

# 2.4GHz low voltage RF transceiver

# SA2420

## DESCRIPTION

The SA2420 transceiver is a combined low-noise amplifier, receive mixer, transmit mixer and LO buffer IC designed for high-performance low-power communication systems for 2.4-2.5GHz. The LNA has a 2.5dB noise figure at 2.4GHz with 13dB gain and an IP3 intercept of -6dBm at the input. The gain is stabilized by on-chip compensation to vary less than  $\pm 0.2$ dB over the -40 to +85°C temperature range. The wide-dynamic-range receive mixer has a 10dB noise figure and an input IP3 of -3dBm at 2.4GHz. The nominal current drawn from a single 3V supply is 34mA in transmit mode and 21mA in receive mode.

## FEATURES

- Low current consumption: 34mA nominal transmit mode and 21mA nominal receive mode
- Fabricated on a high volume, rugged BiCMOS technology
- High system power gain: 24dB (LNA + Mixer) at 2.45GHz
- TSSOP24 package
- Excellent gain stability versus temperature and supply voltage
- -10dBm LO can be used to drive the mixer
- Operates with either full or half frequency LO
- Wide IF range: 50–500MHz
- ESD protected

## ORDERING INFORMATION

DESCRIPTION	TEMPERATURE RANGE	ORDER CODE	DWG #
24-Pin Plastic Thin Shrink Small Outline Package (Surface-mount, TSSOP)	-40 to +85°C	SA2420DH	SOT355-1

## BLOCK DIAGRAM

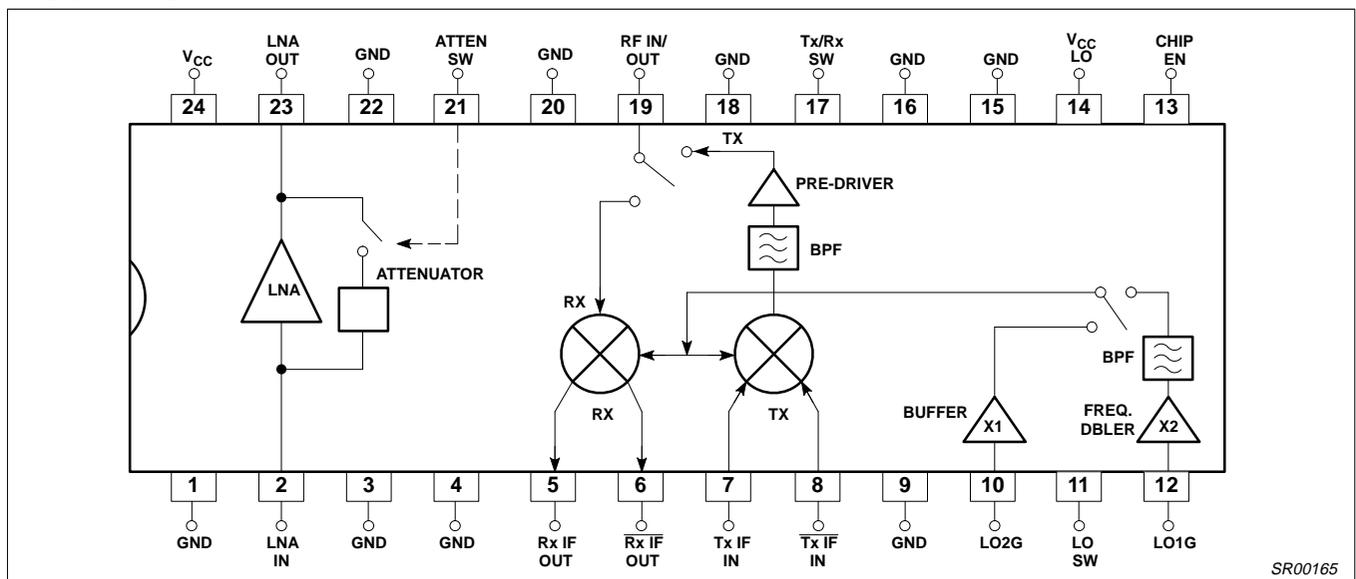


Figure 2. SA2420 Block Diagram

## PIN CONFIGURATION

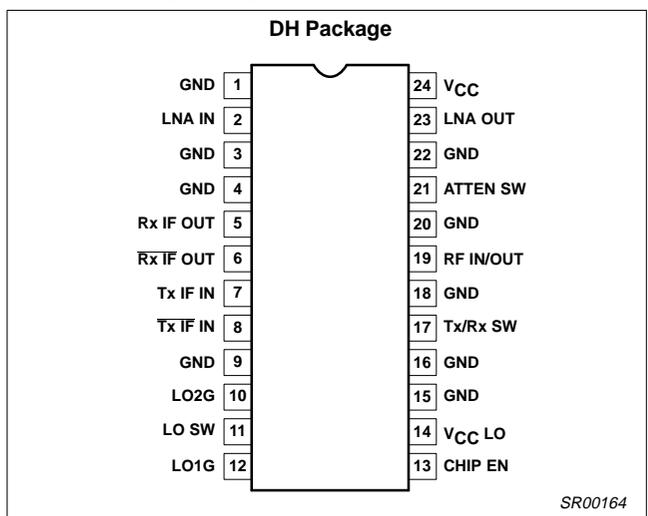


Figure 1. Pin Configuration

## APPLICATIONS

- 2.45GHz WLAN front-end (802.11, ISM)

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## ABSOLUTE MAXIMUM RATINGS

SYMBOL	PARAMETER	RATING	UNITS
$V_{CC}$	Supply voltage <sup>1</sup>	-0.3 to +6	V
$V_{IN}$	Voltage applied to any pin	-0.3 to ( $V_{CC} + 0.3$ )	V
$P_D$	Power dissipation, $T_A = 25^\circ\text{C}$ (still air) <sup>2</sup> 24-Pin Plastic TSSOP	555	mW
$T_{JMAX}$	Maximum operating junction temperature	150	$^\circ\text{C}$
$P_{MAX}$	Maximum power (RF/IF/LO pins)	+20	dBm
$T_{STG}$	Storage temperature range	-65 to +150	$^\circ\text{C}$

## NOTE:

- Transients exceeding 8V on  $V_{CC}$  pin may damage the product.
- Maximum dissipation is determined by the operating ambient temperature and the thermal resistance,  
 $\theta_{JA}$ : 24-Pin TSSOP = 117 $^\circ\text{C}/\text{W}$

## RECOMMENDED OPERATING CONDITIONS

SYMBOL	PARAMETER	RATING	UNITS
$V_{CC}$	Supply voltage	2.7 to 5.5	V
$T_A$	Operating ambient temperature range	-40 to +85	$^\circ\text{C}$
$T_J$	Operating junction temperature	-40 to +105	$^\circ\text{C}$

## DC ELECTRICAL CHARACTERISTICS

 $V_{CC} = +3\text{V}$ ,  $T_A = 25^\circ\text{C}$ ; unless otherwise stated.

SYMBOL	PARAMETER	TEST CONDITIONS	LIMITS					UNITS
			MIN	-3 $\sigma$	TYP	+3 $\sigma$	MAX	
$I_{CCTX}$	Supply current, Transmit	LO mode = Hi	25		34		42	mA
$I_{CCRX}$	Supply current, Receive	LO mode = Hi	15		21		26	mA
$I_{CC\ OFF}$	Power down mode (Tx/Rx SW = Low)	LO mode = Hi, LNA gain = Hi			0		10	$\mu\text{A}$
$V_{LNA-IN}$	LNA input voltage	Receive mode			0.855			V
$I_{LNA-OUT}$	LNA output bias current	Receive mode			4.0			mA
$V_{LO\ 2.1}$	LO buffer DC input voltage	LO mode = Hi			2.1			V
$V_{LO\ 1.05}$	LO buffer DC input voltage	LO mode = Low			2.1			V
$V_{TX\ IF}$	Tx Mixer input voltage	Transmit mode			1.7			V
$V_{TX\ IFB}$	Tx Mixer input voltage	Transmit mode			1.7			V

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## AC ELECTRICAL CHARACTERISTICS

 $V_{CC} = +3V$ ,  $T_A = 25^{\circ}C$ ;  $LO_{IN} = -10dBm$  @ 2.2GHz;  $f_{RF} = 2.4GHz$ ; unless otherwise stated.

SYMBOL	PARAMETER	TEST CONDITIONS	LIMITS					UNITS
			MIN	-3 $\sigma$	TYP	+3 $\sigma$	MAX	
<b>Low Noise Amplifier (In = Pin 2; Out = 23)</b>								
$S_{21}$	Amplifier gain	LNA gain = Hi			13.0			dB
$\Delta S_{21}/\Delta T$	Gain temperature sensitivity	LNA gain = Hi			-0.002			dB/ $^{\circ}C$
$\Delta S_{21}/\Delta V_{CC}$	Gain $V_{CC}$ drift	LNA gain = Hi			0.3			dB/V
$S_{12}$	Amplifier reverse isolation	LNA gain = Hi			-27			dB
$S_{11}$	Amplifier input match <sup>1</sup>	LNA gain = Hi			-10			dB
$S_{22}$	Amplifier output match <sup>1</sup>	LNA gain = Hi			-10			dB
ISO	Isolation: $LO_1$ to $LNA_{IN}$	LO mode = Hi, LNA gain = Hi			-36			dB
$P_{-1dB}$	Amplifier input 1dB gain compression	LNA gain = Hi			-15			dBm
IP3	Amplifier input third order intercept	$f_1 - f_2 = 1MHz$ , LNA gain = Hi			-6			dBm
NF	Amplifier noise figure (50 $\Omega$ )	LNA gain = Hi			2.5			dB
<b>LNA High Overload Mode</b>								
$S_{21}$	Amplifier gain	LNA gain = Low			-11			dB
$\Delta S_{21}/\Delta T$	Gain temperature sensitivity	LNA gain = Low			-0.01			dB/ $^{\circ}C$
$\Delta S_{21}/\Delta V_{CC}$	Gain $V_{CC}$ drift	LNA gain = Low			0.3			dB/V
$S_{12}$	Amplifier reverse isolation	LNA gain = Low			-24			dB
$S_{11}$	Amplifier input match <sup>1</sup>	LNA gain = Low			-10			dB
$S_{22}$	Amplifier output match <sup>1</sup>	LNA gain = Low			-10			dB
ISO	Isolation: $LO_1$ to $LNA_{IN}$	LO mode = Hi, LNA gain = Low			-38			dB
$P_{-1dB}$	Amplifier input 1dB gain compression	LNA gain = Low			5			dBm
IP3	Amplifier input third order intercept	$f_1 - f_2 = 1MHz$ , LNA gain = Low			13			dBm
NF	Amplifier noise figure (50 $\Omega$ )	LNA gain = Low			17			dB
<b>Rx Mixer (RF = Pin 19, IF = Pins 5 and 6, LO = Pin 10 or 12, <math>P_{LO} = -10dBm</math>)</b>								
$PG_{C1}$	Power conversion gain: $R_L = 240\Omega$ diff. (120 $\Omega$ per side), $R_S = 50\Omega$	$f_S = 2.4GHz$ , $f_{LO} = 2.2GHz$ , $f_{IF} = 200MHz$			0			dB
$PG_{C2}$	Power conversion gain: $R_L = 1200\Omega$ diff. (600 $\Omega$ per side), $R_S = 50\Omega$	$f_S = 2.4GHz$ , $f_{LO} = 2.2GHz$ , $f_{IF} = 200MHz$			7			dB
$\Delta G_C/\Delta T$	Gain temperature drift				-0.016			dB/ $^{\circ}C$
$\Delta G_C/\Delta V_{CC}$	Gain $V_{CC}$ drift				0.34			dB/V
$S_{11-RF}$	Input match at RF (2.4GHz) <sup>1</sup>				-10			dB
$NF_M$	SSB noise figure (2.4GHz) (50 $\Omega$ )				10			dB
$P_{-1dB}$	Mixer input 1dB gain compression				-6			dBm
IP3	Input third order intercept	$f_1 - f_2 = 1MHz$			3			dBm
IP <sub>2INT</sub>	Mixer input second order intercept				TBD			dBm
$f_{RF}$	RF frequency range		2.2		2.45		2.7	GHz
$f_{IF}$	IF frequency range		50		200		500	MHz

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## AC ELECTRICAL CHARACTERISTICS (continued)

SYMBOL	PARAMETER	TEST CONDITIONS	LIMITS					UNITS
			MIN	-3 $\sigma$	TYP	+3 $\sigma$	MAX	
<b>Rx Mixer Spurious Components</b>								
P <sub>RF-IF</sub>	RF feedthrough to IF	C <sub>L</sub> = 2pF per side			-50			dBc
P <sub>LO-IF</sub>	LO feedthrough to IF	C <sub>L</sub> = 2pF per side			-35			dBc
<b>Tx Mixer (RF = Pin 19, IF = Pins 7 and 8, LO = Pin 10 or 12, P<sub>LO</sub> = -10dBm)</b>								
P <sub>GC</sub>	Power conversion gain: R <sub>L</sub> = 50 $\Omega$ R <sub>S</sub> = 50 $\Omega$	f <sub>S</sub> = 2.4GHz, f <sub>LO</sub> = 2.2GHz, f <sub>IF</sub> = 200MHz			15			dB
$\Delta G_C/\Delta T$	Gain temperature drift				-0.007			dB/°C
$\Delta G_C/\Delta V_{CC}$	Gain voltage drift				0.4			dB/V
S <sub>11-RF</sub>	Output match at RF (2.4GHz) <sup>1</sup>				-10			dB
NF <sub>M</sub>	SSB noise figure (2.4GHz) (50 $\Omega$ )				12			dB
P <sub>-1dB</sub>	Output 1dB gain compression				0			dBm
IP <sub>3</sub>	Output third order intercept	f <sub>1</sub> - f <sub>2</sub> = 1MHz			7			dBm
IP <sub>2INT</sub>	Output second order intercept				TBD			dBm
f <sub>RF</sub>	RF frequency range		2.2		2.45		2.7	GHz
f <sub>IF</sub>	IF frequency range		50		200		500	MHz
<b>Tx Mixer Spurious Components</b>								
P <sub>IF-RF</sub>	IF feedthrough to RF				-29			dBc
P <sub>LO-RF</sub>	LO feedthrough to RF				-22			dBc
P <sub>2LO-RF</sub>	2*LO feedthrough to RF				-22			dBc
P <sub>IMAGE-RF</sub>	Image feedthrough to RF				-3			dBc
<b>LO Buffer: Full and Half Frequency inputs</b>								
P <sub>LO</sub>	LO drive level		-10		-7		5	dBm
S <sub>11-LO1</sub>	Mixer input match (LO = 2.2GHz)	LO mode = Hi			-11			dB
S <sub>11-LO2</sub>	Mixer input match (LO = 1.1GHz)	LO mode = Low			-11			dB
f <sub>LO2G</sub>	LO2G frequency range	LO mode = Hi	2		2.2		2.4	GHz
f <sub>LO1G</sub>	LO1G frequency range	LO mode = Low	1		1.1		1.2	GHz
<b>Switching<sup>2</sup></b>								
t <sub>Rx-Tx</sub>	Receive-to-transmit switching time				1			$\mu$ s
t <sub>Tx-Rx</sub>	Transmit-to-Receive switching time				1			$\mu$ s
t <sub>POWER UP</sub>	Chip enable time				1			$\mu$ s
t <sub>PWR DWN</sub>	Chip disable time				1			$\mu$ s

## NOTES:

1. With simple external matching
2. With 50pF coupling capacitors on all RF and IF ports

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Table 8. Truth Table

Chip-En	TxRx-SW	LNA-SW	LO-SW	Mode	LNA Gain	LO Freq. (Typ)
0	X	X	X	Sleep	N/S	N/S
1	0	1	1	Receive	+13dB	2.1GHz
1	0	0	1	Receive	-11dB	2.1GHz
1	0	1	0	Receive	+13dB	1.05GHz
1	0	0	0	Receive	-11dB	1.05GHz
1	1	X	1	Transmit	N/S	2.1GHz
1	1	X	0	Transmit	N/S	1.05GHz

**FUNCTIONAL DESCRIPTION**

The SA2420 is a 2.4GHz transceiver front-end available in the TSSOP-24 package. This integrated circuit (IC) consists of a low noise amplifier (LNA) and up- and down-converters. The injection of the local oscillator (LO) signal has two options: 1) direct injection of the LO signal at approximately 2GHz, or 2) injection of an LO signal at approximately 1GHz through an on-chip doubler. The SA2420 functions with a supply voltage range of 3 – 5 V (nominally). There is an enable/disable switch available to power up/down the entire chip in 1 $\mu$ s, typically. This transceiver has several unique features.

The LNA has two operating modes: 1) high gain mode with a gain = +13dB; and 2) low gain mode with a gain <-10dB. The switch for this option is internal and is controlled externally by high and low logic to the pin. When the LNA is switched into the attenuation mode, active matching circuitry (on-chip) is switched in (reducing the number of off-chip components required). The input and output are internally matched to 50 $\Omega$ . To reduce power consumption when the chip is transmitting, the LNA is automatically switched into a "sleep" mode (internally) without the use of external circuitry.

The up and down frequency converters are single-ended at the RF port of the mixers. The up and down converters share the same

(RF) pin and use an internal switch for transmitting (up-converting) or receiving (down-converting) modes. The switch is controlled externally by high and low logic states. The RF port is matched to 50 $\Omega$  and has an input IP3 of +3dBm (mixer only). The down-convert mixer is buffered and has open collectors at the pins to allow for matching to common SAW filters. The up-convert mixer has differential inputs (IF port) and single-ended output (RF port), with an input pin to output pin gain of 10dB. The output of the up-converter is designed for a power level = 0dBm (P<sub>-1dB</sub>). The mixers are fed by the two LO options.

The available LO options are: direct injection (2.1GHz at the pin) or through an on-chip doubler. The doubler has a simple LC bandpass filter (internal) at its output which passes the second harmonic to the mixers. Through an internal switch (controlled externally), either LO can be used depending on the designer's application. If an application requires the use of a 1.05GHz VCO, then the doubler option would be used to double the frequency ( $2 \times 1.05\text{GHz} = 2.1\text{GHz}$ ) before being injected into the mixers. For a 2.1GHz VCO, the direct option would be used. With this option, the signal passes through an on-chip buffer and is then injected into the mixers.