

# DATA SHEET

**TDA8772**

Triple 8-bit video digital-to-analog  
converter

Product specification

1997 Mar 06

Supersedes data of 1995 Mar 09

File under Integrated Circuits, IC02

# Triple 8-bit video digital-to-analog converter

**TDA8772**

## FEATURES

- 8-bit resolution
- Sampling rate up to:
  - 35 MHz for TDA8772H/3
  - 85 MHz for TDA8772H/8
- Internal reference voltage regulator
- No deglitching circuit required
- SYNC, BLANK control inputs
- 3 independent clock inputs (one per DAC)
- 1 V output voltage range
- $75\ \Omega$  output load
- TDA8772 has BLANK control on the 3 channels
- Single 5 V power supply
- 44-pin QFP package.

## APPLICATIONS

- General purpose high-speed digital-to-analog conversion
- Digital TV
- Graphic display
- Desktop video processing.

## ORDERING INFORMATION

TYPE NUMBER	PACKAGE			SAMPLING FREQUENCY
	NAME	DESCRIPTION	VERSION	
TDA8772H/3	QFP44	plastic quad flat package; 44 leads (lead length 1.3 mm); body 10 × 10 × 1.75 mm	SOT307-2	35 MHz
TDA8772H/8				85 MHz

## GENERAL DESCRIPTION

The TDA8772 is a triple 8-bit video Digital-to-Analog Converter (DAC). It converts the digital input signals into analog voltage outputs at a maximum conversion rate of 35 MHz (TDA8772H/3) and 85 MHz (TDA8772H/8).

The DAC is based on resistor-string architecture with integrated output buffers. The output voltage range is determined by a built-in reference source.

The device is fabricated in a 5 V CMOS process that ensures high functionality with low power dissipation.

# Triple 8-bit video digital-to-analog converter

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## QUICK REFERENCE DATA

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
$V_{DDA}$	analog supply voltage		4.5	5.0	5.5	V
$V_{DDD}$	digital supply voltage		4.5	5.0	5.5	V
$I_{DDA}$	analog supply current	$R_L = 75 \Omega$ ; note 1	40	65	100	mA
$I_{DDD}$	digital supply current TDA8772H/3 TDA8772H/8		—	7	16	mA
			—	16	27	mA
INL	integral non-linearity <sup>2</sup>	$f_{clk} = 35 \text{ MHz}$ ; ramp input	—	$\pm 0.5$	$\pm 1$	LSB
		$f_{clk} = 85 \text{ MHz}$ ; ramp input	—	$\pm 0.75$	$\pm 1.2$	LSB
DNL	differential non-linearity	$f_{clk} = 35 \text{ MHz}$ ; ramp input	—	$\pm 0.25$	$\pm 0.5$	LSB
		$f_{clk} = 85 \text{ MHz}$ ; ramp input	-	$\pm 0.5$	$\pm 0.75$	LSB
$f_{clk(max)}$	maximum clock frequency TDA8772H/3 TDA8772H/8		35	—	—	MHz
			85	—	—	MHz
$P_{tot}$	total power dissipation TDA8772H/3 TDA8772H/8	note 1				
		$R_L = 75 \Omega$ ; $f_{clk} = 35 \text{ MHz}$	180	360	640	mW
		$R_L = 75 \Omega$ ; $f_{clk} = 85 \text{ MHz}$	180	405	700	mW

### Note

1. Minimum and maximum data of current and power consumption are measured in worse case conditions: for minimum data, all digital inputs are at logic level 0 while for maximum data, all digital inputs are at logic level 1.

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## BLOCK DIAGRAM

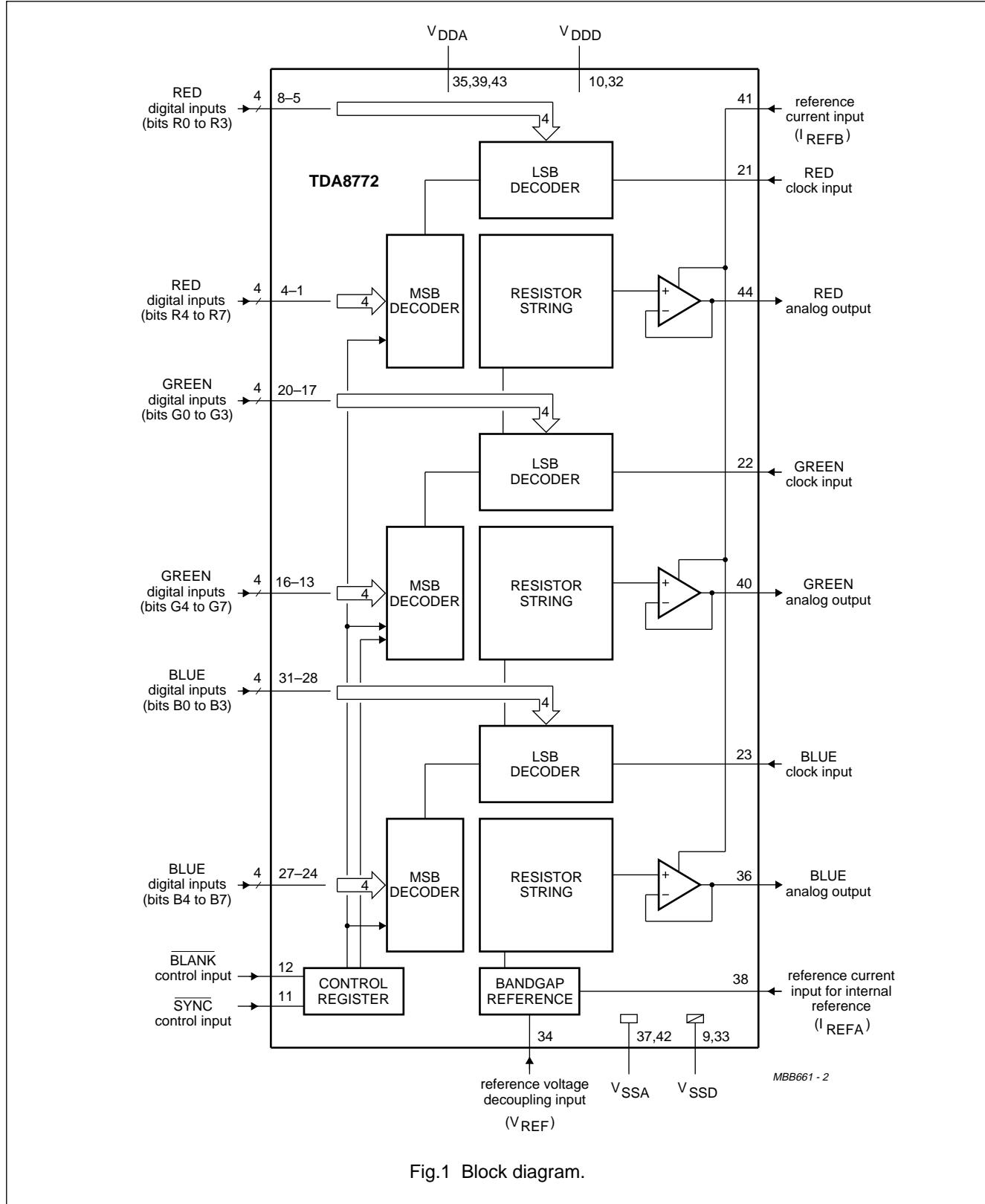


Fig.1 Block diagram.

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## PINNING

SYMBOL	PIN	DESCRIPTION
R7	1	RED digital input data; bit 7 (MSB)
R6	2	RED digital input data; bit 6
R5	3	RED digital input data; bit 5
R4	4	RED digital input data; bit 4
R3	5	RED digital input data; bit 3
R2	6	RED digital input data; bit 2
R1	7	RED digital input data; bit 1
R0	8	RED digital input data; bit 0 (LSB)
V <sub>SSD1</sub>	9	digital supply ground 1
V <sub>DDD1</sub>	10	digital supply voltage 1
SYNC	11	composite sync control input; for GREEN channel only (active LOW)
BLANK	12	composite blank control input (active LOW)
G7	13	GREEN digital input data; bit 7 (MSB)
G6	14	GREEN digital input data; bit 6
G5	15	GREEN digital input data; bit 5
G4	16	GREEN digital input data; bit 4
G3	17	GREEN digital input data; bit 3
G2	18	GREEN digital input data; bit 2
G1	19	GREEN digital input data; bit 1
G0	20	GREEN digital input data; bit 0 (LSB)
CLKR	21	RED clock input
CLKG	22	GREEN clock input

SYMBOL	PIN	DESCRIPTION
CLKB	23	BLUE clock input
B7	24	BLUE digital input data; bit 7 (MSB)
B6	25	BLUE digital input data; bit 6
B5	26	BLUE digital input data; bit 5
B4	27	BLUE digital input data; bit 4
B3	28	BLUE digital input data; bit 3
B2	29	BLUE digital input data; bit 2
B1	30	BLUE digital input data; bit 1
B0	31	BLUE digital input data; bit 0 (LSB)
V <sub>DDD2</sub>	32	digital supply voltage 2
V <sub>SSD2</sub>	33	digital supply ground 2
V <sub>REF</sub>	34	decoupling input for reference voltage
V <sub>DDA1</sub>	35	analog supply voltage 1
OUTB	36	BLUE analog output
V <sub>SSA1</sub>	37	analog supply ground 1
I <sub>REFA</sub>	38	reference current input for internal reference
V <sub>DDA2</sub>	39	analog supply voltage 2
OUTG	40	GREEN analog output
I <sub>REFB</sub>	41	reference current input for output buffers
V <sub>SSA2</sub>	42	analog supply ground 2
V <sub>DDA3</sub>	43	analog supply voltage 3
OUTR	44	RED analog output

# Triple 8-bit video digital-to-analog converter

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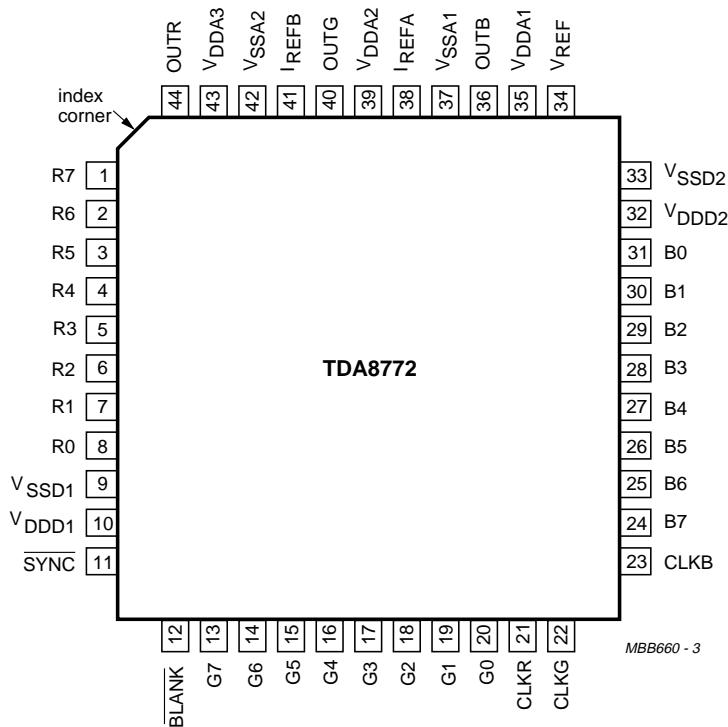


Fig.2 Pin configuration.

## LIMITING VALUES

In accordance with the Absolute Maximum Rating System (IEC 134).

SYMBOL	PARAMETER	MIN.	MAX.	UNIT
$V_{DDA}$	analog supply voltage	-0.5	+6.5	V
$V_{DDD}$	digital supply voltage	-0.5	+6.5	V
$\Delta V_{DD}$	supply voltage difference between $V_{DDA}$ and $V_{DDD}$	-1.0	+1.0	V
$T_{stg}$	storage temperature	-55	+150	°C
$T_{amb}$	operating ambient temperature	0	70	°C
$T_j$	junction temperature	-	125	°C

## THERMAL CHARACTERISTICS

SYMBOL	PARAMETER	VALUE	UNIT
$R_{th\ j-a}$	thermal resistance from junction to ambient in free air	75	K/W

## HANDLING

Inputs and outputs are protected against electrostatic discharges in normal handling. However, to be totally safe, it is desirable to take normal precautions appropriate to handling integrated circuits.

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## CHARACTERISTICS

TDA8772H/3 operating at 35 MHz and TDA8772H/8 operating at 85 MHz unless otherwise specified.

$V_{DDA} = V_{DDD} = 4.5 \text{ V}$  to  $5.5 \text{ V}$ ;  $V_{SSA}$  and  $V_{SSD}$  shorted together;  $V_{DDA} - V_{DDD} = -0.5 \text{ V}$  to  $+0.5 \text{ V}$ ;  $T_{amb} = 0$  to  $+70 \text{ }^{\circ}\text{C}$ ; typical values measured at  $V_{DDA} = V_{DDD} = 5 \text{ V}$  and  $T_{amb} = 25 \text{ }^{\circ}\text{C}$ ; unless otherwise specified.

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
<b>Supplies</b>						
$V_{DDA}$	analog supply voltage		4.5	5.0	5.5	V
$V_{DDD}$	digital supply voltage		4.5	5.0	5.5	V
$I_{DDA}$	analog supply current	$R_L = 75 \Omega$ ; note 1	40	65	100	mA
$I_{DDD}$	digital supply current TDA8772H/3 TDA8772H/8		—	7 16	16 27	mA mA
<b>Inputs</b>						
CLOCK INPUTS (PINS 21 TO 23)						
$V_{IL}$	LOW level input voltage		$V_{SSD} - 0.5$	—	0.8	V
$V_{IH}$	HIGH level input voltage		2.0	—	$V_{DDD} + 0.5$	V
BLANK, SYNC INPUTS (PINS 12 AND 11; ACTIVE LOW)						
$V_{IL}$	LOW level input voltage		$V_{SSD} - 0.5$	—	0.8	V
$V_{IH}$	HIGH level input voltage		2.0	—	$V_{DDD} + 0.5$	V
R, G, B DIGITAL INPUTS (PINS 1 TO 8, 13 TO 20 AND 24 TO 31)						
$V_{IL}$	LOW level input voltage		$V_{SSD} - 0.5$	—	0.8	V
$V_{IH}$	HIGH level input voltage		2.0	—	$V_{DDD} + 0.5$	V
$I_{REFA}$ INTERNAL REFERENCE SUPPLY CURRENT (PIN 38)						
$I_{i(REFA)}$	input current		—	0.17	0.25	mA
$I_{REFB}$ OUTPUT BUFFER SUPPLY CURRENT (PIN 41)						
$I_{i(REFB)}$	input current		—	0.5	0.7	mA
<b>Timing (<math>C_L = 25 \text{ pF}</math>; <math>R_L = 75 \Omega</math>); see Fig.3</b>						
$f_{clk(max)}$	maximum clock frequency TDA8772H/3 TDA8772H/8		35 85	— —	— —	MHz MHz
$t_{CPH}$	clock pulse width HIGH		5	—	—	ns
$t_{CPL}$	clock pulse width LOW		5	—	—	ns
$t_r$	clock rise time TDA8772H/3 TDA8772H/8		— —	— —	5 3	ns ns
$t_f$	clock fall time TDA8772H/3 TDA8772H/8		— —	— —	5 3	ns ns
$t_{SU;DAT}$	input data set-up time		4	—	—	ns
$t_{HD;DAT}$	input data hold time		2.5	—	—	ns

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SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
<b>Voltage reference (pin 34, referenced to V<sub>SSA</sub>)</b>						
V <sub>ref</sub>	output reference voltage		1.180	1.242	1.305	V
<b>Outputs</b>						
OUTB, OUTR, OUTG ANALOG OUTPUTS (PINS 36, 44 AND 40, REFERENCED TO V <sub>SSA</sub> ) R <sub>L</sub> = 75 Ω; SEE TABLES 1 AND 2						
FSR	full-scale output voltage range		0.9	1.0	1.1	V
V <sub>os</sub>	offset of analog voltage output		0.75	0.83	0.95	V
V <sub>OUTmax</sub>	maximum output voltage	data inputs = logic 1; note 2	1.65	1.83	2.05	V
V <sub>OUTmin</sub>	minimum output voltage	data inputs = logic 0; note 2	0.75	0.83	0.95	V
THD	total harmonic distortion	f <sub>i</sub> = 4.43 MHz; f <sub>clk</sub> = 35 MHz	–	–45	–	dB
		f <sub>i</sub> = 4.43 MHz; f <sub>clk</sub> = 85 MHz	–	–43	–	dB
Z <sub>L</sub>	output load impedance		60	75	90	Ω
<b>Transfer function</b>						
INL	integral non-linearity	f <sub>clk</sub> = 35 MHz; ramp input	–	±0.5	±1	LSB
		f <sub>clk</sub> = 85 MHz; ramp input	–	±0.75	±1.2	LSB
DNL	differential non-linearity	f <sub>clk</sub> = 35 MHz; ramp input	–	±0.25	±0.5	LSB
		f <sub>clk</sub> = 85 MHz; ramp input	–	±0.5	±0.75	LSB
α <sub>CT</sub>	crosstalk DAC to DAC		–45	–	–	dB
m	DAC to DAC matching		–	1.0	2.0	%
<b>Switching characteristics (R<sub>L</sub> = 75 Ω); see Fig.4</b>						
t <sub>d</sub>	input to 50% output delay time	full-scale change	–	10	–	ns
t <sub>s1</sub>	settling time	10% to 90% full-scale change	–	6	–	ns
t <sub>s2</sub>	settling time	to ±1 LSB	–	30	–	ns
<b>Output transients (glitches)</b>						
V <sub>g</sub>	area for 1 LSB change		–	1	–	LSB.ns

**Notes**

1. Minimum and maximum data of current and power consumption are measured in worse case conditions: for minimum data, all digital inputs are at logic level 0 while for maximum data, all digital inputs are at logic level 1.
2. V<sub>OUT</sub> is directly proportional to V<sub>ref</sub>.

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**Table 1** Input coding and DAC output voltages (typical values)

<b>BINARY INPUT DATA (SYNC = BLANK = 0)</b>	<b>CODE</b>	<b>DAC OUTPUT VOLTAGES (V) OUTB, OUTR, OUTG <math>R_L = 75 \Omega</math></b>
0000 0000	0	0.830
0000 0001	1	0.834
.....	.	.
1000 0000	128	1.330
.....	.	.
1111 1110	254	1.826
1111 1111	255	1.830

**Table 2** Input coding and DAC output voltages (typical values)

<b>BINARY INPUT DATA</b>	<b>SYNC (PIN 11)</b>	<b>BLANK (PIN 12)</b>	<b>DAC OUTPUT VOLTAGES (V)</b>	
			<b>OUTG (PIN 40)</b>	<b>OUTR/B (PINS 44, 46)</b>
.....	x	1	see Table 1	see Table 1
.....	1	0	0.830	0.830
.....	0	0	0.440	

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## TIMING

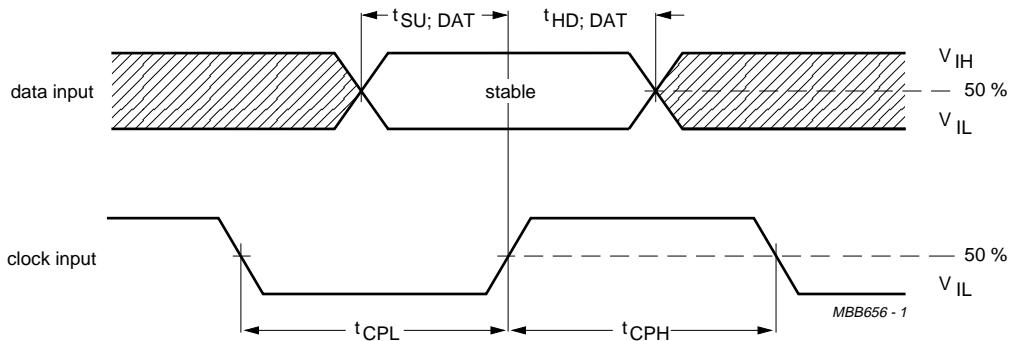


Fig.3 Input timing.

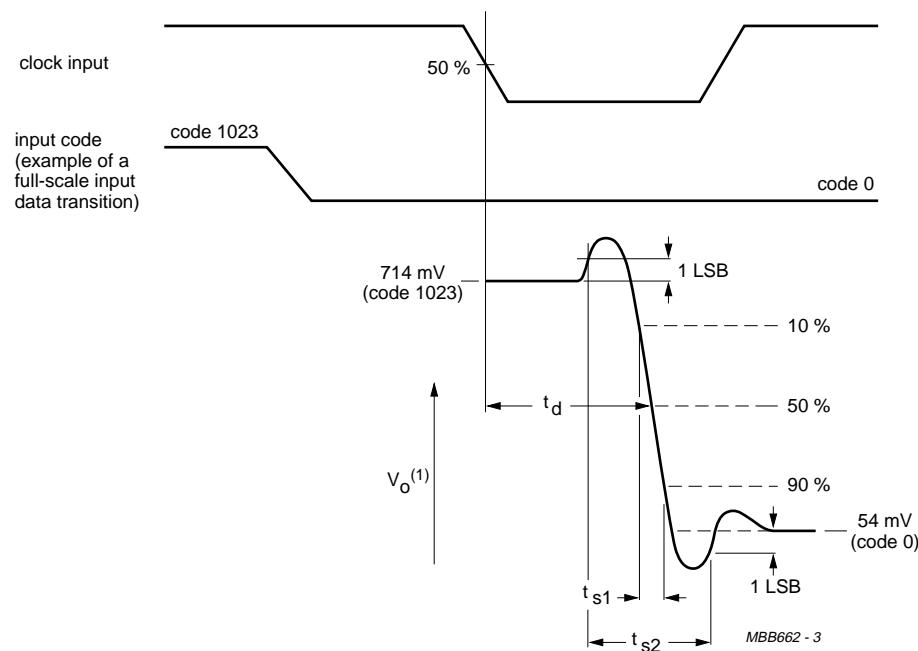
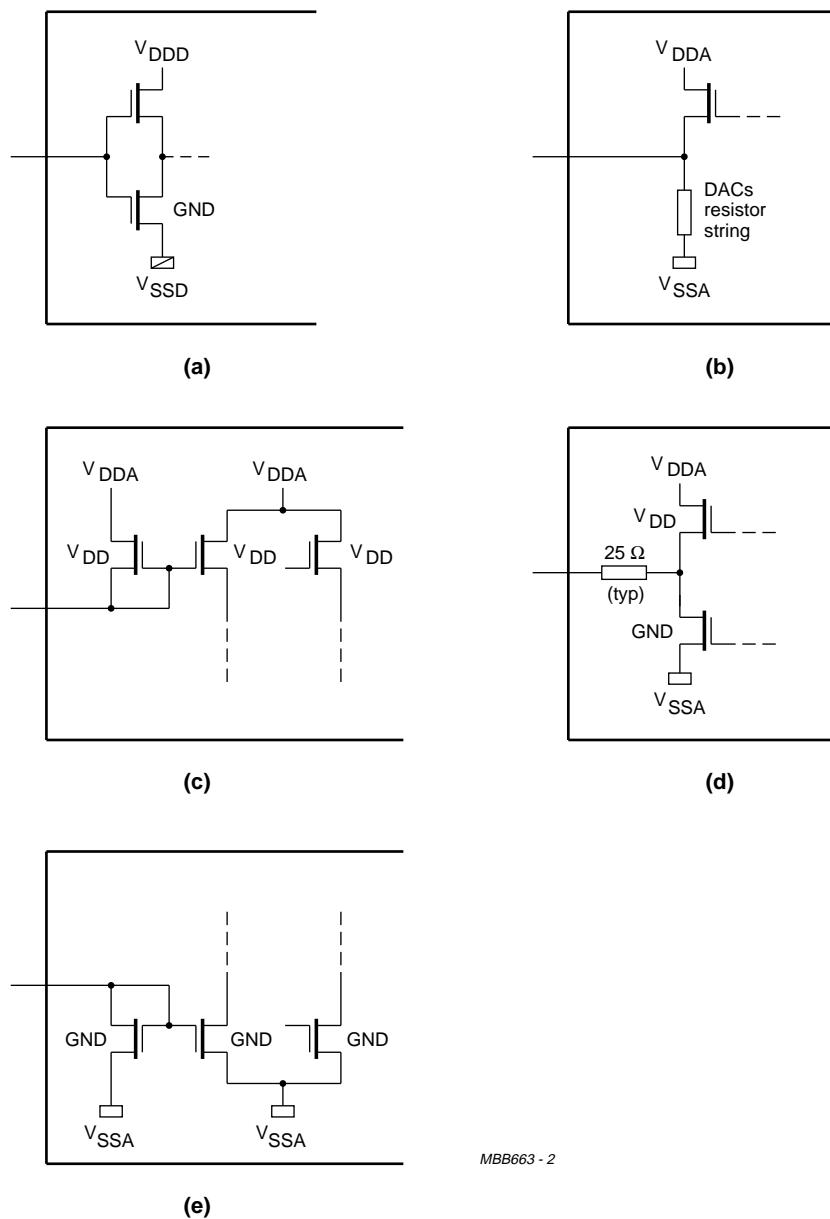


Fig.4 Switching timing.

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## INTERNAL CIRCUITRY



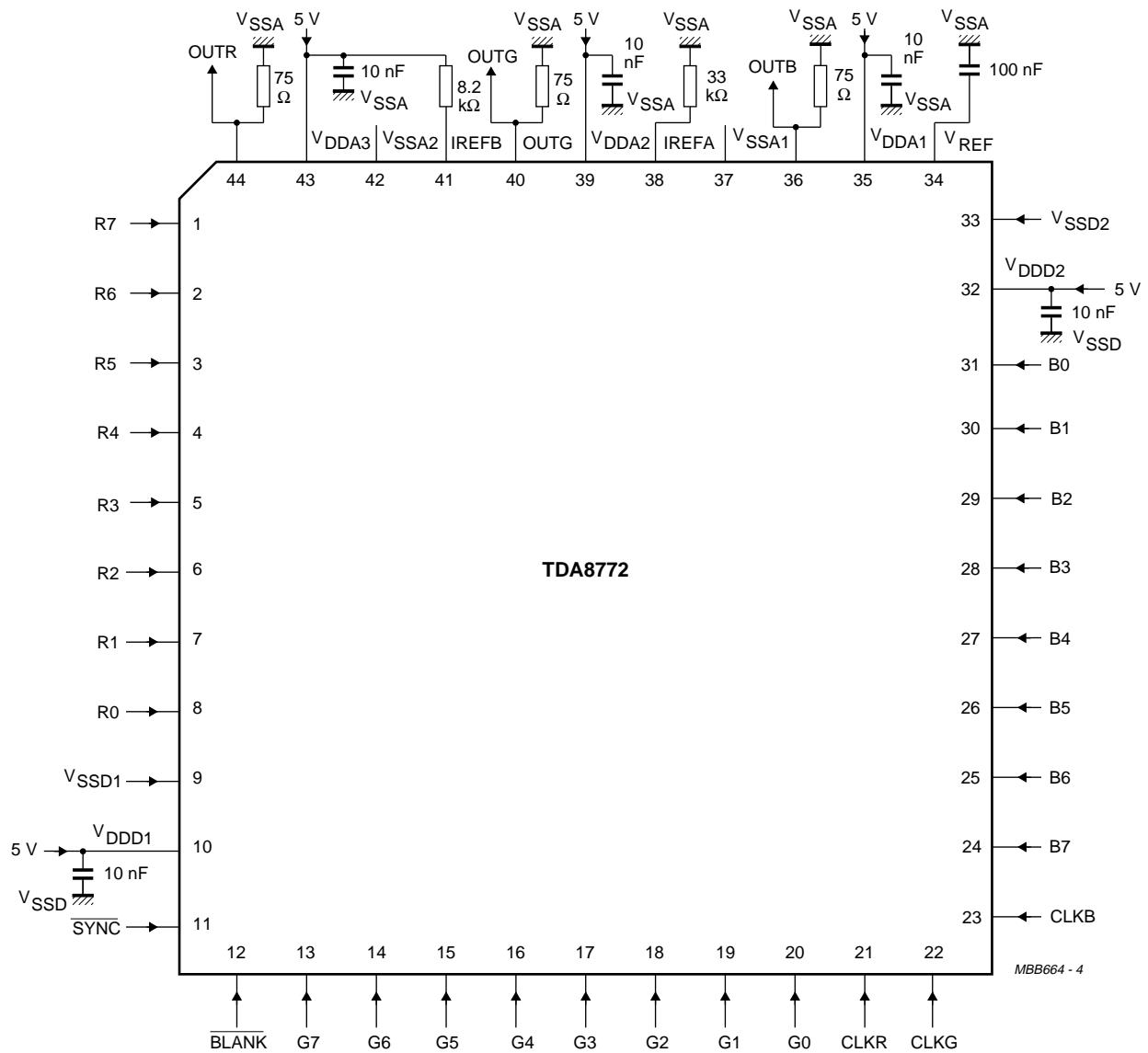
- (a) Digital inputs; pins 1 to 8 and 11 to 31.
- (b)  $V_{ref}$ ; pin 34.
- (c)  $I_{REFA}$ ; pin 38.
- (d) OUTR, G, B; pins 44, 40 and 36.
- (e)  $I_{REFB}$ ; pin 41.

Fig.5 Internal circuitry.

# Triple 8-bit video digital-to-analog converter

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## APPLICATION INFORMATION



Analog and digital supplies should be separated and decoupled.

Supplies are not connected internally.

All ground pins must be connected. One ground plane is preferred although it depends on the application.

See Fig.7 for example of anti-aliasing filter.

Fig.6 Application diagram.

# Triple 8-bit video digital-to-analog converter

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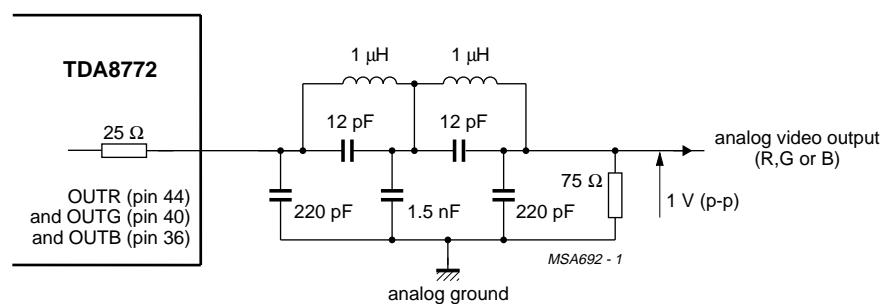


Fig.7 Example of anti-aliasing filter for 1 V output swing.

## Characteristics of Fig.8

- Order 5; adapted CHEBYSHEV
- Ripple  $\rho \geq 0.6$  dB
- Frequency at  $-3$  dB = 6.5 MHz
- $f_{NOTCH} = 46$  MHz.

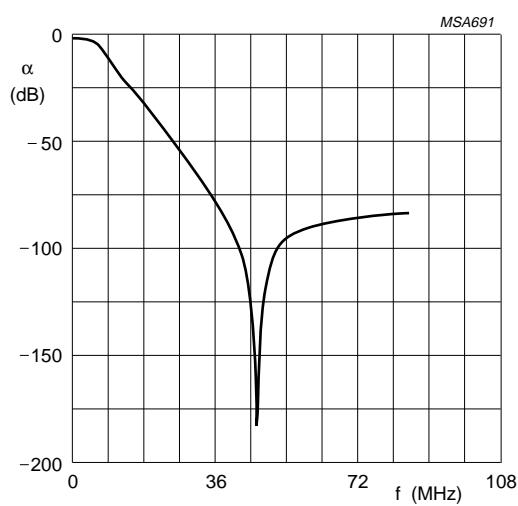


Fig.8 Frequency response for filter shown in Fig.7.

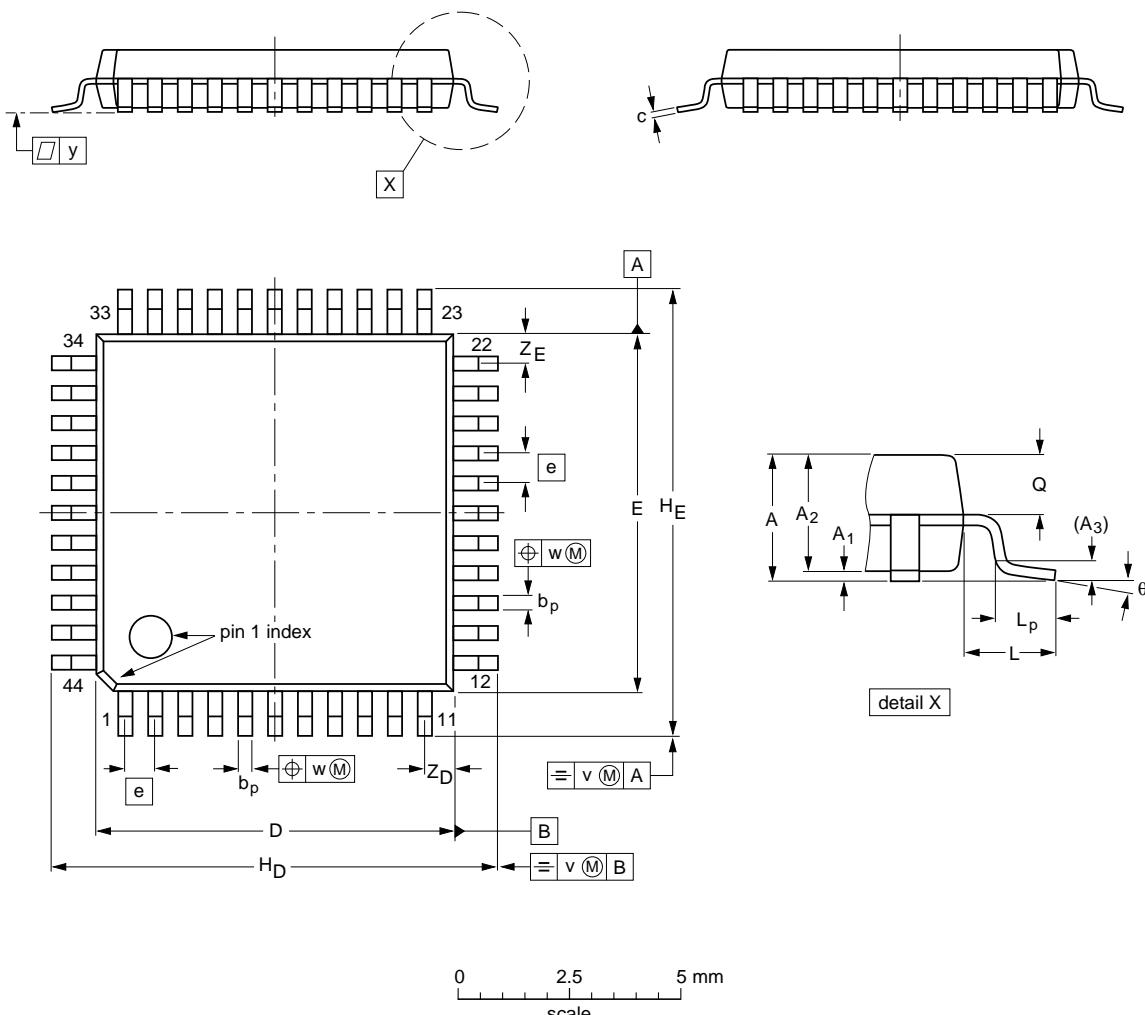
# Triple 8-bit video digital-to-analog converter

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## PACKAGE OUTLINE

QFP44: plastic quad flat package; 44 leads (lead length 1.3 mm); body 10 x 10 x 1.75 mm

SOT307-2



## DIMENSIONS (mm are the original dimensions)

UNIT	A max.	A <sub>1</sub>	A <sub>2</sub>	A <sub>3</sub>	b <sub>p</sub>	c	D <sup>(1)</sup>	E <sup>(1)</sup>	e	H <sub>D</sub>	H <sub>E</sub>	L	L <sub>p</sub>	Q	v	w	y	Z <sub>D</sub> <sup>(1)</sup>	Z <sub>E</sub> <sup>(1)</sup>	θ
mm	2.10 0.05	0.25 1.65	1.85	0.25	0.40 0.20	0.25 0.14	10.1 9.9	10.1 9.9	0.8	12.9 12.3	12.9 12.3	1.3	0.95 0.55	0.85 0.75	0.15	0.15	0.1	1.2 0.8	1.2 0.8	10° 0°

### Note

- Plastic or metal protrusions of 0.25 mm maximum per side are not included.

OUTLINE VERSION	REFERENCES				EUROPEAN PROJECTION	ISSUE DATE
	IEC	JEDEC	EIAJ			
SOT307-2						92-11-17 95-02-04

# Triple 8-bit video digital-to-analog converter

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## SOLDERING

### Introduction

There is no soldering method that is ideal for all IC packages. Wave soldering is often preferred when through-hole and surface mounted components are mixed on one printed-circuit board. However, wave soldering is not always suitable for surface mounted ICs, or for printed-circuits with high population densities. In these situations reflow soldering is often used.

This text gives a very brief insight to a complex technology. A more in-depth account of soldering ICs can be found in our "IC Package Databook" (order code 9398 652 90011).

### Reflow soldering

Reflow soldering techniques are suitable for all QFP packages.

The choice of heating method may be influenced by larger plastic QFP packages (44 leads, or more). If infrared or vapour phase heating is used and the large packages are not absolutely dry (less than 0.1% moisture content by weight), vaporization of the small amount of moisture in them can cause cracking of the plastic body. For more information, refer to the Drypack chapter in our "Quality Reference Handbook" (order code 9397 750 00192).

Reflow soldering requires solder paste (a suspension of fine solder particles, flux and binding agent) to be applied to the printed-circuit board by screen printing, stencilling or pressure-syringe dispensing before package placement.

Several techniques exist for reflowing; for example, thermal conduction by heated belt. Dwell times vary between 50 and 300 seconds depending on heating method. Typical reflow temperatures range from 215 to 250 °C.

Preheating is necessary to dry the paste and evaporate the binding agent. Preheating duration: 45 minutes at 45 °C.

### Wave soldering

Wave soldering is **not** recommended for QFP packages. This is because of the likelihood of solder bridging due to closely-spaced leads and the possibility of incomplete solder penetration in multi-lead devices.

If wave soldering cannot be avoided, the following conditions must be observed:

- A double-wave (a turbulent wave with high upward pressure followed by a smooth laminar wave) soldering technique should be used.
- The footprint must be at an angle of 45° to the board direction and must incorporate solder thieves downstream and at the side corners.

Even with these conditions, do not consider wave soldering the following packages: QFP52 (SOT379-1), QFP100 (SOT317-1), QFP100 (SOT317-2), QFP100 (SOT382-1) or QFP160 (SOT322-1).

During placement and before soldering, the package must be fixed with a droplet of adhesive. The adhesive can be applied by screen printing, pin transfer or syringe dispensing. The package can be soldered after the adhesive is cured.

Maximum permissible solder temperature is 260 °C, and maximum duration of package immersion in solder is 10 seconds, if cooled to less than 150 °C within 6 seconds. Typical dwell time is 4 seconds at 250 °C.

A mildly-activated flux will eliminate the need for removal of corrosive residues in most applications.

### Repairing soldered joints

Fix the component by first soldering two diagonally-opposite end leads. Use only a low voltage soldering iron (less than 24 V) applied to the flat part of the lead. Contact time must be limited to 10 seconds at up to 300 °C. When using a dedicated tool, all other leads can be soldered in one operation within 2 to 5 seconds between 270 and 320 °C.

## Triple 8-bit video digital-to-analog converter

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### DEFINITIONS

<b>Data sheet status</b>	
Objective specification	This data sheet contains target or goal specifications for product development.
Preliminary specification	This data sheet contains preliminary data; supplementary data may be published later.
Product specification	This data sheet contains final product specifications.
<b>Limiting values</b>	
Limiting values given are in accordance with the Absolute Maximum Rating System (IEC 134). Stress above one or more of the limiting values may cause permanent damage to the device. These are stress ratings only and operation of the device at these or at any other conditions above those given in the Characteristics sections of the specification is not implied. Exposure to limiting values for extended periods may affect device reliability.	
<b>Application information</b>	
Where application information is given, it is advisory and does not form part of the specification.	

### LIFE SUPPORT APPLICATIONS

These products are not designed for use in life support appliances, devices, or systems where malfunction of these products can reasonably be expected to result in personal injury. Philips customers using or selling these products for use in such applications do so at their own risk and agree to fully indemnify Philips for any damages resulting from such improper use or sale.

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**NOTES**

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