Dual Precision Retriggerable/Resettable Monostable Multivibrator

The MC14538B is a dual, retriggerable, resettable monostable multivibrator. It may be triggered from either edge of an input pulse, and produces an accurate output pulse over a wide range of widths, the duration and accuracy of which are determined by the external timing components, C_X and R_X .

- Unlimited Rise and Fall Time Allowed on the A Trigger Input
- Pulse Width Range = 10 µs to 10 s
- Latched Trigger Inputs
- Separate Latched Reset Inputs
- 3.0 Vdc to 18 Vdc Operational Limits
- Triggerable from Positive (A Input) or Negative–Going Edge (B–Input)
- Capable of Driving Two Low–power TTL Loads or One Low–power Schottky TTL Load Over the Rated Temperature Range
- Pin-for-pin Compatible with MC14528B and CD4528B (CD4098)
- Use the MC54/74HC4538A for Pulse Widths Less Than 10 μs with Supplies Up to 6 V.

MAXIMUM RATINGS* (Voltages Referenced to VSS)

Symbol	Parameter	Value	Unit
V _{DD}	DC Supply Voltage	– 0.5 to + 18.0	V
V _{in} , V _{out}	Input or Output Voltage (DC or Transient)	– 0.5 to V _{DD} + 0.5	V
l _{in} , l _{out}	Input or Output Current (DC or Transient), per Pin	± 10	mA
PD	Power Dissipation, per Package†	500	mW
T _{stg}	Storage Temperature	– 65 to + 150	°C
ΤL	Lead Temperature (8–Second Soldering)	260	°C

* Maximum Ratings are those values beyond which damage to the device may occur. †Temperature Derating:

Plastic "P and D/DW" Packages: – 7.0 mW/°C From 65°C To 125°C Ceramic "L" Packages: – 12 mW/°C From 100°C To 125°C



ONE-SHOT SELECTION GUIDE



L SUFFIX CERAMIC CASE 620 P SUFFIX PLASTIC CASE 648



DW SUFFIX SOIC CASE 751G

ORDERING INFORMATION

MC14XXXBCP MC14XXXBCL *MC14XXXBDW Plastic Ceramic SOIC

 $T_A = -55^\circ$ to 125°C for all packages.



* Consult factory for possible "D" suffix SOIC Case 751B.



ELECTRICAL CHARACTERISTICS (Voltages Referenced to VSS)

		V _{DD}	- 5	5°C		25°C		125	5°C	
Characteristic	Symbol	Vdc	Min	Max	Min	Тур #	Max	Min	Max	Unit
Output Voltage "0" Level $V_{in} = V_{DD}$ or 0	VOL	5.0 10 15	_ _ _	0.05 0.05 0.05	_ _ _	0 0 0	0.05 0.05 0.05	_ _ _	0.05 0.05 0.05	Vdc
"1" Level V _{in} = 0 or V _{DD}	VOH	5.0 10 15	4.95 9.95 14.95		4.95 9.95 14.95	5.0 10 15		4.95 9.95 14.95		Vdc
$\label{eq:VO} \begin{array}{ll} \mbox{Input Voltage} & \mbox{``0" Level} \\ \mbox{(V}_{O} = 4.5 \mbox{ or } 0.5 \mbox{ Vdc)} \\ \mbox{(V}_{O} = 9.0 \mbox{ or } 1.0 \mbox{ Vdc)} \\ \mbox{(V}_{O} = 13.5 \mbox{ or } 1.5 \mbox{ Vdc)} \end{array}$	VIL	5.0 10 15		1.5 3.0 4.0		2.25 4.50 6.75	1.5 3.0 4.0		1.5 3.0 4.0	Vdc
"1" Level (V _O = 0.5 or 4.5 Vdc) (V _O = 1.0 or 9.0 Vdc) (V _O = 1.5 or 13.5 Vdc)	VIH	5.0 10 15	3.5 7.0 11		3.5 7.0 11	2.75 5.50 8.25		3.5 7.0 11		Vdc
$\begin{array}{ll} \text{Output Drive Current} \\ (V_{OH} = 2.5 \ \text{Vdc}) & \text{Source} \\ (V_{OH} = 4.6 \ \text{Vdc}) \\ (V_{OH} = 9.5 \ \text{Vdc}) \\ (V_{OH} = 13.5 \ \text{Vdc}) \end{array}$	lон	5.0 5.0 10 15	- 3.0 - 0.64 - 1.6 - 4.2		- 2.4 - 0.51 - 1.3 - 3.4	- 4.2 - 0.88 - 2.25 - 8.8		- 1.7 - 0.36 - 0.9 - 2.4	 	mAdc
(V _{OL} = 0.4 Vdc) Sink (V _{OL} = 0.5 Vdc) (V _{OL} = 1.5 Vdc)	lOL	5.0 10 15	0.64 1.6 4.2		0.51 1.3 3.4	0.88 2.25 8.8		0.36 0.9 2.4		mAdc
Input Current, Pin 2 or 14	l _{in}	15	—	±0.05	—	±0.00001	±0.05	—	±0.5	μAdc
Input Current, Other Inputs	l _{in}	15	—	±0.1	—	±0.00001	±0.1	—	±1.0	μAdc
Input Capacitance, Pin 2 or 14	C _{in}	_	—	_	—	25	_	_	—	pF
Input Capacitance, Other Inputs (V _{in} = 0)	C _{in}	-	-	_	-	5.0	7.5	-	—	pF
Quiescent Current (Per Package) $Q = Low, \overline{Q} = High$	IDD	5.0 10 15		5.0 10 20		0.005 0.010 0.015	5.0 10 20		150 300 600	μAdc
Quiescent Current, Active State (Both) (Per Package) $Q = High, \overline{Q} = Low$	IDD	5.0 10 15		2.0 2.0 2.0		0.04 0.08 0.13	0.20 0.45 0.70		2.0 2.0 2.0	mAdc
**Total Supply Current at an external load capacitance (C_L) and at external timing network (R_X , C_X)	ΙŢ	5.0 10		IT = (8.0 x IT = (1.25 where:	: 10 ²) R; x 10 ¹) F Ι _Τ in μΑ (o Cχ in μF, ((Cχf + 4Cχf χCχf + 9Cχf RχCχf + 12C ne monostat CL in pF, Rχ he input freq	+ 2 x 10 ⁻⁴ χf + 3 x 1 ble switchi in k ohms	⁵ CLf 0 ^{–5} CLf ng only),	-	μAdc

#Data labelled "Typ" is not to be used for design purposes but is intended as an indication of the IC's potential performance. **The formulas given are for the typical characteristics only at 25°C.

This device contains protection circuitry to guard against damage due to high static voltages or electric fields. However, precautions must be taken to avoid applications of any voltage higher than maximum rated voltages to this high-impedance circuit. For proper operation, V_{in} and V_{out} should be constrained to the range $V_{SS} \leq (V_{in} \text{ or } V_{out}) \leq V_{DD}$. Unused inputs must always be tied to an appropriate logic voltage level (e.g., either V_{SS} or V_{DD}). Unused outputs must be left open.

SWITCHING CHARACTERISTICS* (CL = 50 pF, TA = 25° C)

		v _{DD}				
Characteristic	Symbol	Vdc	Min	Тур #	Max	Unit
Output Rise Time t _{TLH} = (1.35 ns/pF) C _L + 33 ns t _{TLH} = (0.60 ns/pF) C _L + 20 ns t _{TLH} = (0.40 ns/pF) C _L + 20 ns	tτLH	5.0 10 15		100 50 40	200 100 80	ns
Output Fall Time t _{THL} = (1.35 ns/pF) C _L + 33 ns t _{THL} = (0.60 ns/pF) C _L + 20 ns t _{THL} = (0.40 ns/pF) C _L + 20 ns	ΫΤΗL	5.0 10 15		100 50 40	200 100 80	ns
Propagation Delay Time A or B to Q or \overline{Q} t_{PLH} , t_{PHL} = (0.90 ns/pF) C _L + 255 ns t_{PLH} , t_{PHL} = (0.36 ns/pF) C _L + 132 ns t_{PLH} , t_{PHL} = (0.26 ns/pF) C _L + 87 ns	^t PLH, ^t PHL	5.0 10 15		300 150 100	600 300 220	ns
Reset to Q or Q tpLH, tpHL = (0.90 ns/pF) CL + 205 ns tpLH, tpHL = (0.36 ns/pF) CL + 107 ns tpLH, tpHL = (0.26 ns/pF) CL + 82 ns		5.0 10 15		250 125 95	500 250 190	ns
Input Rise and Fall Times Reset	t _r , t _f	5 10 15			15 5 4	μs
B Input		5 10 15		300 1.2 0.4	1.0 0.1 0.05	ms
A Input		5 10 15		No Limit	1	-
Input Pulse Width A, B, or Reset	t₩H, ^t ₩L	5.0 10 15	170 90 80	85 45 40		ns
Retrigger Time	t _{rr}	5.0 10 15	0 0 0		_ _ _	ns
Output Pulse Width — Q or \overline{Q} Refer to Figures 8 and 9 $C_X = 0.002 \ \mu\text{F}, R_X = 100 \ \text{k}\Omega$	Т	5.0 10 15	198 200 202	210 212 214	230 232 234	μs
$C_X = 0.1 \ \mu\text{F}, \ \text{R}_X = 100 \ \text{k}\Omega$		5.0 10 15	9.3 9.4 9.5	9.86 10 10.14	10.5 10.6 10.7	ms
C_X = 10 µF, R _X = 100 kΩ		5.0 10 15	0.91 0.92 0.93	0.965 0.98 0.99	1.03 1.04 1.06	s
Pulse Width Match between circuits in the same package. CX = 0.1 μ F, RX = 100 k Ω	100 [(T ₁ – T ₂)/T ₁]	5.0 10 15	_ _ _	± 1.0 ± 1.0 ± 1.0	$\pm 5.0 \\ \pm 5.0 \\ \pm 5.0 \\ \pm 5.0$	%

* The formulas given are for the typical characteristics only at 25°C.

#Data labelled "Typ" is not to be used for design purposes but is intended as an indication of the IC's potential performance.

OPERATING CONDITIONS

External Timing Resistance	Rχ	_	5.0	—		kΩ
External Timing Capacitance	CX	—	0	—	No Limit†	μF

* The maximum usable resistance R_X is a function of the leakage of the capacitor C_X , leakage of the MC14538B, and leakage due to board layout and surface resistance. Susceptibility to externally induced noise signals may occur for $R_X > 1 M\Omega$.

+If C_X > 15 μ F, use discharge protection diode per Fig. 11.



Figure 2. Power Dissipation Test Circuit and Waveforms



Figure 3. Switching Test Circuit



Figure 4. Switching Test Waveforms



Figure 5. Typical Normalized Distribution of Units for Output Pulse Width



Figure 7. Typical Total Supply Current versus Output Duty Cycle



Figure 6. Typical Pulse Width Variation as a Function of Supply Voltage V_{DD}

FUNCTION TABLE

	Inputs	Outputs		
Reset	Α	В	Q	Q
H	_/	H	л	디디
H	L	∼	Л	
H	ノ へ	L		iggered
H	H	上		iggered
H	L, H, へ	H		iggered
H	L	L, H, <i>-/</i>		iggered
L	X	X	L	H
~	X	X	Not Tr	iggered





TRIGGER OPERATION

The block diagram of the MC14538B is shown in Figure 1, with circuit operation following.

As shown in Figure 1 and 10, before an input trigger occurs, the monostable is in the quiescent state with the Q output low, and the timing capacitor C_X completely charged to V_{DD}. When the trigger input A goes from V_{SS} to V_{DD} (while inputs B and Reset are held to V_{DD}) a valid trigger is recognized, which turns on comparator C1 and N–channel transistor N1 \odot . At the same time the output latch is set. With transistor N1 on, the capacitor C_X rapidly discharges toward V_{SS} until V_{ref1} is reached. At this point the output of comparator C1 then turns off while at the same time

comparator C2 turns on. With transistor N1 off, the capacitor C χ begins to charge through the timing resistor, R χ , toward V_{DD}. When the voltage across C χ equals V_{ref 2}, comparator C2 changes state, causing the output latch to reset (Q goes low) while at the same time disabling comparator C2 @. This ends at the timing cycle with the monostable in the quiescent state, waiting for the next trigger.

In the quiescent state, C_X is fully charged to V_{DD} causing the current through resistor R_X to be zero. Both comparators are "off" with total device current due only to reverse junction leakages. An added feature of the MC14538B is that the output latch is set via the input trigger without regard to the capacitor voltage. Thus, propagation delay from trigger to Q is independent of the value of C_X , R_X , or the duty cycle of the input waveform.

RETRIGGER OPERATION

The MC14538B is retriggered if a valid trigger occurs ③ followed by another valid trigger ④ before the Q output has returned to the quiescent (zero) state. Any retrigger, after the timing node voltage at pin 2 or 14 has begun to rise from V_{ref 1}, but has not yet reached V_{ref 2}, will cause an increase in output pulse width T. When a valid retrigger is initiated ④, the voltage at C_X/R_X will again drop to V_{ref 1} before progressing along the RC charging curve toward V_{DD}. The Q output will remain high until time T, after the last valid retrigger.

RESET OPERATION

The MC14538B may be reset during the generation of the output pulse. In the reset mode of operation, an input pulse on Reset sets the reset latch and causes the capacitor to be fast charged to V_{DD} by turning on transistor P1 ⁽⁶⁾. When the voltage on the capacitor reaches V_{ref 2}, the reset latch will clear, and will then be ready to accept another pulse. It the Reset input is held low, any trigger inputs that occur will be inhibited and the Q and \overline{Q} outputs of the output latch will not



Figure 11. Use of a Diode to Limit Power Down Current Surge

change. Since the Q output is reset when an input low level is detected on the Reset input, the output pulse T can be made significantly shorter than the minimum pulse width specification.

POWER-DOWN CONSIDERATIONS

Large capacitance values can cause problems due to the large amount of energy stored. When a system containing the MC14538B is powered down, the capacitor voltage may discharge from V_{DD} through the standard protection diodes at pin 2 or 14. Current through the protection diodes should be limited to 10 mA and therefore the discharge time of the V_{DD} supply must not be faster than (V_{DD}). (C)/(10 mA). For example, if V_{DD} = 10 V and C_X = 10 μ F, the V_{DD} supply should discharge no faster than (10 V) x (10 μ F)/(10 mA) = 10 ms. This is normally not a problem since power supplies are heavily filtered and cannot discharge at this rate.

When a more rapid decrease of V_{DD} to zero volts occurs, the MC14538B can sustain damage. To avoid this possibility use an external clamping diode, D_X , connected as shown in Fig. 11.

PIN ASSIGNMENT

∨ss [1 •	16] ∨ _{DD}
Cχ/RχΑ [2	15] Vss
RESET A	3	14] C _X /R _X Β
A _A [4	13	RESET B
B _A	5	12] A _B
Q _A [6	11] B _B
Q _A [7	10] Q _B
vss D	8	9]

TYPICAL APPLICATIONS





Figure 13. Non–Retriggerable Monostables Circuitry





Figure 14. Connection of Unused Sections

OUTLINE DIMENSIONS



OUTLINE DIMENSIONS



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