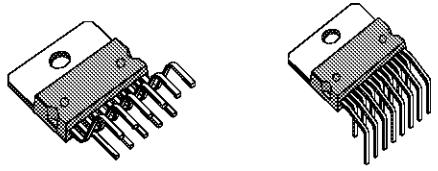


DUAL MULTIFUNCTION VOLTAGE REGULATOR

ADVANCE DATA

- STANDBY OUTPUT VOLTAGE PRECISION 5V $\pm 2\%$
- OUTPUT 2 TRACKED TO THE STANDBY OUTPUT
- OUTPUT 2 DISABLE FUNCTION FOR STANDBY MODE
- VERY LOW QUIESCENT CURRENT, LESS THAN $250\mu\text{A}$, IN STANDBY MODE
- OUTPUT 2 VOLTAGE SETTABLE FROM 5 TO 20V
- OUTPUT CURRENTS : $I_{O1} = 50\text{mA}$, $I_{O2} = 500\text{mA}$
- VERY LOW DROPOUT (max 0.4V/0.6V)
- OPERATING TRANSIENT SUPPLY VOLTAGE UP TO 40V
- POWER-ON RESET CIRCUIT SENSING THE STANDBY OUTPUT VOLTAGE
- POWER-ON RESET DELAY PULSE DEFINED BY THE EXTERNAL CAPACITOR
- EARLY WARNING OUTPUT FOR SUPPLY UNDERVOLTAGE
- THERMAL SHUTDOWN AND SHORT CIRCUIT PROTECTIONS

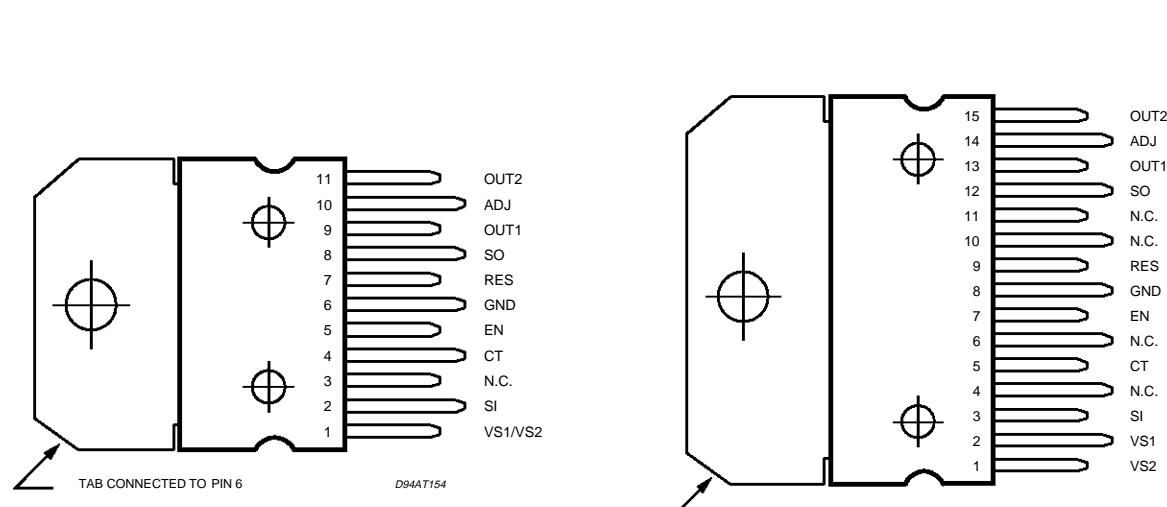

Multiwatt 11
Multiwatt 15

ORDERING NUMBERS: L4936N (Multiwatt 11)
L4936NH (Multiwatt 15)

DESCRIPTION

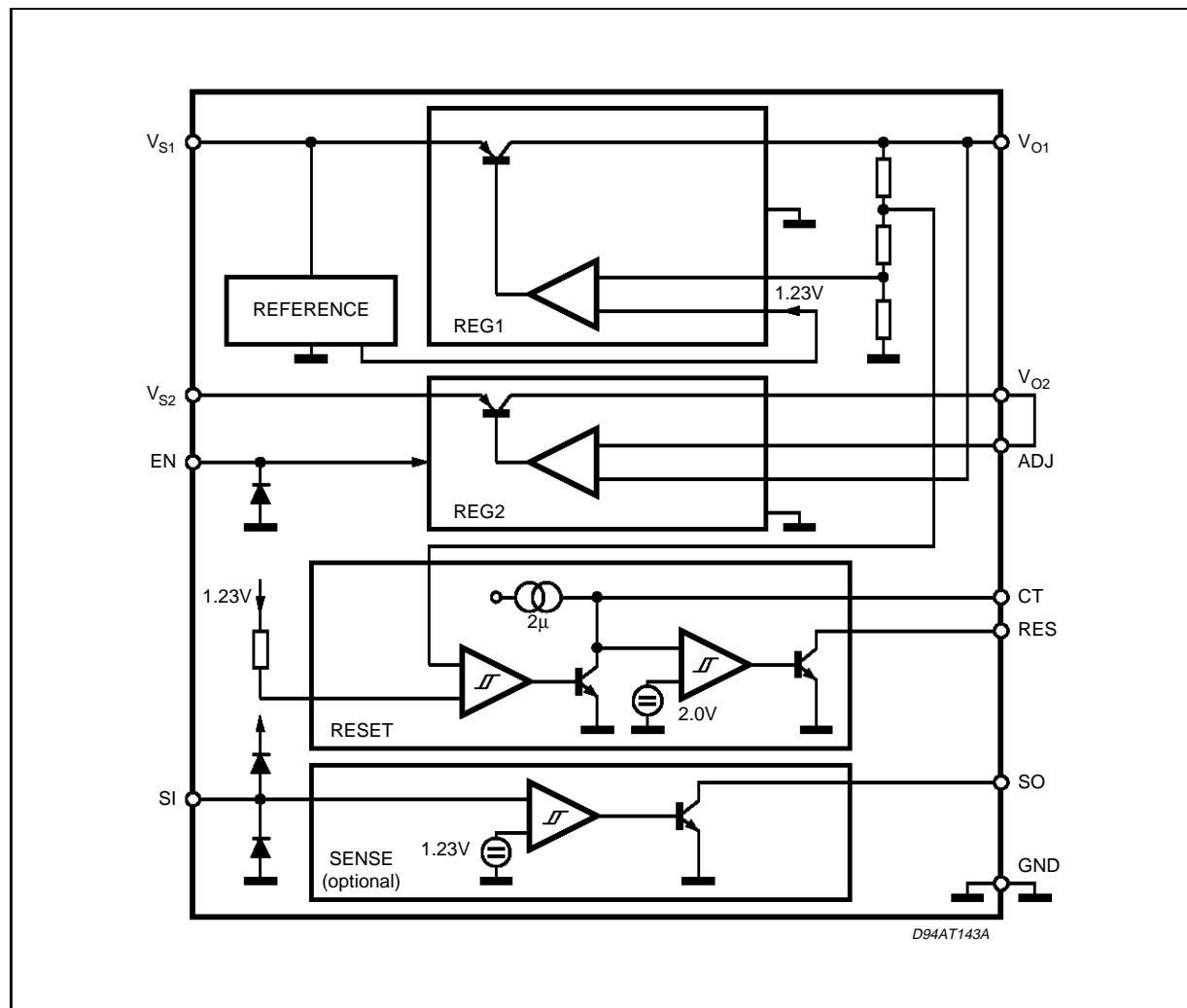
The L4936N/NH are monolithic integrated dual voltage regulators with two very low dropout outputs and additional functions such as power-on reset and input voltage sense. They are designed for supplying microcomputer controlled systems specially in automotive applications.

PIN CONNECTION (top view)



L4936N/NH

BLOCK DIAGRAM



THERMAL DATA

Symbol	Parameter	Value	Unit
$R_{thj-case}$	Thermal Resistance Junction-Case	Max.	3 $^{\circ}\text{C}/\text{W}$

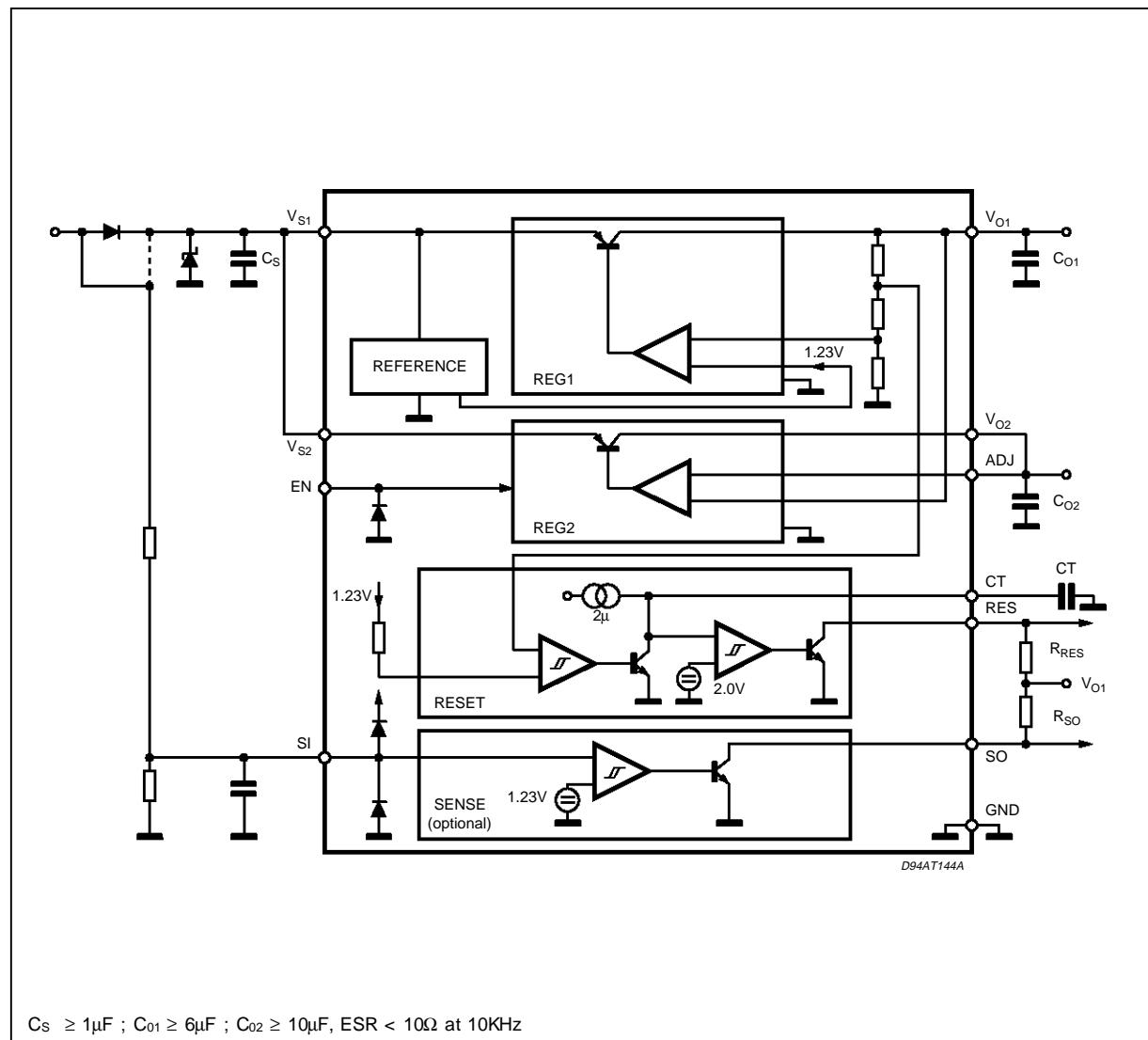
Note: VS1, VS2, SI pins are rated at 1500V versus positive ESD according to human body model; other pins are rated at 2000V.

ABSOLUTE MAXIMUM RATINGS

Symbol	Parameter	Value	Unit
V _S	DC Supply Voltage	28	V
	Transient Supply Voltage (T < 1s)	40	V
T _j , T _{stg}	Junction and Storage Temperature Range	-55 to 150	°C
I _{SI}	Sense Input Current (V _{SI} ≤ 0.3V or V _{SI} > V _S)	±1	mA
I _{EN}	Enable Input Current (V _{EN} ≤ 0.3V)	-1	mA
V _{EN}	Enable Input Voltage	V _S	
V _{RES} , V _{SO}	Reset and Sense Output Voltage	20	V
I _{RES} , I _{SO}	Reset and Sense Output Current	5	mA
P _D	Power Dissipation (T _A = 80°C, R _{th} heatsink = 11°C/W) MW11 POWERDIP	5 1166	W mW

Note : The circuit is ESD protected according to MIL-STD-883C.

APPLICATION CIRCUIT



C_S ≥ 1µF ; C_{O1} ≥ 6µF ; C_{O2} ≥ 10µF, ESR < 10Ω at 10KHz

L4936N/NH

ELECTRICAL CHARACTERISTICS (VS = 14V; -40°C unless otherwise specified)

Symbol	Parameter	Test Conditions	Min.	Typ.	Max.	Unit
VS	Operating Supply Voltage				25	V
VO1	Standby Output Voltage	6V ≤ VS ≤ 25V 1mA ≤ IO1 ≤ 50mA	4.90	5.00	5.10	V
VO2 - VO1	Output Voltage 2 Tracking Error (note 1)	6V ≤ VS ≤ 25V 5mA ≤ IO2 ≤ 500mA Enable = LOW	-25		+25	mV
IADJ	ADJ Input Current	IO1 = 1mA; IO2 = 5mA	-1	0.1	1	μA
VDP1	Dropout Voltage 1	IO1 = 10mA IO1 = 50mA		0.1 0.2	0.25 0.4	V V
VI01	Input to Output Voltage Difference in Undervoltage Condition	VS = 4V, IO1 = 35mA			0.4	V
VDP2	Dropout Voltage 2	IO1 = 100mA IO1 = 500mA		0.2 0.3	0.3 0.6	V V
VI02	Input to Output Voltage Difference in Undervoltage Condition	VS = 4.6V, IO1 = 350mA			0.6	V
VOL 1.2	Line Regulation	6V ≤ VS ≤ 25V IO2 = 1mA; IO2 = 5mA			20	mV
VOLO1	Load Regulation 1	1mA ≤ IO1 ≤ 50mA			25	mV
VOLO2	Load Regulation 2	5mA ≤ IO2 ≤ 500mA			50	mV
ILIM1	Current Limit 1	VO1 = 4.5V VO1 = 0V (note 2)	55 25	100 50	200 100	mA mA
ILIM2	Current Limit 2	VO2 = 0V	550	1000	1700	mA
IQS	Quiescent Current Standby Mode (output 2 disabled)	IO1 = 0.3mA; TJ < 100°C VEN ≥ 2.4V VS = 14V VS = 3.5V		210 340	290 850	μA μA
IQ	Quiescent Current	IO1 = 50mA IO1 = 500mA			30	mA

ENABLE

VENL	Enable Input LOW Voltage (output 2 active)		-0.3		1.5	V
VENH	Enable Input HIGH Voltage		2.4		7	V
VENhyst	Enable Hysteresis		30	75	200	mV
IEN	Enable Input Current	0V < VEN < 1.2V 2.5V < VEN < 7V	-10 -1	-1.5 0	-0.5 +1	μA μA

ELECTRICAL CHARACTERISTICS (continued)**RESET**

Symbol	Parameter	Test Conditions	Min.	Typ.	Max.	Unit
V_{Rt}	Reset Low Threshold Voltage	$V_S = 14V$	$V_{01} - 0.4$	4.7	$V_{01} - 0.1$	V
V_{Rth}	Reset Threshold Hysteresis		50	100	200	mV
t_{RD}	Reset Pulse Delay	$C_T = 100nF; t_R > 100\mu s$	55	100	180	mV
t_{RR}	Reset Reaction Time	$C_T = 100nF$	1	0	50	\mu s
V_{RL}	Reset Output LOW Voltage	$R_{RES} = 10K\Omega$ to V_{01} $V_S = 1.5V$			0.4	V
I_{LRES}	Reset Output HIGH Leakage	$V_{RES} = 5V$			1	\mu A
V_{CTh}	Delay Comparator Threshold			2.0		V
$V_{CTh, hyst}$	Delay Comparator Threshold Hysteresis			100		mV

SENSE

V_{Sth}	Sense Threshold Voltage		1.16	1.23	1.35	V
$V_{Sth, hyst}$	Sense Threshold Hysteresis		20	100	200	mV
V_{SOL}	Sense Output LOW Voltage	$V_{SI} = 0.8V; V_S \geq 3V$ $R_{SO} = 10K\Omega$ to V_{01}			0.4	V
I_{LSO}	Sense Output Leakage	$V_{SO} = 5V; V_{SI} \geq 1.5V$			1	\mu A

Note : 1 : V_{O2} connected to ADJ. V_{O2} can be set to higher values by inserting an external resistor divider.
2 : Foldback characteristic

FUNCTIONAL DESCRIPTION

The L4936N/NH are based on the SGS-THOMSON Microelectronics modular voltage regulator approach. Several out-standing features and auxiliary functions are provided to meet the requirements of supplying the microprocessor systems used in automotive applications.

Furthermore the device is suitable also in other applications requiring two stabilized voltages.

The modular approach allows other features and functions to be realized easily when required.

STANDBY REGULATOR

The standby regulator uses an Isolated Collector Vertical PNP transistor as the regulating element. This structure allows a very low dropout voltage at currents up to 50mA. The dropout operation of the standby regulator is maintained down to 2V input supply voltage. The output voltage is regulated up to the transient input supply voltage of 40V. This feature avoids functional interruptions which could be generated by overvoltage pulses.

The typical curve of the standby output voltage as a function of the input supply voltage is shown in fig. 1.

The current consumption of the device (quiescent current) is less than 250\mu A when output 2 is disabled (standby mode). The dropout voltage is controlled to reduce the quiescent current peak in the undervoltage region and to improve the transient response in this region.

The quiescent current is shown in fig. 2 as a function of the supply input voltage 2.

OUTPUT 2 VOLTAGE

The output 2 regulator uses the same output structure as the standby regulator, but rated for an output current of 500mA.

The output 2 regulator works in tracking mode with the standby output voltage as a reference voltage when the output 2 programming pin ADJ is connected to V_{O2} . By connecting a resistor divider R_1 , R_2 to the pin ADJ as shown in fig. 3, the output voltage 2 can be programmed to the value :

$$V_{O2} = V_{O1} (1 + R_1/R_2)$$

The output 2 regulator can be switched off via the Enable input.

L4936N/NH

Figure 1 : Output Voltage vs. Input Voltage.

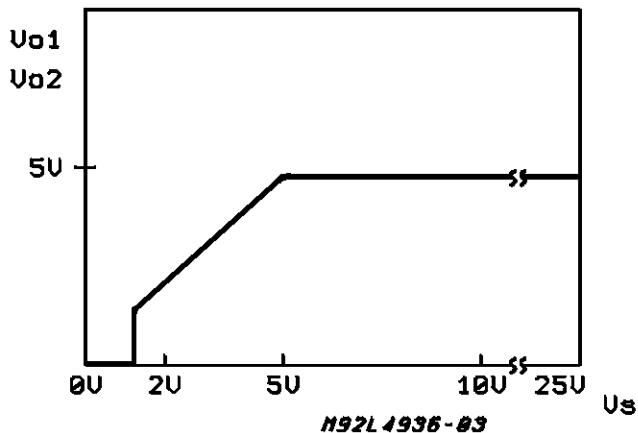


Figure 2 : Quiescent Current vs. Supply Voltage.

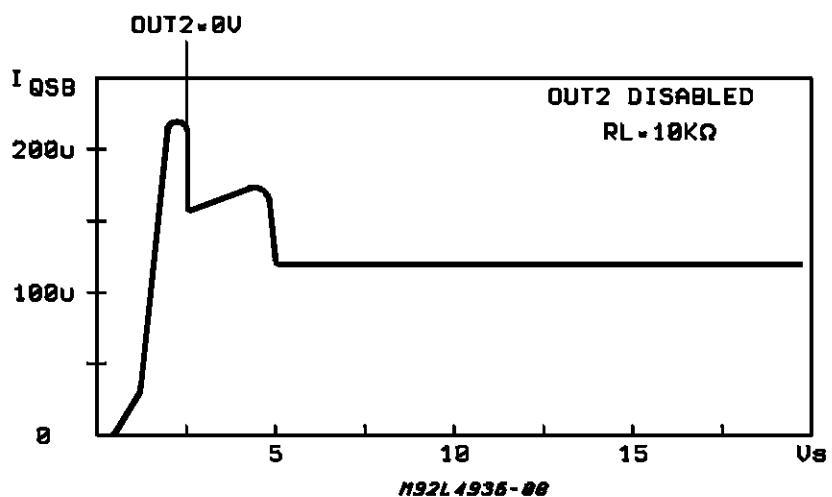
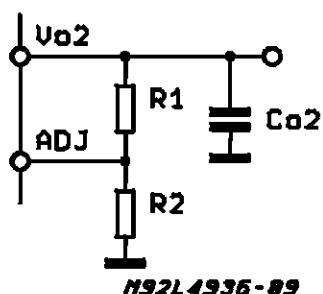


Figure 3 : Programmable Output 2 Voltage with External Resistors.



RESET CIRCUIT

The block circuit diagram of the reset circuit is shown in fig. 4. The reset circuit supervises the standby output voltage. The reset threshold of 4.7V is defined by the internal reference voltage and the standby output divider.

The reset pulse delay time t_{RD} , is defined by the charge time of an external capacitor C_T :

$$t_{RD} = \frac{C_T \times 2V}{2\mu A}$$

The reaction time of the reset circuit depends on the discharge time limitation of the reset capacitor C_T and is proportional to the value of C_T .

The reaction time of the reset circuit increases the noise immunity. In fact, if the standby output voltage

drops below the reset threshold for a time shorter than the reaction time t_{RR} , no reset output variation occurs. The nominal reset delay is generated for standby output voltage drops longer than the time necessary for the complete discharging of the capacitor C_T . This time is typically equal to 50μs if $C_T = 100nF$. The typical reset output waveforms are shown in fig. 5.

SENSE COMPARATOR

This circuit compares an input signal with an internal voltage reference of typically 1.23V. The use of an external voltage divider makes the comparator very flexible in the application. This function can be used to supervise the input voltage - either before or after the protection diode - and to give additional information to the microprocessor such as low voltage warnings.

Figure 4 :Block Diagram of the Reset Circuit.

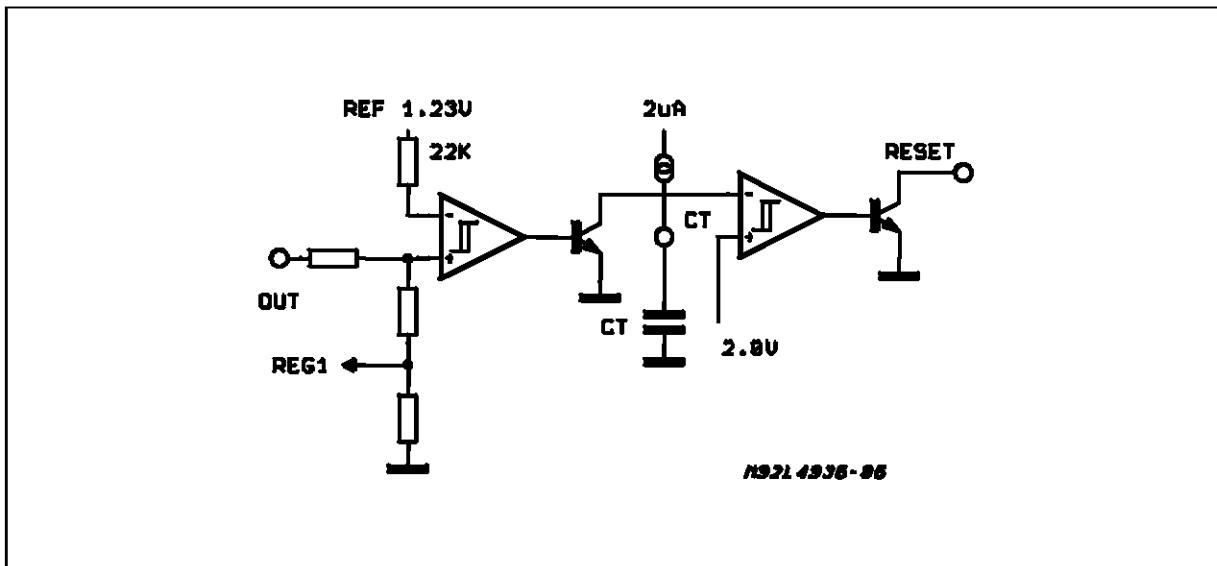
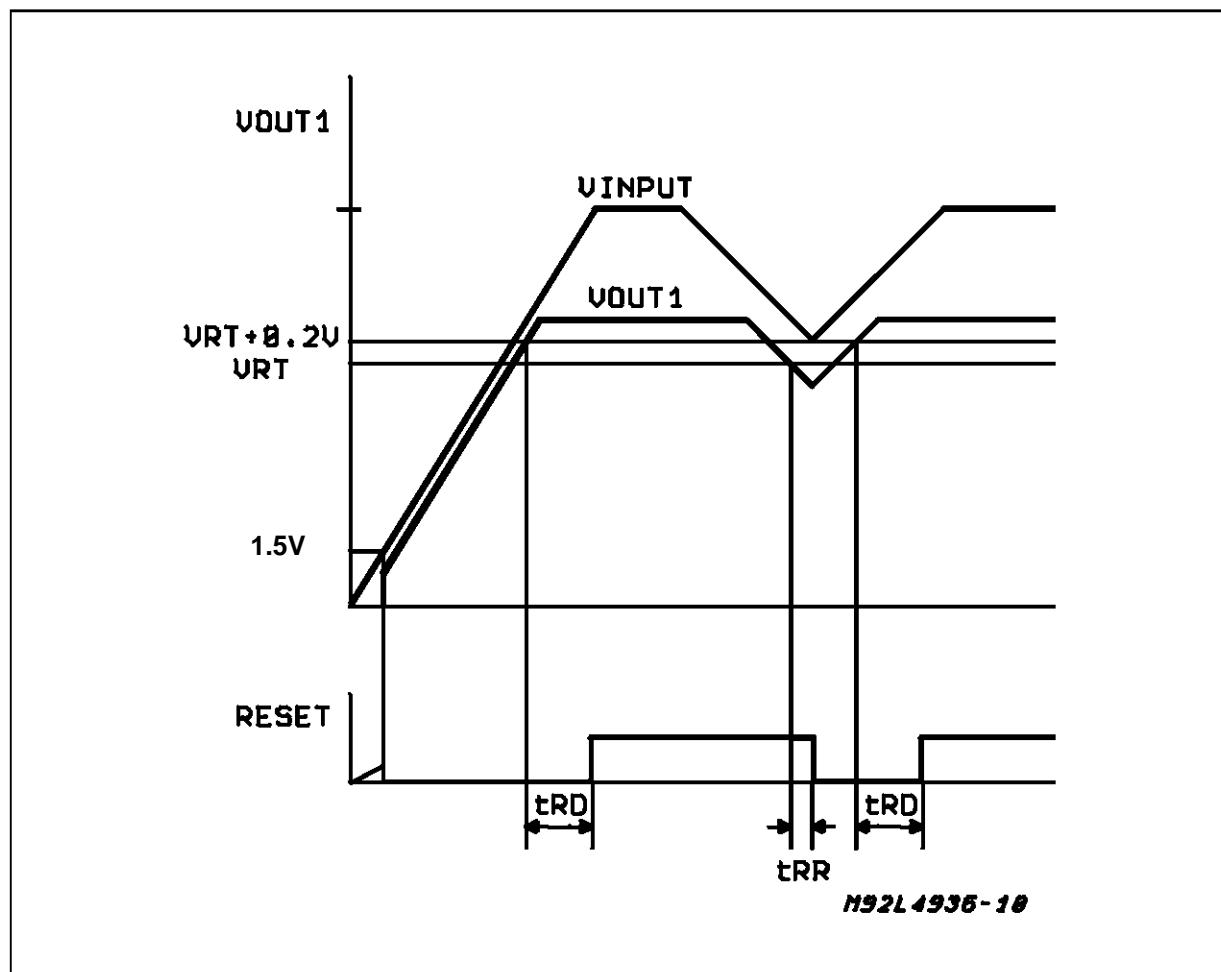
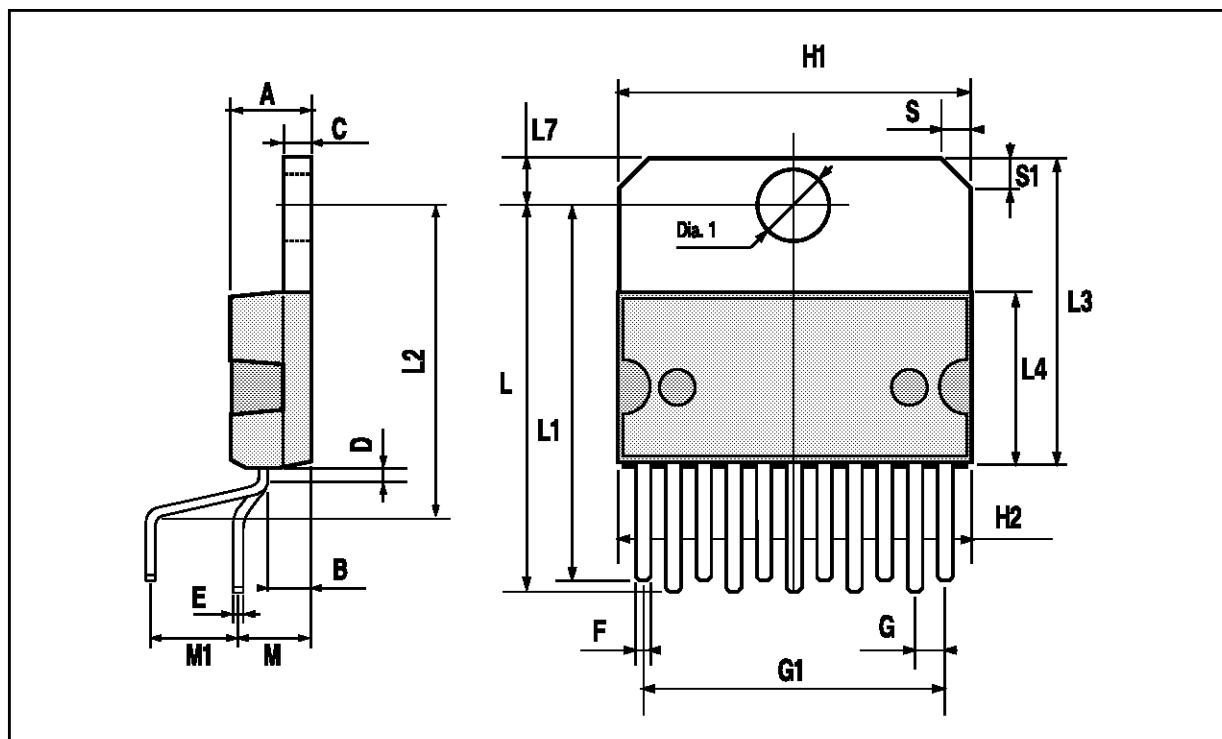


Figure 5 : Typical Reset Output Waveforms.



MULTIWATT11 PACKAGE MECHANICAL DATA

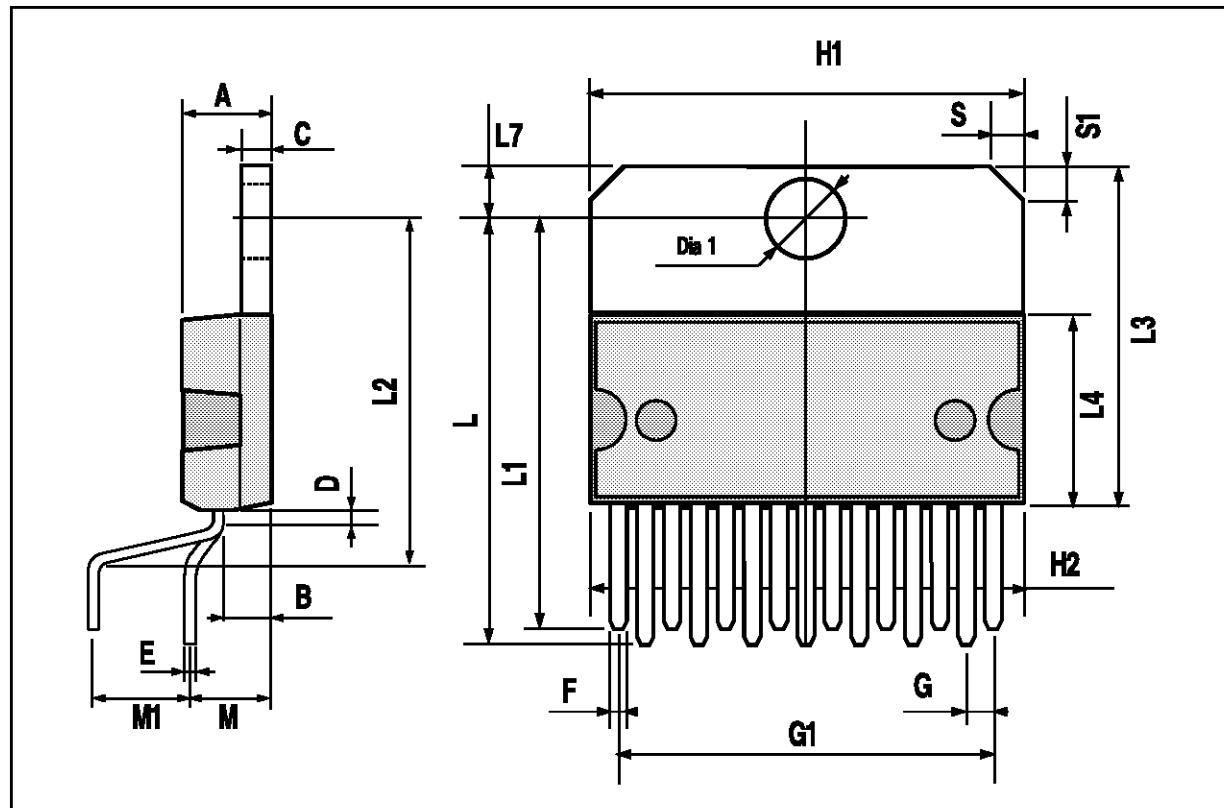
DIM.	mm			inch		
	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
A			5			0.197
B			2.65			0.104
C			1.6			0.063
D		1			0.039	
E	0.49		0.55	0.019		0.022
F	0.88		0.95	0.035		0.037
G	1.45	1.7	1.95	0.057	0.067	0.077
G1	16.75	17	17.25	0.659	0.669	0.679
H1	19.6			0.772		
H2			20.2			0.795
L	21.9	22.2	22.5	0.862	0.874	0.886
L1	21.7	22.1	22.5	0.854	0.87	0.886
L2	17.4		18.1	0.685		0.713
L3	17.25	17.5	17.75	0.679	0.689	0.699
L4	10.3	10.7	10.9	0.406	0.421	0.429
L7	2.65		2.9	0.104		0.114
M	4.25	4.55	4.85	0.167	0.179	0.191
M1	4.73	5.08	5.43	0.186	0.200	0.214
S	1.9		2.6	0.075		0.102
S1	1.9		2.6	0.075		0.102
Dia1	3.65		3.85	0.144		0.152



L4936N/NH

MULTIWATT15 PACKAGE MECHANICAL DATA

DIM.	mm			inch		
	MIN.	TYP.	MAX.	MIN.	TYP.	MAX.
A			5			0.197
B			2.65			0.104
C			1.6			0.063
D		1			0.039	
E	0.49		0.55	0.019		0.022
F	0.66		0.75	0.026		0.030
G	1.02	1.27	1.52	0.040	0.050	0.060
G1	17.53	17.78	18.03	0.690	0.700	0.710
H1	19.6			0.772		
H2			20.2			0.795
L	21.9	22.2	22.5	0.862	0.874	0.886
L1	21.7	22.1	22.5	0.854	0.870	0.886
L2	17.65		18.1	0.695		0.713
L3	17.25	17.5	17.75	0.679	0.689	0.699
L4	10.3	10.7	10.9	0.406	0.421	0.429
L7	2.65		2.9	0.104		0.114
M	4.25	4.55	4.85	0.167	0.179	0.191
M1	4.63	5.08	5.53	0.182	0.200	0.218
S	1.9		2.6	0.075		0.102
S1	1.9		2.6	0.075		0.102
Dia1	3.65		3.85	0.144		0.152



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