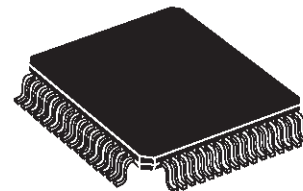




## STEREO AUDIO/MODEM/TELEPHONY CODEC

- STEREO AUDIO CODEC WITH CD QUALITY NOISE AND DISTORTION
  - ADC AND DAC PROGRAMMABLESAMPLING RATES UP TO 48kHz
  - 16-BIT RESOLUTION
  - > 85dB DYNAMIC RANGE
- QUADRAPHONIC MODE FOR SURROUND SOUND APPLICATION (STEREO AUDIO CODEC PLUS BOTH MODEM AND TELEPHONY MONOPHONIC CODECS)
- MONOPHONIC CODEC FOR MODEM APPLICATIONS
  - ADC AND DAC SAMPLING RATES UP TO 48kHz
  - 16-BIT RESOLUTION
  - > 90dB DYNAMIC RANGE (MODEM MODE)
  - V.34BIS MODEM PERFORMANCE CAPABILITY
- SECOND MONOPHONIC CODEC FOR LOCAL HANDSET, ANSWERING MACHINE OR 2<sup>nd</sup> MODEM APPLICATIONS
  - ADC AND DAC SAMPLING RATES UP TO 48kHz
  - 16-BIT RESOLUTION
  - > 85dB DYNAMIC RANGE (MODEM MODE)
- EXTENSIVE AUDIO SWITCHING AND GAIN CONTROLS FOR AUDIO ROUTING
  - A SUPERSET FEATURE LIST OF SOUND-BLASTER PRO AND MPC-2 SPECIFICATIONS
  - 3 STEREO LINE LEVEL INPUTS (LINE IN, CDROM, AUX)
  - 2 MONOPHONIC MICROPHONE INPUTS (DESKTOP AND HEADSET)
  - 2 PAIR OF STEREO OUTPUTS (MAIN AND AUXILIARY)
  - MONOPHONIC INPUT AND OUTPUT
  - PROGRAMMABLE GAIN/ATTENUATION/MUTE BLOCKS ON ALL INPUTS AND OUTPUTS
- INDEPENDENT CAPTURE AND PLAYBACK MIXING CAPABILITY
- MODEM CODEC HAS GAIN/ATTENUATION BLOCKS
  - INPUTS AND OUTPUTS CAN OPERATE IN A BALANCED OR UNBALANCED CONFIGURATION
  - DAC OUTPUT CAN OPTIONALLY BE USED AS REAR CHANNEL SOUND EFFECTS FOR GAMES



**TQFP64** (10 x 10 x 1.4mm)  
(Full Thin Plastic Quad Flat Pack)

**ORDER CODE : STLC7549**

- 2<sup>nd</sup> MODEM CODEC HAS GAIN/ATTENUATION/MUTE BLOCKS ON INPUT AND OUTPUT
  - ADC CAN SELECT HANDSET MOUTHPIECE OR DESKTOP/HEADSET MICROPHONE
  - DAC OUTPUT CAN OPTIONALLY BE USED AS REAR CHANNEL SOUND EFFECTS FOR GAMES
  - DAC OUTPUT CAN OPTIONALLY BE USED TO DRIVE THE HEADSET EARPIECE
  - THIS CODEC CAN ALSO BE USED FOR A SECOND MODEM PORT
- CONTROL PINS TO CONTROL LINE OUT ATTENUATORS BY PUSHBUTTONS UP/DOWN/MUTE
- CONTROL PIN TO CONTROL LINE OUT ATTENUATORS BY A DC-VOLTAGE POTENTIOMETER
- 8 GENERAL PURPOSE DIGITAL I/O LINES FOR CONTROL OF MISCELLANEOUS TELEPHONE CIRCUITS
- EXTENSIVE POWER DOWN OPTIONS WITH LOW-POWER AUDIO LOOPTHRU MODE
- DUAL SYNCHRONOUS SERIAL PORTS FOR INDEPENDENT AUDIO AND COMMUNICATION WITH 64 BITS PER FRAME
- DIGITAL POWER SUPPLY OF 3.3V TO 5.0V
- ANALOG POWER SUPPLY OF 5V
- AVAILABLE IN 64 PIN TQFP PACKAGE

### DESCRIPTION

The STLC7549 codec is a single chip multimedia codec that contains multiple 16-bit sigma-delta codecs and a wide assortment of audio switching and gain controls.

**OVERVIEW**

STLC7549 is the analog component of a 2-chip PC audio solution for multimedia application. It is well suited for CD-Quality Audio/Telephony and DSVD V.34bis Modem/Fax application. Outstanding performances are obtained thanks to 16-bit sigma-delta architecture, giving dynamic ranges exceeding 90dB for modem and 85dB for audio with programmable sample frequency up to 48kHz. With powerdown and 3.3V to 5V digital supply modes, it integrates complete analog-front-end circuits in a 64 pins TQFP package reducing power consumption, size and circuitry complexity.

Figure 1 gives the simplified diagram of the STLC7549 internal structure. It presents three different parts - Audio, Telephony and Modem - that offers real simultaneous audio and communication features, with dual serial port and dual programmable sample rate for more flexibility and efficiency. In that way, previous modem digital controller or DSP software can be used without having to adapt it for additional audio part.

Each part provides a DAC channel and an ADC that respectively convert data from digital interface to analog signals, and convert analog input signals to digital data to the digital interface. ADC digital paths to DAC digital paths loopback is supported.

STLC7549 audio part gives full duplex stereo channels that, thanks to a dual set of mixer, allows independent analog mixing and recording mixing with DAC channel of all the following audio sources:

- Three main stereo input lines from CD player, MPEG audio signal, VCR recording, FM Synthesizer and MIDI application (L1,L2,L3,R1,R2 and R3)
- A stereo microphone input from desktop and/or headset, including a selectable +20dB pre-amplifier (MICL and MICR)
- A mono input for PC speaker (MONO-IN)

Each source can be amplified, attenuated and muted independently. Two stereo outputs (LOUT1,LOUT2,ROUT1 and ROUT2) provide the resulting audio mixing to two stereo line out through a volume control block directly set by pushbuttons, DC potentiometer pins, or by serial port register command. An additional summer provides to the PC speaker (MONO-OUT) a mixing of the stereo outputs with MONO-IN.

STLC7549 Telephony part provides a mono mixing of two mono inputs from second phone line and headset (TP2Rx and TP2Tx) with stereo microphone line to the ADC channel. Input can also be used for mixing music or message in answering machine or hold mode application. DAC channel drives three mono outputs to phone speaker, headset speakers, second line modem, fax or rear game channel (HandTx1,HandTx2,HandTx3).

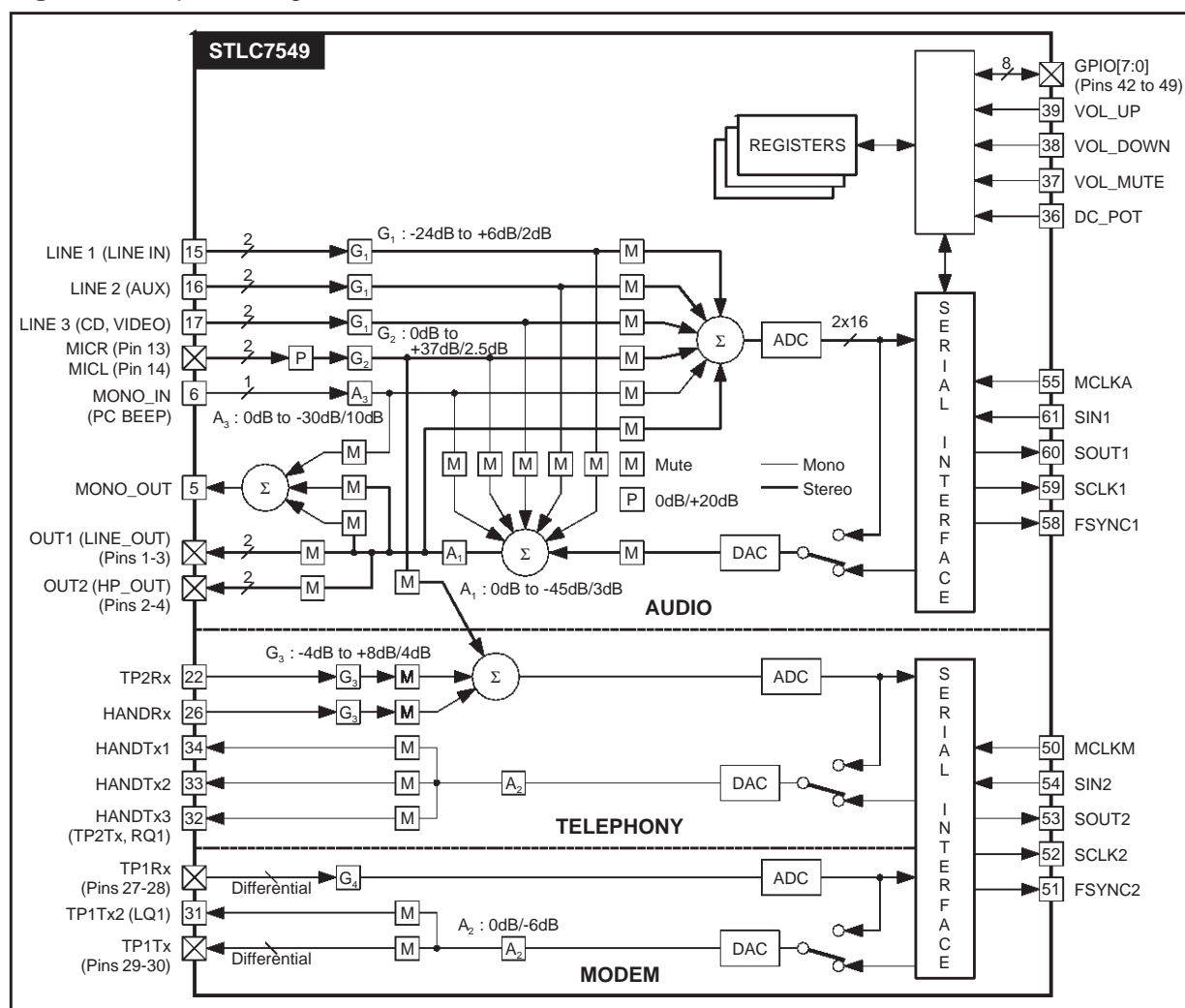
STLC7549 Modem part includes a differential mono input and output line, and an additional output line. It is targeted for V.34bis standard. Telephony and Modem parts are independent of audio parts thanks to a second sampling clock and serial port. However, all clocks can also be controlled from the audio Master clock (MCLKA), providing a quadraphonic mode for surround and 3-D sound applications.

Eight General Programmable Inputs/Outputs are provided for ring detect, LEDs and pushbuttons control, audio or phone source presence.

With full duplex mixing features, quadraphony mode, independent audio and communication serial ports and programmable sample rates, STLC7549 is a simultaneous Modem/Telephony/Stereo Audio Codec well suited for PC multimedia applications.

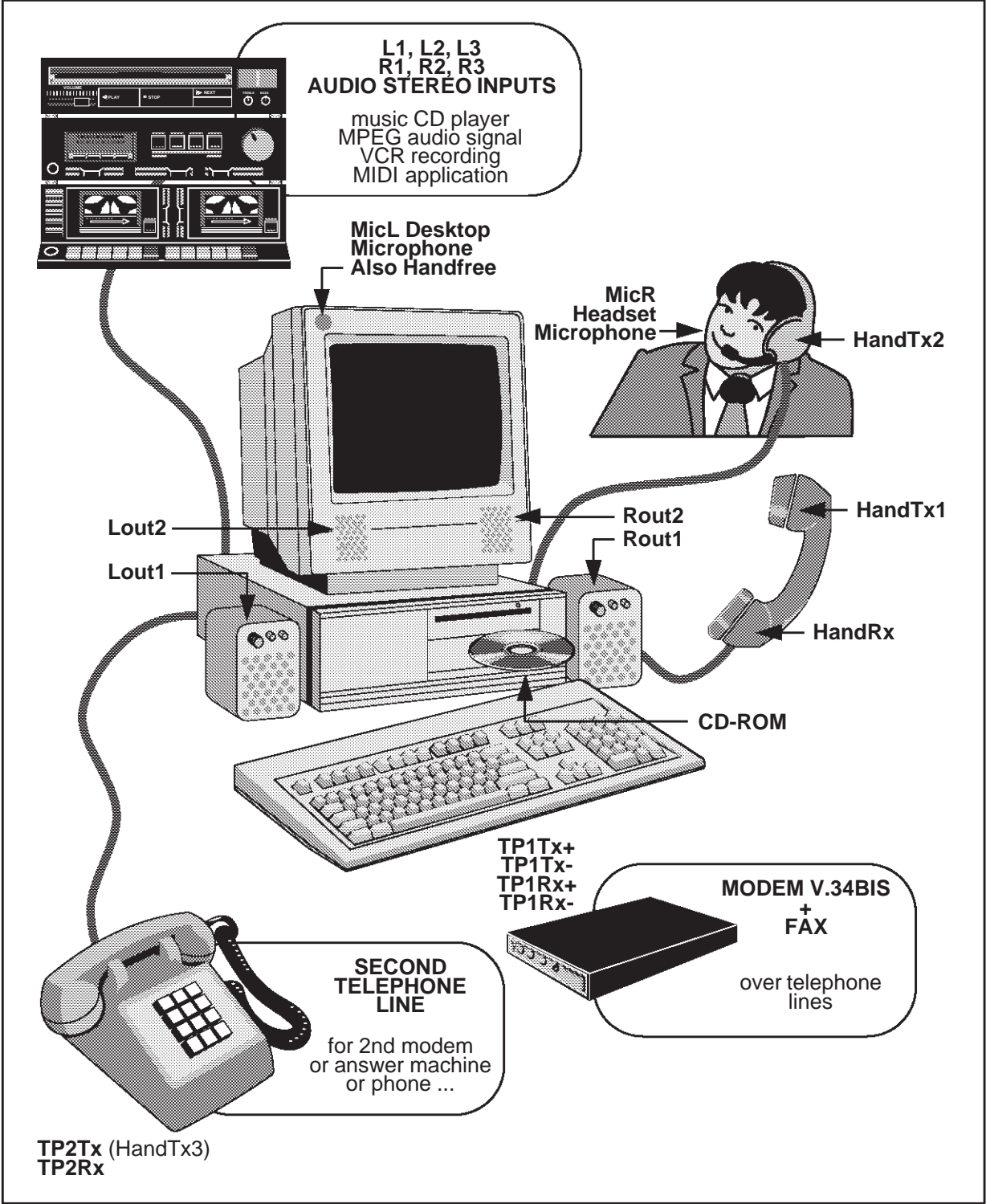
## OVERVIEW

Figure 1 : Simplified Diagram



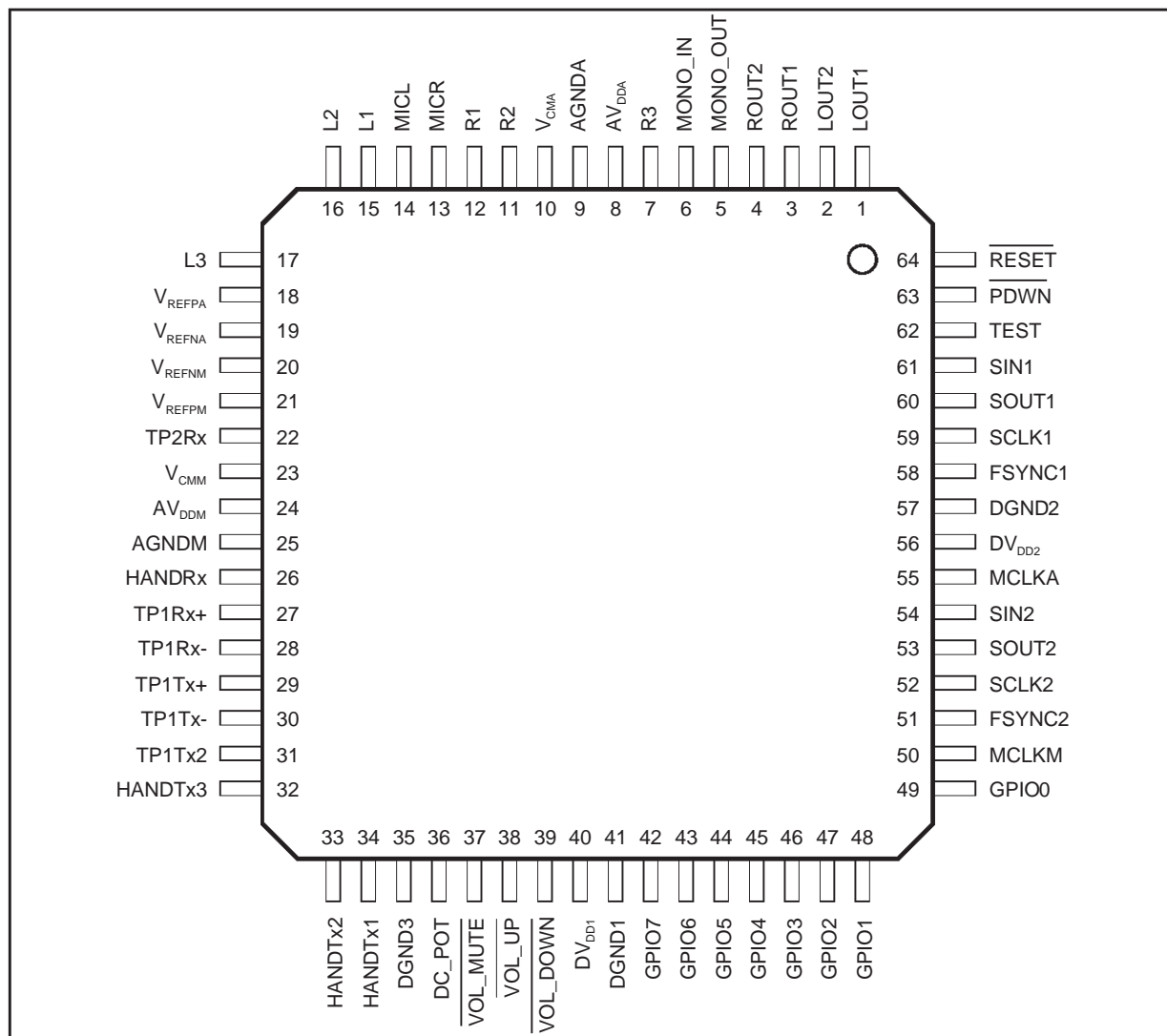
7549-01 EPS

MULTIMEDIA WORLD



7549-02.EPS

## PIN CONNECTIONS



7549-03.EPS

## PIN DESCRIPTION

Pin N°	Symbol	Type	Function
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POWER SUPPLY (9 pins) (see Notes 1 and 2)

8	AVDDA	I	Analog V <sub>DD</sub> Supply. This pin is the positive analog power supply voltage (4.75V to 5.25V) for the DAC and the ADC audio section. The voltage on these pins must be higher or equal to the voltage of the Digital power supply (DV <sub>DD</sub> ).
24	AVDDM	I	Analog V <sub>DD</sub> Supply. This pin is the positive analog power supply voltage (4.75V to 5.25V) for the DAC and the ADC modem and telephony section. The voltage on this pin must be higher or equal to the voltage of the Digital power supply (DV <sub>DD</sub> ).
40,56	DVDD1, DVDD2	I	Digital V <sub>DD</sub> Supply. This pin is the positive digital power supply (3.15V to 5.25V).
9	AGNDA	I	Analog Ground. This pin is the ground return of the analog DAC (ADC) audio section.
25	AGNDM	I	Analog Ground. This pin is the ground return of the analog DAC (ADC) modem and telephony section.
41, 57, 35	DGND1, DGND2, DGND3	I	Digital Ground. These pins are the ground return of the digital section.

**Notes :** 1. To obtain published performance, the analog V<sub>DD</sub> and digital V<sub>DD</sub> should be decoupled with respect to analog ground and digital ground, respectively. The decoupling is intended to isolate digital noise from the analog section ; decoupling capacitors should be as close as possible to the respective analog and digital supply pins (refer to Figure 18).

2. All the ground pins must be tied together. In the following section, the ground and supply pins are referred to as GND and V<sub>DD</sub>, respectively.

## PIN DESCRIPTION (continued)

Pin N°	Symbol	Type	Function
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## STEREO AUDIO CODEC (17 pins)

Voltage Reference			
18	V <sub>REFPA</sub>	O	Stereo Audio DAC and ADC Positive Reference Voltage Output. This pin provides the Positive Reference Voltage used by the 16-bit stereo audio converters. The reference voltage, V <sub>REFA</sub> , is the voltage difference between the V <sub>REFPA</sub> and V <sub>REFNA</sub> outputs. V <sub>REFPA</sub> should be externally decoupled with respect to V <sub>CMA</sub> (see Figure 17). $V_{REFA} = V_{REFPA} - V_{REFNA}$
19	V <sub>REFNA</sub>	O	Stereo Audio DAC and ADC Negative Reference Voltage Output. This pin provides the Negative Reference Voltage used by the 16-bit stereo audio converters, and should be externally decoupled with respect to V <sub>CMA</sub> (see Figure 17).
10	V <sub>CMA</sub>	O	Audio Common Mode Voltage Output. This output pin is the common mode voltage (AV <sub>DDA</sub> -AGNDA)/2. This output must be decoupled with respect to AGNDA (see Figure 17).
Analog Inputs			
15	L1	I	Left Line Input #1. Left analog input #1. Full scale input, with no gain, is 1V <sub>RMS</sub> , centered at V <sub>CMA</sub> .
16	L2	I	Left Line Input #2. Left analog input #2. Full scale input, with no gain, is 1V <sub>RMS</sub> , centered at V <sub>CMA</sub> .
17	L3	I	Left Line Input #3. Left analog input #3. Full scale input, with no gain, is 1V <sub>RMS</sub> , centered at V <sub>CMA</sub> .
14	MICL	I	Left Microphone Input. Microphone input for the left MIC channel, centered at V <sub>CMA</sub> . This signal can be either 1V <sub>RMS</sub> or 0.1V <sub>RMS</sub> depending on the preamp gain.
12	R1	I	Right Line Input #1. Right analog input #1. Full scale input, with no gain, is 1V <sub>RMS</sub> , centered at V <sub>CMA</sub> .
11	R2	I	Right Line Input #2. Right analog input #2. Full scale input, with no gain, is 1V <sub>RMS</sub> , centered at V <sub>CMA</sub> .
7	R3	I	Right Line Input #3. Right analog input #3. Full scale input, with no gain, is 1V <sub>RMS</sub> , centered at V <sub>CMA</sub> .
13	MICR	I	Right Microphone Input. Microphone input for the right MIC channel, centered at V <sub>CMA</sub> . This signal can be either 1V <sub>RMS</sub> or 0.1V <sub>RMS</sub> depending on the preamp gain.
6	MONO_IN	I	Monophonic input, centered at V <sub>CMA</sub> . This signal can be 1V <sub>RMS</sub> . This is a general purpose mono analog input that is normally used to mix the typical "beeper" signal on most computers into the audio system.
Analog Outputs			
1	LOUT1	O	Left Channel Output #1. Left channel analog output. Maximum signal is 1V <sub>RMS</sub> centered at V <sub>CMA</sub> .
2	LOUT2	O	Left Channel Output #2. Left channel analog output. Maximum signal is 1V <sub>RMS</sub> centered at V <sub>CMA</sub> .
3	ROUT1	O	Right Channel Output #1. Right channel analog output. Maximum signal is 1V <sub>RMS</sub> centered at V <sub>CMA</sub> .
4	ROUT2	O	Right Channel Output #2. Right channel analog output. Maximum signal is 1V <sub>RMS</sub> centered at V <sub>CMA</sub> .
5	MONO_OUT	O	Mono Output. This output is a summed analog output from the left and right and Mono input channels. Maximum signal is 1V <sub>RMS</sub> centered at V <sub>CMA</sub> .

## MODEM AND TELEPHONY CODEC (13 pins)

20	V <sub>REFNM</sub>	O	Modem and Telephony DAC and ADC Negative Reference Voltage Output. This pin provides the Negative Reference Voltage used by the 16-bit modem converters and the 16-bits telephony converters, and should be externally decoupled with respect to V <sub>CMM</sub> . (see Figure 17)
21	V <sub>REFPM</sub>	O	Modem and Telephony DAC and ADC Positive Reference Voltage Output. This pin provides the Positive Reference Voltage used by the 16-bit modem converters and the 16-bits telephony converters. The reference voltage, V <sub>REFM</sub> , is the voltage difference between the V <sub>REFPM</sub> and V <sub>REFNM</sub> outputs. V <sub>REFPM</sub> should be externally decoupled with respect to V <sub>CMM</sub> . $V_{REFM} = V_{REFPM} - V_{REFNM}$ (see Figure 17).
23	V <sub>CMM</sub>	O	Modem and Telephony Common Mode Voltage Output. This output pin is the common mode voltage (AV <sub>DDM</sub> -AGNDM)/2. This output must be decoupled with respect to GND (see Figure 17).



## PIN DESCRIPTION (continued)

Pin N°	Symbol	Type	Function
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## MODEM AND TELEPHONY CODEC (13 pins) (continued)

29	TP1Tx+	O	Modem #1 Differential Positive Output. This pin is the noninverting output of the fully differential transmit output modem #1.
30	TP1Tx-	O	Modem #1 Differential Negative Output. This pin is the inverting output of the fully differential transmit output modem #1. Outputs TP1Tx+ and TP1Tx- provide analog signals with maximum peak-to-peak amplitude $2 \times V_{REFM}$ , and must be followed by an external two pole smoothing filter. The external filter follows the internal single pole switch capacitor filter with cut-off frequency of $2 \times \text{FSM}$ . The cut-off frequency of the external filter must be greater than two times the sampling frequency (FSM), so that the combined frequency response of both the internal and external filters is flat in the band-pass.
27	TP1Rx+	I	Modem #1 Differential Positive Analog Input. This pin is the differential non-inverting ADC input.
28	TP1Rx-	I	Modem #1 Differential Negative Analog Input. This pin is the differential inverting receive input. The analog input peak-to-peak differential signal range must be less than $2 \times V_{REFM}$ , and must be preceded by an external single pole anti-aliasing filter. These filters should be set as close as possible to the TP1Rx+ (TP1Rx-) pins.
31	TP1Tx2 (LQ1)	O	Modem #1 Auxiliary Analog Output or left channel output in quadraphonic mode (LQ1). This pin is the single-ended auxiliary modem output. This output provides analog signals with maximum peak-to-peak amplitude $V_{REFM}$ , and must be followed by an external two pole smoothing filter. The external filter follows the internal single pole switch capacitor filter. The cut-off frequency of the external filter must be greater than two times the sampling frequency (FSM), so that the combined frequency response of both the internal and external filters is flat in the band-pass.
34 33 32	HandTx1, HandTx2, HandTx3 (TP2Tx) (RQ1)	O	Telephony Single-ended Outputs or right channel output in quadraphonic mode for TP2Tx (RQ1). These pins are the single-ended outputs of the analog smoothing filter. With maximum peak-to-peak amplitude $V_{REFM}$ , and must be followed by an external two pole smoothing filter. The external filter follows the internal single pole switch capacitor filter. The cut-off frequency of the external filter must be greater than two times the sampling frequency (FSM), so that the combined frequency response of both the internal and external filters is flat in the band-pass.
26 22	HandRx TP2Rx	I	Telephony Single-ended Input. These pins are the single-ended Telephony ADC input. The analog input peak-to-peak single-ended signal range must be less than $V_{REFM}$ , and must be preceded by an external single pole anti-aliasing filter. These filters should be set as close as possible to the TP2Rx, HandRx pins.

## HOST INTERFACE (25 pins)

Serial Interface #1 Dedicated for Stereo Audio Codec. The serial audio interface is synchronous with the audio sampling frequency (FSA).			
61	SIN1	I	Data In. Digital audio data to the DACs, control information, GPIO data are received by the STLC7549 via SIN1. Refer to "Serial Interface Bit Definition" on Page 26.
60	SOUT1	O	Data Out. Digital audio data from the ADCs, status information, GPIO data are output from the STLC7549 via SOUT1. Refer to "Serial Interface Bit Definition" on Page 26.
59	SCLK1	O	Serial Port #1 Bit Clock Output. Clocks the digital data into SIN1 and out of SOUT1 during the frame synchronization interval. The Serial bit clock is generated internally and is equal to the audio Master clock signal frequency $\text{MCLKA}/(4 \times N)$ where N depends on index register 20 contents.
58	FSYNC1	O	Serial Port #1 Frame Synchronization Output. The frame synchronization signal is used to indicate that the device is ready to send and receive data. The data transfer begins on the falling edge of the frame-sync signal. The frame-sync is generated internally and goes low on the rising edge of SCLK1. FSYNC1 and FSA have the same frequency.
55	MCLKA	I	Master Clock A for Audio Codec. Master clock input for audio codecs. This signal is the oversampling clock of the DA and AD convertor. It also provides all the clocks of the audio serial interface #1. This input must be driven by a signal with a frequency from 6.144MHz to 12.288MHz. In quadraphonic mode this input provide all Codecs clocks of the STLC7549.

7549-03.TBL

## PIN DESCRIPTION (continued)

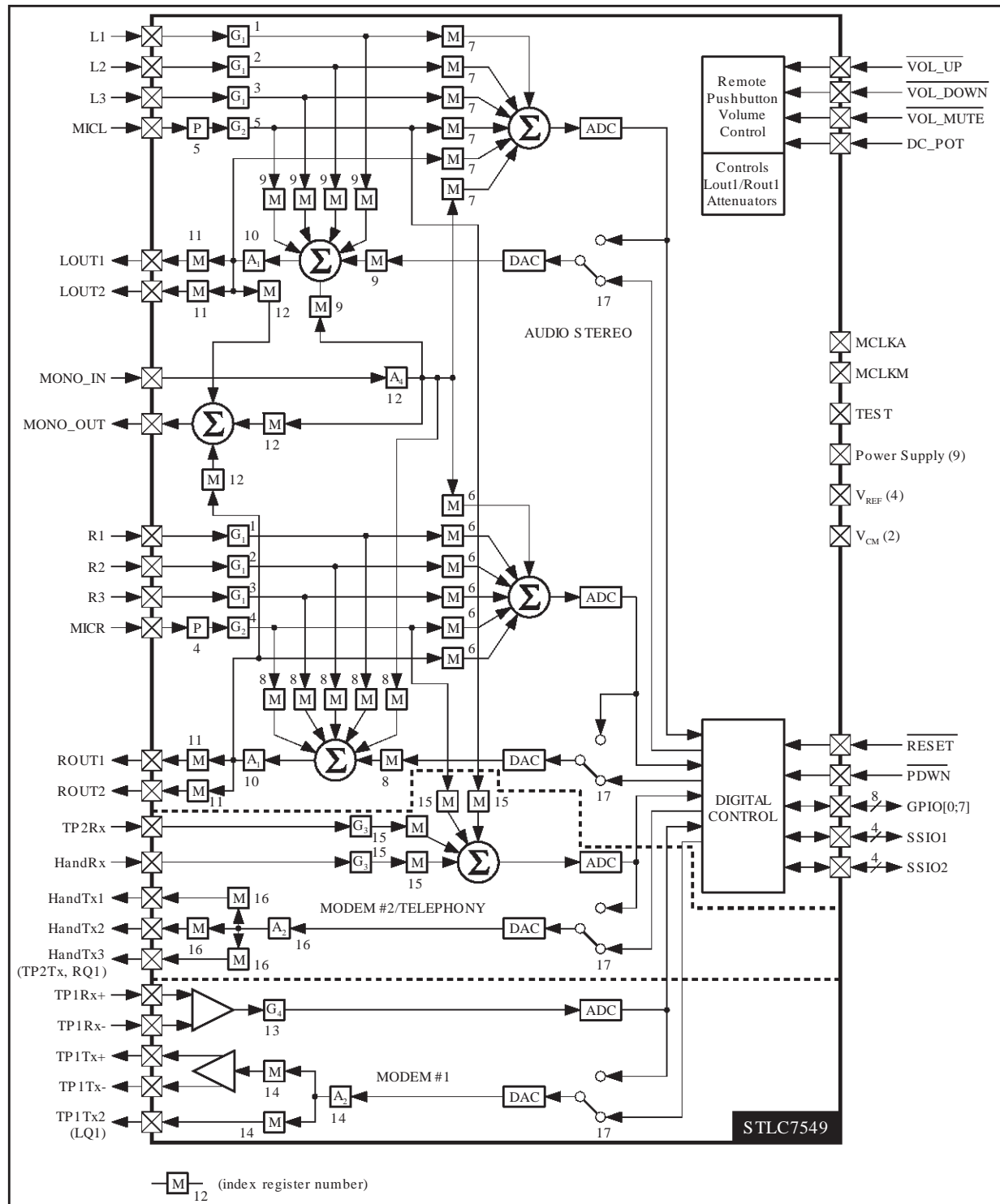
Pin N°	Symbol	Type	Function
HOST INTERFACE (25 pins) (continued)			
Serial Interface #2 Dedicated for Modem and Telephony Codecs. The serial modem interface is synchronous with the modem sampling frequency (FSM).			
54	SIN2	I	Data Input. Digital modem and telephony data to the DACs are received by the STLC7549 via SIN2.
53	SOUT2	O	Data Output. Digital modem and telephony data from the ADCs are output from the STLC7549 via SOUT2.
52	SCLK2	O	Serial Port #2 Bit Clock Output. Clocks the digital data into SIN2 and out of SOUT2 during the frame synchronization interval. The Serial bit clock is generated internally and equal to the Master clock signal frequency MCLKM/4. In quadrasonic mode SCLK2 is equal to MCLKA/4 (refer to Figure 6).
51	FSYNC2	O	Serial Port #2 Frame Synchronization Output. The frame synchronization signal is used to indicate that the device is ready to send and receive data. The data transfer begins on the falling edge of the frame-sync signal. The frame-sync is generated internally and goes low on the rising edge of SCLK2. FSYNC2 and FSM have the same frequency.
50	MCLKM	I	Master Clock M for Modem and Telephony Codecs. Master clock input for modem codecs. In quadrasonic mode this input is ignored. This signal is the oversampling clock of the DA and AD convertor. It also provides all the clocks of the modem serial interface #2. This input may be driven by a signal with a frequency from 1.8432MHz to 3.84MHz (up to 11.2896MHz in Quadrasonic mode).
Miscellaneous			
64	$\overline{\text{RESET}}$	I	Reset Function (active low). A reset function to initialize the internal counters and control register. A minimum low pulse of 100ns is required to reset the chip. This reset function initiates the serial data communications. The reset function will initialize all the registers to their default value and will put the device in a pre-programmed state. Master clocks are not necessary during RESET.
63	$\overline{\text{PDWN}}$	I	Power Down (active low). The Power-Down input powers down the entire chip to 0.5mW. When PDWN pin is taken low, the device is powered down such that the existing internally programmed state is maintained and all analog outputs are in high impedance. When PDWN is driven high, full operation resumes. If the PDWN input is not used, it should be tied to $V_{DD}$ .
62	TEST	I	Test Input. Digital input reserved for test. Should be connected to GND.
38	$\overline{\text{VOL\_UP}}$	I	Pushbutton Lineout Control Volume UP (edge sensitive, active low, internal pull-up). This pin increases the volume and also affects the left and right gain select values in register 10.
39	$\overline{\text{VOL\_DOWN}}$	I	Pushbutton Lineout Control Volume DOWN (edge sensitive, active low, internal pull-up). This pin decreases the volume and also affects the left and right gain select values in the register 10.
37	$\overline{\text{VOL\_MUTE}}$	I	Pushbuttons Lineout Control Volume MUTE (edge sensitive, active low, internal pull-up). This pin mutes and unmutes (toggle function) the Left and Right attenuators overriding the mute bits in registers 11 and 12 (refer to Figure 10).
36	DC_POT	I	DC Potentiometer Control Lineout Volume. An external potentiometer can be attached to this pin to determine a 4-bit value to be used as the main Left/Right output attenuator register value (register 10).
49 : 42	GPIO[0:7]	I/O	General Purpose I/Os. General purpose input/output pins*. These I/Os are configured as inputs at power-up or RESET.

\* Each bit can be configured as input or output (Register 18), set and read via Serial Interface #1 (see Figure 23).

7549-04.TBL



## BLOCK DIAGRAM



7549-04 EP-S

Figure 3 : Simplified Stereo Part Diagram

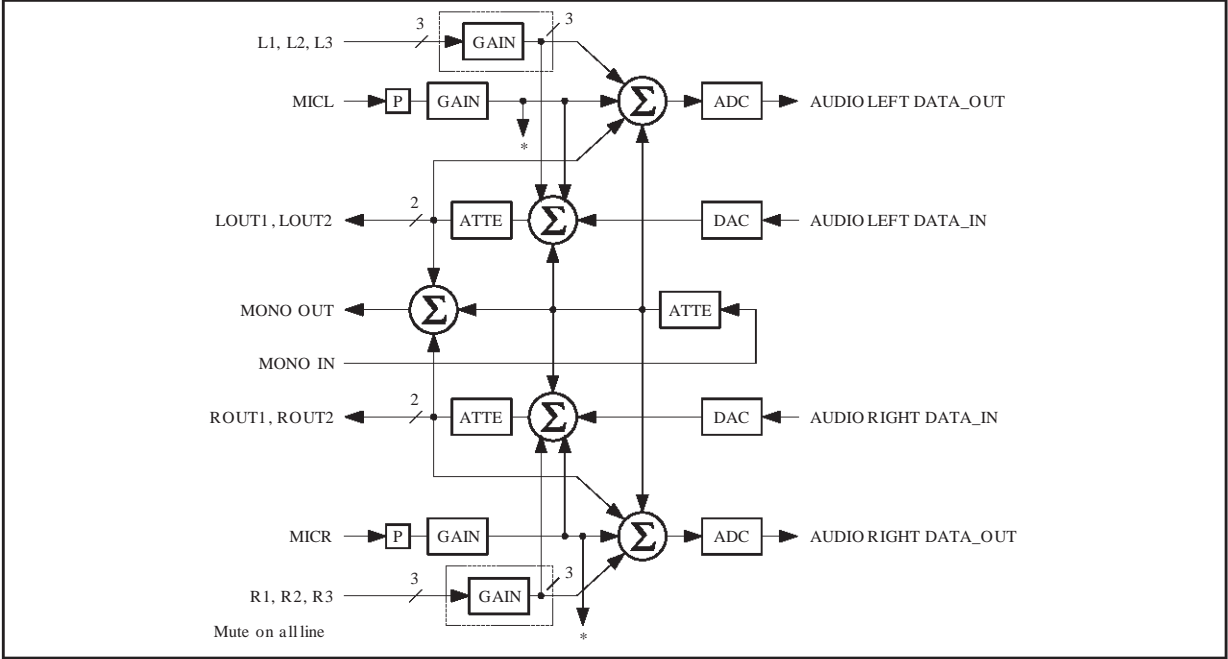


Figure 4 : Simplified Telephony/Modem #2 Part Diagram

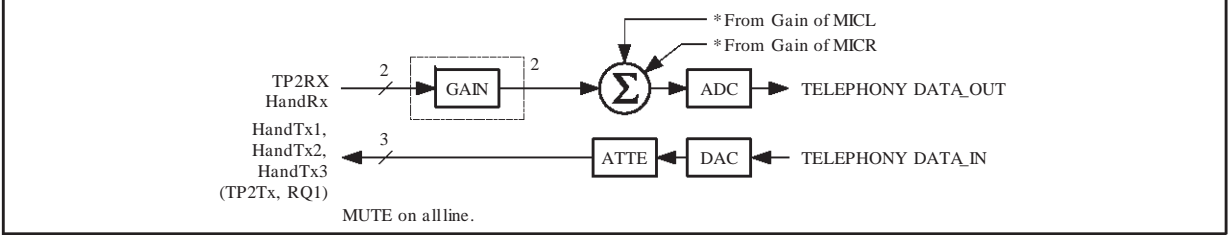


Figure 5 : Simplified Modem #1 Part Diagram

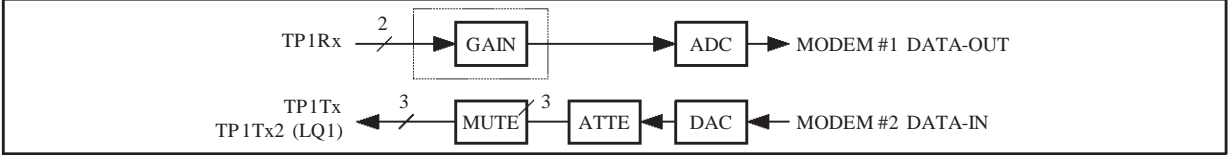
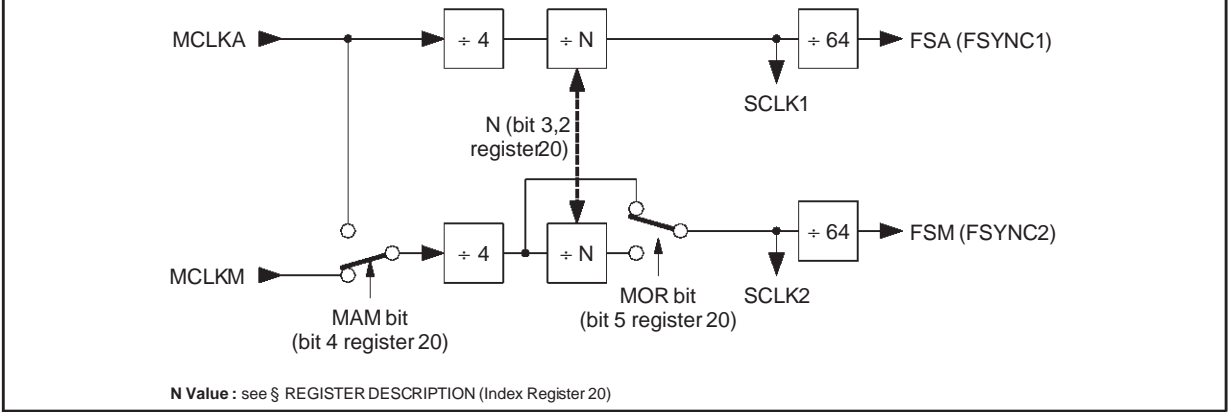


Figure 6 : Clock Generator Diagram



## FUNCTIONAL DESCRIPTION

### 1 - Volume Control

The STLC7549 provides a powerful set of volume control functions via pushbuttons (VOL\_MUTE, VOL\_DOWN and VOL\_UP input pins), a potentiometer (DC\_POT input pin) and control registers (Register 10 : Audio line output attenuator control, Register 11 : Audio output mute control , Register 12 : Mono input/output control, Register 19 : digital control Register #1).

If not used for volume control, pushbuttons and potentiometer can be used for different purpose since pushbutton state is reflected in bit 21, 22 and 23 of the serial output #1, and potentiometer value in index register 19.

#### 1.1 - Index Register 19

By setting the bits +PBVOL (pushbutton volume circuitry) or +DCVOL (DC volume control circuitry) the output attenuators will be affected (increment or decrement by 1 LSB the current 4-bit value of volume control Register 10) by the pushbuttons (VOL\_UP ,VOL\_DOWN) or by the potentiometer (DC voltage range pin DC\_POT to determine a 4-bit value).

The 4-bit value from the DC potentiometer can be read from Register 19 (bits 4 to 7). If the bits +PBVOL and +DCVOL are both set to "1" the priority is given to +DCVOL.

In pushbutton mode the serial interface is still able to modify the volume setting of Register 10. In potentiometer mode the serial interface can not modify Register 10.

Index Register 19		Control Mode	
+PBVOL	+DCVOL	VOLUME (1)	MUTE (2)
0	0	Serial Interface	Serial Interface
1	0	Pushbuttons/ Serial Interface	Pushbuttons/ Serial Interface
0	1	Potentiometer	Serial Interface
1	1	Potentiometer	Pushbuttons/ Serial Interface

Notes : 1. Register 10  
2. Registers 11,12

#### 1.2 - Pushbuttons Circuitry

When active - by setting bit +PB\_VOL in index register 19 - the VOL\_UP and VOL\_DOWN push-buttons affect the main left/right output attenuator in register 10.

VOL\_MUTE affect the following mute blocks :

- OUT1 left/right (MUTR1/MUTL1, register11)

- OUT2 left/right (MUTR2/MUTL2, register12)
- DAC left/right into mono output summer (MUTPCL/MUTPCR, register 12).

Mono\_In into Mono\_Out bit (MUTPCP, register 12) is not affected so that the PC beeper sound will be looped to the mono-output and so PC system sound can be heard during the mute mode.

When switching from serial interface control mode to pushbuttons control mode the contents of the Registers 10,11,12 are kept. If more than 1 push-button is pushed, then the priority is given to the one pushed first.

The action on VOL\_UP and VOL\_DOWN pushbuttons will increment or decrement the contents of the left and right output attenuator in Register 10. The status of the input pins are inverted and reflected in bits 21, 22 and 23 of the serial output #1 interface. The input pins are debounced before acting on registers value. The input pins are edge sensitive and active low.

The push-buttons are debounced as shown in Figures 7 and 8. The internal pull-up resistance is over 20kΩ.

Figure 7

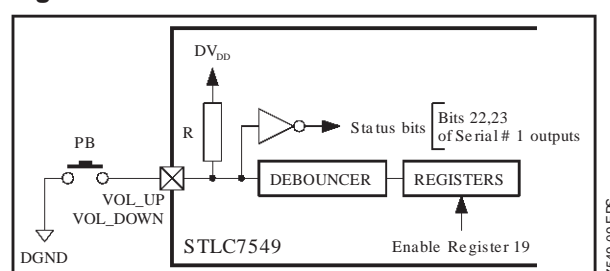
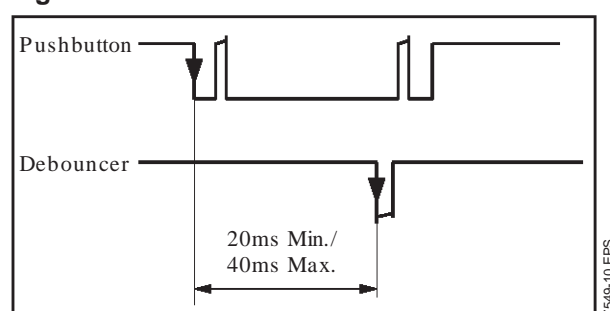


Figure 8



In order to better understand the MUTE pushbutton function please see Figure 10. After RESET the output of the TOGGLE/UNTOGGLE circuitry is in unmute position.

## FUNCTIONAL DESCRIPTION (continued)

## 1.3 - Potentiometer circuitry

With the internal DC to 4-bit value convertor you can control the output volume just with a simple potentiometer of 100k $\Omega$ . When switched to DC volume control mode, the contents of the volume Register 10 are up-dated with the 4-bit convertor value of the potentiometer circuitry. The 4-bit value volume control register is up-dated at least every 100ms. The potentiometer circuitry has an hysteresis of 1/2 LSB. The DC potentiometer circuitry is shown in Figure 9.

Maximum attenuation of -45dB is obtained when DC\_pot pin is set at DVDD potential. When at ground, attenuation is minimal (0dB).

## 2 - Mute Function

The mute function allows each input and output channel to be silenced independently. This function maintains the configuration (gain, attenuation, mixing configuration) during the mute and unmute. The output channels should be muted when the sampling frequency is changed.

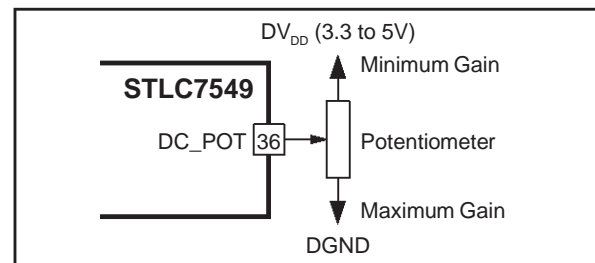
For the complete description of the mute possibi-

ties see :

- Register 6 (Right ADC Summer control)
- Register 7 (Left ADC Summer control)
- Register 8 (Right DAC Summer control)
- Register 9 (Left DAC Summer control)
- Register 11 (Audio output Mute control)
- Register 12 (Mono Input/Output control)
- Register 14 (Mono codec#1 output control)
- Register 15 (Mono codec#2 input control)
- Register 16 (Mono codec#2 output control)

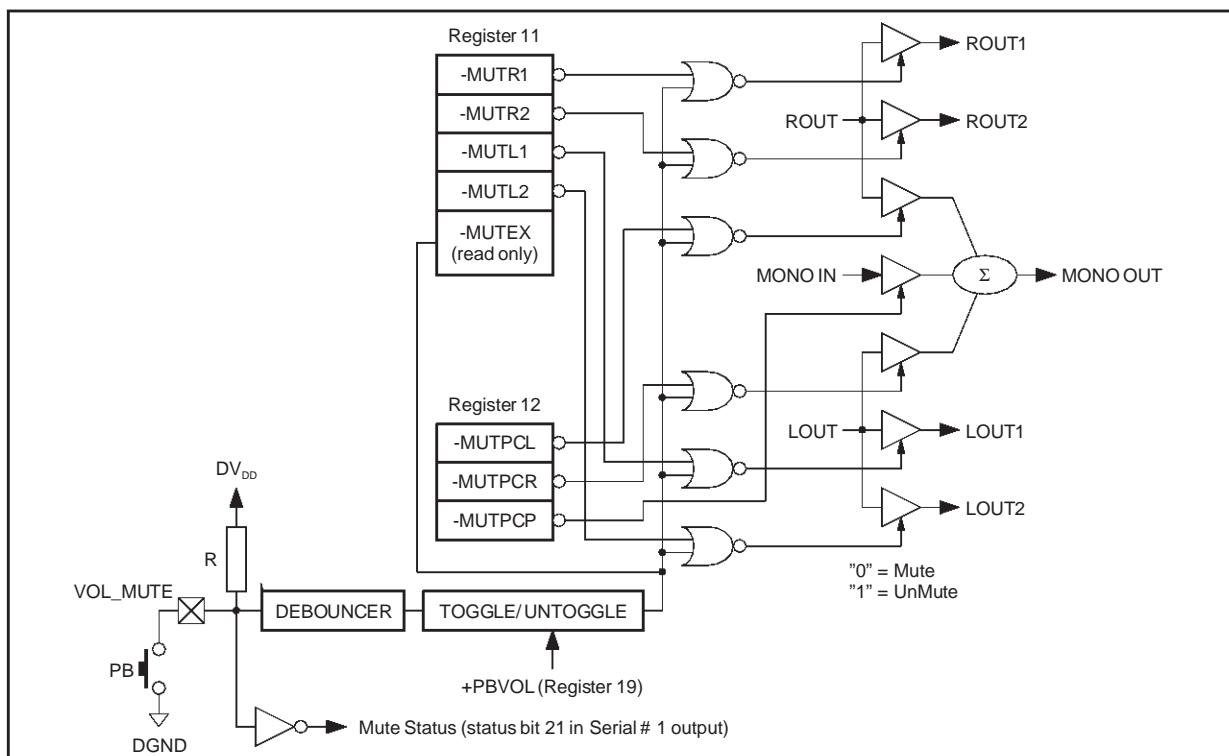
To provide the best mute attenuation for used output, it is advised to mute all unused outputs for the same channel.

Figure 9



7549-11.EPS

Figure 10 : Mute Function



7549-12.EPS

## FUNCTIONAL DESCRIPTION (continued)

### 3 - Input Gain and Output Gain Setting

Both input and output gain setting are internally made on zero crossing of the analog signal to minimize the "zipper" noise. The gain setting on zero crossing of the analog signal is not available in the preamplifier P. The gain change automatically takes effect if zero crossing does not occur within 512 frames. The 512 frames counter is initialized by the register gain access.

### 4 - A/D and D/A Converters

#### 4.1 - Offset Cancellation

The internal input offsets are minimized by internal offset cancellation circuitry. The calibration procedure is forced by setting bit 0 or bit 1 of Register 19 to "1". The length of time required for calibration is 1024 samples at the sampling frequency rate (eg:  $FSx = 44.1\text{kHz}$ ,  $T = 1024 * 1/FSx = 23\text{ms}$ ). When calibration is completed the bit 0 or bit 1 of Register 19 is reset to "0". During calibration the analog inputs are set to a high impedance state, output path are reset, and the data value at SOUT are not valid.

Calibration of all converters are lost during reset operation, and no calibration is done after Reset and Powerdown.

ADC offset is principally depending of chip temperature. It is advised to wait a while at power-up before proceeding to a calibration sequence, to allow the chip to take its typical temperature. Calibration can also be made in case of saturation to minimize distortion.

#### 4.2 - Converter Saturation

ADC and DAC conversion full range is  $1V_{RMS}$ . Each analog-to-digital converter is featured by an overflow flag. These flags are reflected in serial output #1 (bits 16, 17, 18 and 19) and can be cleared by setting bit 0, 1, 2 and 3 in register 21.

When several lines are mixed, signals should be attenuated before mixing in order to provide to the A/D converter a resulting summed signal within  $1V_{RMS}$ .

### 5 - Quadraphonic Mode

The STLC7549 offers the possibility to play music in quadraphonic mode. This mode will allow you to power your multimedia application by providing surround sound using RQ1 (Right Quadraphonic 1 output) and LQ1 (Left Quadraphonic 1 output) along with the normal stereo outputs Lout1 and Rout1.

This mode is programmed by setting the bit MAM (Modem Audio Mode, Register 20) and the bit MOR (Modem Oversampling Ratio, register 20) to "1". This will cause the modem codecs #1 and #2 clocks to be controlled from the audio master clock MCLKA ( $11.2896\text{MHz}$ ) and to have the same oversampling ratio than the audio part. The modem master clock MCLKM will be ignored (see CLOCK GENERATOR DIAGRAM) (Figure 6) and synchronization signals FSYNC1 and FSYNC2 will in phase.

In this mode the FSA must be programmed to  $44.1\text{kHz}$  (NDIV1 and NDIV0 of Register 20 equals zero so  $N=1$ ). FSM will be equal to  $MCLKA/4/64 = 44.1\text{kHz}$ .

### 6 - Analog Inputs and Outputs

#### 6.1 - Audio Analog Input

The audio analog input full scale is  $1V_{RMS}$ . Note that compact discs player output levels is  $2V_{RMS}$  centered around analog ground. To prevent eventual distortion, signal coming from CD player should be externally attenuated by a resistive divider.

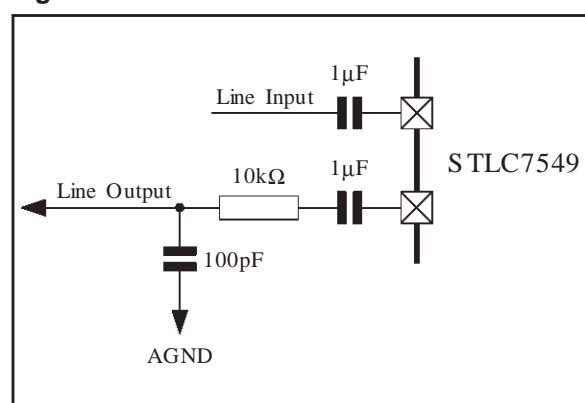
All analog inputs must have an external capacitance as shown on Figure 11.

#### 6.2 - Analog Audio Outputs

Analog audio output full scale is  $1V_{RMS}$  centered at  $V_{CMA}$ , and the minimum load must be  $10\text{k}\Omega$ . An external single pole smoothing filter is advised for large bandpass speakers. Suggested filter is shown on Figure 11.

STLC7549 integrates  $\sin x/x$  error corrector filters.

Figure 11



#### 6.3 - Modem Interface

Suggested duplexors for Modem interface are shown on Figures 12 and 13.

## FUNCTIONAL DESCRIPTION (continued)

Figure 12 : Differential Modem Duplexor

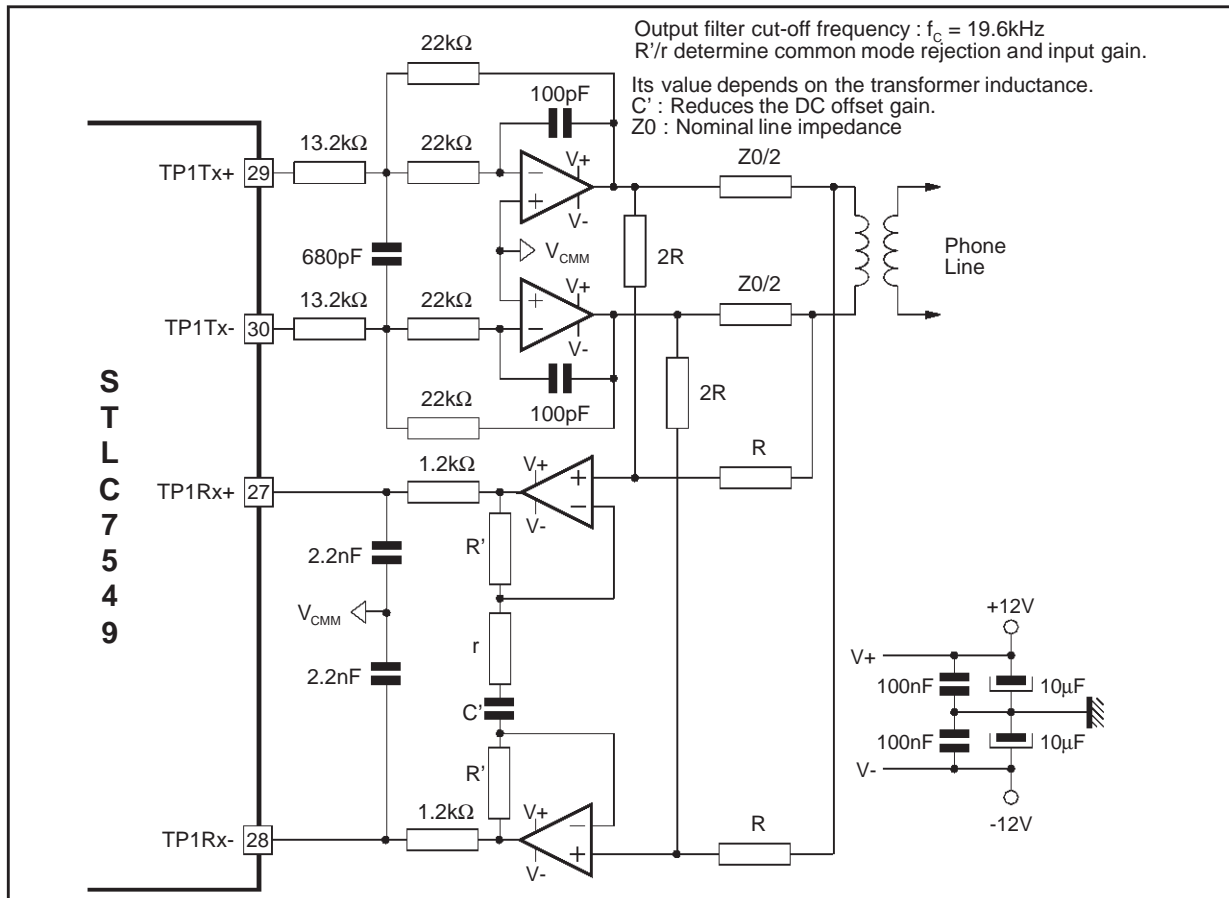
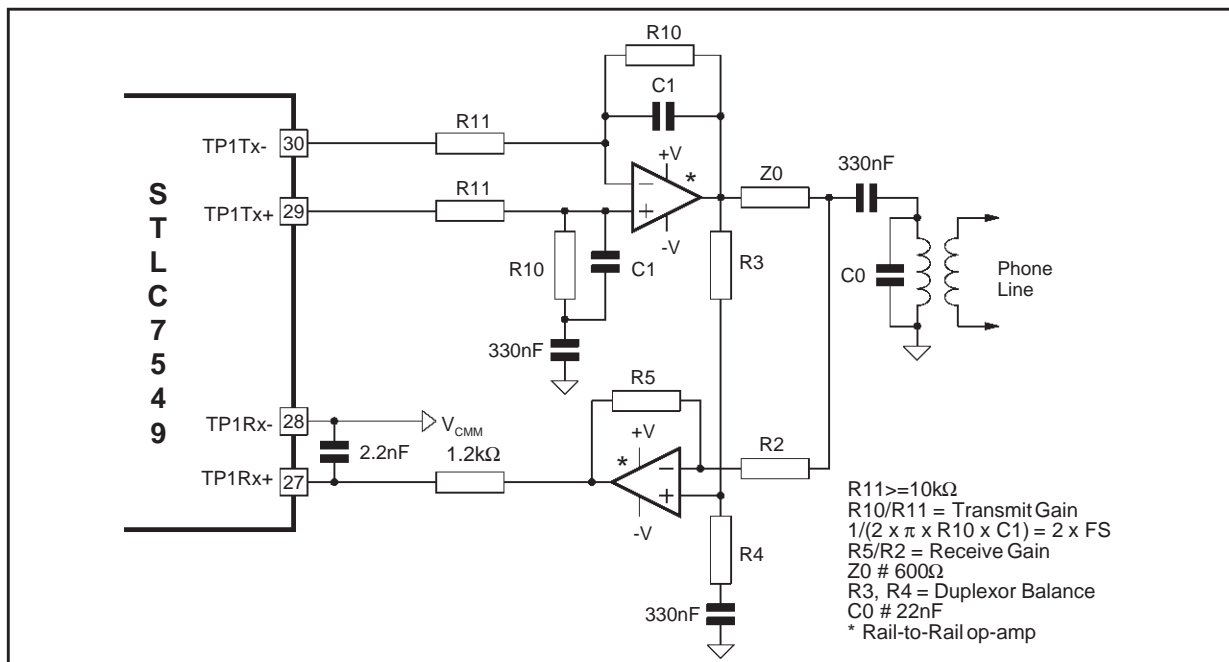


Figure 13 : Low-cost Modem Duplexor Mono Supply





**FUNCTIONAL DESCRIPTION** (continued)**7 - General Purpose Input/Output : GPIO [7:0]**

The STLC7549 offers 8 general purpose Input/Output pins. The setting of the GPIO configuration is done through Register 18 (bit set to "0" for input configuration and bit set to "1" for output configuration). On RESET the pins are configured as Input. The GPIO value is reflected in the serial data stream SOUT1 bits 24-31 (eg : could be used for Ring detect in modem application). When programmed as output the GPIO provides the way to control external devices using bits in the serial data input stream SIN1 bits 56-63 (eg: could be used for controlling relay drivers). The GPIO input voltage is independant of  $DV_{DD}$  and can be from 3.3V to 5V. GPI are latched at the beginning of the frame input word #1 (TSI11) and GPO are switching at the middle of the frame output word #0 (TSO10) as shown in Figure 15.

See also Figures 14 and 22.

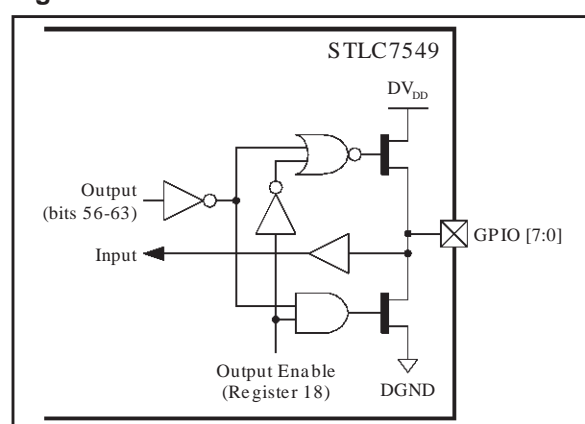
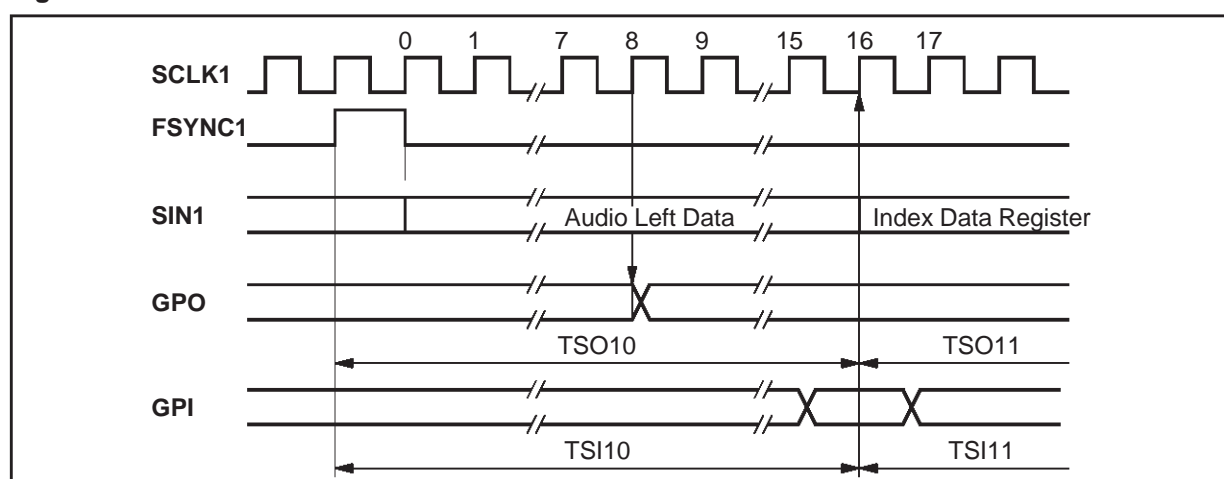
**8 - Clock Generator**

From the Master Clock Audio (MCLKA) and Master Clock Modem (MCLKM) the internal clock generator provides all the synchronous serial interface clocks (SCLK1, FSYNC1, SCLK2 and FSYNC2). The internal sampling frequency (FSA, FSM) used

for A/D and D/A conversion are equal to the synchronous frequency (FSYNC1 and FSYNC2 respectively). FSx and FSYNCx frequencies are not in phase.

In Quadraphonic mode the MCLKM input is ignored and the reference master clock is MCLKA. For proper operation the N divider (Register 20) must be set to 1 in order to have FSA and FSM equals.

The modem codecs can run with master clock higher than 3.84MHz (e.g. : 11.2896MHz) but with reduced performances.

**Figure 14****Figure 15**

## FUNCTIONAL DESCRIPTION (continued)

## 9 - Reset

The reset function initializes the internal counters and control registers. A minimum low pulse of 100ns is required to reset the chip. This reset function initiates the serial data communication.

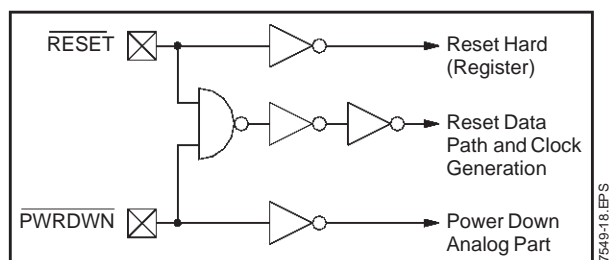
During a reset :

- analog outputs take respective VCMx values
- SCLK1 and SCLK2 are high,
- FSYNC1 and FSYNC2 are low,
- SOUT1 and SOUT2 are in "X" state,
- reference voltages VREFx and VCMx are not affected.

After a reset,

- registers are set to default values :
  - all mute blocks are set to mute except Mono\_In to Mono\_Out path in order to allow PC beep to be heard during configuration,
  - the output of TOGGLE/UNTOGGLE mute function is placed in unmute state,
  - gain and attenuation blocks takes the values shown in Table 1.,
  - sample frequency ratio NDIV (register 20) takes the value 1 (with a master clock frequency of 11.2896MHz), sample frequency is 44.1 kHz),
  - no loopback between ADC and DAC channels,
  - GPIO[7:0] are set to input configuration
  - no quadraphonic mode, Modem and Telephony part use MCLKM clock input without N divider,
  - no blocks in powerdown mode.
- the calibrations of all converters are low,
- all the internal data paths are set to "0" (see Figure 16).

Figure 16



At power-up sequence, a reset is required by the DSP or with a RC network on the NRESET DSP Pin.

Table 1

Default Values	Reg.	Gain/Attenuation
Line inputs #1 gain	1	-24dB
Line inputs #2 gain	2	-24dB
Line inputs #3 gain	3	-24dB
Right microphone input gain	4	0dB
Right microphone pre-amp gain	4	0dB
Left microphone input gain	5	0dB
Left microphone pre-amp gain	5	0dB
Audio line output attenuation	10	0dB
Mono input gain	12	-10dB
Modem Codec #1 input gain	13	0dB
Modem Codec #1 output attenuation	14	0dB
Modem Codec #2 input gain	15	-4dB
Modem Codec #2 output attenuation	16	0dB

## 10 - Powerdown

The PDWN input powers down the entire chip. Minimum power consumption is obtained when the MCLKx clocks are stopped.

When PWDN Pin is taken low,

- the exiting programmed state is maintained,
- data path and clock generation are reset (see Figure 16),
- analog outputs are in high impedance state (MONO\_OUT included),
- SCLK1 and SCLK2 are high,
- FSYNC1 and FSYNC2 are low,
- SCOUT1 and SCOUT2 are in "x" state,
- VCMx Pin takes their respective supply values ( $V_{CMA} \geq AV_{VDA}$  and  $V_{CMM} \geq AV_{DDM}$ ) through an approximatively 1k $\Omega$  resistor,
- VREFx Pin follows respective VCMx values, i.e. respective supply values through approximatively 10k $\Omega$  resistor ( $V_{REFPA}$  and  $V_{REFNA} \geq AV_{DDA}$ ,  $V_{REFPM}$  and  $V_{REFNM} \geq AV_{DDM}$ ).

In order to limit eventual noises, it is advised to reset the chip during exiting powerdown mode until the chip is completely wake up. Times for entering or exiting powerdown mode depends of capacity values C placed on voltage references Pins VCMx and VREFx.

$$T = K \cdot C \text{ with } K = 30.000 \text{sec/F}$$

With recommended value  $C = 10\mu\text{F}$  (see Figure 14),  $T \# 300\text{ms}$ . When not used, this pin should be tied to DVDD.

The STLC7549 provides also 3 software modes detailed in register 20.

## FUNCTIONAL DESCRIPTION (continued)

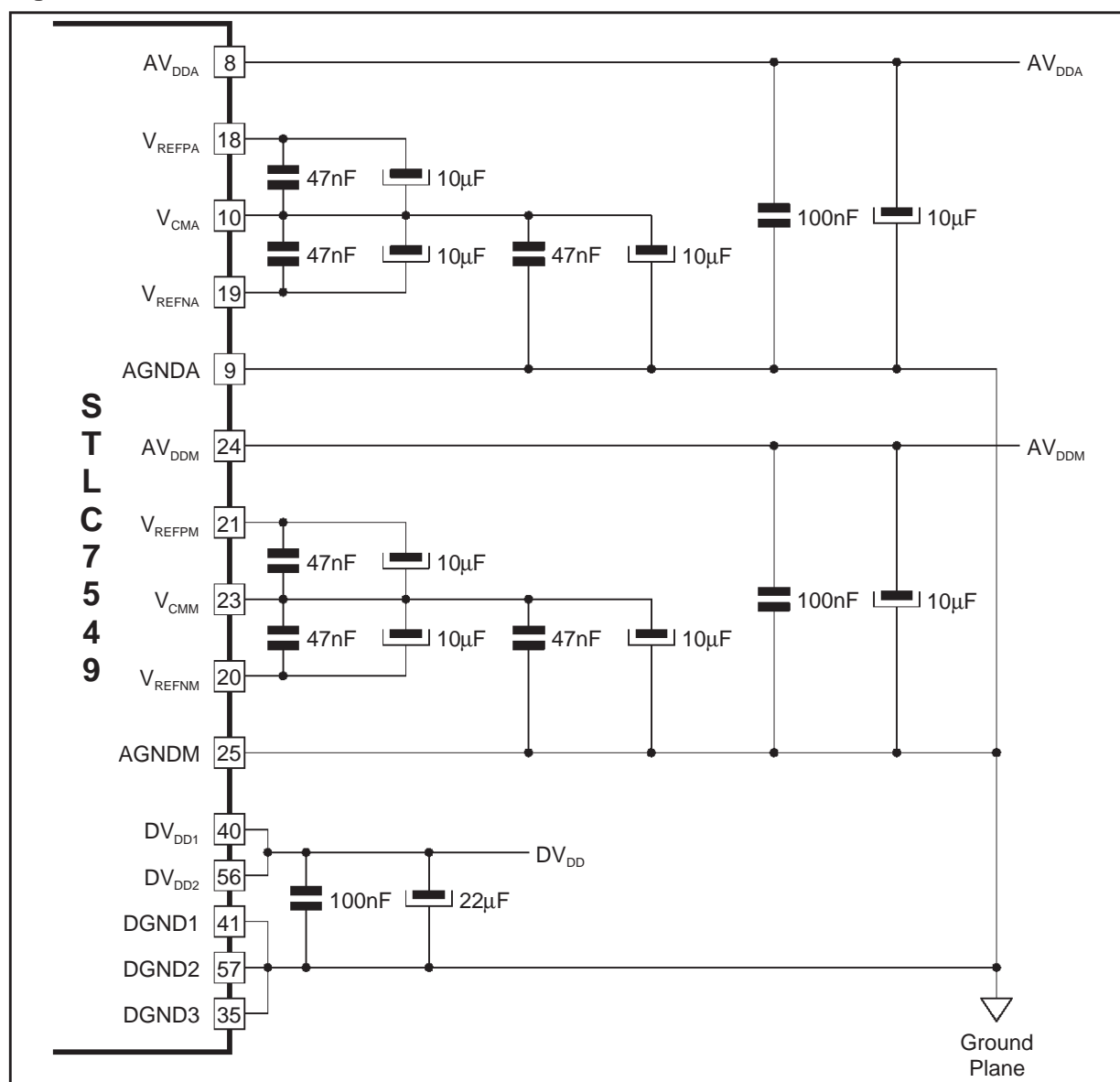
## 11 - Voltage Decoupling

To obtain published performance, the analog  $AV_{DD}$ , digital  $DV_{DD}$  and common mode voltage  $V_{CMM}$  ( $V_{CMA}$ ) should be decoupled with respect to AGND for  $AV_{DD}$  and  $V_{CMM}$  ( $V_{CMA}$ ) and to DGND for  $DV_{DD}$ . The decoupling is intended to isolate digital noise

from the analog section. Decoupling capacitors should be as close as possible to the voltage pin.

All the ground pins must be tied together (see following schematics).

Figure 17



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## ELECTRICAL SPECIFICATION

Unless otherwise noted, Electrical Characteristics are specified over the operating range.

Typical values are given for  $V_{DD} = +5V$ ,  $T_{amb} = 25^{\circ}C$  and for nominal master clocks frequency MCLKA = 11.2896MHz and MCLKM = 2.4576MHz.

### ABSOLUTE MAXIMUM RATINGS (AGND = DGND = 0V, all voltages with respect to 0V)

Symbol	Parameter	Value	Unit
$AV_{DD}$	Analog Power Supply	-0.3, 6.0	V
$DV_{DD}$	Digital Power Supply	-0.3, 6.0	V
$I_I$	Input Current per Pin (except supply pins)	-10, 10	mA
$I_O$	Output Current per Pin (except supply pins)	$\pm 20$	mA
$V_{IA}$	Analog Input Voltage	-0.3, $AV_{DD} + 0.3$	V
$V_{ID}$	Digital Input Voltage	-0.3, $DV_{DD} + 0.3$	V
$V_{IDGPI/O}$	Digital Input Voltage at GPI/O	5.25	V
$T_{oper}$	Operating Temperature	0, +70	$^{\circ}C$
$T_{stg}$	Storage Temperature	-40, +125	$^{\circ}C$
$P_{DMAX}$	Maximum Power Dissipation	1500	mW
ESD	Electrostatic Discharge all Pins (except Pins 40-42 to 49-52-53-56-59 to 62-64 = 1500V)	> 2000	V

**Warning :** Operation beyond these limits may result in permanent damage to the device. Normal operation is not guaranteed at these extremes.

### RECOMMENDED OPERATING CONDITIONS (AGND = DGND = 0V, all voltages with respect to 0V)

Symbol	Parameter	Min.	Typ.	Max.	Unit
--------	-----------	------	------	------	------

#### SINGLE POWER SUPPLY ( $DV_{DD} = AV_{DD}$ )

$V_{DD}$	Supply Voltage	4.75	5	5.25	V
$I_D$	Digital Supply Current		95		mA
$I_A$	Analog Supply Current		55		mA
$I_{LPH}$	Supply Current in Low Power Hardware Mode (PDWN = "0") (MCLKs stops)		10		$\mu A$
$I_{LPSW1}$	Supply Current in Low Power Software Mode 1 (see Register 20)		TBD		$\mu A$
$I_{LPSW2}$	Supply Current in Low Power Software Mode 2		TBD		$\mu A$
$I_{LPSW3}$	Supply Current in Low Power Software Mode 3		TBD		$\mu A$
$P_D$	Power Dissipation		750		mW
$V_{CM}$	Common Mode Voltage Output	$AV_{DD}/2 - 5\%$	$AV_{DD}/2$	$AV_{DD}/2 + 5\%$	V

#### DUAL POWER SUPPLY ( $DV_{DD} \neq AV_{DD}$ )

$DV_{DD}$	Digital Supply Voltage	3.15	3.3	3.45	V
$I_D$	Digital Supply Current		50		mA
$AV_{DD}$	Analog Supply Voltage	4.75	5.0	5.25	V
$I_A$	Analog Supply Current		55		mA
$I_{LPH}$	Supply Current in Low Power Hardware Mode (PDWN = "0") (MCLKs stops)		TBD		$\mu A$
$I_{LPSW1}$	Supply Current in Low Power Software Mode 1 (see Register 20)		TBD		$\mu A$
$I_{LPSW2}$	Supply Current in Low Power Software Mode 2		TBD		$\mu A$
$I_{LPSW3}$	Supply Current in Low Power Software Mode 3		TBD		$\mu A$
$P_D$	Power Dissipation		450		mW
$V_{CM}$	Common Mode Voltage Output	$AV_{DD}/2 - 5\%$	$AV_{DD}/2$	$AV_{DD}/2 + 5\%$	V
$I_{CM}$	Common Mode Current Output (see Note 1)		100		$\mu A$

**Note 1 :** DC current only IF dynamic loading exists, then the common mode voltage output must be buffered or the performance of ADCs and DACs will be degraded

**ELECTRICAL SPECIFICATION (continued)****AUDIO**

$AV_{DD} = 5V$ ,  $DV_{DD} = 3.3V$  ; Input levels : Logic 0 = 0V, Logic 1 =  $DV_{DD}$  ;

1kHz input sine wave ; Conversion rate = 44.1kHz ;

Measurement bandwidth is 20Hz to 20kHz, 16-bit linear coding for audio, microphone inputs.

Gain setting and Attenuation (0dB) ; 0dBr = 1V<sub>RMS</sub> (sine wave) ; Load impedance 10k $\Omega$ , 20pF ;

Unless otherwise specified.

**Analog Input Characteristics**

Symbol	Parameter	Min.	Typ.	Max.	Unit
ADR	ADC Resolution	16			Bits
DNL	ADC Differential Nonlinearity (see Note 1)	-0.9		+0.9	LSB
SID	Signal to Intermodulation Distortion		85		dB
ICl	Interchannel Isolation ( $f = 1\text{kHz}$ , $V_{\text{input}} = 200\text{mV}_{PP}$ )	85			dB
ICGM	Interchannel Gain Mismatch				
	Line Inputs	-0.5		+0.5	dB
	Microphone Inputs	-0.5		+0.5	dB
PIGS	Programmable Input Gain Span				
	Line Inputs	28	30	32	dB
	Microphone Inputs	55.5	57.5	59.5	dB
GSEr	Gain Step Size Error (see Note 3)				
	Line Inputs			0.5	dB
	Microphone Inputs			0.5	dB
OFFr	ADC Offset Error				
	Line Inputs		10	100	LSB
	(microphone pre-amplifier & G2 at maximum gain) Microphone Inputs		600	1000	LSB
FSI	Full Scale Input Voltage				
	Line and Microphone Inputs	2.8	2.9	3.0	V <sub>PP</sub>
R <sub>IN</sub>	Input Resistance (see Note 1)				
	Line Inputs	100			k $\Omega$
	Microphone Inputs	10			k $\Omega$
C <sub>IN</sub>	Input Capacitance (see Note 1)			15	pF
SNDR	Signal / Noise + Distortion at -6dBr ; 1kHz				
	ADC Audio		79		dB
	ADC Micro		70		dB
DR	Dynamic Range (see Note 2)		85		dB

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**Analog Output Characteristics**

Symbol	Parameter	Min.	Typ.	Max.	Unit
DAR	DAC Resolution	16			Bits
DNL	DAC Differential Nonlinearity (see Note 1)	-0.9		+0.9	LSB
SNDR	Signal / Noise + Distortion at -10dBr		75		dB
DR	Dynamic Range (see Note 2)		85		dB
SID	Signal to Intermodulation Distortion		85		dB
ICl	Interchannel Isolation				
	Line Out	85			dB
ICGM	Interchannel Gain Mismatch				
	Line Out	-0.5	0.1	+0.5	dB
PAS	DAC Programmable Attenuation Span	42	45	48	dB
ASS	DAC Attenuation Step Size	2.5	3	3.5	dB
FSO	Full Scale Output Voltage	2.8	2.9	3.0	V <sub>PP</sub>
R <sub>L</sub>	Load Resistance	10			k $\Omega$
C <sub>L</sub>	Load Capacitance			20	pF
R <sub>OUT</sub>	Output Resistance		20		$\Omega$
MAtt	Mute Attenuation (0dB gain)		85		dB
OBE	Total Out-of-band Energy (see Note 1)			-45	dBr
	Fs/2 to 100kHz				

7549-08.TBL

**Notes :** 1. This specification is guaranteed by characterization, not production testing.

2. DR measured over the full bandwidth 0Hz to FSA/2 with -20dBr signal extrapolated to full scale.

3. The gain step size error is the deviation of a gain or attenuation step from a straight line passing through the no-gain/attenuation value and the ideal full-gain/attenuation value.

**ELECTRICAL SPECIFICATION (continued)****MODEM AND TELEPHONY**

$AV_{DD} = 5V$ ,  $DV_{DD} = 3.3V$  ; Input levels : Logic 0 = 0V, Logic 1 =  $DV_{DD}$  ; 1kHz input sine wave ;  
 Conversion rate = 9.6kHz. Measurement bandwidth is 100Hz to 4800Hz, 16-bit linear coding for modem inputs and 16-bit linear coding for telephony inputs. For quadraphonic mode the measurement are done with  $MCLKA = 11.2896MHz$  and  $MAMbit$  set to 1 in Register 20. Gain settings and attenuation (0dB) ;  
 $0dBr = 2 \times V_{REFM}$  peak-to-peak ; Load impedance 10k $\Omega$ , 20pF ; unless otherwise specified.

**Analog Input Characteristics**

Symbol	Parameter	Min.	Typ.	Max.	Unit
ADR	ADC Resolution Modem #1 Modem #2	16 16			Bits Bits
DNL	ADC Differential Nonlinearity (see Note 1)	-0.9		+0.9	LSB
SID	Signal to Intermodulation Distortion		85		dB
ICI	Interchannel Isolation ( $f = 1kHz$ , $V_{input} = 200mV_{PP}$ )	85			dB
PIGS	Programmable Input Gain Span Modem #1 Modem #2		6 12		dB dB
GSEr	Gain Step Size Error (see Note 3)			0.5	dB
OFFr	ADC Offset Error Modem #1 Modem #2		10 10	100 100	LSB LSB
$V_{REFM}$	Differential Reference Voltage Output = $V_{REFPM} - V_{REFNM}$	2.4	2.5	2.6	V
FSI	Full Scale Input Voltage Modem #1 (differential) Modem #2 (single-ended)		$2 \times V_{REFM}$ $V_{REFM}$		V V
$R_{IN}$	Input Resistance Modem #1 Modem #2	100 100			k $\Omega$ k $\Omega$
$C_{IN}$	Input Capacitance		15		pF
SNDR	Signal / Noise + Distorsion at -6dBr ; 1kHz Modem #1 Modem #2 Quadraphonic Mode		84 79 70		dB dB dB
DR	Dynamic Range (see Note 2) Modem #1 Modem #2 Quadraphonic Mode		90 85 76		dB dB dB

- Notes :** 1. This specification is guaranteed by characterization, not production testing.  
 2. DR measured over the full bandwidth 0Hz to  $FSA/2$  with -20dBr signal extrapolated to full scale.  
 3. The gain step size error is the deviation of a gain or attenuation step from a straight line passing through the no-gain/attenuation value and the ideal full-gain/attenuation value.

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**ELECTRICAL SPECIFICATION (continued)****MODEM AND TELEPHONY (continued)****Analog Output Characteristics**

Symbol	Parameter	Min.	Typ.	Max.	Unit
DAR	DAC Resolution	Modem #1 16 Modem #2 16			Bits Bits
DNL	DAC Differential Nonlinearity (see Note 1)	-0.9		+0.9	LSB
SNDR	Signal / Noise + Distorsion at -6dBr Signal / Noise + Distorsion at -10dBr	Modem #1 Modem #2 Quadraphonic Mode	84 75 70		dB dB dB
DR	Dynamic Range (see Note 2)	Modem #1 Modem #2 Quadraphonic Mode	90 85 80		dB dB dB
SID	Signal to Intermodulation Distortion		85		dB
Ici	Interchannel Isolation	85			dB
Gabs	Absolute Gain at 1kHz	-0.5	0	+0.5	dB
PAS	DAC Programmable Attenuation Span	Modem #1 Modem #2	6 6		dB dB
VOFF	DAC Offset Voltage	-100		100	mV
FSO	Full Scale Output Voltage	Modem #1 (differential output) Modem #2 (single ended output)	$2 \times V_{REFM}$ $V_{REFM}$		V V
R <sub>L</sub>	Load Resistance	10			kΩ
C <sub>L</sub>	Load Capacitance			20	pF
R <sub>OUT</sub>	Output Resistance		20		Ω
MAtt	Mute Attenuation		85		dB
OBE	Total Out-of-band Energy (see Note 1)	Fs/2 to 100kHz		-45	dBr

**Notes :** 1. This specification is guaranteed by characterization, not production testing.

2. DR measured over the full bandwidth 0Hz to FSA/2 with -20dBr signal extrapolated to full scale.

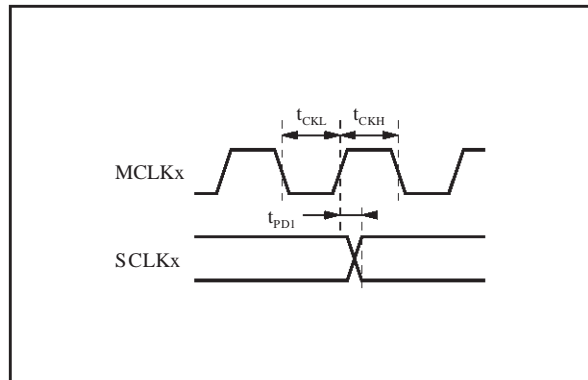
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**ELECTRICAL SPECIFICATION (continued)**

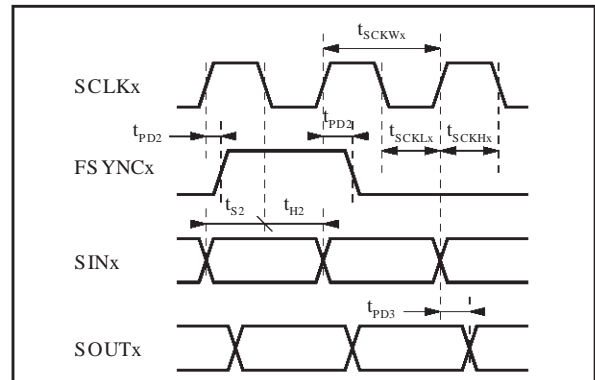
**SWITCHING CHARACTERISTICS** ( $AV_{DD} = DV_{DD} = +5V$ ,  $AGND = DGND = 0V$ ,  
Outputs loaded with 30pF, Input Levels : Logic 0 = 0V, Logic 1 =  $DV_{DD}$ )

Symbol	Parameter	Min.	Typ.	Max.	Unit
MCLKA MCLKM	Input Clock (MCLKx) Frequency Modem Mode Quadraphonic Mode	6.144 1.8432	11.2896 2.4576 11.2896	12.288 3.84	MHz MHz MHz
$t_{PW}$	Master Clock Period		$1/MCLKx$		-
$t_{PW}/t_{CKH}$ $t_{PW}/t_{CKL}$	Master Clock (MCLKx) Duty Cycle Pulse Width High Pulse Width Low	20 20		80 80	% %
$t_{PD1}$	SCLKx Output Delay from MCLKx Rising Edge			35	ns
$t_{PD2}$	FSYNCx Delay Time			15	ns
$t_{S2}$	SINx Set-up Time	15			ns
$t_{H2}$	SINx Hold Time from SCLKx Edge	10			ns
$t_{PD3}$	SOUTx Delay from SCLKx Edge			25	ns
$t_{SCKW1}$	SCLK 1 Period		$1/(64 \times FSA)$		s
$t_{SCKH1}$	SCLK 1 High Time	36			ns
$t_{SCKL1}$	SCLK 1 Low Time	36			ns
$t_{SCKW2}$	SCLK 2 Period Modem Mode Quadraphonic Mode		$1/(64 \times FSM)$ $1/(64 \times FSA)$		s s
$t_{SCKH2}$	SCLK 2 High Time Modem Mode Quadraphonic Mode	1730 36			ns ns
$t_{SCKL2}$	SCLK 2 Low Time Modem Mode Quadraphonic Mode	1730 36			ns ns
FSA	Audio Sample Frequency (N defined in Reg 20)		$MCLKA / (4 \times N \times 64)$		Hz
FSM	Modem & Telephony Sample Frequency		$MCLKM / (4 \times 64)$		Hz

7549-11.TBL

**Figure 18 : SCLKx Output Timing**

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**Figure 19**

7549-21.EPS

**ELECTRICAL SPECIFICATION (continued)****DIGITAL FILTER CHARACTERISTICS**

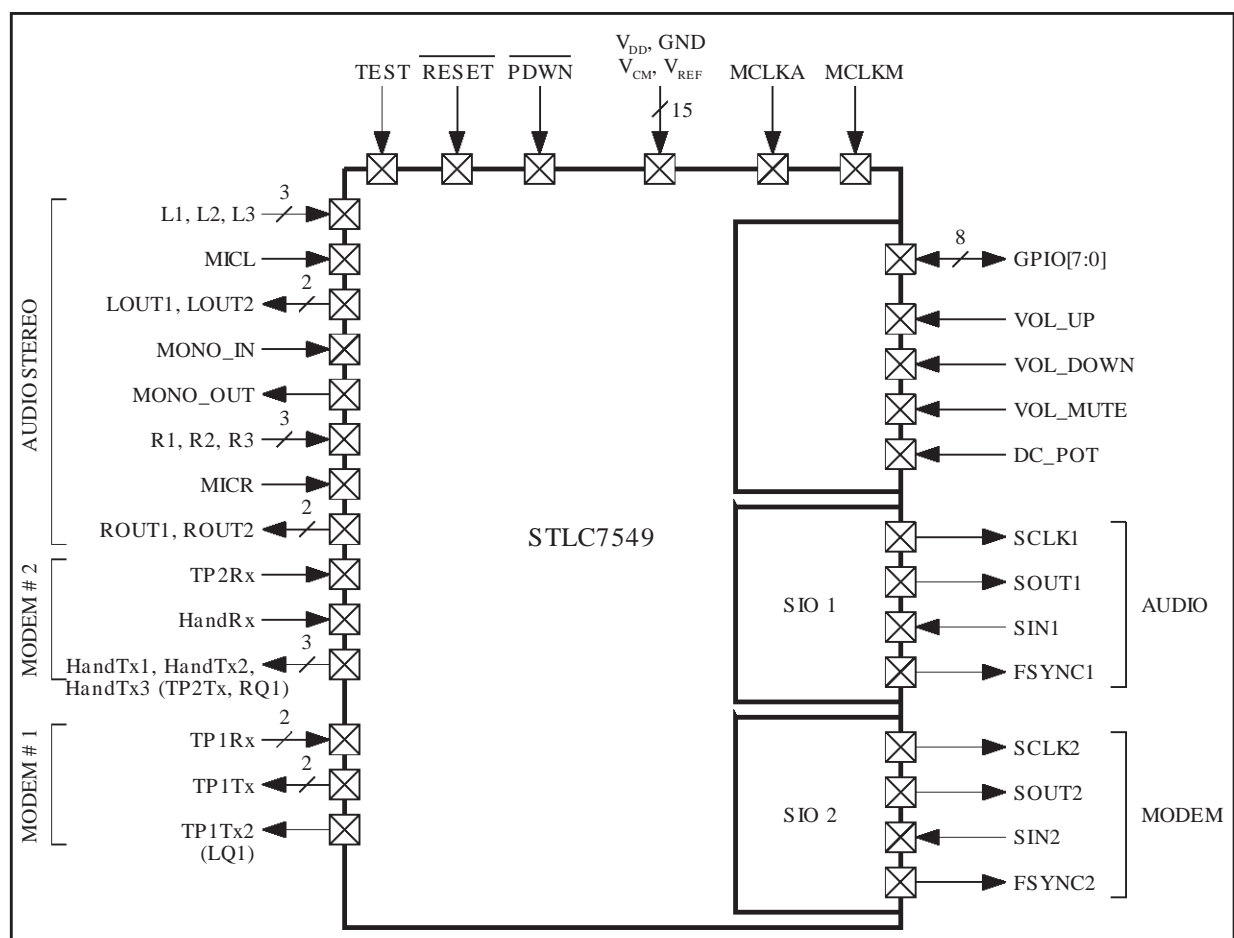
Symbol	Parameter	Min.	Typ.	Max.	Unit
	Passband	0		$0.42 \times F_s$	Hz
	Abs Gain at 1kHz	-0.5		+0.5	dB
	Passband Ripple			+0.2	dB
	(0 - 0.4 x $F_s$ )	-0.2			
	Transition Band	$0.4 \times F_s$		$0.6 \times F_s$	Hz
	Stop Band	$0.6 \times F_s$			Hz
	Stop Band Rejection	65			dB

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**DIGITAL CHARACTERISTICS** ( $AV_{DD} = DV_{DD} = +5V$  ;  $AGND = DGND = 0V$ )

Symbol	Parameter	Min.	Typ.	Max.	Unit
$V_{IH}$	High Level Input Voltage	2.2		$V_{DD} + 0.3$	V
$V_{IL}$	Low level Input Voltage	-0.3		0.8	V
$V_{OH}$	High Level Output Voltage	2.4		$V_{DD}$	V
$V_{OL}$	Low Level Output Voltage			0.4	V
	Input Leakage Current	-10	$\pm 1$	10	$\mu A$

7549-13.TBL

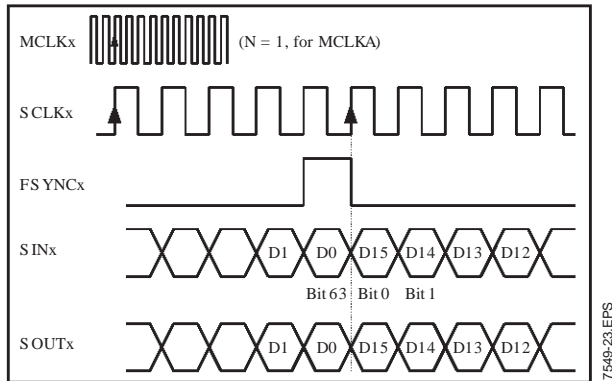
**INPUT/OUTPUT DESCRIPTION**

7549-22.EPS

## SERIAL INTERFACE

### Serial Interface Operation

**Figure 20 : Serial Interface Timing**



Serial data input is initiated by a frame synchro signal (FSYNC). The data is clocked from SIN into the input shift register (ISR) on the falling edge of SCLK and transferred to the input buffer register (IBR) when a complete 16-bit word has been received, the register is loaded 8 bits clocks later. Data is assumed to be received MSB first.

Serial data output is initiated by a frame synchro signal (FSYNC). The 16-bit data word is loaded into the output shift register (OSR) and serially clocked out of OSR to SOUT on the rising edge of SCLK.

The data/index register of SOUT1 refer to the index register set of SIN1 within the same frame.

The SCLK is 64 times FSYNC. This mean that the frame contains four slots of 16 bits. The time slots used for circuit function are indicated on the next paragraph.

#### Serial port 1 and 2

- Frame Synchronization Output (FSYNC)
- Serial Bit Clock Output (SCLK)
- Input Serial Register (ISR)
- Input Buffer Register (IBR)
- Serial Input Data (SIN)
- Output Buffer Register
- Output Serial Register (OSR)

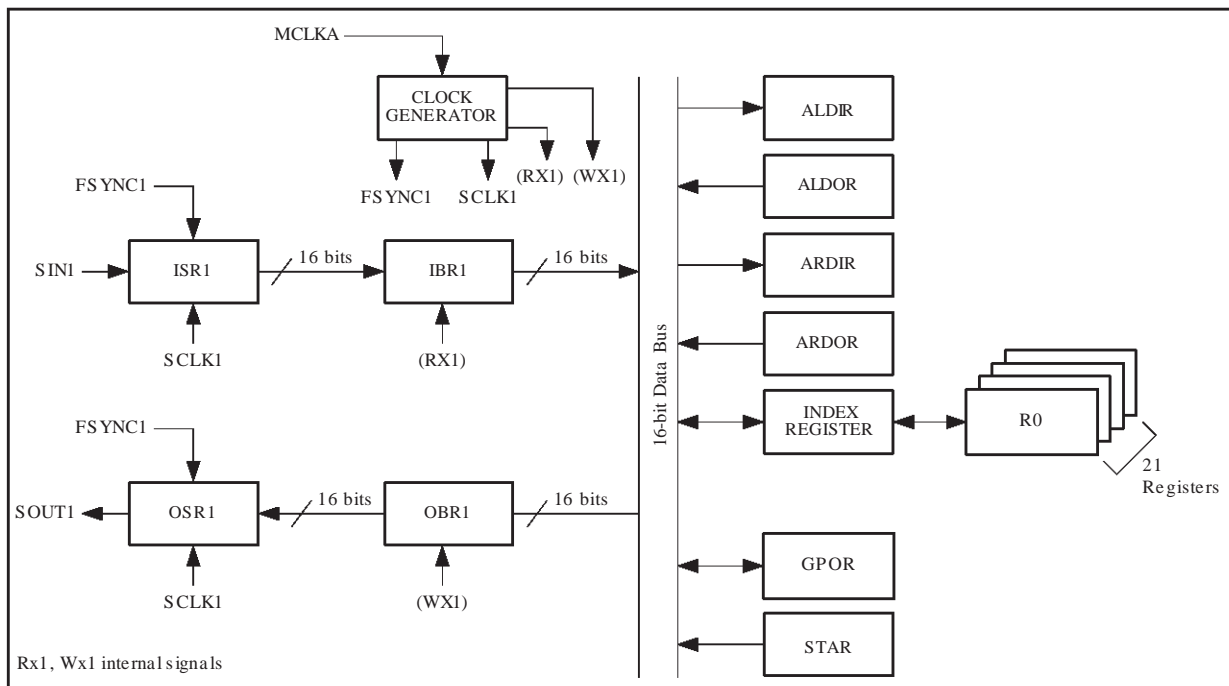
#### Audio Serial Interface One (Figures 21 and 22)

- Audio Left Data Input Register (ALDIR)
- Audio Left Data Output Register (ALDOR)
- Audio Right Data Input Register (ARDIR)
- Audio Right Data Output Register (ARDOR)
- Index Register (R0 to R20)
- General Purpose Output Register (GPOR)
- Status Audio Register

#### Modem and Telephone Serial Interface Two (Figures 23 and 24)

- Modem Data Input Register (MDIR)
- Modem Data Output Register (MDOR)
- Telephone Data Input Register (TDIR)
- Telephone Data Output Register (TDOR)

**Figure 21 : Audio Serial Interface Block Diagram**



SERIAL INTERFACE (continued)

Figure 22 : Audio Codec Serial Interface (SIO 1) Timing

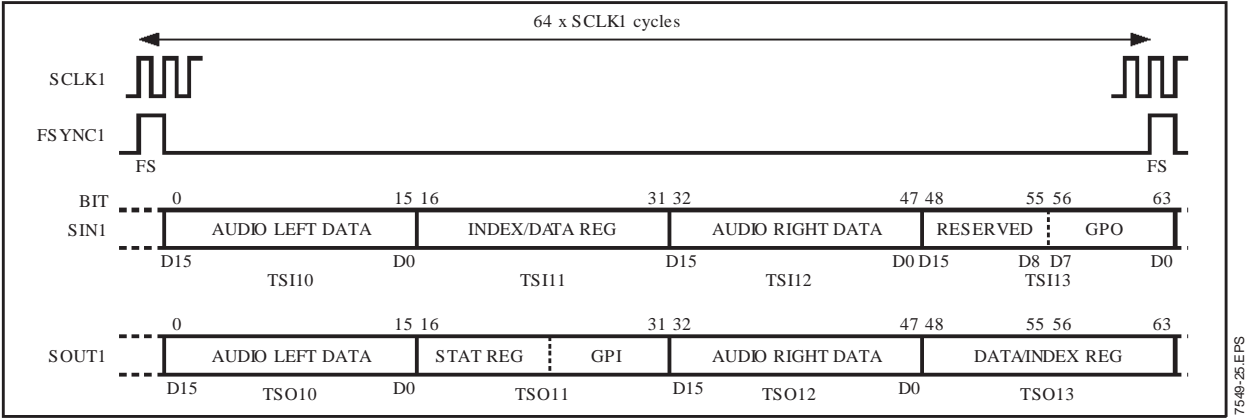


Figure 23 : Modem and Telephone Serial Interface Block Diagram

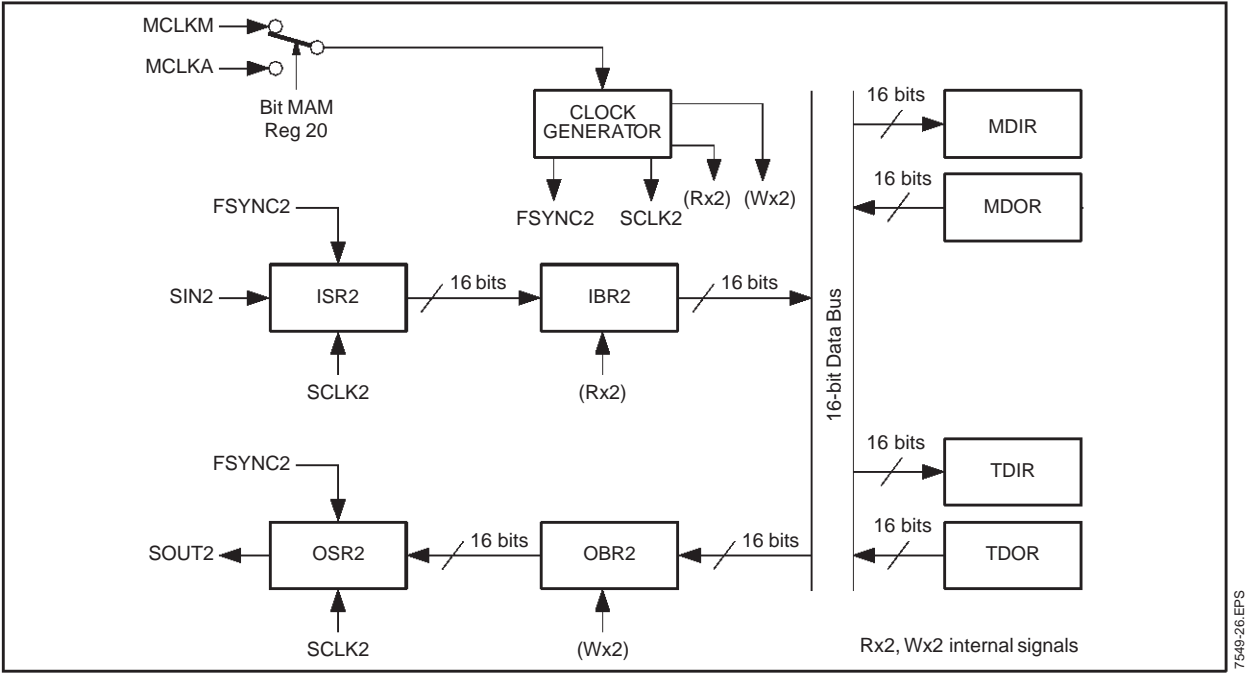
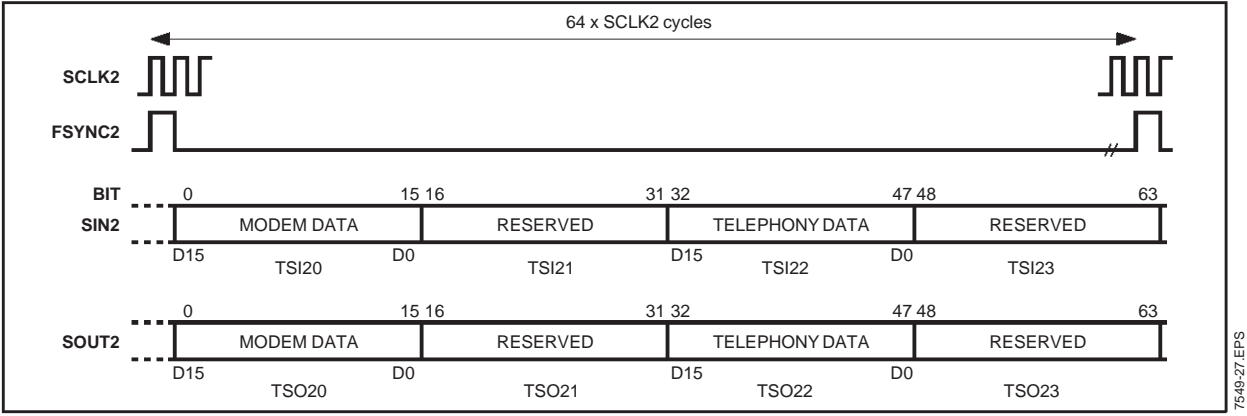


Figure 24 : Modem and Telephony Codecs Serial Interface (SIO 2) Timing



## SERIAL INTERFACE BIT DEFINITION

Bit N°	Word N°	Definition
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## Serial Input #1 Bit Definition (DSP-to-Codec)

0 - 15	0	Audio Left DAC Data (16 bits, MSB first)
16	1	If = "1", write index register set. If = "0", no write , just read index register set
17		Reserved (see Note 1)
18		Reserved (see Note 1)
19 - 23		Index Register Address (5 bits, MSB first)
24 - 31		Index Register Data (8 bits, MSB first)
32 - 47	2	Audio Right DAC Data (16 bits, MSB first)
48 - 55	3	Reserved
56 - 63		GPIO Output Data [7:0] (8 bits, MSB first). These bits are sent to the pins GPIO [7:0] if their respective configuration bits (CFGGPIO [7] thru CFGGPIO[0]) are set to the OUTPUT state. Refer to the GPIO configuration register (index Reg # 18) for more information.

## Serial Output #1 Bit Definition (Codec-to-DSP)

0 - 15	0	Audio Left ADC Data (16 bits, MSB first)
16	1	+ Mono Codec #2 ADC Overflow (indicates saturation of ADC)
17		+ Mono Codec #1 ADC Overflow (indicates saturation of ADC)
18		+ ADC Left Overflow (indicates saturation of ADC)
19		+ ADC Right Overflow (indicates saturation of ADC)
20		Reserved
21		The Inverted Digital Status Value of Input Pin <u>VOL_MUTE</u>
22		The Inverted Digital Status Value of Input Pin <u>VOL_UP</u>
23		The Inverted Digital Status Value of Input Pin <u>VOL_DOWN</u>
24 - 31		Codec Pins GPIO[7:0] Status (8 bits, MSB first). These bits are a reflection of the digital values on pins GPIO[7:0]. If the GPIO Pin is configured as an OUTPUT, the current setting of that OUTPUT bit will be returned in these status bits. If the GPIO is set to be an INPUT, the digital value on the GPIO input Pin will be read. Refer to the GPIO configuration register (index register 18) for more information.
32 - 47	2	Audio Right ADC Data (16 bits, MSB first)
48 - 50	3	Reserved
51 - 55		Index Register Address Returned (5 bits, MSB first)
56 - 63		Index Register Data Return (8 bits, MSB first)

## Serial Input #2 Bit Definition (DSP-to-Codec)

0 - 15	0	Mono Codec #1 DAC Data (16 bits, MSB first)
16 - 31	1	Reserved (see Note 1)
32 - 47	2	Mono Codec #2 DAC Data (16 bits, MSB first)
48 - 63	3	Reserved (see Note 1)

## Serial Output #2 Bit Definition (Codec-to-DSP)

0 - 15	0	Mono Codec #1 ADC Data (16 bits, MSB first)
16 - 31	1	Reserved
32 - 47	2	Mono Codec #2 ADC Data (16 bits, MSB first)
48 - 63	3	Reserved

**Note 1 :** Reserved for future expansion. Should be set to zero.



## REGISTER DESCRIPTION

### Index Register Map

Register	Name
0	Identification Register
1	Line Input #1 Gain Control
2	Line Input #2 Gain Control
3	Line Input #3 Gain Control
4	Right Microphone Input Gain Control
5	Left Microphone Input Gain Control
6	Right Channel Audio ADC Summer Control
7	Left Channel Audio ADC Summer Control
8	Right Channel Audio DAC Summer Control
9	Left Channel Audio DAC Summer Control
10	Audio Line Output Attenuators
11	Audio Output Mute Control
12	Mono Input/Output Control
13	Mono Codec #1 Input Control
14	Mono Codec #1 Output Control
15	Mono Codec #2 Input Control
16	Mono Codec #2 Output Control
17	ADC/DAC Loopback Control Register
18	GPIO Configuration Register
19	Digital Control Register #1
20	Digital Control Register #2
21	Saturation Clear Register

Note : Bit not used in registers must be set to 0 for future compatibility.

### Index Register 0 : Identification Register

7	6	5	4	3	2	1	0
Chip Identification				Revision Code			

Read only register. 0001-0001 : version 1.0

0001-0010 : version 1.1

0001-0011 : version 1.2

0001-0011 : version 1.3

### Index Register 1 : Line Input #1 Gain Control Register

7	6	5	4	3	2	1	0
LI1G3	LI1G2	LI1G1	LI1G0	RI1G3	RI1G2	RI1G1	RI1G0

RI1G3 to RI1G0 : Reset value : 0000.

Right line input #1 gain select.

LI1G3 to LI1G0 : Reset value : 0000.

Left line input #1 gain select.

Refer to the line input gain select table (Table 2) to understand how these bits relate to the actual gain/attenuation value.

### Index Register 2 : Line Input #2 Gain Control Register

7	6	5	4	3	2	1	0
LI2G3	LI2G2	LI2G1	LI2G0	RI2G3	RI2G2	RI2G1	RI2G0

RI2G3 to RI2G0 : Reset value : 0000.

Right line input #2 gain select.

LI2G3 to LI2G0 : Reset value : 0000.

Left line input #2 gain select.

Refer to the line input gain select table (Table 2) to understand how these bits relate to the actual gain/attenuation value.

**REGISTER DESCRIPTION** (continued)**Index Register 3 : Line Input #3 Gain Control Register**

7	6	5	4	3	2	1	0
LI3G3	LI3G2	LI3G1	LI3G0	RI3G3	RI3G2	RI3G1	RI3G0

RI3G3 to RI3G0 : Reset value : 0000.

Right line input #3 gain select.

LI3G3 to LI3G0 : Reset value : 0000.

Left line input #3 gain select.

Refer to the line input gain select table (Table 2) to understand how these bits relate to the actual gain/attenuation value.

**Table 2 : Line Input Gain Select (G1)**

Bit Code	Decimal	Gain (dB)	Bit Code	Decimal	Gain (dB)
0000	0	-24	1000	8	-8
0001	1	-22	1001	9	-6
0010	2	-20	1010	10	-4
0011	3	-18	1011	11	-2
0100	4	-16	1100	12	+0
0101	5	-14	1101	13	+2
0110	6	-12	1110	14	+4
0111	7	-10	1111	15	+6

**Index Register 4 : Right Microphone Input Gain Control Register**

7	6	5	4	3	2	1	0
0	0	0	+MPRER	RIMG3	RIMG2	RIMG1	RIMG0

RIMG3 : Reset value : 0000.

to RIMG0 : Right input microphone gain select.  
Refer to the microphone gain table (Table 3) to understand how these bits relate to the actual gain/attenuation value.

+MPRER : Reset value : 0

Setting this bit to a 1 will place a +20dB pre-amplifier in the microphone right signal path. If this bit is a 0, the pre-amplifier gain will be +0dB (see Table 4).

**Index Register 5 : Left Microphone Input Gain Control Register**

7	6	5	4	3	2	1	0
0	0	0	+MPREL	LIMG3	LIMG2	LIMG1	LIMG0

LIMG3 : Reset value : 0000.

to LIMG0 : Left input microphone gain select.  
Refer to the microphone gain table (Table 3) to understand how these bits relate to the actual gain/attenuation value.

+MPREL : Reset value : 0.

Setting this bit to a 1 will place a +20dB pre-amplifier in the microphone left signal path. If this bit is a 0, the gain will be +0dB.

**Table 3 : Microphone Gain Select (G2)**

Bit Code	Decimal	Gain (dB)	Gain with +20dB preamp. (dB)	Bit Code	Decimal	Gain (dB)	Gain with +20dB preamp. (dB)
0000	0	+0.0	+20.0	1000	8	+20.0	+40.0
0001	1	+2.5	+22.5	1001	9	+22.5	+42.5
0010	2	+5.0	+25.0	1010	10	+25.0	+45.0
0011	3	+7.5	+27.5	1011	11	+27.5	+47.5
0100	4	+10.0	+30.0	1100	12	+30.0	+50.0
0101	5	+12.5	+32.5	1101	13	+32.5	+52.5
0110	6	+15.0	+35.0	1110	14	+35.0	+55.0
0111	7	+17.5	+37.5	1111	15	+37.5	+57.5

**Table 4 : Pre-amplifier (P)**

MPRER / MPREL	Pre-amplifier Gain (P)
0	+0dB
1	+20dB

**REGISTER DESCRIPTION** (continued)**Index Register 6 : Right ADC Summer Control**

7	6	5	4	3	2	1	0
0	0	-MUTRAD	-MUTRAP	-MUTRAM	-MUTRA3	-MUTRA2	-MUTRA1

- MUTRA1 : Reset value : 0. Mute right line input #1 of the ADC right summer. Setting this bit to 0 mutes the signal.
- MUTRA2 : Reset value : 0. Mute right line input #2 of the ADC right summer. Setting this bit to 0 mutes the signal.
- MUTRA3 : Reset value : 0. Mute right line input #3 of the ADC right summer. Setting this bit to 0 mutes the signal.
- MUTRAM : Reset value : 0. Mute right microphone input of the ADC right summer. Setting this bit to 0 mutes the signal.
- MUTRAP : Reset value : 0. Mute mono input of the ADC right summer. Setting this bit to 0 mutes the signal.
- MUTRAD : Reset value : 0. Mute right DAC signal into the ADC right summer. This signal is the right DAC output of the codec wrapped back around into the ADC input summer. Setting this bit to 0 mutes the signal.

**Index Register 7 : Left ADC Summer Control**

7	6	5	4	3	2	1	0
0	0	-MUTLAD	-MUTLAP	-MUTLAM	-MUTLA3	-MUTLA2	-MUTLA1

- MUTLA1 : Reset value : 0. Mute left line input #1 of the ADC left summer. Setting this bit to 0 mutes the signal.
- MUTLA2 : Reset value : 0. Mute left line input #2 of the ADC left summer. Setting this bit to 0 mutes the signal.

- MUTLA3 : Reset value : 0. Mute left line input #3 of the ADC left summer. Setting this bit to 0 mutes the signal.
- MUTLAM : Reset value : 0. Mute left microphone input of the ADC left summer. Setting this bit to 0 mutes the signal.
- MUTLAP : Reset value : 0. Mute mono input of the ADC left summer. Setting this bit to 0 mutes the signal.
- MUTLAD : Reset value : 0. Mute left DAC signal into the ADC left summer. This signal is the left DAC output of the codec wrapped back around into the ADC input summer. Setting this bit to 0 mutes the signal.

**Index Register 8 : Right DAC Summer Control**

7	6	5	4	3	2	1	0
0	0	-MUTRDD	-MUTRDP	-MUTRDM	-MUTRD3	-MUTRD2	-MUTRD1

- MUTRD1 : Reset value : 0. Mute right line input #1 of the DAC right summer. Setting this bit to 0 mutes the signal.
- MUTRD2 : Reset value : 0. Mute right line input #2 of the DAC right summer. Setting this bit to 0 mutes the signal.
- MUTRD3 : Reset value : 0. Mute right line input #3 of the DAC right summer. Setting this bit to 0 mutes the signal.
- MUTRDM : Reset value : 0. Mute right microphone input of the DAC right summer. Setting this bit to 0 mutes the signal.
- MUTRDP : Reset value : 0. Mute mono input of the DAC right summer. Setting this bit to 0 mutes the signal.
- MUTRDD : Reset value : 0. Mute right DAC signal going into the DAC right summer. This input is the main right DAC output of the sigma-delta codec. Setting this bit to 0 mutes the signal.

**REGISTER DESCRIPTION** (continued)**Index Register 9 : Left DAC Summer Control**

7	6	5	4	3	2	1	0
0	0	-MUTLDD	-MUTLDP	-MUTLDM	-MUTLD3	-MUTLD2	-MUTLD1

- MUTLD1 : Reset value : 0. Mute left line input #1 of the DAC left summer. Setting this bit to 0 mutes the signal.
- MUTLD2 : Reset value : 0. Mute left line input #2 of the DAC left summer. Setting this bit to 0 mutes the signal.
- MUTLD3 : Reset value : 0. Mute left line input #3 of the DAC left summer. Setting this bit to 0 mutes the signal.
- MUTLDM : Reset value : 0. Mute left microphone input of the DAC left summer. Setting this bit to 0 mutes the signal.
- MUTLDP : Reset value : 0. Mute mono input of the DAC left summer. Setting this bit to 0 mutes the signal.
- MUTLDD : Reset value : 0. Mute left DAC signal into the DAC left summer. This input is the main left DAC output of the sigma-delta codec. Setting this bit to 0 mutes the signal.

**Index Register 10 : Audio Line Output Attenuator Control**

7	6	5	4	3	2	1	0
ROUTG3	ROUTG2	ROUTG1	ROUTG0	ROUTG3	ROUTG2	ROUTG1	ROUTG0

- ROUTG3 to ROUTG0 : Reset value : 0000. Right line output gain select.
- ROUTG3 to ROUTG0 : Reset value : 0000. Left line output gain select.

Refer to the line output attenuator select Table (Table 5) to understand how these bits relate to the actual attenuation value.

**Table 5 : Line Output Attenuator Select (A1)**

Bit Code	Decimal	Gain (dB)	Bit Code	Decimal	Gain (dB)
0000	0	0	1000	8	-24
0001	1	-3	1001	9	-27
0010	2	-6	1010	10	-30
0011	3	-9	1011	11	-33
0100	4	-12	1100	12	-36
0101	5	-15	1101	13	-39
0110	6	-18	1110	14	-42
0111	7	-21	1111	15	-45

**Index Register 11 : Audio Output Mute Control**

7	6	5	4	3	2	1	0
0	0	0	-MUTEX	-MUTL2	-MUTL1	-MUTR2	-MUTR1

- MUTR1 : Reset value : 0. Mute right line output #1. Setting this bit to 0 mutes the output Pin ROUT1.
- MUTR2 : Reset value : 0. Mute right line output #2. Setting this bit to 0 mutes the output Pin ROUT2.
- MUTL1 : Reset value : 0. Mute left line output #1. Setting this bit to 0 mutes the output Pin LOUT1.
- MUTL2 : Reset value : 0. Mute left line output #2. Setting this bit to 0 mutes the output Pin LOUT2.
- MUTEX : Reset value : 1 (unmute). This bit is read only and is the output of toggle/untoggled debounced external mute.

**Index Register 12 : Mono Input/Output Control**

7	6	5	4	3	2	1	0
0	0	0	-MUTPCP	-MUTPCR	-MUTPCL	PGAIN1	PGAIN0

- PGAIN1, PGAIN0 : Reset value : 01. Mono input attenuator select (A4). These two bits select the input attenuation value as shown in the Table 6.
- MUTPCL : Reset value : 0. Mute DAC left into mono output summer. Setting this bit to 0 mutes the signal.
- MUTPCR : Reset value : 0. Mute DAC right into mono output summer. Setting this bit to 0 mutes the signal.
- MUTPCP : Reset value : 1. Mute mono input into mono output summer. This bit will power on to a one (1) so that the PC system "beeper" sounds coming into the mono input pin will be looped to the mono output pin so that PC system sounds can be heard during power-up. Setting this bit to 0 mutes this loop path.

**Table 6 : Mono Input Attenuator**

PGAIN1	PGAIN0	Gain (A4)
0	0	0dB
0	1	-10dB
1	0	-20dB
1	1	-30dB

**REGISTER DESCRIPTION** (continued)

**Index Register 13** : Modem Codec #1  
Input Control

7	6	5	4	3	2	1	0
0	0	0	0	0	0	0	M1G0

M1G0 : Reset value : 0.

Gain select for mono codec #1 input amplifier (G4) (see Table 7).

**Table 7** : Codec #1 Amplifier Gain

M1G0	Amplifier Gain (G4)
0	+0dB
1	+6dB

**Note** : If single-ended hardware, M1G0 must be set to 1 to +0dB gain.

**Index Register 14** : Mono Codec #1  
Output Control

7	6	5	4	3	2	1	0
0	0	0	0	0	M1TXG	-M1TX2	-M1TX1

-M1TX1 : Reset value : 0. Setting this bit to 1 will enable the mono codec #1 DAC output to the TP1TX1+/TP1TX1- output Pins. If set to 0, the pins are muted.

-M1TX2 : Reset value : 0. Setting this bit to 1 will enable the mono codec #1 DAC output to the TP1TX2 output Pin. If set to 0, the pin is muted.

M1TXG : Reset value : 0. Setting this bit to 1 will place a -6dB attenuator in the output DAC path. Setting this bit to 0 will result in a gain path of 0dB (A2).

**Table 8** : Transmit Gain (A2)

M1TXG	Transmit Amplifier Gain (A2)
0	+0dB
1	-6dB

**Index Register 15** : Mono Codec #2  
Input Control

7	6	5	4	3	2	1	0
TP2RXG1	TP2RXG0	-M2MICL	-M2MICR	-M2TP2	-M2HRX	HRXG1	HRXG0

TP2RXG1, TP2RXG0 : Reset value : 00.  
TP2Rx gain select. These two bits determine the gain of the input amplifier of the TP2RX input Pin of mono codec #2. The Table 9 illustrates the bit definitions of the four possible gain settings. This amplifier is only applied to the TP2RX input of mono codec #2.

HRXG1, HRXG0 : Reset value : 00.  
HandRx gain select. These two bits determine the gain of the input amplifier of the HandRx input Pin of mono codec #2. The Table 9 illustrates the bit definitions of the four possible gain settings. This amplifier is only applied to the HandRx input of mono codec #2.

-M2HRX : Reset value : 0.  
Setting this bit to 1 will enable the HandRx input into the ADC summer of the mono codec #2. Setting the bit to 0 will mute the signal.

-M2TP2 : Reset value : 0.  
Setting this bit to 1 will enable the TP2Rx input into the ADC summer of the mono codec #2. Setting the bit to 0 will mute the signal.

-M2MICR : Reset value : 0.  
Setting this bit to 1 will enable the internally amplified microphone right signal into the ADC summer of the mono codec #2. Setting the bit to 0 will mute the signal.

-M2MICL : Reset value : 0.  
Setting this bit to 1 will enable the internally amplified microphone left signal into the ADC summer of the mono codec #2. Setting the bit to 0 will mute the signal.

**Table 9** : TP2Rx and HANDRx Input Gain (G3)

Bit Code		Gain (G3)
0	0	-4dB
0	1	0dB
1	0	+4dB
1	1	+8dB

**REGISTER DESCRIPTION** (continued)**Index Register 16** : Modem Codec #2  
Output Control

7	6	5	4	3	2	1	0
0	0	0	0	M2TXG	-M2TX3	-M2TX2	-M2TX1

-M2TX1 : Reset value : 0. Setting this bit to 1 will enable the mono codec #2 DAC output to the HandTx1 output Pin. Setting the bit to 0 will mute the signal.

-M2TX2 : Reset value : 0. Setting this bit to 1 will enable the mono codec #2 DAC output to the HandTx2 output Pin. Setting the bit to 0 will mute the signal.

-M2TX3 : Reset value : 0. Setting this bit to 1 will enable the mono codec #2 DAC output to the HandTx3 output Pin. Setting the bit to 0 will mute the signal.

M2TXG : Reset value : 0. Setting this bit to 1 will place a -6dB attenuator in the output DAC path. Setting this bit to 0 will result in a gain path of 0dB.

**Table 10** : Transmit Gain (A2)

M2TXG	Transmit Amplifier Gain (A2)
0	+0dB
1	-6dB

**Index Register 17** : Loopback Control Register

7	6	5	4	3	2	1	0
0	0	0	0	+LOOPM2	+LOOPM1	+LOOPL	+LOOPR

+LOOPR : Reset value : 0. Setting this bit to 1 will loop the ADC digital data to the DAC converter on the audio codec right converter. Setting this bit to 0 will cause the DAC converter to receive its digital data from the serial port which is normal operation.

+LOOPL : Reset value : 0. Setting this bit to 1 will loop the ADC digital data to the DAC converter on the audio codec left converter. Setting this bit to 0 will cause the DAC converter to receive its digital data from the serial port which is normal operation.

+LOOPM1 : Reset value : 0. Setting this bit to 1 will loop the ADC digital data to the DAC converter on the mono codec #1 converter. Setting this bit to 0 will cause the DAC converter to receive its digital data from the serial port which is normal operation.

+LOOPM2 : Reset value : 0. Setting this bit to 1 will loop the ADC digital data to the DAC converter on the mono codec #2 converter. Setting this bit to 0 will cause the DAC converter to receive its digital data from the serial port which is normal operation.

**Index Register 18** : GPIO Configuration Register

7	6	5	4	3	2	1	0
CFGPIO7	CFGPIO6	CFGPIO5	CFGPIO4	CFGPIO3	CFGPIO2	CFGPIO1	CFGPIO0

CFGPIO0 : Reset value : 0. If this bit is set to 0, then the Pin GPIO0 is an input. If this bit is set to 1, then the Pin is an output.

CFGPIO1 : Reset value : 0. If this bit is set to 0, then the Pin GPIO1 is an input. If this bit is set to 1, then the Pin is an output.

CFGPIO2 : Reset value : 0. If this bit is set to 0, then the Pin GPIO2 is an input. If this bit is set to 1, then the Pin is an output.

CFGPIO3 : Reset value : 0. If this bit is set to 0, then the Pin GPIO3 is an input. If this bit is set to 1, then the Pin is an output.

CFGPIO4 : Reset value : 0. If this bit is set to 0, then the Pin GPIO4 is an input. If this bit is set to 1, then the Pin is an output.

CFGPIO5 : Reset value : 0. If this bit is set to 0, then the Pin GPIO5 is an input. If this bit is set to 1, then the Pin is an output.

CFGPIO6 : Reset value : 0. If this bit is set to 0, then the Pin GPIO6 is an input. If this bit is set to 1, then the Pin is an output.

CFGPIO7 : Reset value : 0. If this bit is set to 0, then the Pin GPIO7 is an input. If this bit is set to 1, then the pin is an output.



**REGISTER DESCRIPTION** (continued)**Index Register 19 : Digital Control Register #1**

7	6	5	4	3	2	1	0
DCVOL3	DCVOL2	DCVOL1	DCVOL0	+DCVOL	+PBVOL	CALIBM	CALIBA

DCVOL3-0 : Read only DC volume control value (see +DCVOL bit for explanation).

+DCVOL : Reset value : 0. + Enable DC volume control circuitry. If the DCVOL bit is enabled, the codec will use a DC voltage range from a potentiometer attached to the external chip pin DC\_POT to determine a 4-bit value to be used as the main left/right output attenuator register value. Both left and right attenuation levels have the same value. Regardless of the setting of this DCVOL bit, this 4-bit value can be read by DSP software since the 4-bit value is reflected in the Register 19 (bits [7:4]). The DSP software can use this 4-bit value for unique functions. Setting this bit to 1 will enable the circuitry on the codec which permits the external potentiometer to control the main left/right output attenuator register 4-bit value in Register 10.

+PBVOL : Reset value : 0. + Enable pushbutton volume circuitry. Setting this bit to 1 will enable the pushbutton digital volume control input pins to influence the main left/right output attenuator register values in Register 10. The output attenuators have 16 possible states and the VOL\_UP and VOL\_DOWN input pins will make the attenuator register value increment or decrement from 0 to 15. The VOL\_MUTE input pin will affect the main left/right output mute blocks in Register 11. When pushbuttons are used, left and Right attenuation levels can have different values. Regardless of the value of this control bit, the three digital input values of the UP, DOWN and MUTE pins are reflected in the STATUS word that is sent to the DSP on the codec SOUT1 line.

CALIBA, CALIBM : Reset value : 0. Setting the bit CALIBA to 1 will force a calibration of all audio converters and the bit CALIBM to 1 will force a calibration of all modem converters. The length of time required for calibration is 23ms min. at 44.1kHz (section 4 of the functional description) and after CALIBx is automatically reset to 0 when calibration is completed.

**Note :** If +DCVOL and +PBVOL are set to 1, the priority is given to +DCVOL.

**REGISTER DESCRIPTION** (continued)**Index Register 20 : Digital Control Register #2**

7	6	5	4	3	2	1	0
0	0	MOR	MAM	NDIV1	NDIV0	SWPDN1	SWPDN0

MOR : Reset value : 0. When this bit is set to 1 the mono codecs #1 and #2 have the same ratio than audio part, (N value, register 20). When set to 0, mono codecs have an oversampling ratio fixed to 128.

MAM : Reset value : 0. Mono codecs #1 and #2 clocks derived from MCLKM when set to 0. If this bit is set to 1, all codecs clocks are issued from MCLKA (quadraphonic mode) (see Figure 6).

	MOR	MAM	SCLK1	FSA	SCLK2	FSM
Modem mode	0	0	$MCLKA/(4 \cdot N)$	$MCLKA/(256 \cdot N)$	$MCLKM/4$	$MCLKM/256$
Common master clock	0	1	$MCLKA/(4 \cdot N)$	$MCLKA/(256 \cdot N)$	$MCLKA/4$	$MCLKA/256$
Common ratio	1	0	$MCLKA/(4 \cdot N)$	$MCLKA/(256 \cdot N)$	$MCLKM/(4 \cdot N)$	$MCLKM/(256 \cdot N)$
Quadraphony	1	1	$MCLKA/(4 \cdot N)$	$MCLKA/(256 \cdot N)$	$MCLKA/(4 \cdot N)$	$MCLKA/(256 \cdot N)$

See also Figure 6 for clock generator diagram.

NDIV1, NDIV0 : Reset value : 00. Ratio FSA versus MCLKA : see Table 11.

Typical sampling frequency obtained with  $MCLK = 11.2896\text{MHz}$  : see Table 12.

SWPDN1, SWPDN0 : Reset value : 00. Software power-down mode options (see Table 13).

**Table 11 : Sampling Frequency Setting**

Bit 3	Bit 2	N	FSA
0	0	1	24kHz $\rightarrow$ 48kHz
0	1	2	12kHz $\rightarrow$ 24kHz
1	0	4	6kHz $\rightarrow$ 12kHz
1	1	8	3kHz $\rightarrow$ 6kHz

**Table 12 : Typical FSA with MCLKA=11.2896MHz**

Bit 3	Bit 2	N	FSA
0	0	1	44.1kHz
0	1	2	22.05kHz
1	0	4	11.025kHz
1	1	8	5.5125kHz

**Table 13**

Bit 1	Bit 0	Power-down Mode
0	0	SWPDN Mode 0 : all codec circuitry is active.
0	1	SWPDN Mode 1 : Audio left/right ADC/DAC are powered down. Mono codecs #1 and #2 are active. All input and output analog mixers, GPIO and pushbuttons are active. Serial port #1 active. Serial port #2 active.
1	0	SWPDN Mode 2 : Audio left/right ADC/DAC are active. Mono codecs #1 and #2 are powered down. All input and output analog mixers, GPIO and pushbuttons are active. Serial port #1 active. Serial port #2 in-active.

**Index Register 21 : Saturation Clear Register**

7	6	5	4	3	2	1	0
0	0	0	0	ADC MODEM #2	ADC MODEM #1	ADC LEFT	ADC RIGHT

ADC LEFT, ADC RIGHT, ADC MODEM #1, ADC MODEM #2 : Reset value : 0000.

When bit is set to 1, the STLC7549 will clear the respective ADC overflow bit in the serial output datastream periods. This is a write only register.

