

PLL Frequency Synthesizer for Electronic Tuning



Overview

The LC72133M and LC72133V are a phase-locked loop frequency synthesizer LSI circuits for use in radio tuners. It supports low-voltage (2.7 to 3.6 V) operation and can implement high-performance AM/FM tuners easily.

Functions

- High speed programmable dividers
 - FMIN: 10 to 120 MHzpulse swallower (built-in divide-by-two prescaler), $V_{DD} \ge 2.7 \text{ V}$ 10 to 130 MHzpulse swallower (built-in divide-by-two prescaler), $V_{DD} \ge 3.0 \text{ V}$
 - AMIN: 2 to 40 MHzpulse swallower 0.5 to 10 MHzdirect division
- · IF counter
 - IFIN: 0.4 to 12 MHzAM/FM IF counter
- · Reference frequencies
 - Twelve selectable frequencies
 (4.5 or 7.2 MHz crystal)
 1,3,5,9,10,3.125,6.25,12.5,15,25,50 and 100 kHz

- · Phase comparator
 - Dead zone control
 - Unlock detection circuit
 - Deadlock clear circuit
- Built-in MOS transistor for forming an active low-pass filter
- I/O ports
 - Dedicated output ports: 4
 - Input or output ports: 2
 - Support clock time base output
- · Serial data I/O
- Support CCB format communication with the system controller. (Compatible with LC72131)
- · Operating ranges
 - Supply voltage......2.7 to 3.6 V
 - Operating temperature.....-20 to +70°C
- Package

MFP20

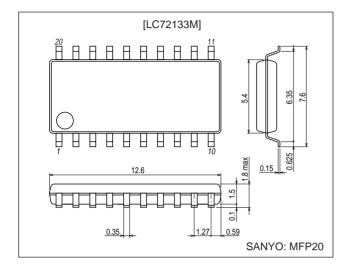
SSOP20

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Package Dimensions

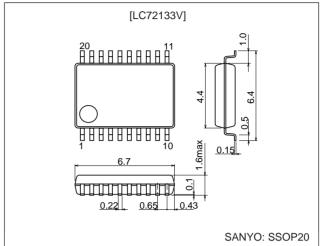
unit: mm

3036B-MFP20



unit: mm

3179A-SSOP20



Specifications

Absolute Maximum Ratings at $Ta = 25^{\circ}C$, $V_{SS} = 0$ V

Parameter	Symbol	Pins	Ratings	Unit
Supply voltage	V _{DD} max	V_{DD}	-0.3 to +5.5	V
	V _{IN} 1 max	CE, CL, DI, AIN	-0.3 to +5.5	V
Maximum input voltage	V _{IN} 2 max	XIN, FMIN, AMIN, IFIN	-0.3 to V _{DD} + 0.3	V
	V _{IN} 3 max	<u>101</u> , <u>102</u>	-0.3 to +15	V
	V _O 1 max	DO	-0.3 to +5.5	V
Maximum output voltage	V _O 2 max	XOUT, PD	-0.3 to V _{DD} + 0.3	V
	V _O 3 max	BO1 to BO4, IO1, IO2, AOUT	-0.3 to +15	V
	I _O 1 max	BO1	0 to 3.0	mA
Maximum output current	I _O 2 max	AOUT, DO	0 to 6.0	mA
	I _O 3 max	BO2 to BO4, IO1, IO2	0 to 6.0	mA
Allowable newer discination	Dd may	Ta ≤ 70°C: LC72133M	180	mW
Allowable power dissipation	Pd max	Ta ≤ 70°C: LC72133V	160	mW
Operating temperature	Topr		-20 to +70	°C
Storage temperature	Tstg		-40 to +125	°C

Allowable Operating Ranges at Ta = –20 to +70 $^{\circ}C,\,V_{SS}$ = 0 V

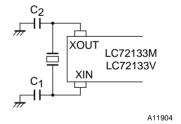
Parameter	Symbol	Pins	Conditions	min	typ	max	Unit
Supply voltage	V _{DD}	V _{DD}		2.7		3.6	V
longs high lovel veltage	V _{IH} 1	CE, CL, DI		0.7 V _{DD}		5.5	V
Input high-level voltage	V _{IH} 2	ĪO1, ĪO2		0.7 V _{DD}		13	V
Input low-level voltage	V _{IL}	CE, CL, DI, IO1, IO2		0		0.3 V _{DD}	V
	V _O 1	DO		0		5.5	V
Output voltage	V _O 2	BO1 to BO4, IO1, IO2, AOUT		0		13	V
	f _{IN} 1	XIN	V _{IN} 1	1		8	MHz
	f _{IN} 2-1	FMIN	V _{IN} 2-1	10		90	MHz
	f _{IN} 2-2	FMIN	V _{IN} 2-2	10		120	MHz
Input frequency	f _{IN} 2-3	FMIN	$V_{IN}2-1, V_{DD} \ge 3.0 \text{ V}$	10		130	MHz
	f _{IN} 3	AMIN	V _{IN} 3, SNS = 1	2		40	MHz
	f _{IN} 4	AMIN	V _{IN} 4, SNS = 0	0.5		10	MHz
	f _{IN} 5	IFIN	V _{IN} 5	0.4		12	MHz
	V _{IN} 1	XIN	f _{IN} 1	400		900	mVrms
	V _{IN} 2-1	FMIN	f _{IN} 2-1, f _{IN} 2-3	70		900	mVrms
	V _{IN} 2-2	FMIN	f _{IN} 2-2	100		900	mVrms
Input amplitude	V _{IN} 3	AMIN	f _{IN} 3, SNS = 1	70		900	mVrms
	V _{IN} 4	AMIN	f _{IN} 4, SNS = 0	70		900	mVrms
	V _{IN} 5-1	IFIN	f _{IN} 5, IFS = 1	70		900	mVrms
	V _{IN} 5-2	IFIN	f _{IN} 6, IFS = 0	100		900	mVrms
Supported crystals	Xtal	XIN, XOUT	*	4.0		8.0	MHz

Note: * Recommended crystal oscillator CI values: $\text{CI} \leq 120\Omega \text{ (For a 4.5 MHz crystal)}$

 $CI \le 70\Omega$ (For a 7.2 MHz crystal)

<Sample Oscillator Circuit> Crystal oscillator: HC-49/U (manufactured by Kinseki, Ltd.), $CL = 12 \ pF$

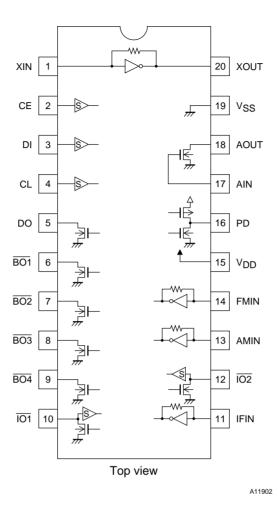
The circuit constants for the crystal oscillator circuit depend on the crystal used, the printed circuit board pattern, and other items. Therefore we recommend consulting with the manufacturer of the crystal for evaluation and reliability.



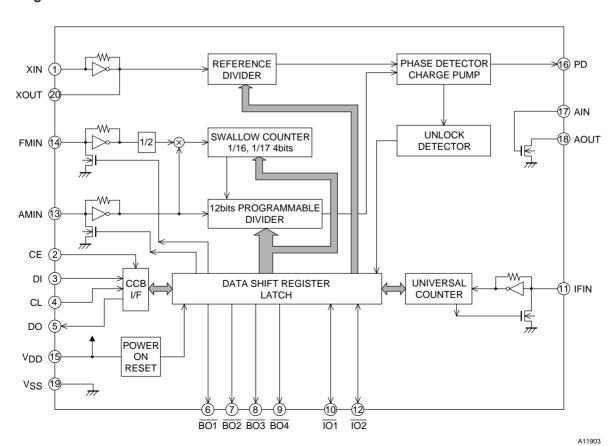
Electrical Characteristics for the Allowable Operating Ranges at Ta = -20 to $+70^{\circ}$ C, V_{SS} = 0 V

Parameter	Symbol	Pins	Conditions	min	typ	max	Unit
	Rf1	XIN			1.0		ΜΩ
	Rf2	FMIN			500		kΩ
Built-in feedback resistance	Rf3	AMIN			500		kΩ
	Rf4	IFIN			250		kΩ
	Rpd1	FMIN			200		kΩ
Built-in pull-down resistor	Rpd2	AMIN			200		kΩ
Hysteresis	V _{HIS}	CE, CL, DI, IO1, IO2			0.1 V _{DD}		V
Output high level voltage	V _{OH} 1	PD	I _O = -1 mA	V _{DD} – 1.0			V
	V _{OL} 1	PD	I _O = 1 mA			1.0	V
		504	I _O = 0.5 mA			0.6	V
	V _{OL} 2	BO1	I _O = 1 mA			1.2	V
		20	I _O = 1 mA			0.25	V
Output low level voltage	V _{OL} 3	DO	I _O = 3 mA			0.75	V
		<u> </u>	I _O = 1 mA			0.25	V
	V _{OL} 4	BO2 to BO4, IO1, IO2	I _O = 5 mA			1.25	V
	V _{OL} 5	AOUT	I _O = 1 mA, AIN = 1.3 V			0.5	V
	I _{IH} 1	CE, CL, DI	V _I = 5.5 V			5.0	μA
	I _{IH} 2	101, 102	V _I = 13 V			5.0	μA
Input high level current	I _{IH} 3	XIN	$V_I = V_{DD}$	1.3		8	μA
	I _{IH} 4	FMIN, AMIN	$V_I = V_{DD}$	2.7		15	μA
	I _{IH} 5	IFIN	$V_I = V_{DD}$	5.4		30	μA
	I _{IH} 6	AIN	V _I = 5.5 V			200	nA
	I _{IL} 1	CE, CL, DI	V _I = 0 V			5.0	μA
	I _{IL} 2	IO1, IO2	V _I = 0 V			5.0	μA
Innut love lovel overent	I _{IL} 3	XIN	V _I = 0 V	1.3		8	μA
Input low level current	I _{IL} 4	FMIN, AMIN	V _I = 0 V	2.7		15	μA
	I _{IL} 5	IFIN	V _I = 0 V	5.4		30	μA
	I _{IL} 6	AIN	V _I = 0 V			200	nA
Output off leakage current	I _{OFF} 1	BO1 to BO4, AOUT, IO1, IO2	V _O = 13 V			5.0	μA
-	I _{OFF} 2	DO	V _O = 5.5 V			5.0	μΑ
High level three-state off leakage current	I _{OFFH}	PD	$V_O = V_{DD}$		0.01	200	nA
Low level three-state off leakage current	I _{OFFL}	PD	V _O = 0 V		0.01	200	nA
Input capacitance	C _{IN}	FMIN			6		pF
	I _{DD} 1	V _{DD}	Xtal = 7.2 MHz, $f_{IN}2 = 130 \text{ MHz},$ $V_{IN}2 = 70 \text{ mVrms}$		2	5	mA
Current drain	I _{DD} 2	V _{DD}	PLL block stopped (PLL INHIBIT), Xtal oscillator operating (Xtal = 7.2 MHz)		0.3		mA
	I _{DD} 3	V _{DD}	PLL block stopped Xtal oscillator stopped			30	μА

Pin Assignment



Block Diagram



No. 5427-5/23

Pin Functions

Symbol	Pin No.	Туре	Functions	Circuit configuration
XIN XOUT	1 20	Xtal OSC	Crystal resonator connection (4.5/7.2 MHz)	A11905
FMIN	14	Local oscillator signal input	FMIN is selected when the serial data input DVS bit is set to 1. The input frequency range is from 10 to 130 MHz. The input signal passes through the internal divide-bytwo prescaler and is input to the swallow counter. The divisor can be in the range 272 to 65535. However, since the signal has passed through the divide-by-two prescaler, the actual divisor is twice the set value. Operating FMIN input frequency The first of the set value	A11906
AMIN	13	Local oscillator signal input	AMIN is selected when the serial data input DVS bit is set to 0. When the serial data input SNS bit is set to 1: — The input frequency range is 2 to 40 MHz. — The signal is directly input to the swallow counter. — The divisor can be in the range 272 to 65535, and the divisor used will be the value set. When the serial data input SNS bit is set to 0: — The input frequency range is 0.5 to 10 MHz. — The signal is directly input to a 12-bit programmable divider. — The divisor can be in the range 4 to 4095, and the divisor used will be the value set.	A11907
CE	2	Chip enable	Set this pin high when inputting (DI) or outputting (DO) serial data.	A11908
CL	4	Clock	Used as the synchronization clock when inputting (DI) or outputting (DO) serial data.	☐————————————————————————————————————
DI	3	Data input	Inputs serial data transferred from the controller to the LC72133.	A11910
DO	5	Data output	Outputs serial data transferred from the LC72133 to the controller. The content of the output data is determined by the serial data DOC0 to DOC2.	A11911
V _{DD}	15	Power supply	The LC72133 power supply pin (V _{DD} = 2.7 to 3.6 V) The power on reset circuit operates when power is first applied.	

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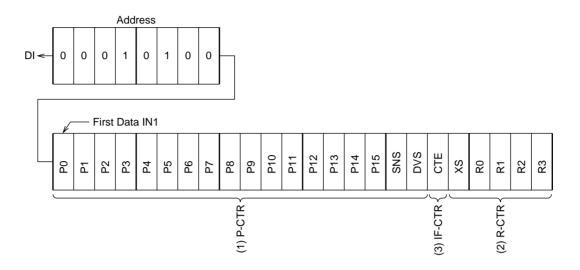
Symbol	Pin No.	Туре	Functions	Circuit configuration
V _{SS}	19	Ground	The LC72133 ground	-
BO1 BO2 BO3 BO4	6 7 8 9	Output port	Dedicated output pins The output states are determined by BO1 to BO4 bits in the serial data. Data: 0 = open, 1 = low A time base signal (8 Hz) can be output from the BO1 pin. (When the serial data TBC bit is set to 1.) Care is required when using the BO1 pin, since it has a higher on impedance than the other output ports (pins BO2 to BO4). The data = 0 (open) state is selected after the power-on reset.	A11912
101 102	10 12	I/O port	I/O dual-use pins Interview of the interview of the serial data. Interview of the input port of the input port over the DO pin. Input state: low = 0 data value high = 1 data value When specified for use as output ports: The output state: low = 0 data value high = 1 data value When specified for use as output ports: The output states are determined by the IO1 and IO2 bits in the serial data. Data: 0 = open, 1 = low These pins function as input pins following a power on reset.	A11913
PD	16	Charge pump output	PLL charge pump output When the frequency generated by dividing the local oscillator frequency by N is higher than the reference frequency, a high level is output from the PD pin. Similarly, when that frequency is lower, a low level is output. The PD pin goes to the high impedance state when the frequencies match.	A11914
AIN AOUT	17 18	LPF amplifier transistor	The n-channel MOS transistor used for the PLL active low-pass filter.	A11915
IFIN	11	IF counter	Accepts an input in the frequency range 0.4 to 12 MHz. The input signal is directly transmitted to the IF counter. The result is output starting the MSB of the IF counter using the DO pin. Four measurement periods are supported: 4, 8, 32, and 64 ms.	A11916

Serial Data I/O Methods

The LC72133 inputs and outputs data using the Sanyo CCB (computer control bus) audio LSI serial bus format. This LSI adopts an 8-bit address format CCB.

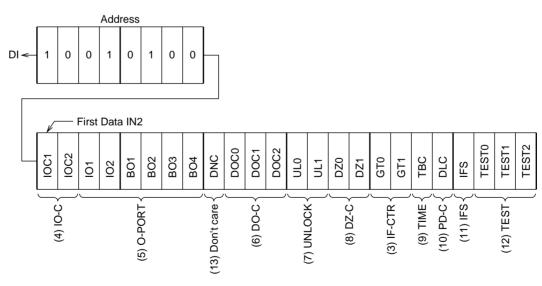
	Address											
	I/O mode	В0	B1	B2	В3	A0	A1	A2	АЗ	Function		
1	IN1 (82)	0	0	0	1	0	1	0	0	Control data input mode (serial data input) 24 data bits are input. See the "DI Control Data (serial data input) Structure" item for details on the meaning of the input data.		
2	IN2 (92)	1	0	0	1	0	1	0	0	Control data input mode (serial data input)		
3	OUT (A2)	0	1	0	1	0	1	0	0	Data output mode (serial data output) The number of bits output is equal to the number of clock cycles. See the "DO Output Data (Serial Data Output) Structure" item for details on the meaning of the output data.		
	CE	B1)	B2 Norma	B:	3 \ \ .		A1	A2		A3 W First Data IN1/2 First Data OUT First Data OUT A11917		

- 1. DI Control Data (Serial Data Input) Structure
 - IN1 Mode



A11918

• IN2 Mode



A11919

2. DI Control Data Functions

No.	Control block/data				Related data				
	Programmable divider data	Data that	Data that sets the programmable divider.						
	P0 to P15	A binary value in which P15 is the MSB. The LSB changes depending on							
		DVS and	SNS. (*: 0	don't care)					
		DVS	SNS	LSB	Divisor	setting (N)	Actual divisor		
		1	*	P0	P0 272 to 65535 T		Twice the value of the setting		
		0	1	P0	272 t	o 65535	The value of the setting		
		0	0	P4	4	to 4095	The value of the setting		
(1)		Note: P0	to P3 are	ignored wl	hen P4 is t	he LSB.		·	
	DVS, SNS		0			, ,	grammable divider, switches		
		the input	frequency	range. (*:	don't care	2)			
		DVS	SNS	Input	pin	ļ	nput frequency range		
		1	*	FMI	N		10 to 130 MHz		
		0	1	AMI	N		2 to 40 MHz		
		0	0	AMI			0.5 to 10 MHz		
						tem for more in	nformation.		
	Reference divider data	Reference	e frequenc	y (fref) sel	ection data	Э.			
	R0 to R3	R3	R2	R1	R0	Re	ference frequency (kHz)		
		0	0	0	0		100		
		0	0	0	1		50		
		0	0	1	0		25 25		
		0	1	0	0		12.5		
		0	1	0	1		6.25		
		0	1	1	1 0 1		3.125 3.125		
		1	0	0			10		
		1	0	0			9		
(2)		1	0	1			5		
		1	0	0	0		3		
			1	0	1		3 15		
		1	1	1	0	PLL I	NHIBIT + Xtal OSC STOP		
		1	1	1	1		PLL INHIBIT		
		Note: PL	LINHIBIT						
							r block are stopped, the FMIN,		
			n, and IFIN to the hig	•		Duii-down State	e (ground), and the charge pump		
	XS	Crystal re	_	-					
		XS = 0: 4	.5 MHz						
		XS = 1: 7		novio colo	atad aftar	the newer on	rooot		
	IF counter control data	• IF counte				the power-on	reset.		
	CTE		Counter st		uala				
		CTE = 0:	Counter re	eset					
	GT0, GT1	Determine	es the IF c	ounter me	asurement	t period.		IFS	
		GT1	GT0	Mea	surement t	time (ms)	Wait time (ms)		
(3)		0	0		Measurement time (ms) Wait time (ms) 4 3 to 4				
		0	1		8				
		1	0	8 3 to 4 32 7 to 8					
		1	1						
		1 1 64 7 to 8 Note: See the "IF Counter" item for more information.							
1.0	I/O port specification data					tional pins IO1	and IO2.		
(4)	IOC1, IOC2	Data: 0 =	input mod	e, 1 = outp	out mode				
	Output port data				ut from the	BO1 to BO4,	IO1 and IO2 output ports	IOC1	
(5)	BO1 to BO4, IO1, IO2		open, 1 =					IOC1	
		The data	= 0 (open)	state is se	elected afte	er the power-o	n reset.		

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No.	Control block/data				Functions		Related data		
	DO pin control data	Data that	determine	s the DO	pin output				
	DOC0, DOC1, DOC2	DOC2	DOC1	DOC0		O pin state			
		0	0	0	Open	70 pin state			
			0	1	Low when the unlock s	tate is detected			
		0	1	0	end-UC*1				
		0	1	1	Open				
		1 1	0	0	Open The IO1 pin state*2				
			1	0	The IO2 pin state*2				
		1	1	1	Open				
		The open	state is se	elected aft	er the power-on reset.				
		· ·			IF counter measurement	completion			
(6)))	_	UL0, UL1, CTE,		
(0)		DO p	in	_X	((<u>/</u>	IOC1, IOC2		
			1	Counter		unter ③ CE: high			
		2.							
		hi Al w	gh) will be so, the D0 ill output tl	e open, reg O pin durir he conten	gardless of the state of thing a data output period (ats of the internal DO serial	id (an IN1 or IN2 mode period with CE e DO control data (DOC0 to DOC2). In OUT mode period with CE high) al data in synchronization with the D control data (DOC0 to DOC2).			
	Unlock detection data	1		. ,	detection width for checki	0			
	UL0, UL1	A phase e	error in exc	cess of the	e specified detection width	h is seen as an unlocked state.			
		UL1	UL0	Ø	E detection width	Detector output			
-		0	0	Stopped		Open	DOC0,		
(7)		0	1	0		øE is output directly	DOC1, DOC2		
		1	0	±0.55 µs	5	øE is extended by 1 to 2 ms	5002		
		1	1	±1.11	00	øE is extended by 1 to 2 ms			
		1	ne uniock comes zer		ne DO pin goes low and t	he UL bit in the serial data			
	Phase comparator	Controls t	he phase	comparate	or dead zone.				
	control data DZ0, DZ1	DZ1	DZ0		Dead zo	ne mode			
	,	0	0	DZA					
(0)		0	1	DZB					
(8)		1	0	DZC					
		1	1	DZD					
				•					
	0				B < DZC < DZD				
(9)	Clock time base TBC	Setting TE from the E	ne base signal to be output	BO1					
	Charge pump control data	Forcibly c	ontrols the	e charge p	oump output.				
	DLC	D	ımp output						
(10)			1	Forced I	OW				
			Note: If deadlock occurs due to the VCO control voltage (Vtune) going to zero and the VCO						
					adlock can be cleared by a cook of V _{CC} . (This is the deadle	forcing the charge pump output to			
		l low	and setti	ng viune i	VCC. (Triis is the deadi	ook oleaning ollean.)			

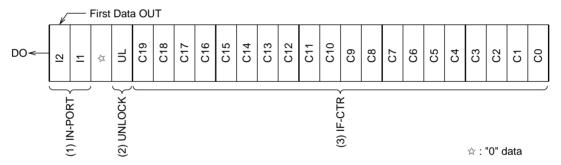
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No.	Control block/data	Functions	Related data					
(11)	IF counter control data IFS	Note that if this value is set to zero the system enters input sensitivity degradation mode, and the sensitivity is reduced to 10 to 30 mV rms. See the "IF Counter Operation" item for details.						
(12)	LSI test data TEST 0 to TEST 2	LSI test data TEST0 TEST1 TEST2 These values must all be set to 0. TEST2 These test data are set to 0 automatically after the power-on reset.						
(4.0)	5110	, ,						
(13)	DNC	Don't care. This data must be set to 0.	Don't care. This data must be set to 0.					

3. DO Output Data (Serial Data Output)

• OUT Mode



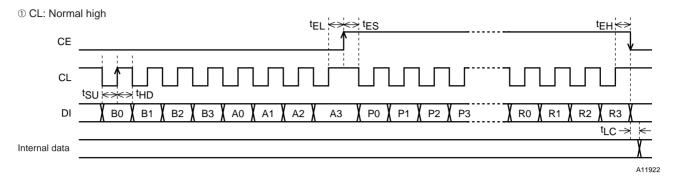


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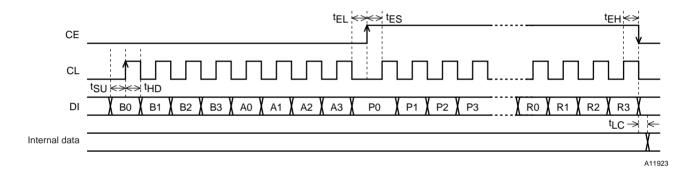
4. DO Output Data

No.	Control block/data	Functions	Related data
(1)	I/O port data I2, I1	Latched from the pin states of the lol and lol l/O ports. These values follow the pin states regardless of the input or output setting. Data is latched at the point where the circuit enters data output mode (OUT mode). I1 ← lol pin state High: 1 I2 ← lol pin state Low: 0	IOC1, IOC2
(2)	PLL unlock data UL	Latched from the state of the unlock detection circuit. UL ← 0: Unlocked UL ← 1: Locked or detection stopped mode	ULO, UL1
(3)	IF counter binary data C19 to C0	Latched from the value of the IF counter (20-bit binary counter). C19 ← MSB of the binary counter C0 ← LSB of the binary counter	CTE, GT0, GT1

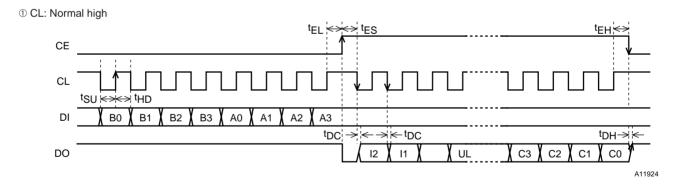
5. Serial Data Input (IN1/IN2) t_{SU} , t_{HD} , t_{EL} , t_{ES} , $t_{EH} \ge 0.75 \,\mu s$, $t_{LC} < 0.75 \,\mu s$

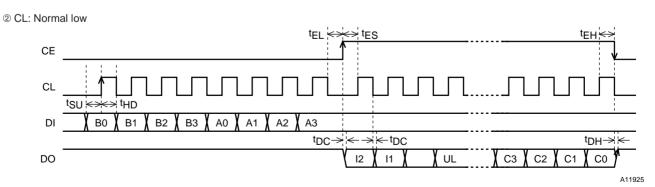


@ CL: Normal low



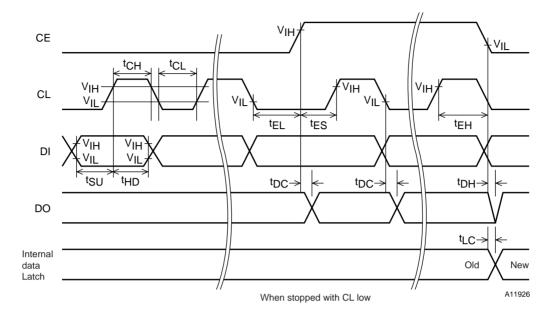
6. Serial Data Output (OUT) t_{SU} , t_{HD} , t_{EL} , t_{ES} , $t_{EH} \ge 0.75~\mu s$, t_{DC} , $t_{DH} < 0.35~\mu s$

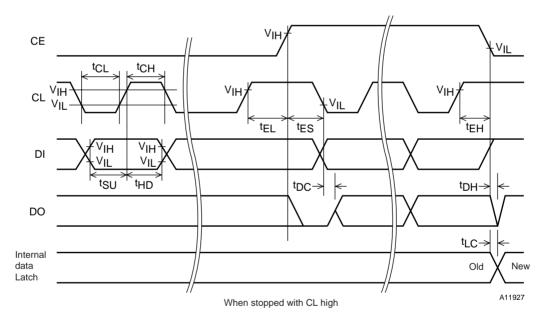




Note: Since the DO pin is an n-channel open-drain circuit, the time for the data to change (t_{DC} and t_{DH}) will differ depending on the value of the pull-up resistor and printed circuit, board capacitance.

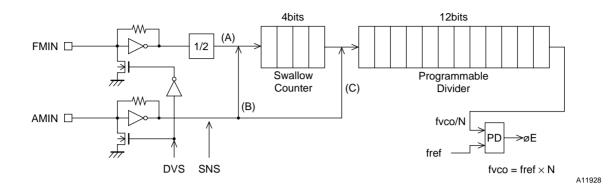
7. Serial Data Timing





Parameter	Symbol	Pins	Conditions	min	typ	max	Unit
Data setup time	t _{SU}	DI, CL		0.75			μs
Data hold time	t _{HD}	DI, CL		0.75			μs
Clock low-level time	t _{CL}	CL		0.75			μs
Clock high-level time	t _{CH}	CL		0.75			μs
CE wait time	t _{EL}	CE, CL		0.75			μs
CE setup time	t _{ES}	CE, CL		0.75			μs
CE hold time	t _{EH}	CE, CL		0.75			μs
Data latch change time	t _{LC}					0.75	μs
Data output time	t _{DC}	DO, CL	Differs depending on the value of the pull-up resistor			0.35	
Data output time	t _{DH}	DO, CE	and the printed circuit board capacitances.			0.33	μs

Programmable Divider Structure



	DVS	SNS	Input pin	Set divisor	Actual divisor: N	Input frequency range (MHz)
А	1	*	FMIN	272 to 65535	Twice the set value	10 to 130
В	0	1	AMIN	272 to 65535	The set value	2 to 40
С	0	0	AMIN	4 to 4095	The set value	0.5 to 10

Note: * Don't care.

- 1. Programmable Divider Calculation Examples
 - FM, 50 kHz steps (DVS = 1, SNS = *, FMIN selected)

FM RF = 80.0 MHz (IF = -10.7 MHz)

FM VCO = 69.3 MHz

PLL fref = 25 kHz (R0 to R1 = 1, R2 to R3 = 0)

69.3 MHz (FM VCO) \div 25 kHz (fref) \div 2 (FMIN: divide-by-two prescaler) = 1386 \rightarrow 056A (HEX)

_		<i></i>	_	_			_				_	_	(_								
0	1	0	1	0	1	1	0	1	0	1	0	0	0	0	0	*	1			1	1	0	0
Po	7	P2	Р3	P4	P5	9e	Ь7	P8	P9	P10	P11	P12	P13	P14	P15	SNS	DVS	CTE	XS	RO	R1	R2	R3

A11929

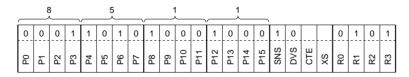
• SW, 5 kHz steps (DVS = 0, SNS = 1, AMIN high speed side selected)

SW RF = 21.75 MHz (IF = +450 kHz)

SW VCO = 22.20 MHz

PLL fref = 5 kHz (R0 = R2 = 0, R1 = R3 = 1)

22.2 MHz (SW VCO) \div 5 kHz (fref) = 4440 \rightarrow 1158 (HEX)



A11930

• MW, 10 kHz steps (DVS = 0, SNS = 0, AMIN low-speed side selected)

MW RF = 1000 kHz (IF = +450 kHz)

MW VCO = 1450 kHz

PLL fref = 10 kHz (R0 to R2 = 0, R3 = 1)

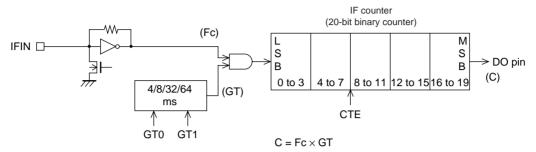
 $1450 \text{ kHz} \text{ (MW VCO)} \div 10 \text{ kHz (fref)} = 145 \rightarrow 091 \text{ (HEX)}$

							_				_				_								
*	*	*	*	1	0	0	0	1	0	0	1	0	0	0	0	0	0			0	0	0	1
P0	7	P2	Р3	P4	PS	P6	Ь7	P8	P3	P10	P11	P12	P13	P14	P15	SNS	DVS	CTE	XS	R0	R1	R2	R3

A11931

IF Counter Structure

The LC72133 IF counter is a 20-bit binary counter. The result, i.e., the counter's MSB, can be read serially from the DO pin.



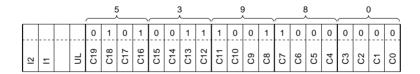
A11932

GT1	GT0	Measurement time								
GII	GIO	Measurement period (GT) (ms)	Wait time (twu) (ms)							
0	0	4	3 to 4							
0	1	8	3 to 4							
1	0	32	7 to 8							
1	1	64	7 to 8							

The IF frequency (Fc) is measured by determining how many pulses were input to an IF counter in a specified measurement period, GT.

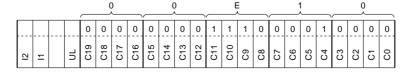
$$Fc = \frac{C}{GT}$$
 (C = Fc × GT) C: Count value (number of pulses)

- 1. IF Counter Frequency Calculation Examples
 - When the measurement period (GT) is 32 ms, the count (C) is 53980 hexadecimal (342400 decimal): IF frequency (Fc) = $342400 \div 32$ ms = 10.7 MHz



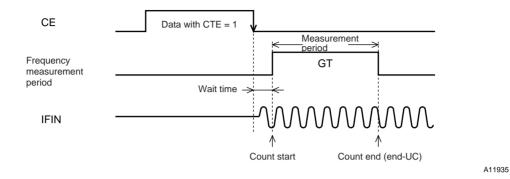
A11933

• When the measurement period (GT) is 8 ms, the count (C) is E10 hexadecimal (3600 decimal): IF frequency (Fc) = $3600 \div 8 \text{ ms} = 450 \text{ kHz}$



A11934

2. IF Counter Operation



Before starting the IF count, the IF counter must be reset in advance by setting CTE in the serial data to 0.

The IF count is started by changing the CTE bit in the serial data from 0 to 1. The serial data is latched by the LC72133 when the CE pin is dropped from high to low. The IF signal must be supplied to the IFIN pin in the period between the point the CE pin goes low and the end of the wait time at the latest. Next, the value of the IF counter at the end of the measurement period must be read out during the period that CTE is 1. This is because the IF counter is reset when CTE is set to 0.

Note: When operating the IF counter, the control microprocessor must first check the state of the IF-IC SD (station detect) signal and only after determining that the SD signal is present turn on IF buffer output and execute an IF count operation. Autosearch techniques that use only the IF counter are not recommended, since it is possible for IF buffer leakage output to cause incorrect stops at points where there is no station.

IFIN minimum input sensitivity standard

f (MHz)

IFS	0.4 ≤ f < 0.5	0.5 ≤ f < 8	8 ≤ f ≤ 12			
1: Normal mode	70 mVrms (0.5 to 5 mVrms)	70 mVrms	70 mVrms (2 to 10 mVrms)			
0: Degradation mode	100 mVrms (10 to 15 mVrms)	100 mVrms	100 mVrms (30 to 50 mVrms)			

Note: Values in parentheses are actual performance values presented as reference data.

Unlock Detection Timing

1. Unlock Detection Determination Timing

Unlocked state detection is performed in the reference frequency (fref) period (interval). Therefore, in principle, unlock determination requires a time longer than the period of the reference frequency. However, immediately after changing the divisor N (frequency) unlock detection must be performed after waiting at least two periods of the reference frequency.

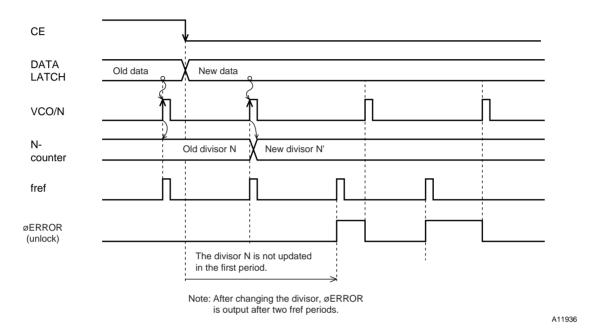


Figure 1 Unlocked State Detection Timing

For example, if fref is 1 kHz, i.e., the period is 1 ms, after changing the divisor N, the system must wait at least 2 ms before checking for the unlocked state.

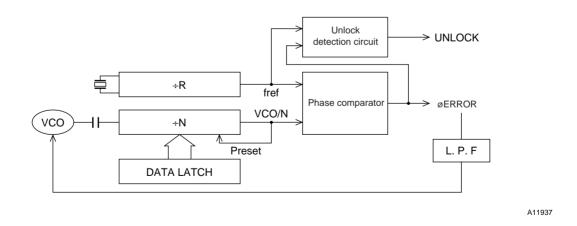
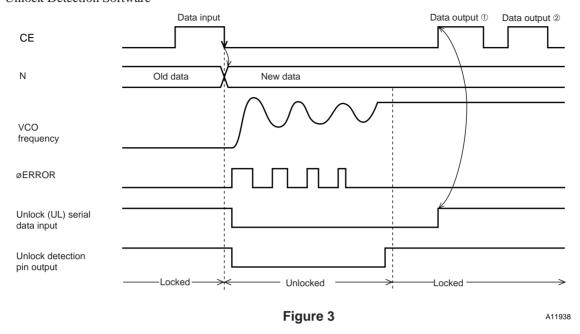


Figure 2 Circuit Structure

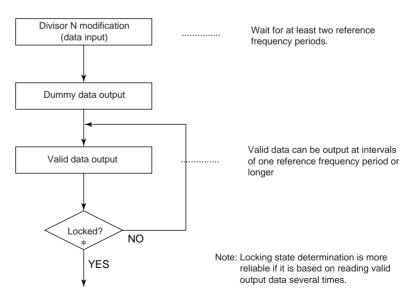
2. Unlock Detection Software



3. Unlocked State Data Output Using Serial Data Output

In the LC72133, once an unlocked state occurs, the unlocked state serial data (UL) will not be reset until a data input (or output) operation is performed. At the data output ① point in Figure 3, although the VCO frequency has stabilized (locked), since no data output has been performed since the divisor N was changed the unlocked state data remains in the unlocked state. As a result, even though the frequency has stabilized (locked), the system remains (from the standpoint of the data) in the unlocked state.

Therefore, the unlocked state data acquired at data output 1, which occurs immediately after the divisor N was changed, should be treated as a dummy data output and ignored. The second data output (data output 2) and following outputs are valid data.



Locked State Determination Flowchart

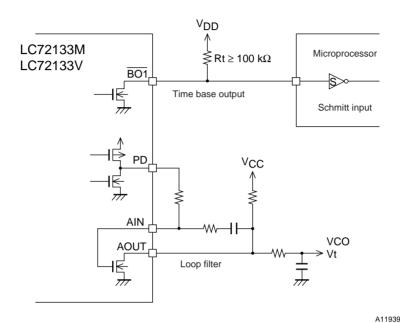
4. Directly Outputting Unlocked State Data from the DO Pin (Set by the DO pin control data)

Since the locking state (high = locked, low = unlocked) is output directly from the DO pin, the dummy data processing described in section 3 above is not required. After changing the divisor N, the locking state can be checked after waiting at least two reference frequency periods.

Clock Time Base Usage Notes

The pull-up resistor used on the clock time base output pin $(\overline{BO1})$ should be at least 100 k Ω . Also, to prevent chattering we recommend using a Schmitt input at the controller (microprocessor) that receives this signal.

This is to prevent degrading the VCO C/N characteristics when a loop filter is formed using the built-in low-pass filter transistor. Since the clock time base output pin and the low-pass filter have a common ground internal to the IC, it is necessary to minimize the time base output pin current fluctuations and to suppress their influence on the low-pass filter.



Other Items

1. Notes on the Phase Comparator Dead Zone

DZ1	DZ0	Dead zone mode	Charge pump	Dead zone
0	0	DZA	ON/ON	0 s
0	1	DZB	ON/ON	−0 s
1	0	DZC	OFF/OFF	+0 s
1	1	DZD	OFF/OFF	+ +0 s

Since correction pulses are output from the charge pump even if the PLL is locked when the charge pump is in the ON/ON state, the loop can easily become unstable. This point requires special care when designing application circuits.

The following problems may occur in the ON/ON state.

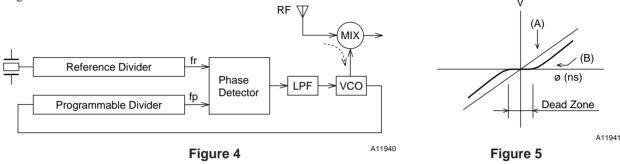
- Side band generation due to reference frequency leakage
- Side band generation due to both the correction pulse envelope and low frequency leakage

Schemes in which a dead zone is present (OFF/OFF) have good loop stability, but have the problem that acquiring a high C/N ratio can be difficult. On the other hand, although it is easy to acquire a high C/N ratio with schemes in which there is no dead zone, it is difficult to achieve high loop stability. Therefore, it can be effective to select DZA or DZB, which have no dead zone, in applications which require an FM S/N ratio in excess of 90 to 100 dB, or in which an increased AM stereo pilot margin is desired. On the other hand, we recommend selecting DZC or DZD, which provide a dead zone, for applications which do not require such a high FM signal-to-noise ratio and in which either AM stereo is not used or an adequate AM stereo pilot margin can be achieved.

Dead Zone

The phase comparator compares fp to a reference frequency (fr) as shown in Figure 4. Although the characteristics of this circuit (see Figure 5) are such that the output voltage is proportional to the phase difference \emptyset (line A), a region (the dead zone) in which it is not possible to compare small phase differences occurs in actual ICs due to internal circuit delays and other factors (line B). A dead zone as small as possible is desirable for products that must provide a high S/N ratio.

However, since a larger dead zone makes this circuit easier to use, a larger dead zone is appropriate for popularly-priced products. This is because it is possible for RF signals to leak from the mixer to the VCO and modulate the VCO in popularly-priced products in the presence of strong RF inputs. When the dead zone is narrow, the circuit outputs correction pulses and this output can further modulate the VCO and generate beat frequencies with the RF signal.



2. Notes on the FMIN, AMIN, and IFIN Pins

Coupling capacitors must be placed as close as possible to their respective pin. A capacitance of about 100 pF is desirable. In particular, if a capacitance of 1000 pF or over is used for the IF pin, the time to reach the bias level will increase and incorrect counting may occur due to the relationship with the wait time.

3. Notes on IF Counting → SD must be used in conjunction with the IF counting time When using IF counting, always implement IF counting by having the microprocessor determine the presence of the IF-IC SD (station detect) signal and turn on the IF counter buffer only if the SD signal is present. Schemes in which auto-searches are performed with only IF counting are not recommended, since they can stop at points where there is no signal due to leakage output from the IF counter buffer.

4. DO Pin Usage Techniques

In addition to data output mode times, the DO pin can also be used to check for IF counter count completion and for unlock detection output. Also, an input pin state can be output unchanged through the DO pin and input to the controller.

5. Power Supply Pins

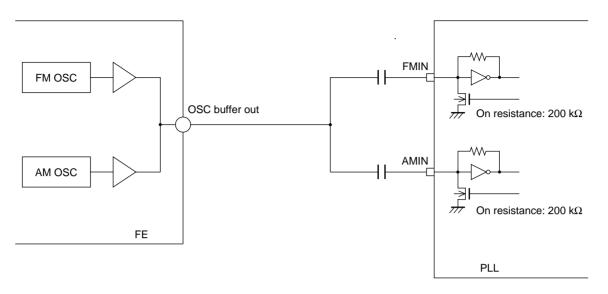
A capacitor of at least 2000 pF must be inserted between the power supply V_{DD} and V_{SS} pins for noise exclusion. This capacitor must be placed as close as possible to the V_{DD} and V_{SS} pins.

6. VCO setup

Applications must be designed so that the VCO (local oscillator) does not stop, even if the control voltage (Vtune) goes to 0V. If it is possible for the oscillator to stop, the application must use the control data (DLC) to temporarily force Vtune to V_{CC} to prevent the deadlock from occurring. (Deadlock clear circuit)

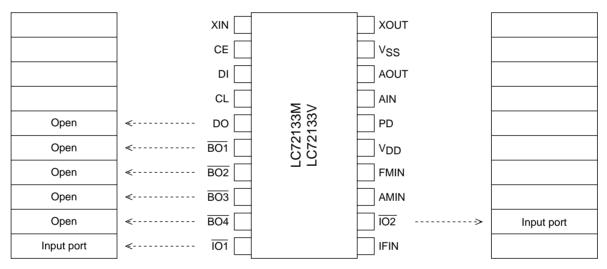
7. Front end connection example

Since this product is designed with the relatively high resistance of 200 k Ω for the pull-down (on) resistors built in to the FMIN and AMIN pins, a common AM/FM local oscillator buffer can be used as shown in the following circuit.



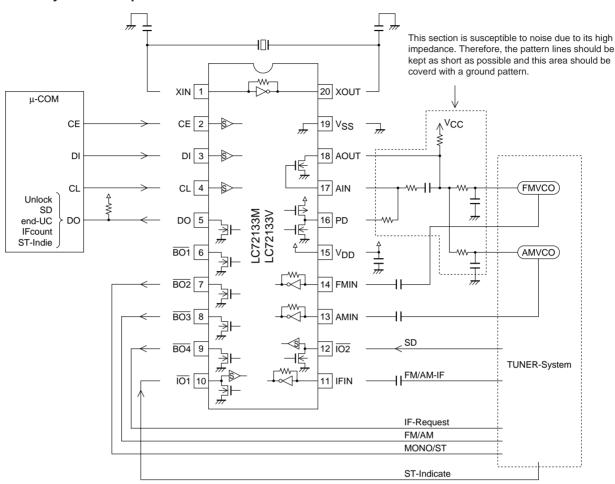
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Pin States After the Power ON Reset



A11943

Application System Example



A11944

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