MOTOROLA

MC33102

Advance Information

Sleep-Mode™ Two-State, Micropower Operational Amplifier

The MC33102 dual operational amplifier is an innovative design concept employing Sleep–Mode™ technology. Sleep–Mode amplifiers have two separate states, a sleepmode and an awakemode. In sleepmode, the amplifier is active and waiting for an input signal. When a signal is applied causing the amplifier to source or sink 160 μA (typically) to the load, it will automatically switch to the awakemode which offers higher slew rate, gain bandwidth, and drive capability.

- Two States: "Sleepmode" (Micropower) and "Awakemode" (High Performance)
- Switches from Sleepmode to Awakemode in 4.0 μs when Output Current Exceeds the Threshold Current (R_L = 600 Ω)
- Independent Sleepmode Function for Each Op Amp
- Standard Pinouts – No Additional Pins or Components Required
- Sleepmode State – Can Be Used in the Low Current Idle State as a Fully Functional Micropower Amplifier
- Automatic Return to Sleepmode when Output Current Drops Below Threshold
- No Deadband/Crossover Distortion; as Low as 1.0 Hz in the Awakemode
- Drop–in Replacement for Many Other Dual Op Amps
- ESD Clamps on Inputs Increase Reliability without Affecting Device Operation
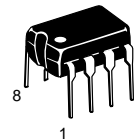
DUAL SLEEP–MODE™ OPERATIONAL AMPLIFIER

SEMICONDUCTOR TECHNICAL DATA

D SUFFIX
PLASTIC PACKAGE
CASE 751
(SO–8)

8
1

P SUFFIX
PLASTIC PACKAGE
CASE 626

8
1

PIN CONNECTIONS

Output 1 — 1   8 — V_{CC}
2   7 — Output 2
Inputs 1   3   1   6 } Inputs 2
V_{EE} — 4   2   5

(Dual, Top View)

TYPICAL SLEEPMODE/AWAKEMODE PERFORMANCE

Characteristic	Sleepmode (Typical)	Awakemode (Typical)	Unit
Low Current Drain	45	750	μA
Low Input Offset Voltage	0.15	0.15	mV
High Output Current Capability	0.15	50	mA
Low T.C. of Input Offset Voltage	1.0	1.0	$\mu V/°C$
High Gain Bandwidth (@ 20 kHz)	0.33	4.6	MHz
High Slew Rate	0.16	1.7	$V/\mu s$
Low Noise (@ 1.0 kHz)	28	9.0	$nV/\sqrt{Hz}$

MAXIMUM RATINGS

Ratings	Symbol	Value	Unit
Supply Voltage (V_{CC} to V_{EE})	V_S	+36	V
Input Differential Voltage Range Input Voltage Range	V_{IDR} V_{IR}	(Note 1)	V
Output Short Circuit Duration (Note 2)	t_{SC}	(Note 2)	sec
Maximum Junction Temperature Storage Temperature	T_J T_{stg}	+150 −65 to +150	°C
Maximum Power Dissipation	P_D	(Note 2)	mW

NOTES: 1. Either or both input voltages should not exceed V_{CC} or V_{EE}.
      2. Power dissipation must be considered to ensure maximum junction temperature (T_J) is not exceeded (refer to Figure 1).

ORDERING INFORMATION

Device	Operating Temperature Range	Package
MC33102D	T_A = − 40° to +85°C	SO–8
MC33102P		Plastic DIP

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Simplified Block Diagram
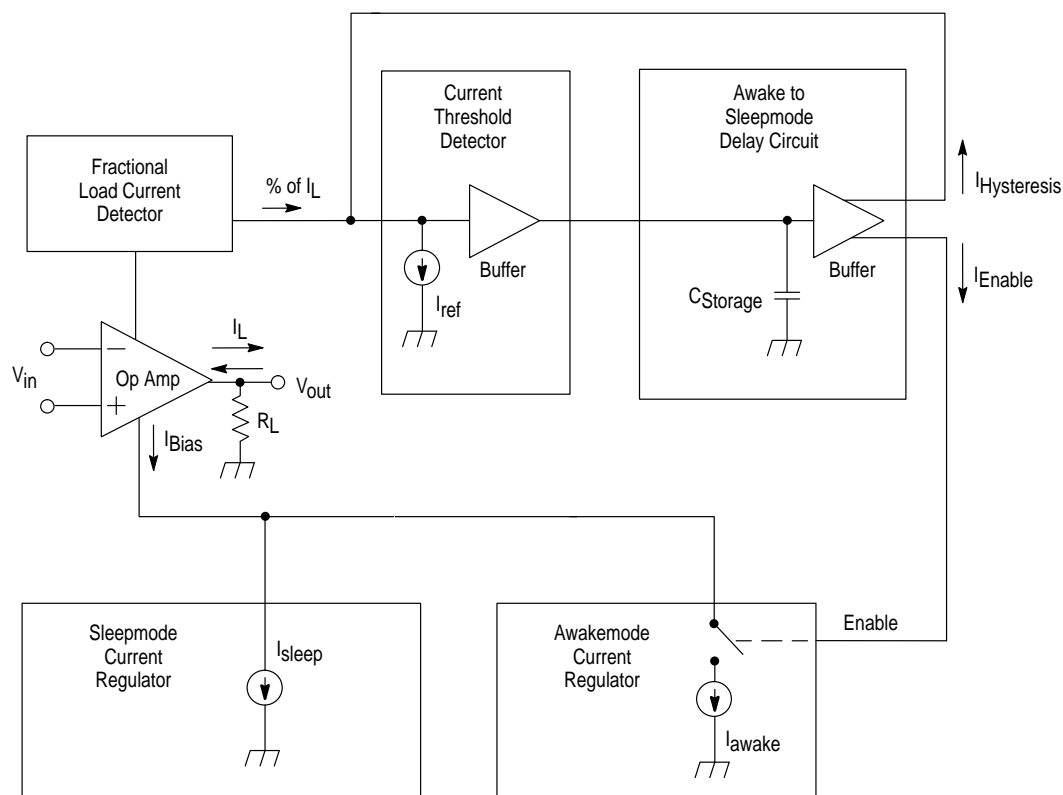
DC ELECTRICAL CHARACTERISTICS (V_{CC} = +15 V, V_{EE} = −15 V, T_A = 25°C, unless otherwise noted.)

Characteristics	Figure	Symbol	Min	Typ	Max	Unit

Input Offset Voltage (R_S = 50 Ω, V_{CM} = 0 V, V_O = 0 V)
  Sleepmode
    T_A = +25°C
    T_A = −40° to +85°C
  Awakemode
    T_A = +25°C
    T_A = −40° to +85°C
	2	$\lvert V_{IO} \rvert$	
			
			—
			—
			
			—
			—
				
				0.15
				—
				
				0.15
				—
					
					2.0
					3.0
					
					2.0
					3.0
						mV

Input Offset Voltage Temperature Coefficient
  (R_S = 50 Ω, V_{CM} = 0 V, V_O = 0 V)
  T_A = −40° to +85°C (Sleepmode and Awakemode)
	3	$\Delta V_{IO}/\Delta T$	
			
			—
				
				1.0
					
					—
						μV/°C

Input Bias Current (V_{CM} = 0 V, V_O = 0 V)
  Sleepmode
    T_A = +25°C
    T_A = −40° to +85°C
  Awakemode
    T_A = +25°C
    T_A = −40° to +85°C
	4, 6	I_{IB}	
			
			—
			—
			
			—
			—
				
				8.0
				—
				
				100
				—
					
					50
					60
					
					500
					600
						nA

Input Offset Current (V_{CM} = 0 V, V_O = 0 V)
  Sleepmode
    T_A = +25°C
    T_A = −40° to +85°C
  Awakemode
    T_A = +25°C
    T_A = −40° to +85°C
	—	$\lvert I_{IO} \rvert$	
			
			—
			—
			
			—
			—
				
				0.5
				—
				
				5.0
				—
					
					5.0
					6.0
					
					50
					60
						nA

DC ELECTRICAL CHARACTERISTICS (V_{CC} = +15 V, V_{EE} = −15 V, T_A = 25°C, unless otherwise noted.)

Characteristics	Figure	Symbol	Min	Typ	Max	Unit
Common Mode Input Voltage Range (ΔV_{IO} = 5.0 mV, V_O = 0 V) Sleepmode and Awakemode	5	V_{ICR}	 −13 —	 −14.8 +14.2	 — +13	V

Large Signal Voltage Gain
    Sleepmode (R_L = 1.0 MΩ)
        T_A = +25°C
        T_A = −40° to +85°C
    Awakemode (V_O = ±10 V, R_L = 600 Ω)
        T_A = +25°C
        T_A = −40 to +85°C
	7	A_{VOL}	
			
			25
			15
			
			50
			25
				
				200
				—
				
				700
				—
					
					—
					—
					
					—
					—
						kV/V

Output Voltage Swing (V_{ID} = ±1.0 V)
    Sleepmode (V_{CC} = +15 V, V_{EE} = −15 V)
        R_L = 1.0 MΩ
        R_L = 1.0 MΩ
    Awakemode (V_{CC} = +15 V, V_{EE} = −15 V)
        R_L = 600 Ω
        R_L = 600 Ω
        R_L = 2.0 kΩ
        R_L = 2.0 kΩ
    Awakemode (V_{CC} = +2.5 V, V_{EE} = −2.5 V)
        R_L = 600 Ω
        R_L = 600 Ω
	8, 9, 10	
		
		V_{O+}
		V_{O-}
		
		V_{O+}
		V_{O-}
		V_{O+}
		V_{O-}
		
		V_{O+}
		V_{O-}
			
			+13.5
			—
			
			+12.5
			—
			+13.3
			—
			
			+1.1
			—
				
				+14.2
				−14.2
				
				+13.6
				−13.6
				+14
				−14
				
				+1.6
				−1.6
					
					—
					−13.5
					
					—
					−12.5
					—
					−13.3
					
					—
					−1.1
						V
						
						V

Common Mode Rejection (V_{CM} = ±13 V) Sleepmode and Awakemode	11	CMR	 80	 90	 —	dB

Power Supply Rejection (V_{CC}/V_{EE} = +15 V/−15 V,
    5.0 V/−15 V, +15 V/−5.0 V)
    Sleepmode and Awakemode
	12	PSR	
			
			80
				
				100
					
					—
						dB

Output Transition Current
    Sleepmode to Awakemode (Source/Sink)
        (V_S = ±15 V)
        (V_S = ±2.5 V)
    Awakemode to Sleepmode (Source/Sink)
        (V_S = ±15 V)
        (V_S = ±2.5 V)
	13, 14	
		$|I_{TH1}|$
		
		$|I_{TH2}|$
			
			200
			250
			
			—
			—
				
				160
				200
				
				142
				180
					
					—
					—
					
					90
					140
						µA

Output Short Circuit Current (Awakemode)
    (V_{ID} = ±1.0 V, Output to Ground)
        Source
        Sink
	15, 16	$|I_{SC}|$	
			
			50
			50
				
				110
				110
					
					—
					—
						mA

Power Supply Current (per Amplifier) (A_{CL} = 1, V_O = 0V)
    Sleepmode (V_S = ±15 V)
        T_A = +25°C
        T_A = −40° to +85°C
    Sleepmode (V_S = ±2.5 V)
        T_A = +25°C
        T_A = −40° to +85°C
    Awakemode (V_S = ±15 V)
        T_A = +25°C
        T_A = −40° to +85°C
	17	I_D	
			
			—
			—
			
			—
			—
			
			—
			—
				
				45
				48
				
				38
				42
				
				750
				800
					
					65
					70
					
					65
					—
					
					800
					900
						µA

AC ELECTRICAL CHARACTERISTICS (V_{CC} = +15 V, V_{EE} = −15 V, T_A = 25°C, unless otherwise noted.)

| Characteristics	Figure	Symbol	Min	Typ	Max	Unit
Slew Rate (V_{in} = −5.0 V to +5.0 V, C_L = 50 pF, A_V = 1.0) Sleepmode (R_L = 1.0 MΩ) Awakemode (R_L = 600 Ω)	18	SR	 0.10 1.0	 0.16 1.7	 — —	V/µs
Gain Bandwidth Product Sleepmode (f = 10 kHz) Awakemode (f = 20 kHz)	19	GBW	 0.25 3.5	 0.33 4.6	 — —	MHz
Sleepmode to Awakemode Transition Time (A_{CL} = 0.1, V_{in} = 0 V to +5.0 V) R_L = 600 Ω R_L = 10 kΩ	20, 21	t_{tr1}	 — —	 4.0 15	 — —	µs
Awakemode to Sleepmode Transition Time	22	t_{tr2}	—	1.5	—	sec
Unity Gain Frequency (Open Loop) Sleepmode (R_L = 100 kΩ, C_L = 0 pF) Awakemode (R_L = 600 Ω, C_L = 0 pF)		f_U	 — —	 200 2500	 — —	kHz
Gain Margin Sleepmode (R_L = 100 kΩ, C_L = 0 pF) Awakemode (R_L = 600 Ω, C_L = 0 pF)	23, 25	A_M	 — —	 13 12	 — —	dB
Phase Margin Sleepmode (R_L = 100 kΩ, C_L = 0 pF) Awakemode (R_L = 600 Ω, C_L = 0 pF)	24, 26	$\varnothing_M$	 — —	 60 60	 — —	Degrees
Channel Separation (f = 100 Hz to 20 kHz) Sleepmode and Awakemode	29	CS	 —	 120	 —	dB
Power Bandwidth (Awakemode) (V_O = 10 V_{pp}, R_L = 100 kΩ, THD ≤ 1%)		BW_P	 —	 20	 —	kHz
Total Harmonic Distortion (V_O = 2.0 V_{pp}, A_V = 1.0) Awakemode (R_L = 600 Ω) f = 1.0 kHz f = 10 kHz f = 20 kHz	30	THD	 — — —	 0.005 0.016 0.031	 — — —	%
DC Output Impedance (V_O = 0 V, A_V = 10, I_Q = 10 µA) Sleepmode Awakemode	31	R_O	 — —	 1.0 k 96	 — —	Ω
Differential Input Resistance (V_{CM} = 0 V) Sleepmode Awakemode		R_{in}	 — —	 1.3 0.17	 — —	MΩ
Differential Input Capacitance (V_{CM} = 0 V) Sleepmode Awakemode		C_{in}	 — —	 0.4 4.0	 — —	pF
Equivalent Input Noise Voltage (f = 1.0 kHz, R_S = 100 Ω) Sleepmode Awakemode	32	e_n	 — —	 28 9.0	 — —	nV/$\sqrt{\text{Hz}}$
Equivalent Input Noise Current (f = 1.0 kHz) Sleepmode Awakemode	33	i_n	 — —	 0.01 0.05	 — —	pA/$\sqrt{\text{Hz}}$

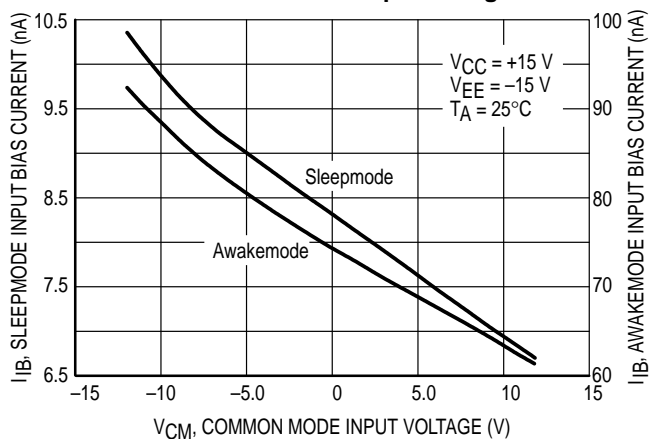**Figure 1. Maximum Power Dissipation versus Temperature**
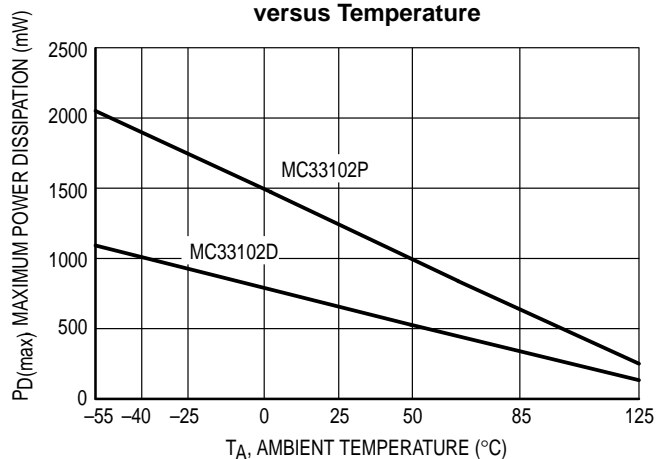
Figure 2. Distribution of Input Offset Voltage (MC33102D Package)
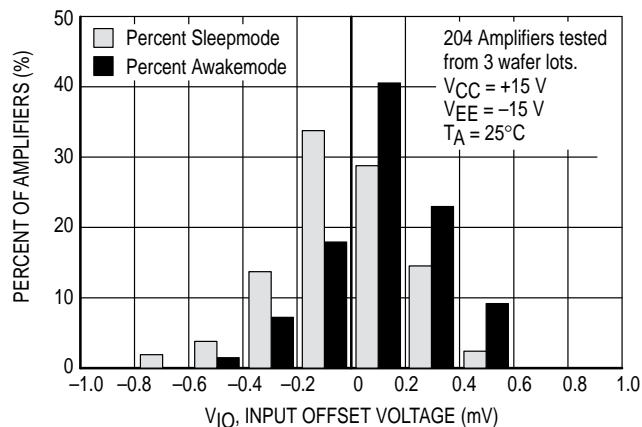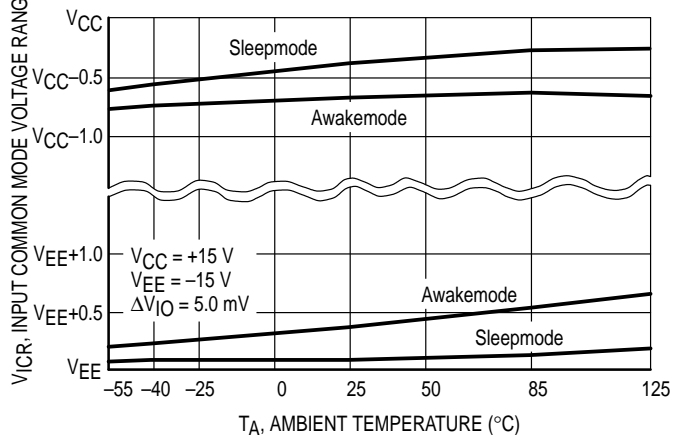

Figure 3. Input Offset Voltage Temperature Coefficient Distribution (MC33102D Package)
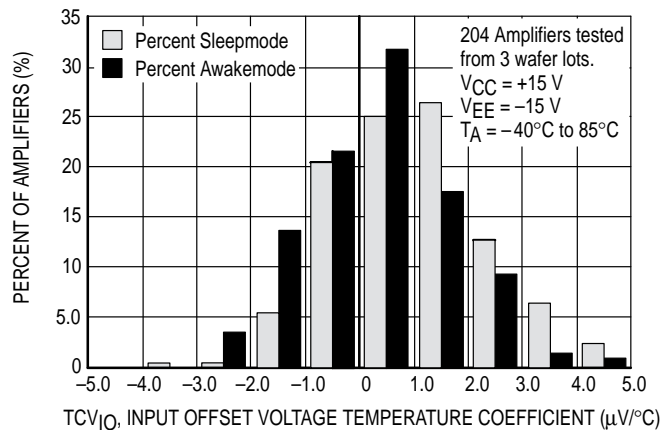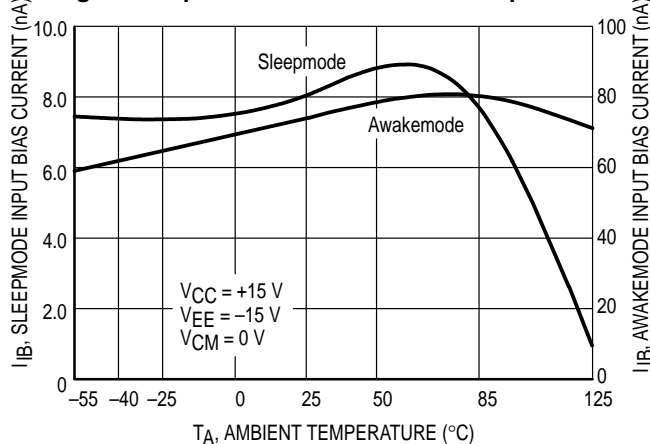

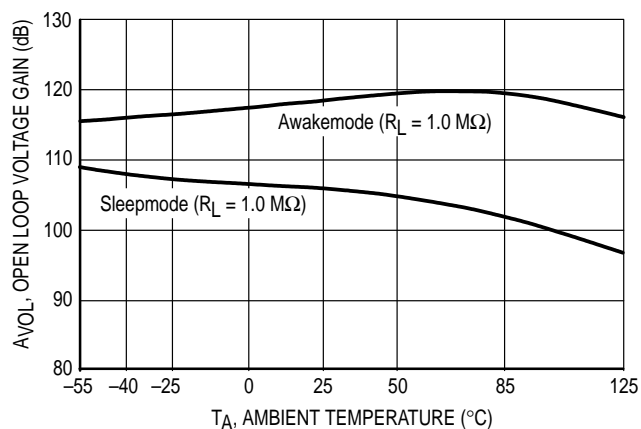**Figure 4. Input Bias Current versus Common Mode Input Voltage**

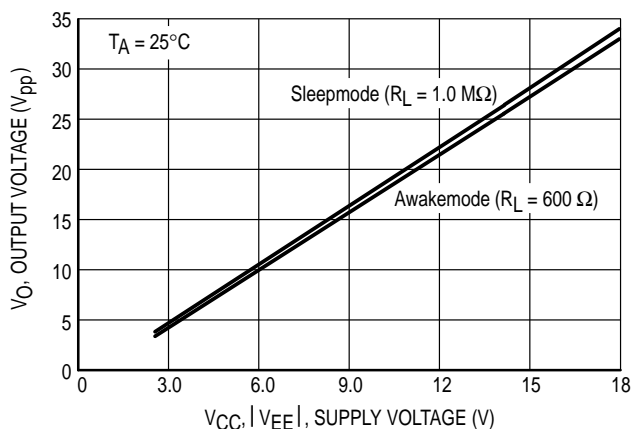**Figure 5. Input Common Mode Voltage Range versus Temperature**

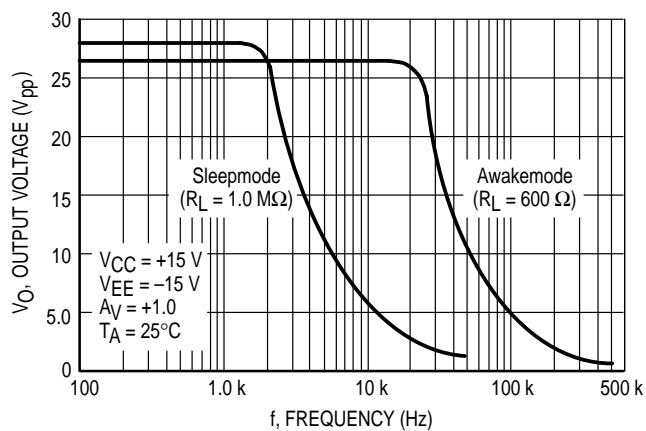**Figure 6. Input Bias Current versus Temperature**
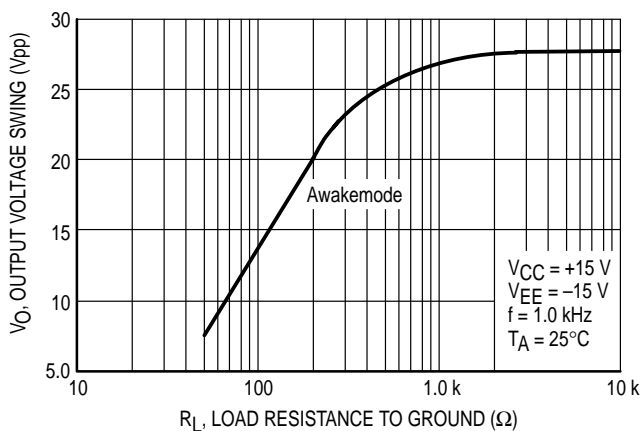
MC33102

Figure 7. Open Loop Voltage Gain versus Temperature

Figure 7. Open Loop Voltage Gain versus Temperature

A_{VOL}, OPEN LOOP VOLTAGE GAIN (dB) versus T_A, AMBIENT TEMPERATURE (°C)

Awakemode (R_L = 1.0 MΩ)
Sleepmode (R_L = 1.0 MΩ)

Figure 8. Output Voltage Swing versus Supply Voltage

Figure 8. Output Voltage Swing versus Supply Voltage

V_O, OUTPUT VOLTAGE (V_{pp}) versus V_{CC}, $|V_{EE}|$, SUPPLY VOLTAGE (V)

T_A = 25°C

Sleepmode (R_L = 1.0 MΩ)
Awakemode (R_L = 600 Ω)

Figure 9. Output Voltage versus Frequency

Figure 9. Output Voltage versus Frequency

V_O, OUTPUT VOLTAGE (V_{pp}) versus f, FREQUENCY (Hz)

Sleepmode (R_L = 1.0 MΩ)
Awakemode (R_L = 600 Ω)

V_{CC} = +15 V
V_{EE} = −15 V
A_V = +1.0
T_A = 25°C

Figure 10. Maximum Peak–to–Peak Output Voltage Swing versus Load Resistance

Figure 10. Maximum Peak–to–Peak Output Voltage Swing versus Load Resistance

V_O, OUTPUT VOLTAGE SWING (V_{pp}) versus R_L, LOAD RESISTANCE TO GROUND (Ω)

Awakemode

V_{CC} = +15 V
V_{EE} = −15 V
f = 1.0 kHz
T_A = 25°C

Figure 11. Common Mode Rejection versus Frequency

Figure 11. Common Mode Rejection versus Frequency

CMR, COMMON MODE REJECTION (dB) versus f, FREQUENCY (Hz)

Awakemode
Sleepmode

V_{CC} = +15 V
V_{EE} = −15 V
V_{CM} = 0 V
ΔV_{CM} = ±1.5 V
T_A = 25°C

Figure 12. Power Supply Rejection versus Frequency

Figure 12. Power Supply Rejection versus Frequency

PSR, POWER SUPPLY REJECTION (dB) versus f, FREQUENCY (Hz)

+PSR Sleepmode
+PSR Awakemode
−PSR Awakemode
−PSR Sleepmode

V_{CC} = +15 V
V_{EE} = −15 V
ΔV_{CC} = ±1.5 V
T_A = 25°C

MC33102
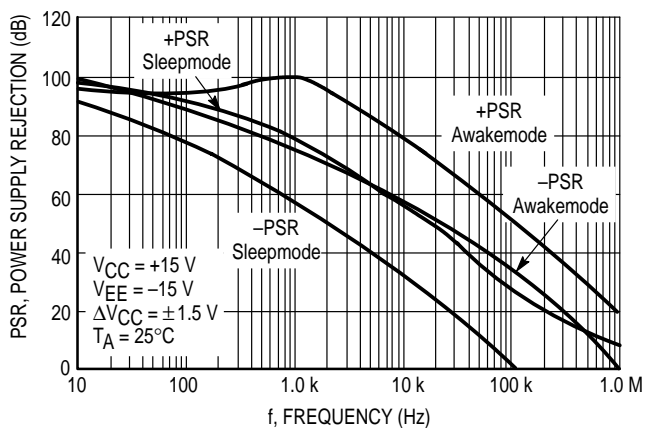
**Figure 13. Sleepmode to Awakemode
Current Threshold versus Supply Voltage**
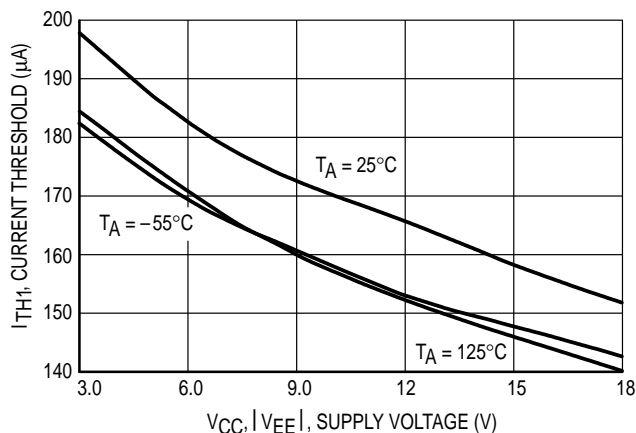
**Figure 14. Awakemode to Sleepmode
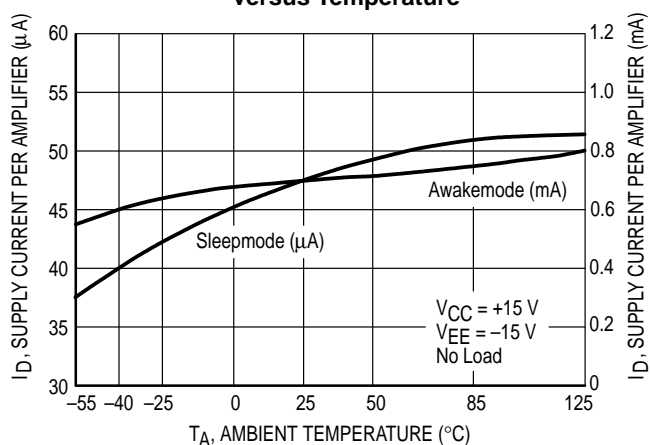Current Threshold versus Supply Voltage**
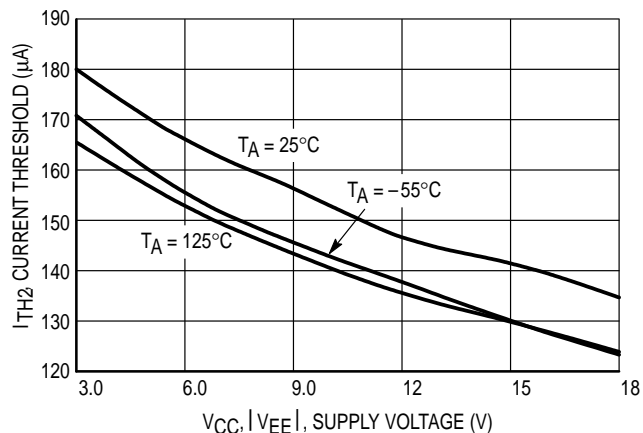
**Figure 15. Output Short Circuit Current
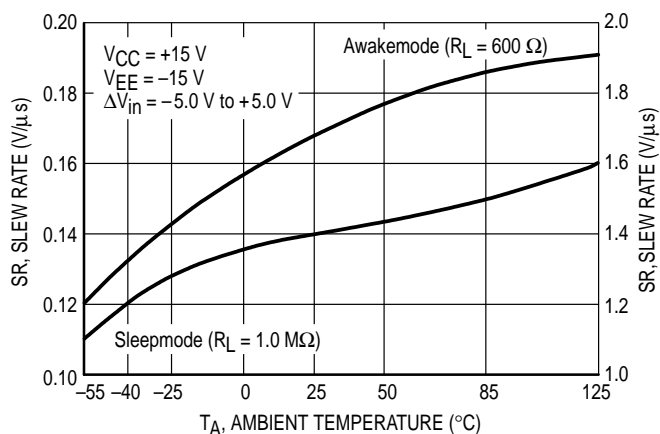versus Output Voltage**
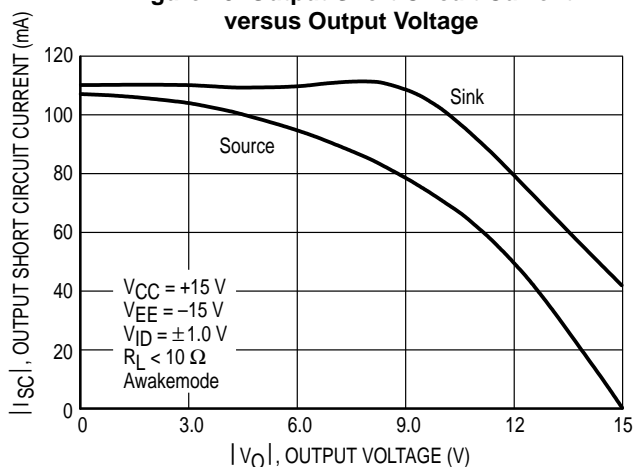
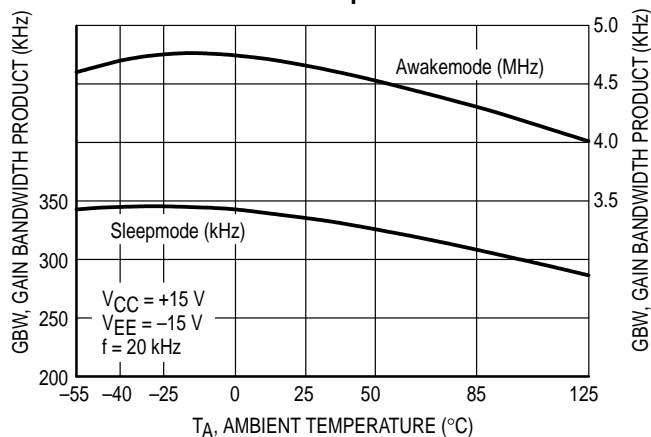**Figure 16. Output Short Circuit Current
versus Temperature**
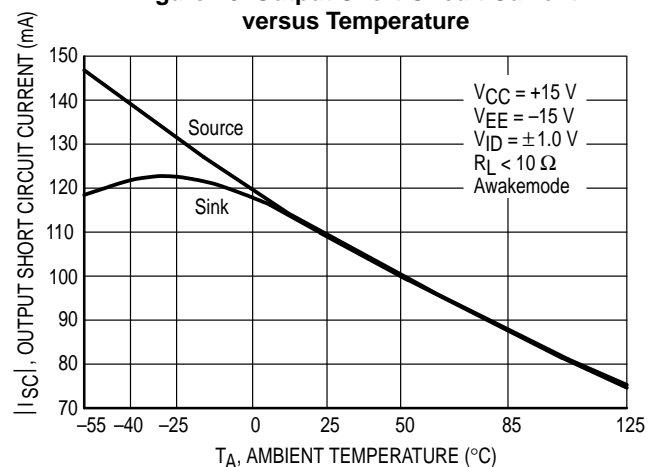
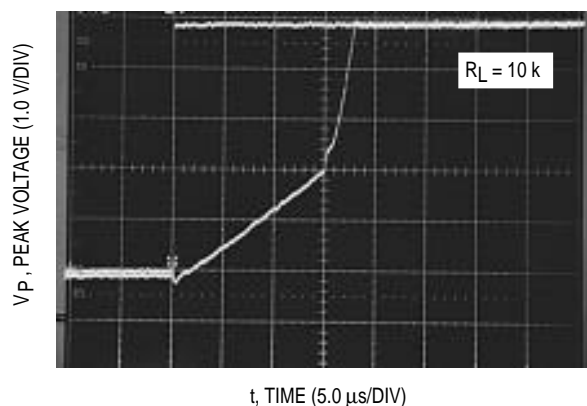**Figure 17. Power Supply Current Per Amplifier
versus Temperature**

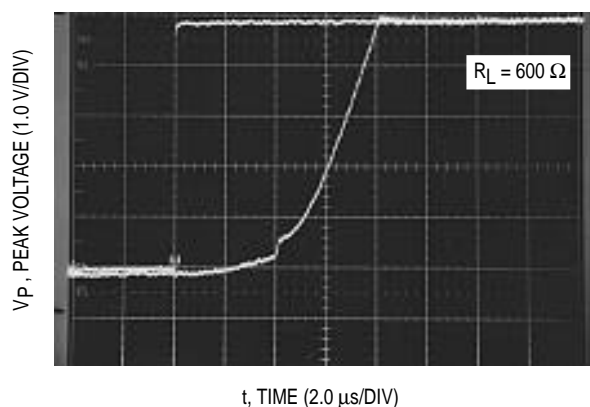**Figure 18. Slew Rate versus Temperature**
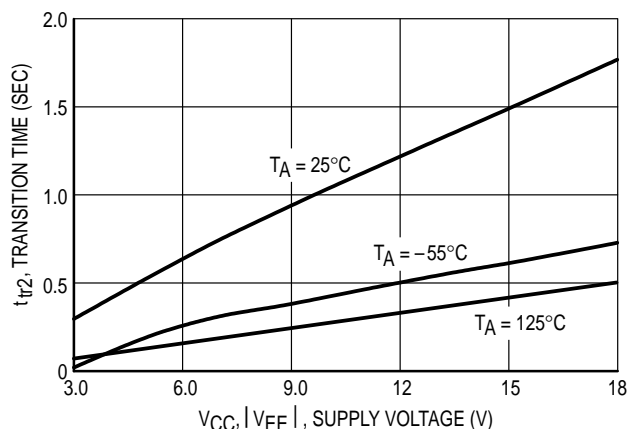
Figure 19. Gain Bandwidth Product versus Temperature

GBW, GAIN BANDWIDTH PRODUCT (KHz)

GBW, GAIN BANDWIDTH PRODUCT (KHz)

Awakemode (MHz)

Sleepmode (kHz)

$V_{CC} = +15$ V
$V_{EE} = -15$ V
f = 20 kHz

T_A, AMBIENT TEMPERATURE (°C)

Figure 20. Sleepmode to Awakemode Transition Time

V_P, PEAK VOLTAGE (1.0 V/DIV)

$R_L = 10$ k

t, TIME (5.0 µs/DIV)

Figure 21. Sleepmode to Awakemode Transition Time

V_P, PEAK VOLTAGE (1.0 V/DIV)

$R_L = 600$ Ω

t, TIME (2.0 µs/DIV)

Figure 22. Awakemode to Sleepmode Transition Time versus Supply Voltage

t_{tr2}, TRANSITION TIME (SEC)

$T_A = 25°C$

$T_A = -55°C$

$T_A = 125°C$

$V_{CC}, |V_{EE}|$, SUPPLY VOLTAGE (V)

Figure 23. Gain Margin versus Differential Source Resistance

A_m, GAIN MARGIN (dB)

Sleepmode

Awakemode

$V_{CC} = +15$ V
$V_{EE} = -15$ V
$R_T = R1 + R2$
$V_O = 0$ V
$T_A = 25°C$

R1
R2
V_O

R_T, DIFFERENTIAL SOURCE RESISTANCE (Ω)
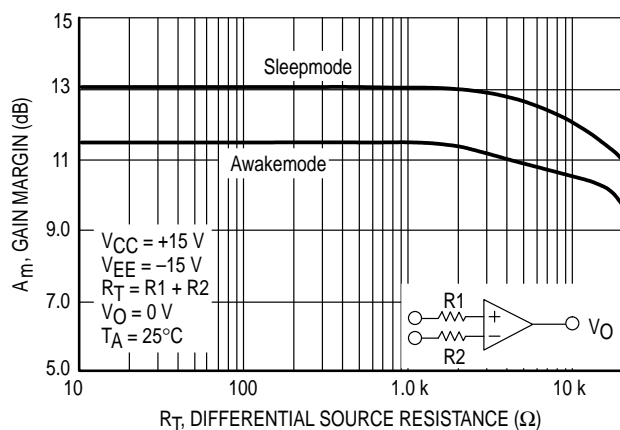
Figure 24. Phase Margin versus Differential Source Resistance

$\varnothing_m$, PHASE MARGIN (DEG)

Sleepmode

$V_{CC} = +15$ V
$V_{EE} = -15$ V
$R_T = R1 + R2$
$V_O = 0$ V
$T_A = 25°C$

Awakemode

R1
R2
V_O

R_T, DIFFERENTIAL SOURCE RESISTANCE (Ω)

MC33102
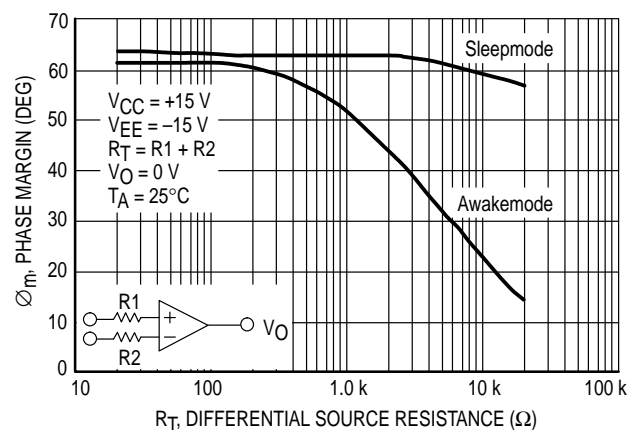
Figure 25. Open Loop Gain Margin versus Output Load Capacitance
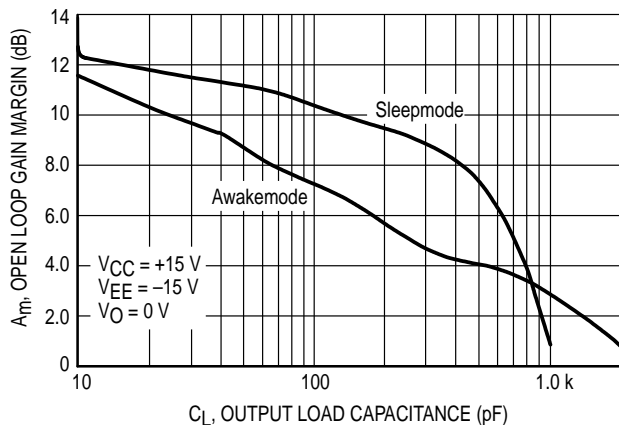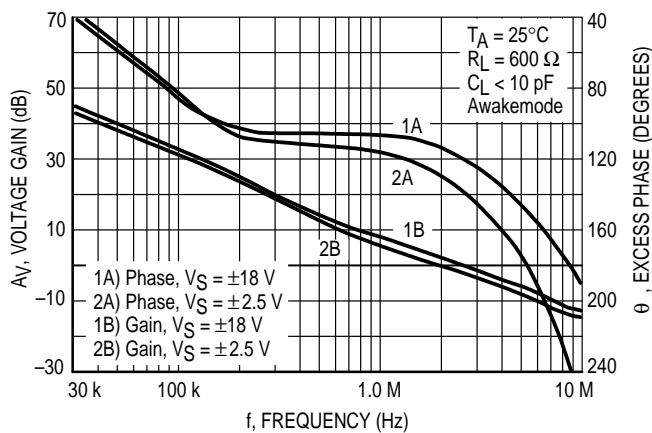

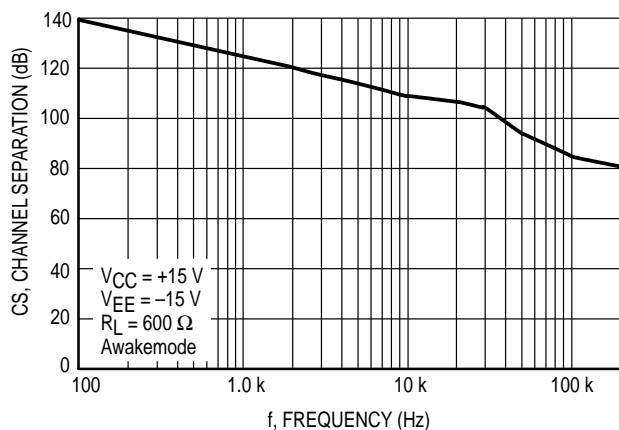**Figure 26. Phase Margin versus Output Load Capacitance**
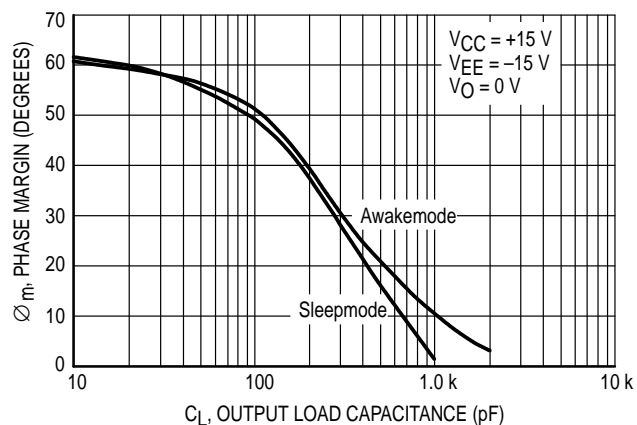
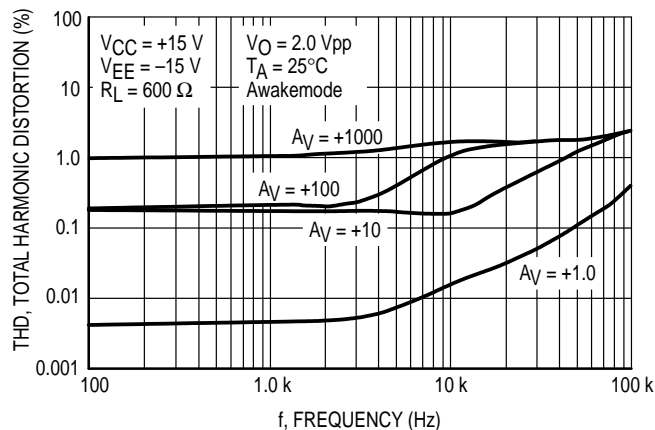**Figure 27. Sleepmode Voltage Gain and Phase versus Frequency**
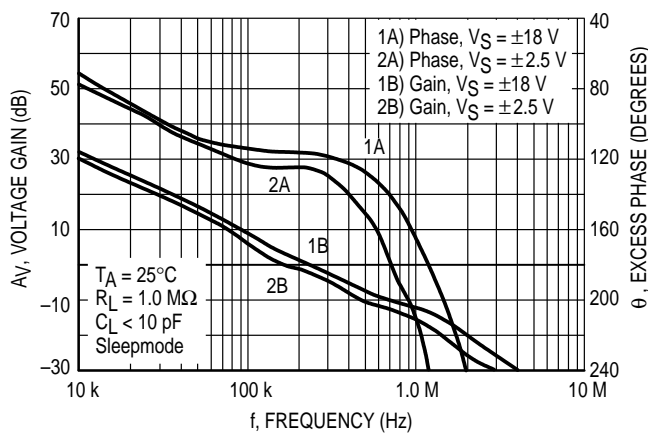
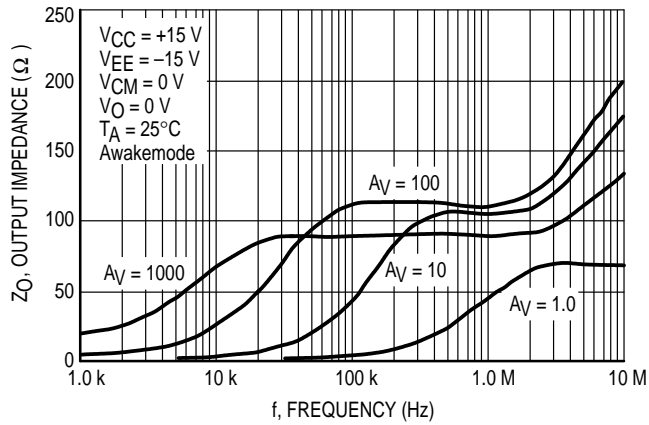**Figure 28. Awakemode Voltage Gain and Phase versus Frequency**

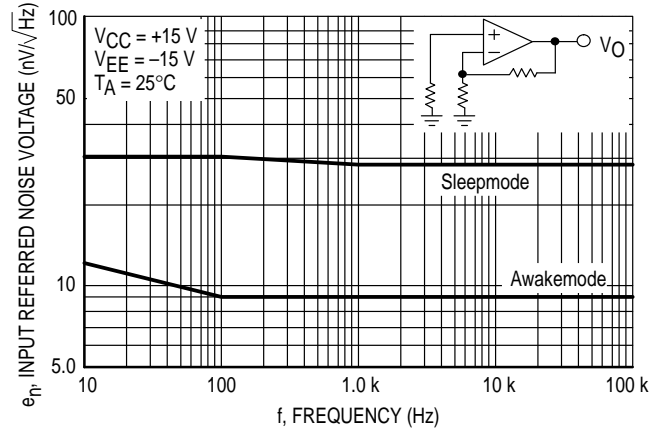**Figure 29. Channel Separation versus Frequency**

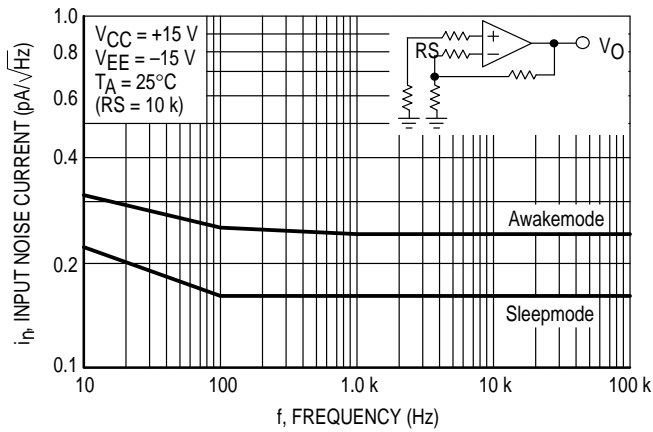**Figure 30. Total Harmonic Distortion versus Frequency**
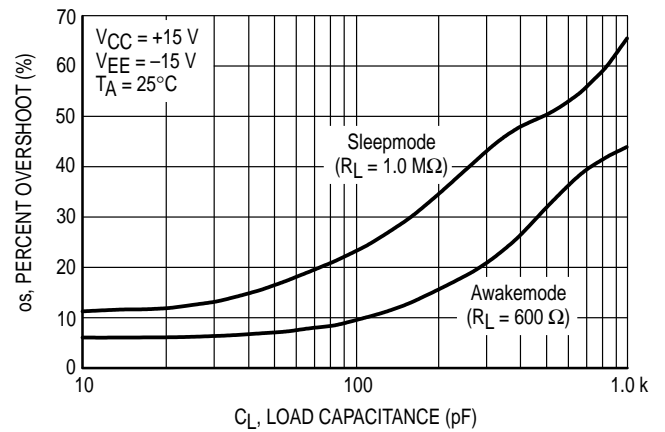
**Figure 31. Awakemode Output Impedance
versus Frequency**

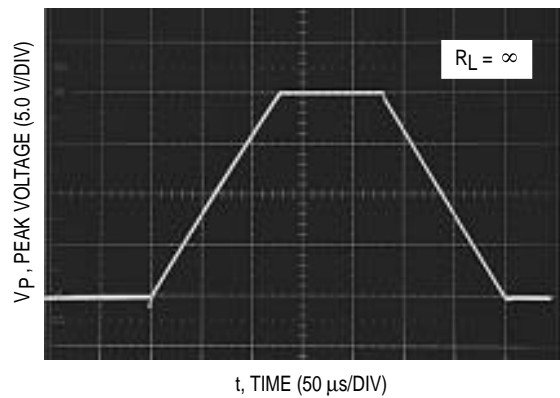**Figure 32. Input Referred Noise Voltage
versus Frequency**

Figure 33. Current Noise versus Frequency

**Figure 34. Percent Overshoot
versus Load Capacitance**

**Figure 35. Sleepmode Large Signal
Transient Response**

**Figure 36. Awakemode Large Signal
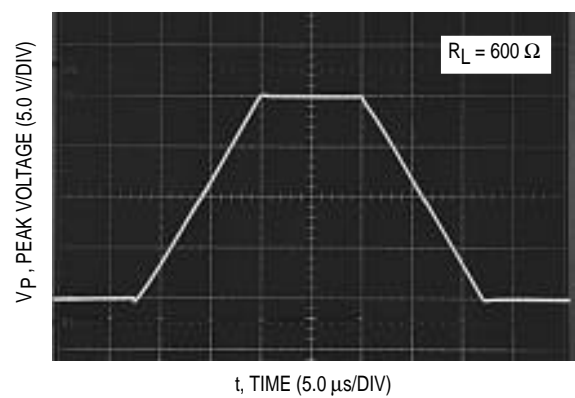Transient Response**
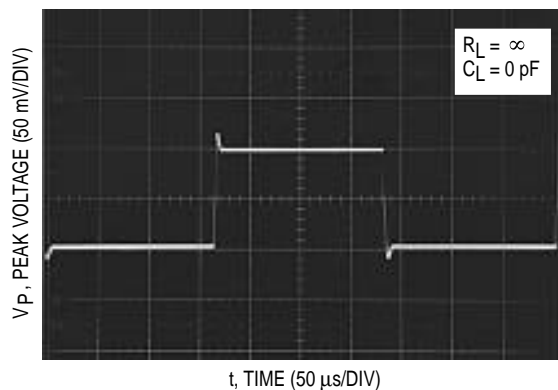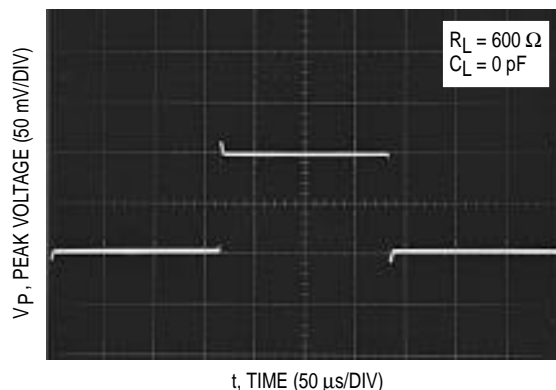
**Figure 37. Sleepmode Small Signal
Transient Response**

$R_L = \infty$
$C_L = 0\ pF$

V_P, PEAK VOLTAGE (50 mV/DIV)

t, TIME (50 μs/DIV)

**Figure 38. Awakemode Small Signal
Transient Response**

$R_L = 600\ \Omega$
$C_L = 0\ pF$

V_P, PEAK VOLTAGE (50 mV/DIV)

t, TIME (50 μs/DIV)

CIRCUIT INFORMATION

The MC33102 was designed primarily for applications where high performance (which requires higher current drain) is required only part of the time. The two–state feature of this op amp enables it to conserve power during idle times, yet be powered up and ready for an input signal. Possible applications include laptop computers, automotive, cordless phones, baby monitors, and battery operated test equipment. Although most applications will require low power consumption, this device can be used in any application where better efficiency and higher performance is needed.

The Sleep–Mode™ amplifier has two states; a sleepmode and an awakemode. In the sleepmode state, the amplifier is active and functions as a typical micropower op amp. When a signal is applied to the amplifier causing it to source or sink sufficient current (see Figure 13), the amplifier will automatically switch to the awakemode. See Figures 20 and 21 for transition times with 600 Ω and 10 kΩ loads.

The awakemode uses higher drain current to provide a high slew rate, gain bandwidth, and output current capability. In the awakemode, this amplifier can drive 27 Vpp into a 600 Ω load with $V_S = \pm15$ V.

An internal delay circuit is used to prevent the amplifier from returning to the sleepmode at every zero crossing. This delay circuit also eliminates the crossover distortion commonly found in micropower amplifiers. This amplifier can process frequencies as low as 1.0 Hz without the amplifier returning to sleepmode, depending on the load.

The first stage PNP differential amplifier provides low noise performance in both the sleep and awake modes, and an all NPN output stage provides symmetrical source and sink AC frequency response.

APPLICATIONS INFORMATION

The MC33102 will begin to function at power supply voltages as low as $V_S = \pm1.0$ V at room temperature. (At this voltage, the output voltage swing will be limited to a few hundred millivolts.) The input voltages must range between V_{CC} and V_{EE} supply voltages as shown in the maximum rating table. Specifically, **allowing the input to go more negative than 0.3 V below V_{EE} may cause product damage.** Also, exceeding the input common mode voltage range on either input may cause phase reversal, even if the inputs are between V_{CC} and V_{EE}.

When power is initially applied, the part may start to operate in the awakemode. This is because of the currents generated due to charging of internal capacitors. When this occurs and the sleepmode state is desired, the user will have to wait approximately 1.5 seconds before the device will switch back to the sleepmode. To prevent this from occurring, ramp the power supplies from 1.0 V to full supply. Notice that the device is more prone to switch into the awakemode when V_{EE} is adjusted than with a similar change in V_{CC}.

The amplifier is designed to switch from sleepmode to awakemode whenever the output current exceeds a preset current threshold (I_{TH}) of approximately 160 μA. As a result, the output switching threshold voltage (V_{ST}) is controlled by the output loading resistance (R_L). This loading can be a load resistor, feedback resistors, or both. Then:

$$V_{ST} = (160\ \mu A) \times R_L$$

Large valued load resistors require a large output voltage to switch, but reduce unwanted transitions to the awakemode. For instance, in cases where the amplifier is connected with a large closed loop gain (A_{CL}), the input offset voltage (V_{IO}) is multiplied by the gain at the output and could produce an output voltage exceeding V_{ST} with no input signal applied.

Small values of R_L allow rapid transition to the awakemode because most of the transition time is consumed slewing in the sleepmode until V_{ST} is reached (see Figures 20, 21). The output switching threshold voltage V_{ST} is higher for larger values of R_L, requiring the amplifier to slew longer in the slower sleepmode state before switching to the awakemode.

MC33102

The transition time (t_{tr1}) required to switch from sleep to awake mode is:

$$t_{tr1} = t_D = I_{TH} (R_L/SR_{sleepmode})$$

Where: t_D = Amplifier delay (<1.0 µs)
I_{TH} = Output threshold current for more transition (160 µA)
R_L = Load resistance
$SR_{sleepmode}$ = Sleepmode slew rate (0.16 V/µs)

Although typically 160 µA, I_{TH} varies with supply voltage and temperature. In general, any current loading on the output which causes a current greater than I_{TH} to flow will switch the amplifier into the awakemode. This includes transition currents such as those generated by charging load capacitances. In fact, the maximum capacitance that can be driven while attempting to remain in the sleepmode is approximately 1000 pF.

$$C_{L(max)} = I_{TH}/SR_{sleepmode}$$
$$= 160 \text{ µA}/(0.16 \text{ V/µs})$$
$$= 1000 \text{ pF}$$

Any electrical noise seen at the output of the MC33102 may also cause the device to transition to the awakemode. To minimize this problem, a resistor may be added in series with the output of the device (inserted as close to the device as possible) to isolate the op amp from both parasitic and load capacitance.

The awakemode to sleepmode transition time is controlled by an internal delay circuit, which is necessary to prevent the amplifier from going to sleep during every zero crossing. This time is a function of supply voltage and temperature as shown in Figure 22.

Gain bandwidth product (GBW) in both modes is an important system design consideration when using a sleepmode amplifier. The amplifier has been designed to obtain the maximum GBW in both modes. "Smooth" AC transitions between modes with no noticeable change in the amplitude of the output voltage waveform will occur as long as the closed loop gains (A_{CL}) in both modes are substantially equal at the frequency of operation. For smooth AC transitions:

$$(A_{CLsleepmode}) (BW) < GBW_{sleepmode}$$

Where: $A_{CLsleepmode}$ = Closed loop gain in the sleepmode
BW = The required system bandwidth or operating frequency

TESTING INFORMATION

To determine if the MC33102 is in the awakemode or the sleepmode, the power supply currents (I_D+ and I_D-) must be measured. When the magnitude of **either** power supply current exceeds 400 µA, the device is in the awakemode. When the magnitudes of both supply currents are less than 400 µA, the device is in the sleepmode. Since the total supply current is typically ten times higher in the awakemode than the sleepmode, the two states are easily distinguishable.

The measured value of I_D+ equals the I_D of both devices (for a dual op amp) plus the output source current of device A and the output source current of device B. Similarly, the measured value of I_D- is equal to the I_D- of both devices plus the output sink current of each device. I_{out} is the sum of the currents caused by both the feedback loop and load resistance. The total I_{out} needs to be subtracted from the measured I_D to obtain the correct I_D of the dual op amp.

An accurate way to measure the awakemode I_{out} current on automatic test equipment is to remove the I_{out} current on both Channel A and B. Then measure the I_D values before the device goes back to the sleepmode state. The transition will take typically 1.5 seconds with ±15 V power supplies.

The large signal sleepmode testing in the characterization was accomplished with a 1.0 MΩ load resistor which ensured the device would remain in sleepmode despite large voltage swings.

MC33102

OUTLINE DIMENSIONS

D SUFFIX
PLASTIC PACKAGE
CASE 751–05
ISSUE N
(SO–8)
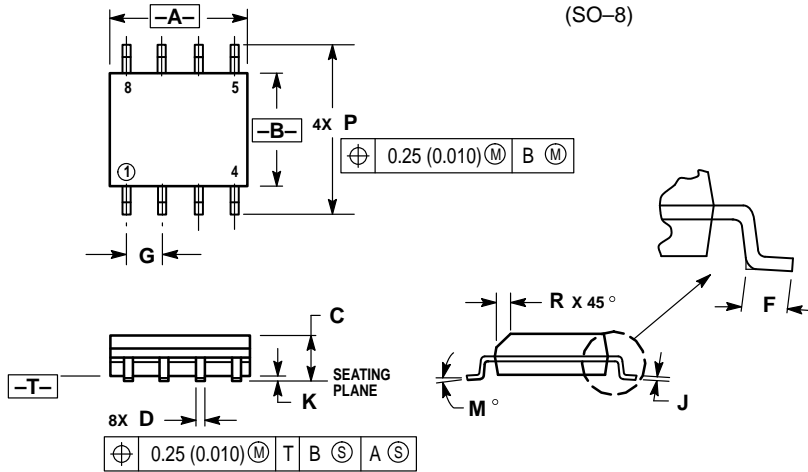
NOTES:
1. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982.
2. CONTROLLING DIMENSION: MILLIMETER.
3. DIMENSIONS A AND B DO NOT INCLUDE MOLD PROTRUSION.
4. MAXIMUM MOLD PROTRUSION 0.15 (0.006) PER SIDE.
5. DIMENSION D DOES NOT INCLUDE DAMBAR PROTRUSION. ALLOWABLE DAMBAR PROTRUSION SHALL BE 0.127 (0.005) TOTAL IN EXCESS OF THE D DIMENSION AT MAXIMUM MATERIAL CONDITION.

DIM	MILLIMETERS		INCHES	
	MIN	MAX	MIN	MAX
A	4.80	5.00	0.189	0.196
B	3.80	4.00	0.150	0.157
C	1.35	1.75	0.054	0.068
D	0.35	0.49	0.014	0.019
F	0.40	1.25	0.016	0.049
G	1.27 BSC		0.050 BSC	
J	0.18	0.25	0.007	0.009
K	0.10	0.25	0.004	0.009
M	0 °	7 °	0 °	7 °
P	5.80	6.20	0.229	0.244
R	0.25	0.50	0.010	0.019

P SUFFIX
PLASTIC PACKAGE
CASE 626–05
ISSUE K
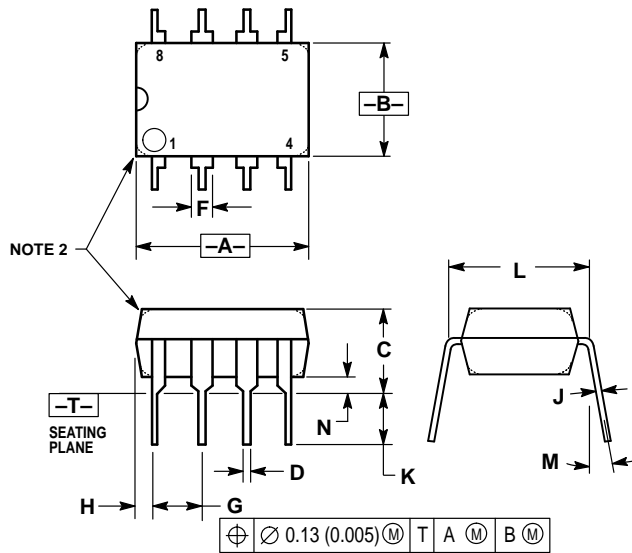
NOTES:
1. DIMENSION L TO CENTER OF LEAD WHEN FORMED PARALLEL.
2. PACKAGE CONTOUR OPTIONAL (ROUND OR SQUARE CORNERS).
3. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982.

DIM	MILLIMETERS		INCHES	
	MIN	MAX	MIN	MAX
A	9.40	10.16	0.370	0.400
B	6.10	6.60	0.240	0.260
C	3.94	4.45	0.155	0.175
D	0.38	0.51	0.015	0.020
F	1.02	1.78	0.040	0.070
G	2.54 BSC		0.100 BSC	
H	0.76	1.27	0.030	0.050
J	0.20	0.30	0.008	0.012
K	2.92	3.43	0.115	0.135
L	7.62 BSC		0.300 BSC	
M	—	10 °	—	10 °
N	0.76	1.01	0.030	0.040

MOTOROLA

◊ CODELINE TO BE PLACED HERE