

JFET Input Operational Amplifiers

These low cost JFET input operational amplifiers combine two state-of-the-art analog technologies on a single monolithic integrated circuit. Each internally compensated operational amplifier has well matched high voltage JFET input devices for low input offset voltage. The BIFET technology provides wide bandwidths and fast slew rates with low input bias currents, input offset currents, and supply currents.

The Motorola BIFET family offers single, dual and quad operational amplifiers which are pin-compatible with the industry standard MC1741, MC1458, and the MC3403/LM324 bipolar devices. The MC34001/ 34002/34004 series are specified from 0° to +70°C.

- Input Offset Voltage Options of 5.0 mV and 10 mV Maximum
- Low Input Bias Current: 40 pA
- Low Input Offset Current: 10 pA
- Wide Gain Bandwidth: 4.0 MHz
- High Slew Rate: 13 V/µs
- Low Supply Current: 1.4 mA per Amplifier
- High Input Impedance: $10^{12} \Omega$
- High Common Mode and Supply Voltage Rejection Ratios: 100 dB
- Industry Standard Pinouts

ORDERING INFORMATION

Op Amp Function	Device	Operating Temperature Range	Package
Single	MC34001BD, D	T _A = 0° to+ 70°C	SO–8
Single	MC34001BP, P	Plastic DIP	
Dual	MC34002BD, D	$T_A = 0^\circ$ to +70°C	SO–8
Duai	MC34002BP, P	IA = 0 10 +70 C	Plastic DIP
Quad	MC34004BP, P	$T_A = 0^\circ$ to +70°C	Plastic DIP

JFET INPUT OPERATIONAL AMPLIFIERS

MC34001, B MC34002, B

MC34004, B



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MAXIMUM RATINGS

Rating	Symbol	Value	Unit
Supply Voltage	V _{CC} , V _{EE}	±18	V
Differential Input Voltage (Note 1)	VID	±30	V
Input Voltage Range	VIDR	±16	V
Open Short Circuit Duration	tSC	Continuous	
Operating Ambient Temperature Range	т _А	0 to +70	°C
Operating Junction Temperature	ТJ	150	°C
Storage Temperature Range	T _{stg}	-65 to +150	°C

NOTES: 1. Unless otherwise specified, the absolute maximum negative input voltage is equal to the

negative power supply.

ELECTRICAL CHARACTERISTICS (V_{CC} = +15 V, V_{EE} = -15 V, T_A = 25°C, unless otherwise noted.)

Characteristics	Symbol	Min	Тур	Max	Unit
Input Offset Voltage ($R_S \le 10 \text{ k}$) MC3400XB MC3400X	VIO		3.0 5.0	5.0 10	mV
Average Temperature Coefficient of Input Offset Voltage $R_S \le 10 \text{ k}, T_A = T_{low} \text{ to } T_{high} \text{ (Note 2)}$	$\Delta V_{IO} / \Delta T$	_	10	-	μV/°C
Input Offset Current (V _{CM} = 0) (Note 3) MC3400XB MC3400X	IIO		25 25	100 100	рА
Input Bias Current (V _{CM} = 0) (Note 3) MC3400XB MC3400X	IIB	_	50 50	200 200	рА
Input Resistance	r _i	-	10 ¹²	—	Ω
Common Mode Input Voltage Range	VICR	±11 —	+15 -12	_	V
Large Signal Voltage Gain (V _O = ±10 V, R _L = 2.0 k) MC3400XB MC3400X	AVOL	50 25	150 100		V/mV
$\begin{array}{l} \mbox{Output Voltage Swing} \\ (R_L \geq 10 \ k) \\ (R_L \geq 2.0 \ k) \end{array}$	Vo	±12 ±10	±14 ±13		V
Common Mode Rejection Ratio ($R_S \le 10 \text{ k}$) MC3400XB MC3400X	CMRR	80 70	100 100		dB
Supply Voltage Rejection Ratio (RS \leq 10 k) (Note 4) MC3400XB MC3400X	PSRR	80 70	100 100		dB
Supply Current (Each Amplifier) MC3400XB MC3400X	۱ _D		1.4 1.4	2.5 2.7	mA
Slew Rate ($A_V = 1.0$)	SR	_	13	-	V/µs
Gain–Bandwidth Product	GBW	_	4.0	-	MHz
Equivalent Input Noise Voltage ($R_S = 100 \Omega$, f = 1000 Hz)	e _n	—	25	_	nV/√Hz
Equivalent Input Noise Current (f = 1000 Hz)	in	_	0.01	_	pA/√H

2. T_{Iow} = 0°C for MC34001/34001B T_{high} = +70°C for MC34001/34001B MC34002 MC34004/34004B MC34002 MC34004/34004B MC34004/34004B 3. The input bias currents approximately double for every 10°C rise in junction temperature, T_J. Due to limited test time, the input bias currents are

correlated to junction temperature. Use of a heatsink is recommended if input bias current is to be kept to a minimum. 4. Supply voltage rejection ratio is measured for both supply magnitudes increasing or decreasing simultaneously, in accordance with common practice.

Characteristics	Symbol	Min	Тур	Max	Unit
Input Offset Voltage ($R_S \le 10 \text{ k}$) MC3400XB MC3400X	VIO			7.0 13	mV
Input Offset Current (V _{CM} = 0) (Note 3) MC3400XB MC3400X	liO			4.0 4.0	nA
Input Bias Current (V _{CM} = 0) (Note 3) MC3400XB MC3400X	IB			8.0 8.0	nA
Common Mode Input Voltage Range	VICR	±11	—	—	V
Large Signal (V _O = ±10 V, R _L = 2.0 k) MC3400XB MC3400X	AVOL	25 15			V/mV
Output Voltage Swing (R \ge 10 k) (R \ge 2.0 k)	Vo	±12 ±10			V
Common Mode Rejection Ratio ($R_S \le 10 \text{ k}$) MC3400XB MC3400X	CMRR	80 70			dB
Supply Voltage Rejection Ratio ($R_S \le 10 \text{ k}$) (Note 4) MC3400XB MC3400X	PSRR	80 70			dB
Supply Current (Each Amplifier) MC3400XB MC3400X	ID			2.8 3.0	mA

ELECTRICAL CHARACTERISTICS (V_{CC} = +15 V, V_{EE} = -15 V, T_A = T_{low} to T_{high} [Note 2].)

NOTES: 2. T_{IOW} = 0°C for MC34001/34001B MC34002

 $T_{high} = +70^{\circ}C \text{ for MC34001/34001B} MC34002$

MC34004/34004B

MC34004/34004B

3. The input bias currents approximately double for every 10°C rise in junction temperature, T_J. Due to limited test time, the input bias currents are

correlated to junction temperature. Use of a heatsink is recommended if input bias current is to be kept to a minimum.

4. Supply voltage rejection ratio is measured for both supply magnitudes increasing or decreasing simultaneously, in accordance with common practice.

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Figure 6. Supply Current per Amplifier versus Temperature







Figure 10. Equivalent Input Noise Voltage versus Frequency







MOTOROLA ANALOG IC DEVICE DATA





Figure 12. Output Current to Voltage Transformation for a D–to–A Converter



Settling time to within 1/2 LSB is approximately $4.0 \,\mu$ s from the time all bits are switched (C = 68 pF).

The value of C may be selected to minimize overshoot and ringing.

Theoretical VO

$$V_{O} = \frac{V_{ref}}{R1} (R_{O}) \left[\frac{A1}{2} + \frac{A2}{4} + \frac{A3}{8} + \frac{A4}{16} + \frac{A5}{32} + \frac{A6}{64} + \frac{A7}{128} + \frac{A8}{256} \right]$$

Figure 13. Positive Peak Detector



Figure 14. Long Interval RC Timer





Figure 15. Isolating Large Capacitive Loads



Figure 16. Wide BW, Low Noise, Low Drift Amplifier





Parasitic input capacitance (C1 \cong 3.0 pF plus any additional layout capacitance) interacts with feedback elements and creates undesirable high–frequency pole. To compensate add C2 such that: R2C2 \cong R1C1.

OUTLINE DIMENSIONS



OUTLINE DIMENSIONS – continued



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