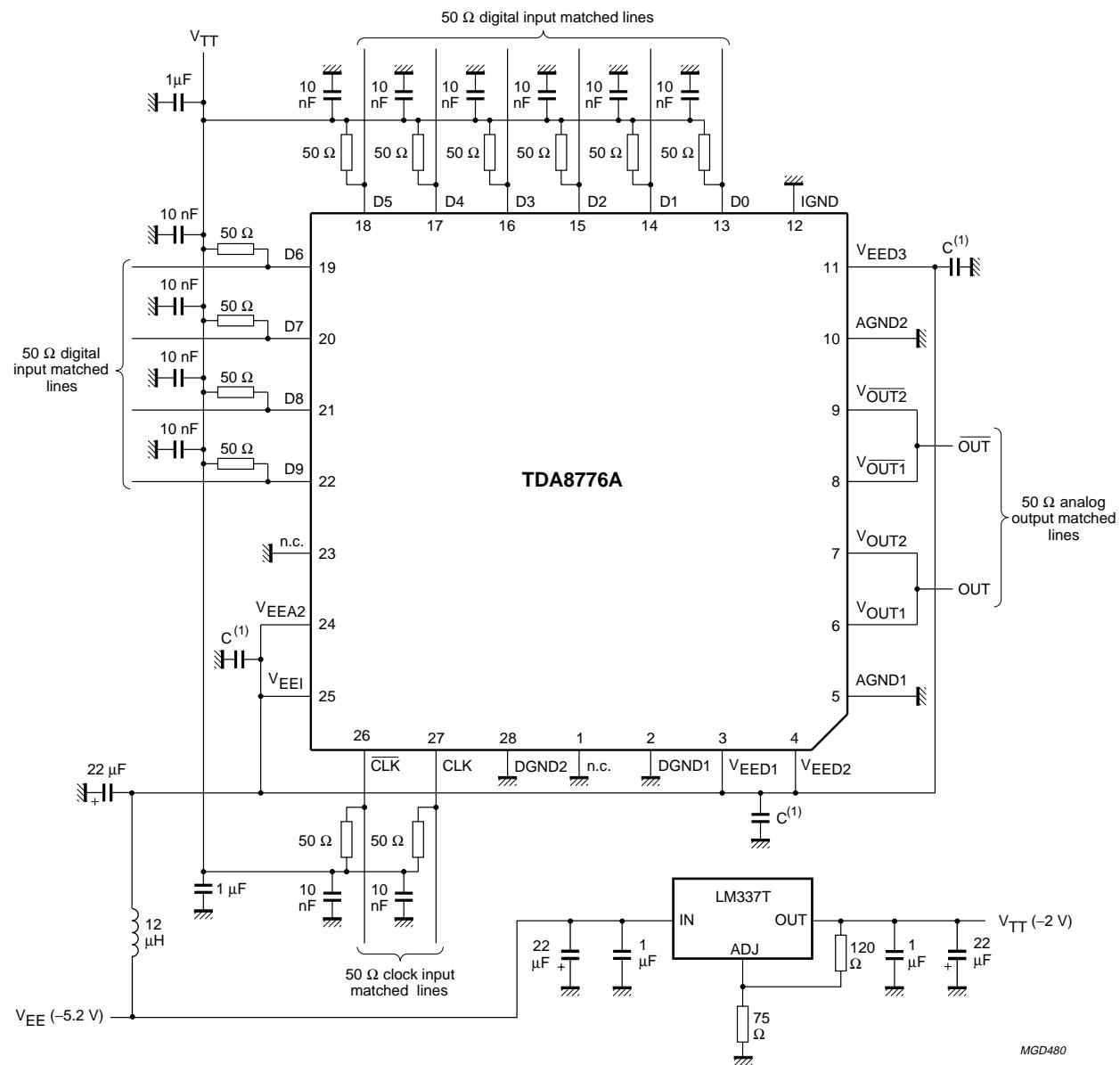


Fig.13 CLK and CLK̄.

# 10-bit, 1000 Msps Digital-to-Analog Converter (DAC)

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## APPLICATION INFORMATION



(1)  $C = 1 \mu\text{F}$  in parallel with  $10 \text{ nF}$  in parallel with  $1 \text{ nF}$ ; all three mounted close to the supply pin of the DAC with  $1 \text{ nF}$  the nearest.

Fig.14 Application diagram.

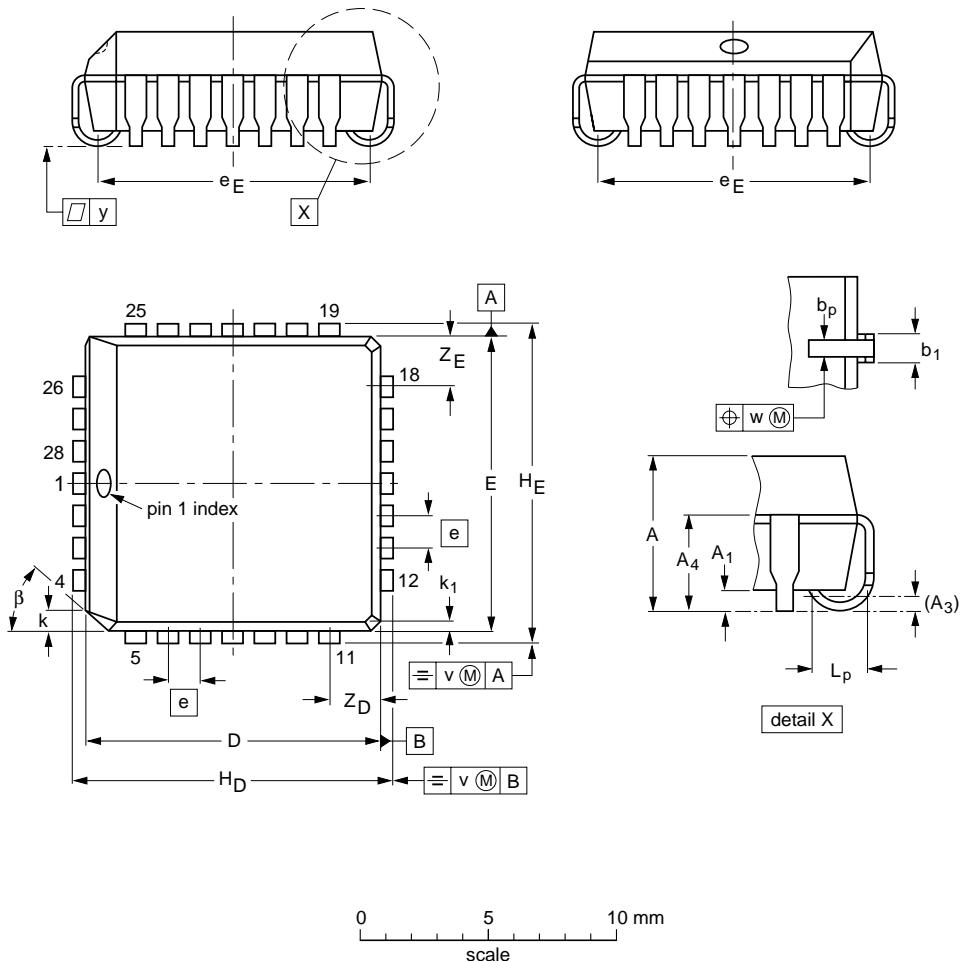
# 10-bit, 1000 Msps Digital-to-Analog Converter (DAC)

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## PACKAGE OUTLINE

PLCC28: plastic leaded chip carrier; 28 leads

SOT261-2



### DIMENSIONS (millimetre dimensions are derived from the original inch dimensions)

UNIT	A	$A_1$ min.	$A_3$	$A_4$ max.	$b_p$	$b_1$	$D^{(1)}$	$E^{(1)}$	$e$	$e_D$	$e_E$	$H_D$	$H_E$	$k$	$k_1$ max.	$L_p$	$v$	$w$	$y$	$Z_D^{(1)}$ max.	$Z_E^{(1)}$ max.	$\beta$
mm	4.57 4.19	0.51	0.25	3.05	0.53 0.33	0.81 0.66	11.58 11.43	11.58 11.43	1.27	10.92 9.91	10.92 9.91	12.57 12.32	12.57 12.32	1.22 1.07	0.51	1.44 1.02	0.18	0.18	0.10	2.16	2.16	45°
inches	0.180 0.165	0.020	0.01	0.12	0.021 0.013	0.032 0.026	0.456 0.450	0.456 0.450	0.05	0.430 0.390	0.430 0.390	0.495 0.485	0.495 0.485	0.048 0.042	0.020	0.057 0.040	0.007	0.007	0.004	0.085	0.085	

### Note

- Plastic or metal protrusions of 0.01 inches maximum per side are not included.

OUTLINE VERSION	REFERENCES				EUROPEAN PROJECTION	ISSUE DATE
	IEC	JEDEC	EIAJ			
SOT261-2						92-11-17 95-02-25

# 10-bit, 1000 Msps Digital-to-Analog Converter (DAC)

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## SOLDERING PLCC

### Introduction

There is no soldering method that is ideal for all IC packages. Wave soldering is often preferred when through-hole and surface mounted components are mixed on one printed-circuit board. However, wave soldering is not always suitable for surface mounted ICs, or for printed-circuits with high population densities. In these cases reflow soldering is often used.

This text gives a very brief insight to a complex technology. A more in-depth account of soldering ICs can be found in our "IC Package Databook2" (order code 9398 65290011).

### Reflow soldering

Reflow soldering techniques are suitable for all PLCC packages.

The choice of heating method may be influenced by larger plastic packages (44 leads, or more). If infrared or vapour phase heating is used and the large packages are not absolutely dry (less than 0.1% moisture content by weight), vaporization of the small amount of moisture in them can cause cracking of the plastic body. For more information, refer to the Drypack chapter in our "Quality Reference Manual" (order code 9398 510 63011).

Reflow soldering requires solder paste (a suspension of fine solder particles, flux and binding agent) to be applied to the printed-circuit board by screen printing, stencilling or pressure-syringe dispensing before package placement.

Several techniques exist for reflowing; for example, thermal conduction by heated belt. Dwell times vary between 50 and 300 seconds depending on heating method. Typical reflow temperatures range from 215 to 250 °C.

Preheating is necessary to dry the paste and evaporate the binding agent. Preheating duration: 45 minutes at 45 °C.

### Wave soldering

Wave soldering techniques can be used for all PLCC packages if the following conditions are observed:

- A double-wave (a turbulent wave with high upward pressure followed by a smooth laminar wave) soldering technique should be used.
- The longitudinal axis of the package footprint must be parallel to the solder flow.
- The package footprint must incorporate solder thieves at the downstream corners.

During placement and before soldering, the package must be fixed with a droplet of adhesive. The adhesive can be applied by screen printing, pin transfer or syringe dispensing. The package can be soldered after the adhesive is cured.

Maximum permissible solder temperature is 260 °C, and maximum duration of package immersion in solder is 10 seconds, if cooled to less than 150 °C within 6 seconds. Typical dwell time is 4 seconds at 250 °C.

A mildly-activated flux will eliminate the need for removal of corrosive residues in most applications.

### Repairing soldered joints

Fix the component by first soldering two diagonally-opposite end leads. Use only a low voltage soldering iron (less than 24 V) applied to the flat part of the lead. Contact time must be limited to 10 seconds at up to 300 °C. When using a dedicated tool, all other leads can be soldered in one operation within 2 to 5 seconds at 270 to 320 °C

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## DEFINITIONS

<b>Data sheet status</b>	
Objective specification	This data sheet contains target or goal specifications for product development.
Preliminary specification	This data sheet contains preliminary data; supplementary data may be published later.
Product specification	This data sheet contains final product specifications.
<b>Limiting values</b>	
Limiting values given are in accordance with the Absolute Maximum Rating System (IEC 134). Stress above one or more of the limiting values may cause permanent damage to the device. These are stress ratings only and operation of the device at these or at any other conditions above those given in the Characteristics sections of the specification is not implied. Exposure to limiting values for extended periods may affect device reliability.	
<b>Application information</b>	
Where application information is given, it is advisory and does not form part of the specification.	

## LIFE SUPPORT APPLICATIONS

These products are not designed for use in life support appliances, devices, or systems where malfunction of these products can reasonably be expected to result in personal injury. Philips customers using or selling these products for use in such applications do so at their own risk and agree to fully indemnify Philips for any damages resulting from such improper use or sale.

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**NOTES**

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10-bit, 1000 Msps Digital-to-Analog  
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**NOTES**

# **Philips Semiconductors – a worldwide company**

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