



# **QPSK Transmitter for Cable TV**

#### Overview

The LA7790M is a QPSK data transmitter for digital cable TV applications. By integrating the I/Q quadrature modulator, RF amplifier, electronic volume control, mute control, and other functions onto a single chip, parts count is reduced and set size is miniaturized.

#### **Features**

- Maximum RF amplifier output level of +10 dBm (75 $\Omega$  terminator), suitable for directly driving the cable.
- RF output frequency range of 5 to 70 MHz. Frequency range selection function permits broadband designs.
- Electronic volume control for direct-current control of RF output level.
- Muting ensures ample attenuation during periods with no transmission.
- Support for both internal and external bias for I/Q modulation inputs.
- Support for I/Q modulation frequencies up to 10 MHz. (typ: 500 mVp-p)

#### **Functions**

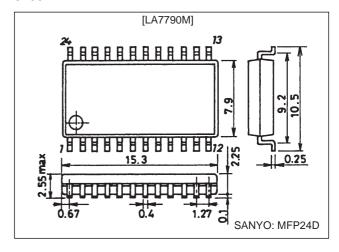
- I/O quadrature modulator
- I/Q input bias power supply
- · RF amplifier

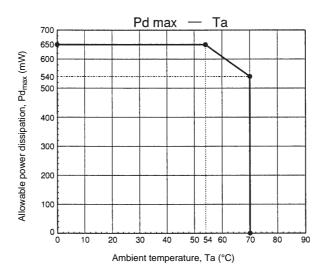
- · Varactor diode-based VCO
- Muting
- Electronic volume control
- · Power-saving modes
- Switchable output frequency range
- Power supply voltage of 5 V (4.5 to 5.5 V)

# **Package Dimensions**

unit: mm

#### 3108-MFP24D





# $\begin{tabular}{ll} \textbf{Specifications} \\ \textbf{Maximum Ratings at } Ta = 25^{\circ}C \\ \end{tabular}$

Parameter	Symbol	Conditions	Ratings	Unit
Maximum power supply voltage	V <sub>CC</sub> max	Pins 1, 10, and 24	7	V
Circuit voltage	Vmax	Pins 1, 12, 17, 20, 21, and 23	V <sub>CC</sub>	V
Circuit current	I <sub>11</sub>	Output lead-in current	1	mA
	I <sub>19</sub>	Output lead-in current	2	mA
Allowable power dissipation	Pd max	Ta ≤ 54°C	650	mW
Operating ambient temperature	Topr		-20 to +70	°C
Storage temperature	Tstg		-55 to +150	°C

### Recommended Conditions at $Ta = 25^{\circ}C$

Parameter	Symbol	Conditions	Ratings	Unit
Operating power supply voltage	V <sub>CC</sub> op	Pins 1, 10, and 24	4.5 to 5.5	V

# Operating Characteristics at $Ta=25^{\circ}C,\,V_{CC}=5.0~V$

Parameter	Sumbal	Conditions	Ratings			Linit
raidifietei	Symbol	Conditions	min	typ	max	Unit
Current drain						
Circuit current 1	I <sub>1</sub>	With no signal, pin 1	26	33	44	mA
Circuit current 2	l <sub>24</sub> + l <sub>10</sub>	With no signal, pins 24 and 10	44	55	73	mA
Modulator fo: f(V19) = 25 MHz	•					
Output frequency range	f <sub>(V19)</sub>		5		70	MHz
Output signal level	V <sub>19(DSB)</sub>	Note 1: V <sub>20</sub> = V <sub>23</sub> = GND, V <sub>12</sub> = V <sub>17</sub> = V <sub>21</sub> = V <sub>CC</sub> , S1 = A	-7	-4	-2	dBm
Output harmonic distortion	V <sub>19(HD2)</sub>	Note 2: V <sub>20</sub> = V <sub>23</sub> = GND,	40			dB
Output Harmonic distortion	V <sub>19(HD3)</sub>	$V_{12} = V_{17} = V_{21} = V_{CC}$ , S1 = A	35			dB
Output secondary harmonic distortion	V <sub>19(2fo)</sub>	Note 3: $V_{20} = V_{23} = GND$ , $V_{12} = V_{17} = V_{21} = V_{CC}$ , S1 = A	20			dB
Output tertiary harmonic distortion	V <sub>19(3fo)</sub>	Note 3: $V_{20} = V_{23} = GND$ , $V_{12} = V_{17} = V_{21} = V_{CC}$ , S1 = A	8			dB
Carrier suppression ratio	V <sub>19(fo)</sub>	Note 4: $V_{20} = V_{23} = GND$ , $V_{12} = V_{17} = V_{21} = V_{CC}$ , S1 = A	30			dB
Sideband suppression ratio	V <sub>19(SSB)</sub>	Note 5: $V_{20} = V_{23} = GND$ , $V_{12} = V_{17} = V_{21} = V_{CC}$ , S1 = A	30			dB
I input level	V <sub>4.5</sub>	$V_{4.5} =  V_4 - V_5 $		500		mVp-p
Q input level	V <sub>8.9</sub>	$V_{8.9} =  V_8 - V_9 $		500		mVp-p
I input DC voltage	V <sub>4, 5</sub>	External DC bias voltage	1.9	2.1	2.3	V
Q input DC voltage	V <sub>8, 9</sub>	External DC bias voltage	1.9	2.1	2.3	V
Reference voltage	V <sub>6</sub>	Internal DC bias voltage	1.9	2.1	2.3	V
Variable attenuator						
Minimum gain control voltage	V <sub>21</sub>	$V_{12} = V_{17} = V_{CC}, V_{20} = V_{23} = GND, S1 = A$	0		0.5	V
Gain range	V <sub>19(GR)</sub>	Note 6: $V_{20} = V_{23} = \text{GND}$ , $V_{12} = V_{17} = V_{CC}$ , S1 = A, $V_{21} = V_{CC} \rightarrow 0.5 \text{ V}$	45			dB
Modulator output impedance	R <sub>19</sub>	$V_{12} = V_{17} = V_{21} = V_{CC},$ $V_{20} = V_{23} = GND, S1 = A$	80	120	160	Ω
VCO						
Oscillator frequency range	f <sub>(osc)</sub>	$V_{12} = V_{17} = V_{21} = V_{CC},$ $V_{20} = V_{23} = GND, S1 = A$	20		280	MHz
VCO output level	V <sub>11(fo)</sub>	$V_{12} = V_{17} = V_{21} = V_{CC},$ $V_{20} = V_{23} = GND, S1 = A$	500	700	900	mVp-p
VCO output impedance	R <sub>11</sub>	$V_{12} = V_{17} = V_{21} = V_{CC},$ $V_{20} = V_{23} = GND, S1 = A$	200	300	400	Ω
Band switch						
Band switch 1 "H" level f(V19) = 25 MHz	V <sub>17H</sub>	Note 7: V <sub>12</sub> = V <sub>21</sub> = V <sub>CC</sub> , V <sub>20</sub> = V <sub>23</sub> = GND, S1 = A	4			V
Band switch 1 "L" level f(V19) = 12.5 MHz	V <sub>17L</sub>	Note 7: $V_{12} = V_{21} = V_{CC}$ , $V_{20} = V_{23} = GND$ , S1 = A			1	V

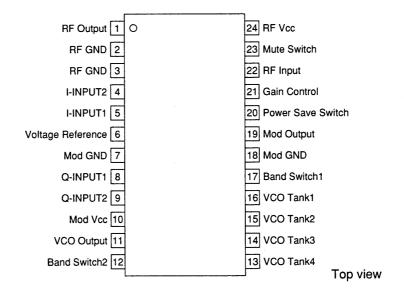
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Parameter	Cumbal	Conditions	Ratings			1.1
	Symbol		min	typ	max	Unit
Band switch 2 "H" level f(V <sub>11</sub> ) = 25 MHz	V <sub>12H</sub>	Note 7: $V_{17} = V_{21} = V_{CC}$ , $V_{20} = V_{23} = GND$ , S1 = A	4			V
Band switch 2 "L" level f(V <sub>11</sub> ) = 12.5 MHz	V <sub>12L</sub>	Note 7: $V_{17} = V_{21} = V_{CC}$ , $V_{20} = V_{23} = GND$ , S1 = A			1	V
RF output amplifier						
Maximum output level	V <sub>1</sub> max	Note 8: $V_{12} = V_{17} = V_{21} = V_{CC}$ , $V_{20} = V_{23} = GND$ , $SG3 = -6$ dBm, $S1 = A$	7	10	13	dBm
Maximum output distortion	V <sub>1(HD2)</sub>	Note 8: $V_{12} = V_{17} = V_{21} = V_{CC}$ , $V_{20} = V_{23} = GND$ , $SG3 = -6$ dBm, $S1 = A$	40			dBc
	V <sub>1(HD3)</sub>		40			dBc
Muting						
Muting on voltage	V <sub>23H</sub>	$V_{12} = V_{17} = V_{21} = V_{CC},$ $V_{20} = \text{GND}, \text{ S1} = \text{B}$	4			V
Muting off voltage	V <sub>23L</sub>	$V_{12} = V_{17} = V_{21} = V_{CC},$ $V_{20} = \text{GND}, \text{S1} = \text{B}$			1	٧
Muting attenuation	V <sub>1(mute)</sub>	Note 9: $V_{23} = 1 \text{ V} \rightarrow 4 \text{ V}$ $V_{12} = V_{17} = V_{21} = V_{CC}, V_{20} = \text{GND}, \text{S1} = \text{B}$	70			dB
Power save function	•		•			
Power save on voltage	V <sub>20H</sub>		4			V
Power save off voltage	V <sub>20L</sub>				1	V
Power save current	I <sub>1</sub>	Note 10: V <sub>20</sub> = 4 V	0		0.1	mA

# Pin Assignment



#### LA7790M

Note 1

Input: SG1 =1 MHz CW, 500 mVp-p, SG2 = No Signal or

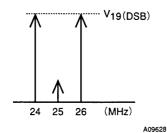
SG2 = 1 MHz CW, 500 mVp-p, SG1 = No Signal

Output:

Note 2

Input: Same as Note 1

Output:

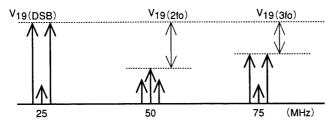


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Note 3

Input: Same as Note 1

Output:



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Note 4

Input: SG1 = SG2 = 1 MHz CW, 500 mVp-p

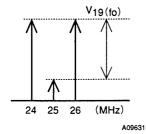
Output:

Note 5

Input: SG1 = 1 MHz CW, 500 mVp-p, 0deg

SG1 = 1 MHz CW, 500 mVp-p, 90deg

Output:



V19(SSB)

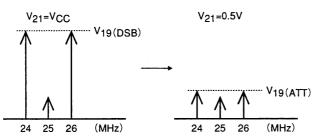
1 24 25 26 (MHz)

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Note 6

Input: SG1 = 1 MHz CW, 500 mVp-p

Output:



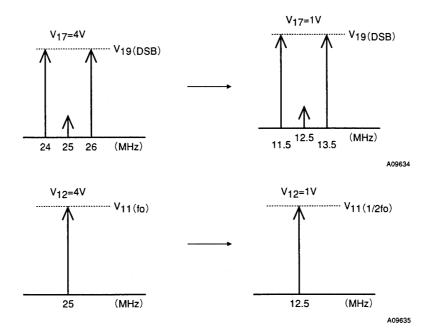
V<sub>19</sub>(GR)=V<sub>19</sub>(DSB)-V<sub>19</sub>(ATT)

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Note 7

Input: SG1 = 1 MHz CW, 500 mVp-p

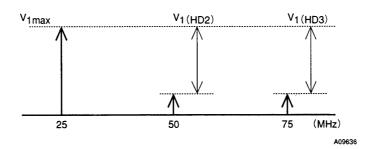
Output:



Note 8

Input: SG3 = 25 MHz CW, -6 dBm

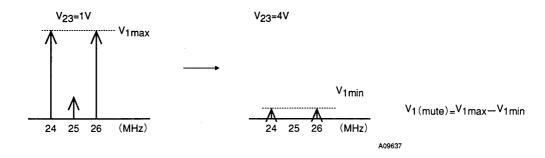
Output:



Note 9

Input: SG1 = 1 MHz CW, 500 mVp-p

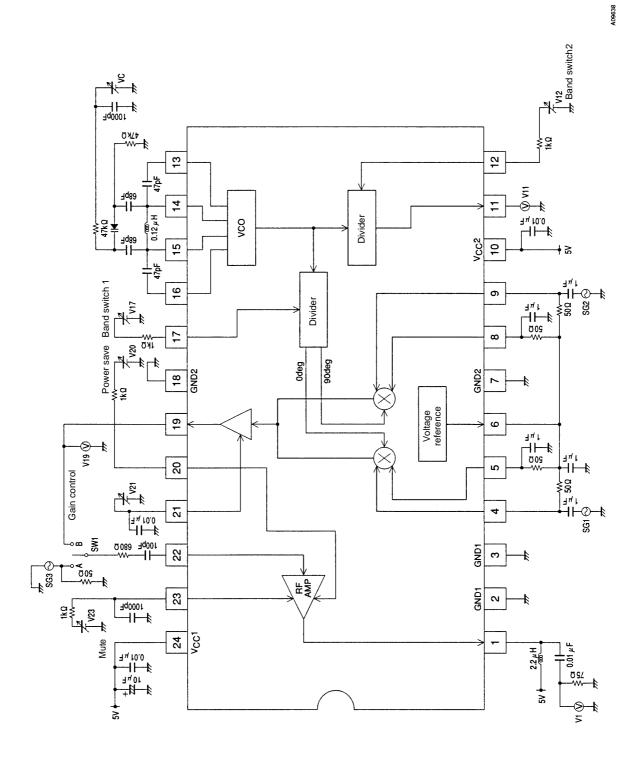
Output:



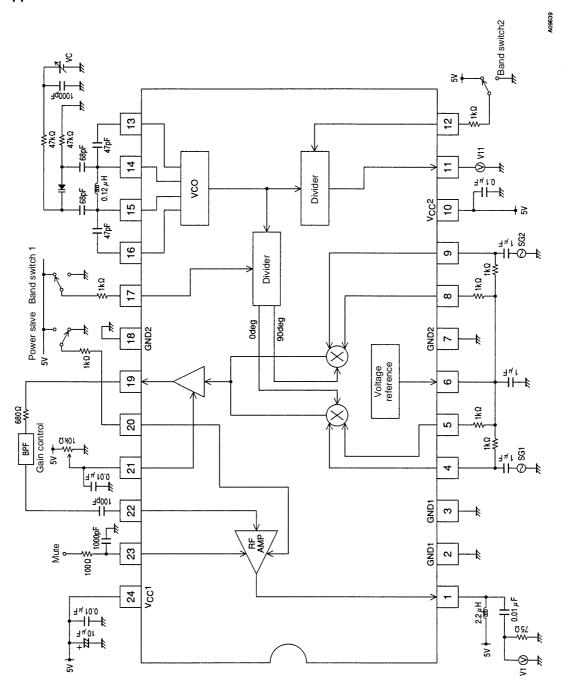
Note 10

 $I_1 = pin 1$  current when V20 = 4 V (power save on).

### **Measurement Circuit**



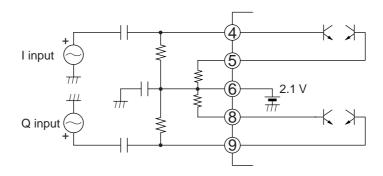
# **Sample Application Circuit**



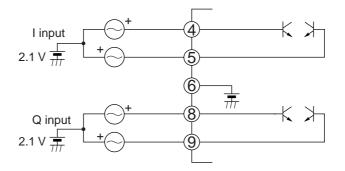
#### 1. Modulator

The modulator consists of two identical multiplier circuits, creating I and Q channels. Pins 4 & 5 and pins 8 & 9 are I and Q channel inputs, respectively. These pins must be biased at  $2.1 \pm 0.2$  V. Pin 6 is an internal 2.1 V bias. This internal bias can be used if the I & Q data inputs are AC coupled, but an external bias must be used in the case of DC coupled data inputs. Carrier suppression is improved if the offset voltages between pins 4 & 5 and pins 8 & 9 are small.

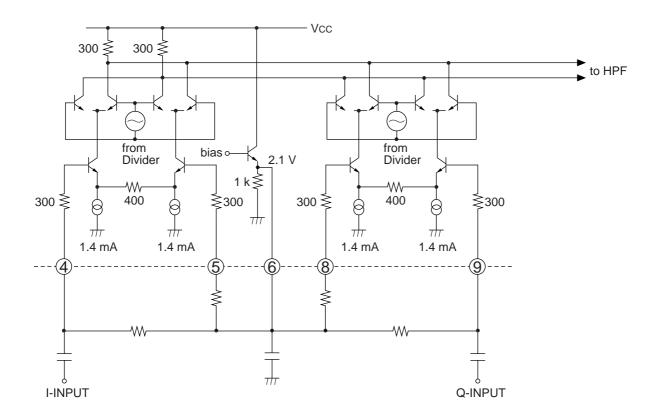
#### (1) AC coupled application



### (2) DC coupled application



# Equivalent circuit of modulator block

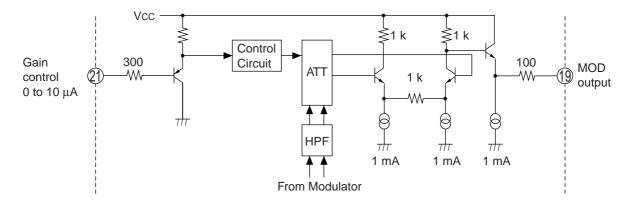


#### 2. Attenuator

The attenuator input signal comes from the modulator output via a high pass filter ( $f_C = 2$  MHz). The modulator output is pin 19. Pin 19 is connected to the emitter of an NPN emitter follower through a 100  $\Omega$  resistor. The emitter sink current is about 1 mA, but can be increased by the addition of an external resistor between pin 19 and ground. The minimum value for the external resistor is 1.5 k $\Omega$ .

The output level of pin 19 can be adjusted using pin 21. Pin 21 is connected to the base of a PNP emitter follower through a 300  $\Omega$  resistor. When pin 21 = 0 V, the base current is about 10  $\mu$ A.

#### Equivalent circuit of attenuator block



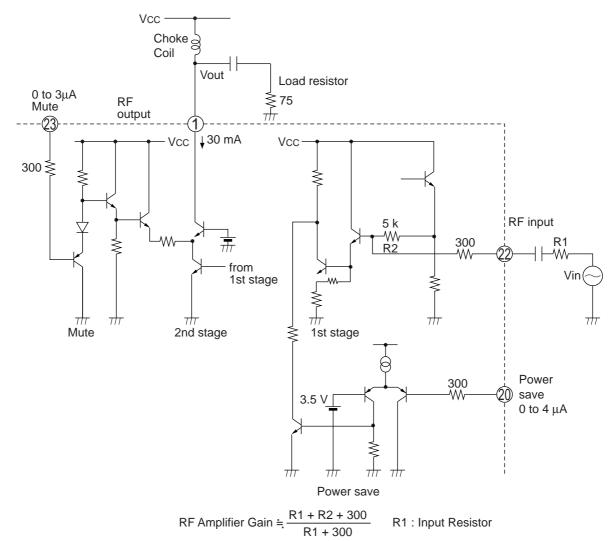
Unit (resistance :  $\Omega$ )

## 3. RF Amplifier

The RF amplifier input signal comes from the attenuator output after the harmonics are removed by the LPF. The RF amplifier is composed of a two-stage amplifier and includes mute and power save functions.

The gain of the RF amplifier can be adjusted using the external input resistor connected to pin 22 and the external load resistor connected to pin 1. The input of pin 22 must be AC coupled. The output of pin 1 must be connected to  $V_{CC}$  via a choke coil. Pin 23 (mute control) and pin 20 (power save control) are both high impedance inputs (base current less than 10  $\mu$ A). For normal operation, the voltage at pins 20 and 23 must be less than 1 V. For Mute and Power Save operation, the voltage at pins 20 are 23 must be between  $V_{CC}$ –1 and  $V_{CC}$ .

# Equivalent circuit of RF amplifier block



The gain of RF amplifier can be adjusted by changing the value R1.

#### 4. Oscillator

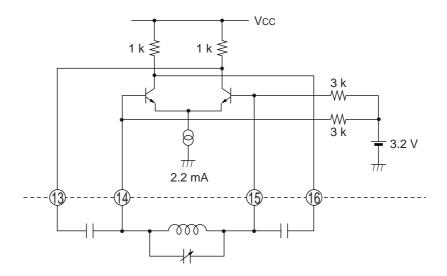
The oscillator signal is divided by either 4 or 8, and supplied to the modulator as  $0^{\circ}$  and  $90^{\circ}$  switching signals. The division mode can be selected using Band Switch 1 (pin 17).

Pin 17 = High  $\rightarrow$  Modulation frequency = 1/4 oscillator frequency

Pin  $17 = \text{Low} \rightarrow \text{Modulation frequency} = 1/8 \text{ oscillator frequency}$ 

The oscillator requires an external coil and capacitors. Pins 14 & 15 should be DC coupled using a coil. A coil with Qu of 30 or greater is required and the impedance between pins 14 and 15 is  $6 \text{ k}\Omega$ . The value of the coupling capacitors between pins 13 & 14 and pins 15 & 16 must be large enough so that the signal phase rotation is small.

#### Equivalent circuit of oscillator block



#### 5. Oscillator Output

The output frequency of the oscillator is equal to or half of the output frequency of the modulator. The mode is set using Band Switch 2 (pin 12).

Pin  $12 = \text{High} \rightarrow \text{Oscillator output frequency} = \text{modulator frequency}$ 

Pin  $12 = \text{Low} \rightarrow \text{Oscillator output frequency} = 1/2 \text{ modulator frequency}$ 

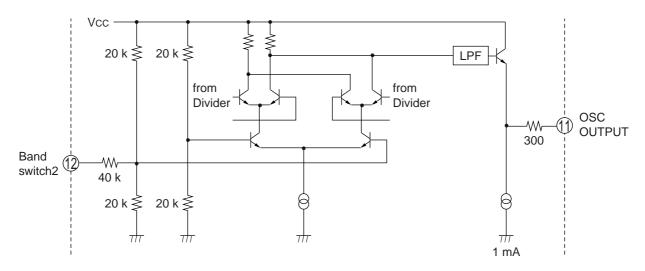
#### 6. Band Swtich

The input circuitry of Band Switch 1 (pin 17) and Band Switch 2 (pin 12) are the same. A low setting requires a voltage of less than 1 V at the pin. A high setting requires a voltage between  $V_{CC}$ -1 and  $V_{CC}$  at the pin. Settings of low or high generate currents of  $\pm 50~\mu A$  at pins 12 and 17.

Pin 12, 17 = High 
$$\rightarrow$$
 Sink current = 50  $\mu$ A

Pin 12, 
$$17 = \text{Low} \rightarrow \text{Source current} = 50 \,\mu\text{A}$$

### Equivalent circuit of osc output and band switch blocks

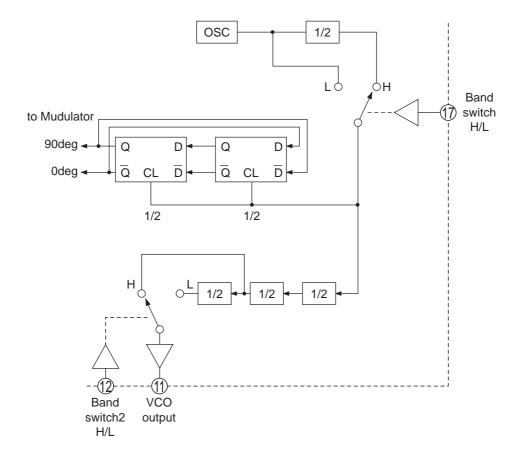


The input circuitry of band switch 1 is same.

#### 7. Digital ECL Circuit

To get the correct  $0^{\circ}/90^{\circ}$  phase switching signal for the modulator, a divide-by-four dual flip-flop is employed as shown in the following figure.

Band Switch 1 can be used to expand the range of the modulation frequency. Band Switch 2 can be used to select the frequency to output to an external PLL.



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