



Preamplifier + Power Amplifier for 1.5 V Headphone Stereos

Overview

The LA4587M is a system IC that includes all of the necessary functions for a playback set on a single chip, reducing the number of external components needed.

Functions

- Stereo preamplifier (supports auto reverse function, switchable between metal and normal tape)
- Stereo power amplifier (OCL, mute function)
- · Ripple filter
- Low boost function (BTL operation in low-frequency range)
- AMSS (Automatic Music Select System)
- · Power switch

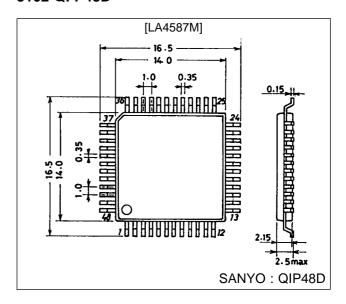
Features

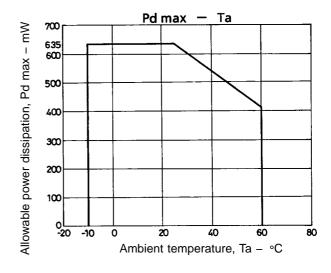
- Preamplifier has a high open-loop gain ($VG_O = 73 \text{ dB}$).
- Preamplifier requires no NF capacitor.
- Virtual ground capacitor can be 1 μF or less. (Lower impedance is achieved by having a V_{REF} amplifier built in.)
- Ripple filter requires no capacitor for preventing oscillation.
- Powerful output is obtained in low boost output (Po = $21 \text{ mW/V}_{CC} = 1.2 \text{ V}$, f = 100 Hz).
- A high-frequency cutoff capacitor is built into the preamplifier and power amplifier inputs; anti-buzz provision.

Package Dimensions

unit: mm

3102-QFP48D





Specifications

Maximum Ratings at $Ta = 25 \,^{\circ}C$

Parameter	Symbol	Conditions	Ratings	Unit
Maximum supply voltage	V _{CC} max		3.0	V
Allowable power dissipation	Pd max		635	mW
Operating temperature	Topr		-10 to +60	∘C
Storage temperature	Tstg		-40 to +125	°C

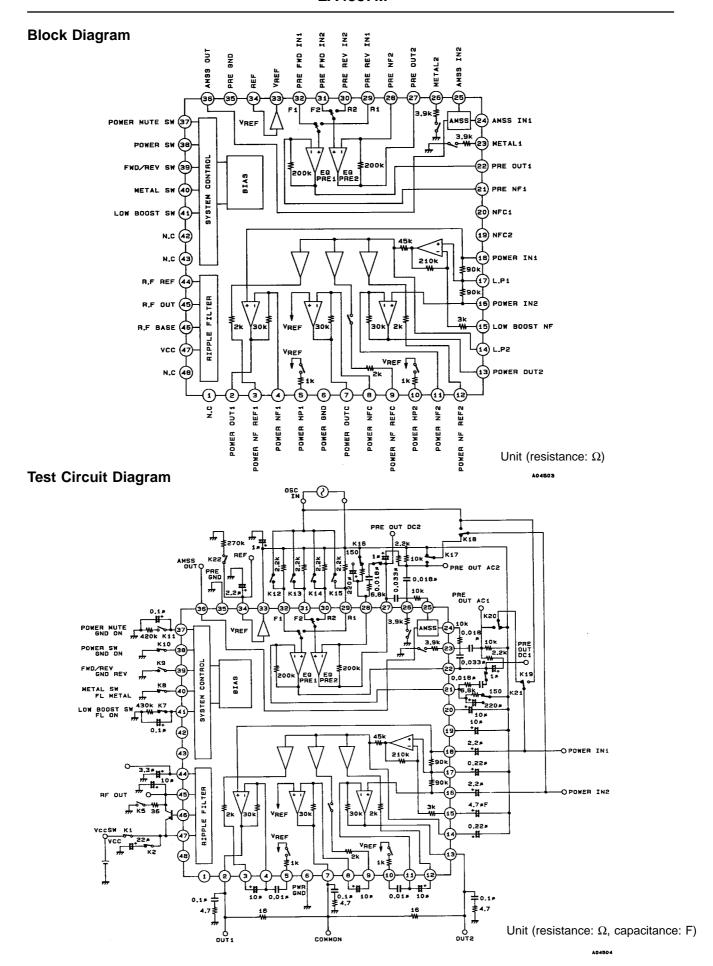
Operating Conditions at $Ta = 25 \,^{\circ}C$

Parameter	Symbol	Conditions	Ratings	Unit
Recommended supply voltage	V _{CC}		1.5	V
Operating supply voltage range	V _{CC} op		0.95 to 2.2	V

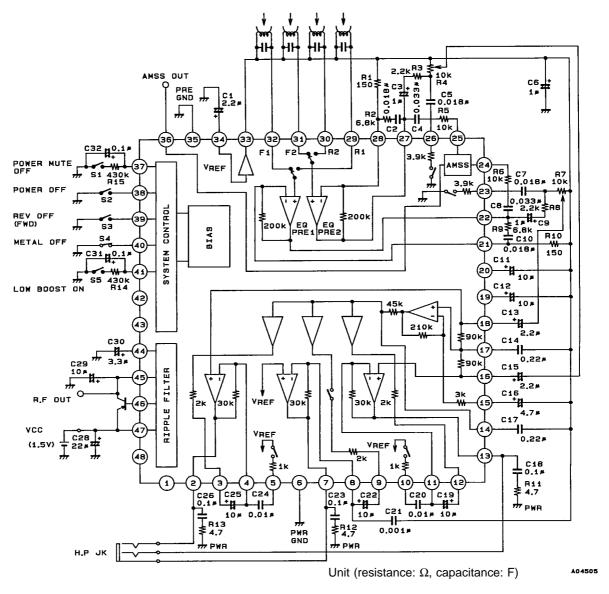
Operation Characteristics at Ta = 25 °C, $V_{\rm CC}$ = 1.2 V, f = 1 kHz, 0.775 V = 0 dBm, $R_{\rm L}$ = 10 k Ω (preamplifier), $R_{\rm L}$ = 16 Ω (power amplifier)

Parameter	Symbol	Conditions	min	typ	max	Unit	
[Preamplifier + Power Amplifier]							
Quiescent current	I _{CCO} 1	$Rg = 2.2 \text{ k}\Omega, Rv = 0 \Omega$	8	15	24	mA	
Quiescent current	I _{CCO} 2	When power switch is off		0.1	5	μA	
Voltage gain (closed)	VG _T	$V_O = -20 \text{ dBm}, R_V = 10 \text{ k}\Omega$	54	57	60	dB	
[Preamplifier]							
Voltage gain (open)	VG ₀	$V_O = -20 \text{ dBm}$	60	73		dB	
Voltage gain (alegad)	VG ₁	$V_O = -20 \text{ dBm}$	34	35.5	37	dB	
Voltage gain (closed)	VG ₂	$V_O = -20$ dBm, f = 10 kHz, metal on	25.5	28	30.5	dB	
Maximum output voltage	V _O max	THD = 1 %	100	210		mV	
Total harmonic distortion	THD ₁	VG = 35.5 dB/NAB, V _O = 100 mV		0.1	0.5	%	
Equivalent input noise voltage	V _{NI}	Rg = 2.2 kΩ, BPF: 20 Hz to 20 kHz		1.3	3.0	μV	
Interchannel crosstalk	CT ₁	Rg = 2.2 kΩ, 1 kHz TUNE, $V_O = -20$ dBm	45	56		dB	
Interchannel crosstalk between F and R	CT ₂	Rg = 2.2 k Ω , 1 kHz TUNE, V $_{0}$ = -20 dBm	65	78		dB	
Ripple rejection ratio	Rr1	Rg = 2.2 k Ω , Vr = -30 dBm, fr = 100 Hz, 100 Hz TUNE	45	52		dB	
[Low Boost + Power Amplifier]			1				
	VG ₃	$V_O = -20 \text{ dBm}$	20.5	23	25.5	dB	
\\alta = \\ \tag{\alpha} \\ \alpha = \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \\ \	VG ₄	$V_{O} = -20 \text{ dBm}, \text{ L.B.} = \text{on}$	20.5	23	25.5	dB	
Voltage gain (closed)	VG ₅	$V_{O} = -20 \text{ dBm}, \text{ L.B.} = \text{on, f} = 10 \text{ kHz}$	24.5	27.5	30.5	dB	
	VG ₆	V _O = -20 dBm, L.B. = on, f = 100 Hz	30	34	38	dB	
Outside source	P _{O1}	THD = 10 %	5	9		mW	
Output power	P _{O2}	THD = 10 %, f = 100 Hz, L.B. = on	13	21		mW	
Total harmonic distortion	THD ₂	P _O = 1 mW		0.5	1.5	%	
Interchannel crosstalk	CT ₃	$V_O = -20$ dBm, $R_V = 0$ Ω	38	43		dB	
Output noise voltage	V _{NO}	$R_V = 0 \Omega$, BPF: 20 Hz to 20 kHz		35	48	μV	
Ripple rejection ratio	Rr ₂	$R_V = 0 \Omega$, $Vr = -30 \text{ dBm}$, $fr = 100 \text{ Hz}$, 100 Hz TUNE	50	74		dB	
Output mute voltage	V _M	V _{IN} = -30 dBm, 1 KHz TUNE, mute on			-85	dBm	
Input resistance	Ri		8	10	12	kΩ	
Voltage gain difference	ΔVG_3			0	+1.5	dB	
[Ripple Filter]			'		•		
Ripple rejection ratio	Rr ₃	$fr = 100 \text{ Hz}, Vr = -30 \text{ dBm}, V_{CC} = 1.0 \text{ V}, I_{RF} = 25 \text{ mA}, 2SB1295, h_{FE}6 \text{ rank used}$	33	39		dB	
Output voltage	V_{RF}	V _{CC} = 1.0 V, I _{RF} = 25 mA	0.89	0.93		V	
[AMSS]							
Operating output voltage	V _{OAMSS}	Preout voltage when AMSS $V_O = 0.6$ Vp-p Pin 34 is short-circuited through 270 kΩ.	1.80	2.55	3.60	mV	

Note: L.B. = Low boost



Sample Application Circuit



Note 1: Transistors equivalent to the 2SB1295 with $h_{\mbox{\scriptsize FE}}6$ rank and upward are recommended.

Note 2: C18, C23, and C26 are oscillation prevention capacitors; a polyester film or ceramic capacitor (which can guarantee the specified capacitance at operating temperatures) is recommended.

Pin Functions

Unit (resistance: Ω , capacitance: F)

* Pin voltage is when $V_{CC} = 1.2 \text{ V}$

	T	1	Unit (resistance: \$2, capacitance: F)	* Pin voltage is when $V_{CC} = 1.2 \text{ V}$
Pin No.	Pin name	Pin voltage [V]	Internal equivalent circuit	Remarks
45	R.F OUT	1.13	45 300 A04506	
2	POWER OUT1	0.6	A A A	A 160 Ω resistor is connected
7	POWER OUTC		150 30k	between individual outputs (between pins 2 and 7, and between pins 13 and 7).
13	POWER OUT2		2 A003B4	between pins 10 and 7).
3	POWER NF REF1	0.75	Λ Λ	Each power NF connection
9	POWER NF REFC		3 * W * W *	
12	POWER NF REF2		A00385	
4	POWER NF1	0.75	^	• Each power NF connection.
8	POWER NFC		30k 300 J	
11	POWER NF2		4 ¥ 100k	
5	POWER H.P1	0.75	Α Α	• Grounded to V_{REF} through a 1 k Ω resistor when low boost
10	POWER H.P2		5 N N N N N N N N N N N N N N N N N N N	is on (pin 41: floating).
14	L.P2	0.75	45K 210K 33K A00988	Low boost secondary LP connection.

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Unit (resistance: Ω , capacitance: F)

Pin No.	Pin name	Pin voltage [V]	Internal equivalent circuit	Remarks
15	Low Boost NF	0.75	3k 210k A00369	Low boost amplifier NF connection.
16	POWER IN2	0.75	^	Each power input connection.The input resistance is
18	POWER IN1		300 10k 90k 15p 0	 10 kΩ. An anti-buzz capacitor is built in.
17	L.P1	0.75	90k 300 300 300 300 300 300 300	Low boost primary LP. connection.
19	NFC2	0.75	^ .	
20	NFC1		43k 300 W A00392	
21	PRE NF1	0.75	† ,	Each preamplifier NF connection.
28	PRE NF2		21 W II W A00393	NF requires no capacitor.
22	PRE OUT1	0.45	A . A	• 200 kΩ is connected between each output pin and
27	PRE OUT2		22 W 300	NF pin.

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Unit (resistance: Ω , capacitance: F)

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Pin No.	Pin name	Pin voltage [V]	Internal equivalent circuit	Remarks
23 26	METAL1 METAL2	0	3.9k W M	- Connected to GND through 3.9 $\ensuremath{k\Omega}$ in metal on mode (pin 40: floating)
24	AMSS IN1	0.75	A A	AMSS inverting input
25	AMSS IN2		150k W 300 A00396	 connection. An external input resistor is required.
29	PRE REV IN1	0.75	٨	Pins 29 and 30 turn on in
30	PRE REV IN2		110p	REV mode (pin 39: GND). • Pins 31 and 32 turn on in
31	PRE FWD IN2		29 W 1 1 300	FWD mode (pin 39: floating) • When not using the head, a
32	PRE FWD IN1		A00397	 bias resistor (2.2 kΩ) is required between these pins and V_{REF} (pin 33). An anti-buzz capacitor is built in.
33	V _{REF}	0.75	33 W A00398	 V_{REF} amplifier output. Low impedance is achieved due to the output resistor (ro = 10 Ω). Inflow/outflow current: 200 μA max.
34	REF	0.75	34 300 \$ A00399	The V _{REF} amplifier is referenced hereto.
36	AMSS OUT		300 300 300 A00400	Outputs a pulse waveform in accordance with the AMSS IN (pins 24 and 25) input level.

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Unit (resistance: Ω , capacitance: F)

Pin No.	Pin name	Pin voltage [V]	Internal equivalent circuit	Remarks
37 41	POWER MUTE SW Low Boost SW		37 W A00401	When pin 37 is grounded, mute is on. When pin 41 is floating, low boost is on.
38	POWER SW		3B 200k ₹22k ₩ 300	Power on when grounded.
39	FWD/REV SW		A A	When pin 39 is floating: FWD
40	METAL SW		39	mode; when grounded: REV mode. • When pin 40 is in FL mode: metal on.
			A00403	
44	R.F REF	1.13	445k 445k 3000	RF is referenced hereto. An external capacitor can be used to vary RF SVRR.
46	R.F BASE	0.5	46 W 100 A00405	Used for external PNP transistor base drive.

Description of External Components

• C_1 (1.0 to 10 μ F): V_{RFF} amplifier is referenced to this decoupling capacitor.

The V_{REF} SVRR depends on the value of this capacitor.

Note that if the capacitance is reduced, the SVRR worsens.

• C₂, C₁₀: Playback preamplifier EQ constant.

• C₃, C₉ (0.47 to 3.3 μF): Preamplifier output capacitor.

• C₄, C₈: AMSS input HPF capacitor.

• C₅, C₇: EQ constant for metal (built-in resistance 3.9 k Ω ±15%).

• C_6 (0.1 to 22 μ F): V_{REF} decoupling capacitor.

For high-frequency noise rejection.

• C_{11} , C_{12} (3.3 to 10 μF): NFC decoupling capacitor.

Note that if the capacitance is reduced, the preamplifier low-frequency gain decreases.

• C_{13} , C_{15} (1.0 to 3.3 μ F): Power amplifier input capacitor (Input resistance: 10 $k\Omega$).

• C₁₄, C₁₇: Capacitor for low boost LPF.

The low boost gain depends on the capacitance.

• C_{16} (1.0 to 4.7 µF): Boost amplifier NF capacitator.

Note that if the capacitance is reduced, the low boost low-frequency gain decreases.

• C_{18} , C_{23} , C_{26} (0.1 to 1.0 μF): Oscillation blocking capacitator.

• C_{19} , C_{22} , C_{25} (3.3 to 10 μ F): Power amplifier NF capacitor.

Note that if the capacitance is reduced, the power amplifier low-frequency gain decreases.

• C₂₀, C₂₄: Bass high boost capacitor.

The high gain depends on the capacitance.

• C₂₁ (100 to 2200 pF): Oscillation blocking capacitator.

• C₂₉ (4.7 to 10 μF): RF output decoupling capacitor. (Also serves as the power supply capacitor and the

oscillation blocking capacitor.)

• C_{28} (22 to 220 μ F): Power supply capacitor.

• C_{30} (2.2 to 10 μF): RF is referenced to this LPF capacitor.

The RF SVRR depends on the capacitance.

• C₃₁, C₃₂ (0.047 to 0.22 μF): Switching circuit smoothing capacitor. Must be adjusted according to the set timing.

• R_1 , R_{10} : For preamplifier gain adjustment.

• R₂, R₉: Playback preamplifier EQ constant.

• R_4 , R_7 : 10 k Ω volume control.

• R₅, R₆: For AMSS gain adjustment and HPF.

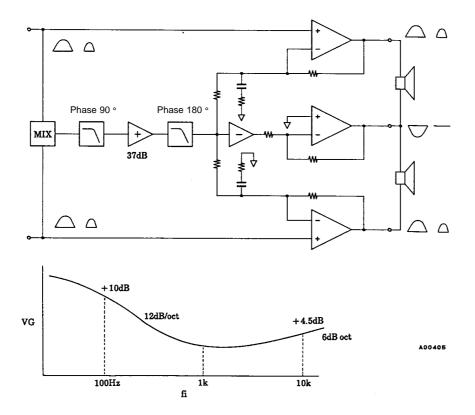
• R₁₁, R₁₂, R₁₃: For oscillation blocking.

• R_{14} , R_{15} (100 to 430 k Ω): For switching circuit smoothing (discharge resistors).

Operation Descprition

· Low boost system

Low-frequency region amplification: 12 dB/oct, high-frequency region amplification: 6 dB/oct.



· Note on low boost

The signals that are applied to each power input are mixed and then passed through a two-stage LPF. Because the signal levels are attenuated by the LPF, level compensation is accomplished by amplifying the signals through a low boost amplifier located in between. The phase of signals that pass through the secondary LPF is inverted relative to the input signal; these signals are then input to each power amplifier.

· Note on channels 1 and 2

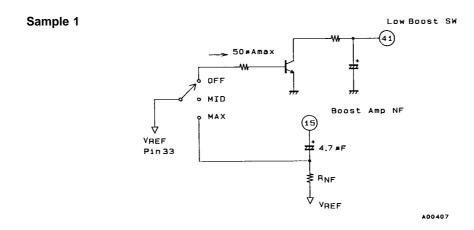
The positive phase signals that were input from the positive ("+") input pins and the reverse phase signals that were input from the negative ("-") input pins and then were passed through the secondary LPF are all input, amplified, and then output.

· Note on the common amplifier

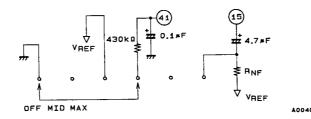
The phase of the signals that passed through the secondary LPF is inverted by the inverting amplifier; the signals (with reversed phases relative to channels 1 and 2) are then input to the negative ("-") inputs. The positive ("+") input signals are grounded to V_{REF} , amplified by the inverting amplifier and then output.

The phase of the channel 1 and 2 amplifier outputs and the common amplifier outputs are made to oscillate with inverted phases, making it possible to obtain the dynamic range efficiently.

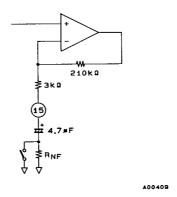
Sample Application Circuits for Low Boost Switching



Sample 2

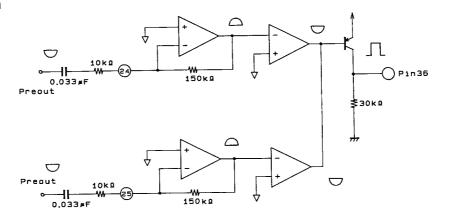


In the above circuits, MID and MAX are switched by changing the gain of the boost amplifier.



The AMSS comparator

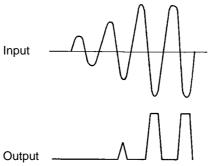
Block Diagram



A00410

Operation Description

- The input amplifiers are inverting amplifiers. The gain and HPF characteristics can be adjusted through an external C-R (input impedance).
- The AMSS comparator outputs pulses for an input waveform that satisfies certain set conditions (frequency and voltage level).



 ullet When AMSS is not used, the input pins (pins 24 and 25) are connected to V_{REF} (pin 33).

Notes on the ripple filter

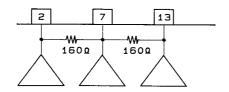
• The RF SVRR can be adjusted by an external capacitor connected to pin 44.

$$\begin{array}{l} 3.3~\mu F \rightarrow 39~dB \\ 4.7~\mu F \rightarrow 42~dB \\ 10~\mu F \rightarrow 47~dB \end{array}$$

• It is recommended that external transistors be equivalent to the 2SB1295 with hFE 6 rank and upward.

Note on power output

•The power amplifier output and the common amplifier output are connected by a resistor of approximately 160 Ω .



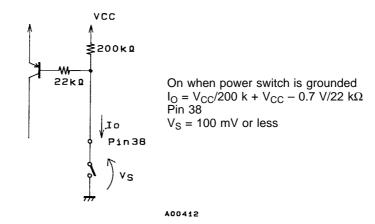
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Notes on power mute

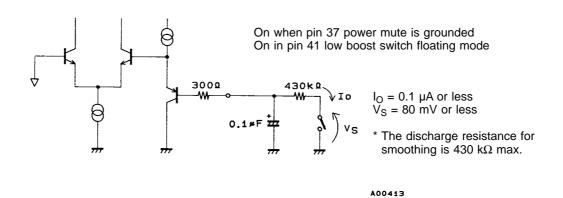
- Power mute turns off the fixed current that is supplied to the power section.
- $^{\bullet}$ The output DC when power mute is on is the V_{REF} electric potential (0.75 V).
- $^{\bullet}$ The output impedance when power mute is on is approximately 10 k $\!\Omega.$

SW Pin Equivalent Circuit Diagram

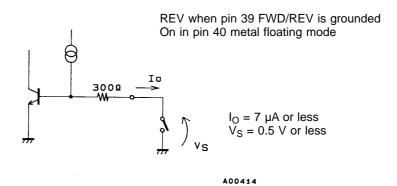
1. Power switch

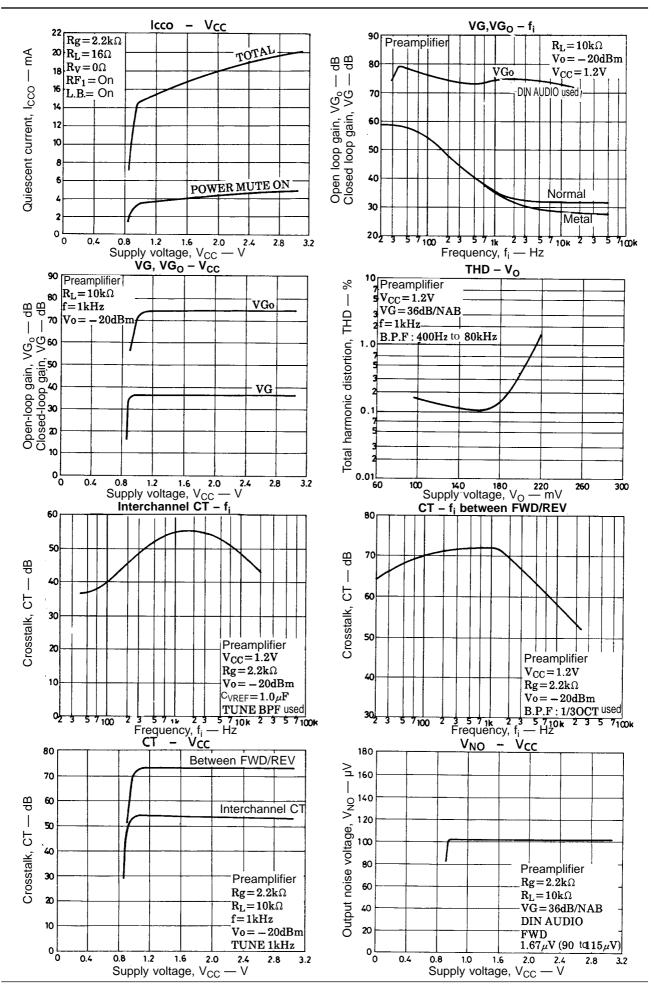


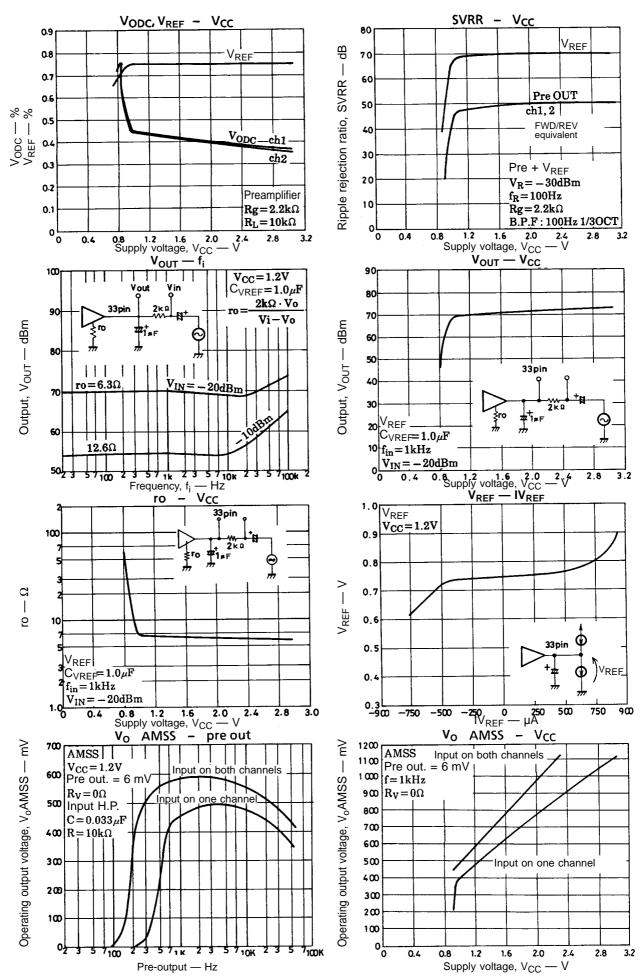
2. Power mute and low boost switch

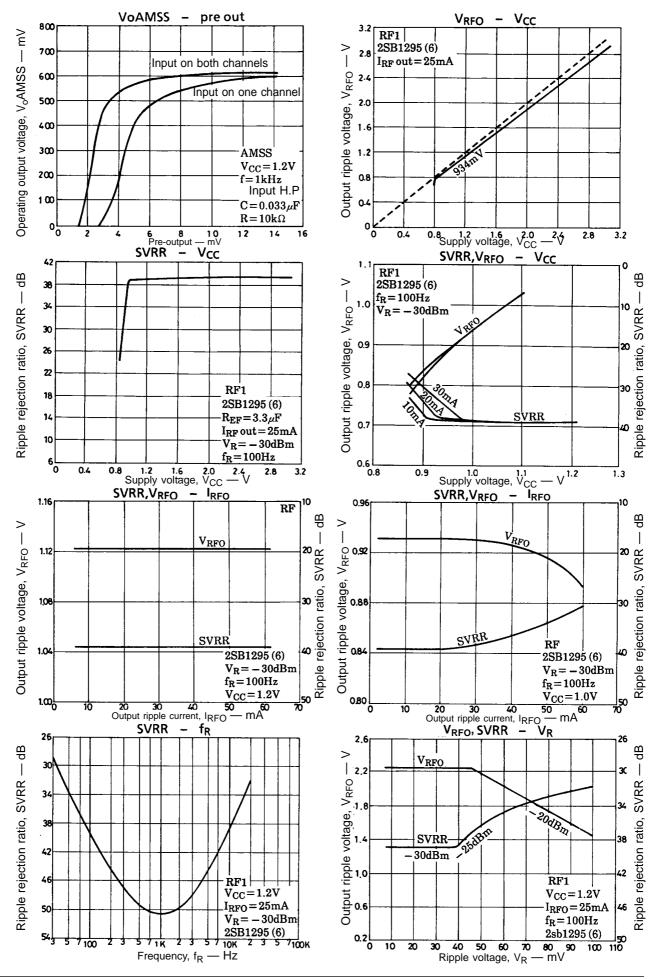


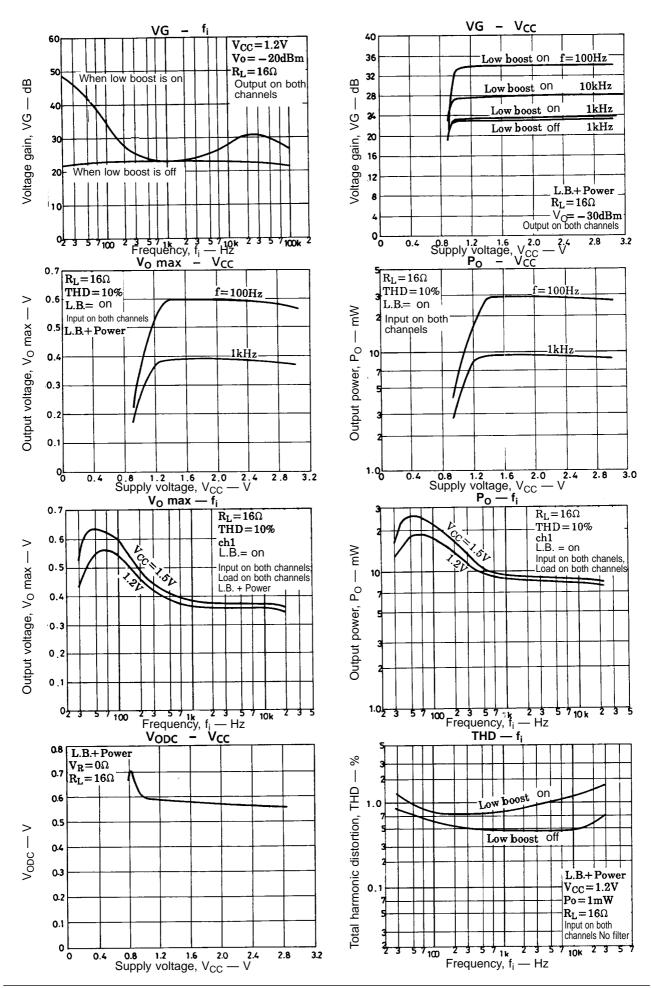
3. FWD/REV, METAL switch

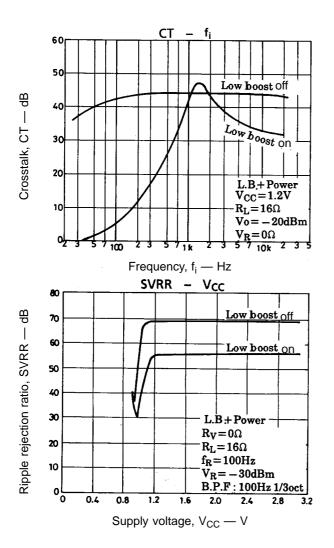


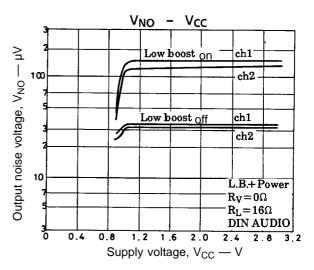












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