



# **DIFFERENTIAL BUS TRANSCEIVER**

## **FEATURES**

- FULL-/HALF-DUPLEX OPERATION
- 1500Vrms ISOLATION (cont)
- 2500Vrms ISOLATION (1 min)
- 2.5Mbps PERFORMANCE
- LOOP-TEST FACILITY

## **APPLICATIONS**

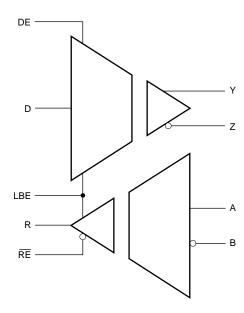
- BUS TRANSMISSION SYSTEMS
- GROUND LOOP ISOLATION

## **DESCRIPTION**

ISO422 provides 1500Vrms isolation for industrial bus transmission systems. ISO422 may be configured in full or half duplex modes providing the user with best flexibility for the application. Transmission rates of 2.5Mbps can be obtained covering most requirements. A loop-back test facility is included. LBE allows data on the D input to be routed to the R output for test purposes.

ISO422 is available in 24-pin PDIP and 24-pin Gull Wing<sup>(1)</sup> packages and is specified over the temperature range -40°C to +85°C.

NOTE: (1) Gull Wing version available Q1'99.



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# **SPECIFICATIONS**

At  $T_A = +25^{\circ}C$ , and  $V_S = +5V$ , unless otherwise noted.

				ISO422P, P-U(1)		_
PARAMETER		CONDITIONS	MIN	TYP	MAX	UNITS
SOLATION						
Rated Continuous Isolation	$V_{ISO}$	50Hz, 60Hz	1500			V
Partial Discharge Voltage	- 130	1s, 5 x 5pC/per cycle <sup>(2)</sup>	2500			V
Barrier Impedance		10, 0 x op o/por oyolo	2000	> 10 <sup>14</sup>    10		Ω    pF
Leakage Current		240V, 60Hz		1		
leakage Current				'	10	μA 
D		2500V, 50Hz		0.0	10	μΑ
Creepage Distance				8.6		mm
nternal Isolation Distance				0.1		mm
Fransient Recovery Time		5kV/μs Edge			1	μs
DRIVER DC CHARACTERISTICS						
High Level Input Voltage	$V_{IH}$	D and DE Inputs(3)	2			V
Low Level Input Voltage	V <sub>IL</sub>	D and DE Inputs <sup>(3)</sup>			0.8	V
nput Leakage Current	ال	D and DE Inputs <sup>(3)</sup>		5	0.0	nA
		D and DE Inputs <sup>(3)</sup>		5		pF
nput Capacitance	C <sub>IN</sub>	•		5	_	
Output Voltage	Vo	$V_{\rm Y}$ or $V_{\rm Z}$	0		5	V
Differential Output Voltage	V <sub>OD</sub>	$I_{OY}$ or $I_{OZ} = 0$	1.5		5	V
		$R_L = 100\Omega$	2	3.6	5	V
		$R_L = 54\Omega$	1.5	2.8	5	V
Change in Mag Diff Out Voltage	$\Delta  V_{OD} $	$R_L = 100\Omega$ or $54\Omega^{(4)}$		±40	±200	mV
Common-Mode Output Voltage	V <sub>oc</sub>	$R_L = 100\Omega$ or $54\Omega$			3	V
Change in Mag CM Out Voltage	$\Delta  V_{OC} $	$R_1 = 100\Omega \text{ or } 54\Omega^{(4)}$		±40	±200	mV
Output Current	I <sub>O</sub>	$V_O = V_{CC2}$ , Output Disabled		±10	±1000	nA
Julput Gurrent	'0					
		$V_O = 0V$ , Output Disabled		±10	±1000	nA
Short-Circuit Output Current		$V_O = V_{CC2}$ , Continuous		100		mA
		$V_O = 0V$ , Continuous		-110		mA
DRIVER SWITCHING CHARACTERIS	TICS (Figur	e 6)				
Differential Output Delay Time	t <sub>DD</sub>	$R_1 = 54\Omega$		120	150	ns
Skew  t <sub>DDH</sub> - t <sub>DDL</sub>	55	$R_1 = 54\Omega$		25	50	ns
Differential Output Transition Time	t <sub>DT</sub>	$R_1 = 54\Omega$			100	ns
Output Enable Time to HIGH		$R_L = 100\Omega$		120	150	ns
•	t <sub>DZH</sub>			120		
Output Enable Time to LOW	t <sub>DZL</sub>	$R_L = 100\Omega$		-	150	ns
Output Disable Time from HIGH	t <sub>DHZ</sub>	$R_L = 100\Omega$		120	150	ns
Output Disable Time from LOW	t <sub>DLZ</sub>	$R_L = 100\Omega$		120	150	ns
RECEIVER DC CHARACTERISTICS						
High Level Output Voltage	$V_{OH}$	$I_{OH} = 6mA$	V <sub>CC</sub> - 1			V
Low Level Output Voltage	$V_{OL}$	$I_{OL} = 6mA$			0.4	V
Output Short-Circuit Current	Ios	1s max		30		mA
Output HI-Z Leakage	I <sub>OZ</sub>	$V_{OUT} = 0V \text{ to } V_{CC1}$		±10	±1000	nA
Enable Input HIGH Threshold	V <sub>IH</sub>	RE Input <sup>(3)</sup>	2		21000	V
•		_ '			0.0	V
Enable Input LOW Threshold	V <sub>IL</sub>	RE Input <sup>(3)</sup>		_	0.8	
nput Leakage Current	Ι <sub>L</sub>	RE Input <sup>(3)</sup>		5		nA
nput Capacitance	$C_{IN}$	RE Input <sup>(3)</sup>		5		pF
Differential Input HIGH Threshold	$V_{TH}$	$V_O = 2.8V$		100	200	mV
Differential Input LOW Threshold	$V_{TL}$	$V_O = 0.4V$	-200	-100		mV
nput Hysteresis		See Note 5		60		mV
ine Input Current	I <sub>BI</sub>	Power On $(GND_B < V_{BI} < V_{SB})$		±10	±1000	nA
ine Voltage	V <sub>BI</sub>	Power Off (I <sub>BI</sub> ±10mA max)		±12		V
nput Resistance	R <sub>IN</sub>		1			ΜΩ
RECEIVER SWITCHING CHARACTER	RISTICS (Fi	qure 7)				
Propagation Delay L to H	t <sub>RLH</sub>	$V_{ID} = -1.5V \text{ to } 1.5V, C_{I} = 10pF$		120	150	ns
Propagation Delay H to L		$V_{ID} = 1.5V \text{ to } 1.5V, C_{L} = 10pF$		120	150	ns
	t <sub>RHL</sub>	v <sub>ID</sub> = 1.5 v to -1.5 v, O <sub>L</sub> = 10pF			130	
Skew  t <sub>RLH</sub> - t <sub>RHL</sub>				40		ns
	t <sub>R</sub>	$C_L = 10pF$		10		ns
Output Rise Time		$C_L = 10pF$	1	10		ns
Dutput Rise Time Dutput Fall Time	t <sub>F</sub>	or .ob.	1			
•	ι <sub>F</sub> t <sub>RZH</sub>	$C_L = 10pF$		15	25	ns
Output Fall Time Output Enable Time to HIGH	$t_{RZH}$	$C_L = 10pF$				
Output Fall Time				15 15 15	25 25 25	ns ns ns



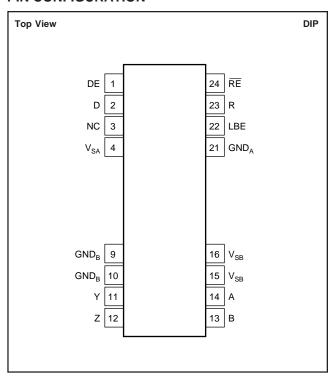
# SPECIFICATIONS (CONT)

At  $T_A = +25$ °C, and  $V_S = +5V$ , unless otherwise noted.

			ISO422P, P-U <sup>(1)</sup>			
PARAMETER		CONDITIONS	MIN	TYP	MAX	UNITS
POWER						
Supply Voltage—Data Side	V <sub>SA</sub>		4.5		5.5	V
Supply Current—Data Side	I <sub>SA</sub>	Output Unloaded, dc		10	13	mA
Supply Current—Data Side	I <sub>SA</sub>	Output Unloaded, max Rate		20		mA
Supply Voltage—Bus Side	V <sub>SB</sub>		4.5		5.5	V
Supply Voltage—Bus Side	I <sub>SB</sub>	Output Unloaded, dc		12	20	mA
		Output Unloaded, max Rate		20		mA
BUS LIMITS						
Input Current					±10	mA
Maximum Differential Input					±5	V
Maximum Data Rate				2.5		Mbps
TEMPERATURE RANGE						
Operating			-40		+85	°C
Storage			-40		+125	°C
Thermal Resistance	$ heta_{\sf JA}$			75		°C/W

NOTES: (1) Gull Wing version available Q1'99. (2) All devices receive a 1s test. Failure criterion is > 5 pulses of > 5pC per cycle. (3) Logic inputs are HCT-type and thresholds are a function of power supply voltage with approximately 100mV hysteresis. (4) Change in magnitude when the input is changed from HIGH to LOW. (5) The difference between the differential low to high and high to low transition points.

#### **PIN CONFIGURATION**



### **ABSOLUTE MAXIMUM RATINGS**

Supply Voltage: V <sub>SA</sub>	0.5V to +6V
V <sub>SB</sub>	0.5V to +6V
Continuous Isolation Voltage	1500Vrms
Storage Temperature	40°C to +125°C
Lead Temperature (soldering, 10s)	+300°C

## **PACKAGE INFORMATION**

PRODUCT	PACKAGE	PACKAGE DRAWING NUMBER <sup>(1)</sup>
ISO422P	24-Pin Plastic DIP	243-4
ISO422P-U	24-Pin Gull Wing Surface Mount	243-5

NOTE: (1) For detailed drawing and dimension table, please see end of data sheet, or Appendix C of Burr-Brown IC Data Book.



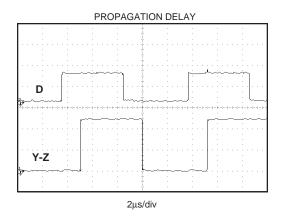
Electrostatic discharge can cause damage ranging from performance degradation to complete device failure. Burr-Brown Corporation recommends that all integrated circuits be handled and stored using appropriate ESD protection methods.

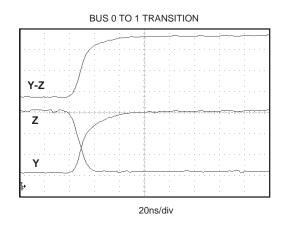
ESD damage can range from subtle performance degradation to complete device failure. Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet published specifications.

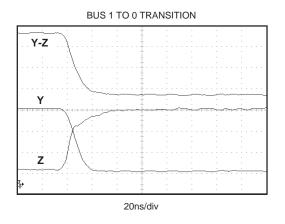
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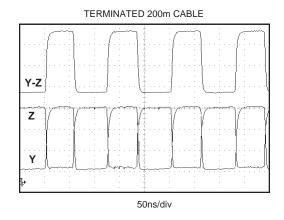
# **TYPICAL PERFORMANCE CURVES**

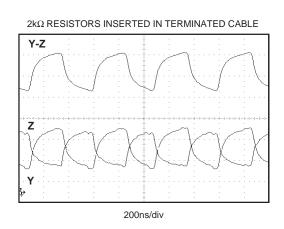
At  $T_A = +25^{\circ}C$ , and  $V_S = +5V$ , unless otherwise noted.











## **OPERATION**

ISO422 is an isolated, full-duplex bus transceiver which is compatible with three-wire data bus systems using EIA standards RS-422-A and RS-485. It is based on Burr-Brown's capacitive barrier technology. The data bus input is designed to present a very high impedance to the data bus, thus allowing a virtually unlimited number of receivers on any data bus section. To allow this feature, the data bus input is limited to a common-mode range within the magnitude of the supplies. This limitation requires that all nodes on the bus are referenced to a common ground. However, systems attached to the bus through ISO422, are isolated up to 1500Vrms and may, therefore, have local floating ground potentials up to this isolation voltage. The circuit encodes all data passed across the barrier to ensure that the input values and control signals are correctly passed across the barrier under all power up conditions. The ISO422 also allows data recovery to the current input state, after any transient upset.

#### **TRANSMIT**

Data is passed from the D input to the data bus outputs after a barrier transmission delay ( $t_{DD}$ ) when the DE input is HIGH. When DE is LOW, the data bus drivers are switched off, and assume the high impedance state. When enabling the data bus output, i.e., switching DE from LOW to HIGH, the enable signal is passed directly across the barrier and enables the output, after a barrier transmission delay and output enable time ( $t_{DLZ}/t_{DHZ}$ ). Similarly, when disabling the data bus output, i.e., switching DE from HIGH to LOW, the disable signal is passed directly across the barrier and disables the output after a barrier transmission delay and output disable time ( $t_{DLZ}/t_{DHZ}$ ).

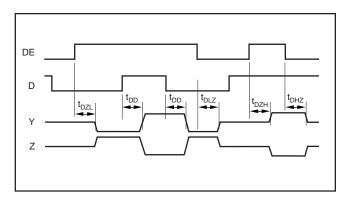


FIGURE 1. ISO422 Data Transmit.

### **RECIEVE**

The receive data is determined by the data bus differential signal after a barrier transmission delay ( $t_{RZL}$ ). When the difference between the A input and the B input (A-B) is greater than  $+200\,\mathrm{mV}$ , the R output will be HIGH. If A-B is more negative than  $-200\,\mathrm{mV}$ , the R output is undefined. Since the reciver has a high impedance input, no disable signal is required for the data bus input, which is always

active. The receive enable/disable time is simply the time to enable/disable the R output  $(t_{RLZ})$  and does not require any additional barrier transmission time.

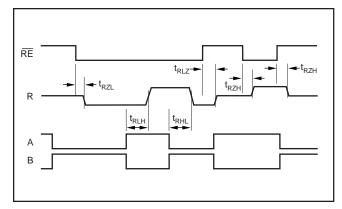


FIGURE 2. ISO422 Data Receive.

### **DATA CORRUPTION**

If, due to transient upset, the data passed across the barrier is corrupted, the data will be restored within 100ns from the end of the corrupting signal.

#### **SYNCHRONIZATION**

The data transmitted across the barrier is coded using an internal clock. This clock also captures the incoming asynchronous data and synchronizes it to the clock edges. This will give rise to an rms propagation delay jitter of approximately 50ns.

### **LOOPBACK**

A loopback function is provided by the LBE input. If this input is HIGH, then enabling both the transmitter and the receiver will cause the device to route the D input to the R output, in addition to the data bus outputs. Data on the incoming bus is ignored. This feature allows a simple connection test to be performed during any application. When LBE is LOW, transmit and receive will operate in the normal full-duplex mode.

### **DATA BUS CONNECTION**

ISO422 can be used in half duplex, or full duplex data communication bus systems. It is capable of continuously driving a  $54\Omega$  load, equivalent to a double-terminated transmission line, at the fully specified data rate. When connecting to the data bus, the voltage on the A and B input lines must remain between  $V_{SB}$  and  $GND_{B}$ . This can be achieved by using a common bus ground connection, such as  $GND_{B}$ , as shown in Figures 5 and 6.

For any system connected to the bus, the isolation provided by ISO422 allows the independent local ground potential to be as high as 1500Vrms with respect to the date bus ground reference. This feature replaces the limited +12V to -7V range of the RS-485 standard with the full-isolation voltage capability of the ISO422.

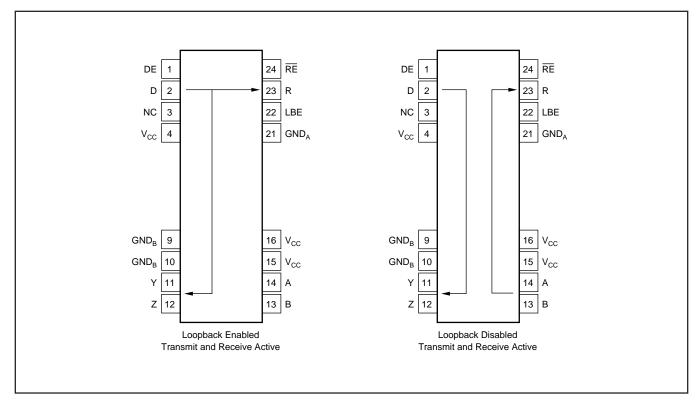


FIGURE 3. Loopback.

### **CONNECTION TO CAN BUS**

Since the bus can be enabled and disabled at the same rate as the data (2.5MHz), it is possible to use ISO422 as an isolated bus driver in CAN systems. Again, the ISO422 bus line must be constrained within the supply voltages.

Figure 4 shows the connections which allow ISO422 to be used in CANbus systems. The DE input of the ISO422 is used as the CAN TX0 input and is used to transmit the data by enabling and disabling the Y and Z outputs. The D and RE inputs of the ISO422 are tied to GND<sub>A</sub>. This ensures that the Y output can only pull down, and the Z output can only pull up. With D tied to GND<sub>A</sub>, the DE input of ISO422 (TX0 of CAN) activates the Y output as an open drain pull-down driver, and activates the Z output as an open drain pull-up driver. Therefore, the Y line acts as CANL and the Z line acts as CANH. When DE (TX0) is HIGH, ISO422 makes the bus state dominant i.e., Y pulls LOW and Z pulls HIGH. With DE (TX0) LOW, Y and Z are high impedance and the bus state is recessive. Data is received in the normal manner which is half duplex. Line A is connected to CANH, and line B is connected to CANL. The R output becomes RX0. RE is tied to GNDA to keep R (RX0) enabled. If required, RE may be used to disable the RX0 output.

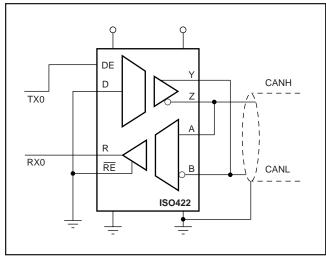


FIGURE 4. CANBus Connection.

TX0	CANH	CANL	BUS	RX0
Н	Н	L	Dominant	L
L	Hi-Z	Hi-Z	Recessive	Н

TABLE I. CAN.

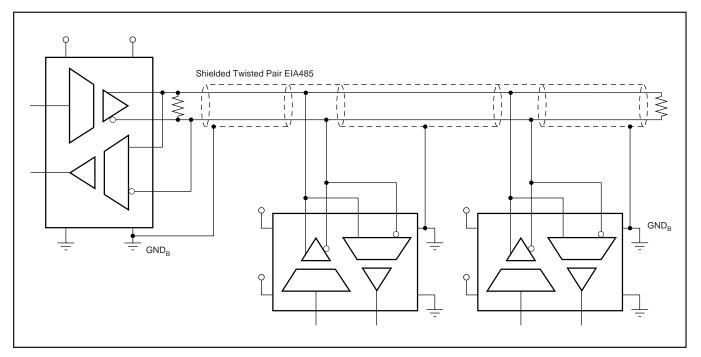
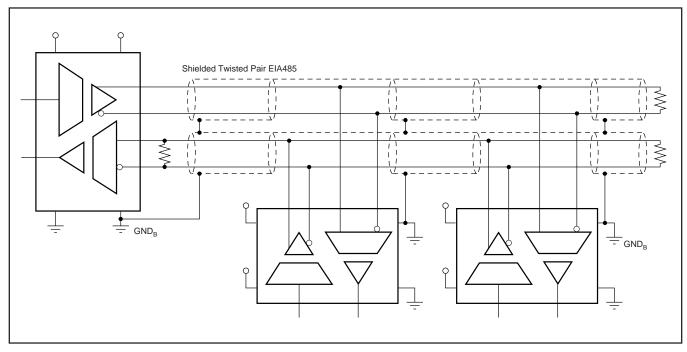


FIGURE 5. Half-Duplex Connection.



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FIGURE 6. Full-Duplex Connection.

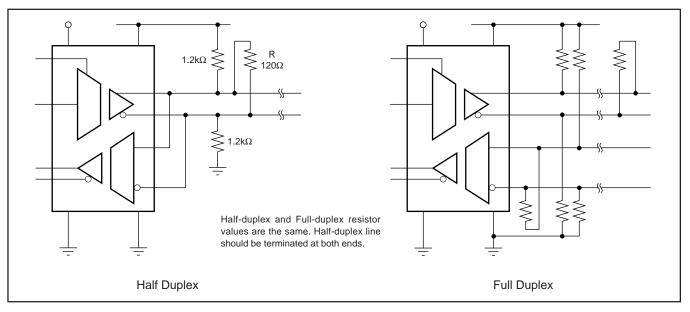


FIGURE 7. Suggested Bus Termination Methods.

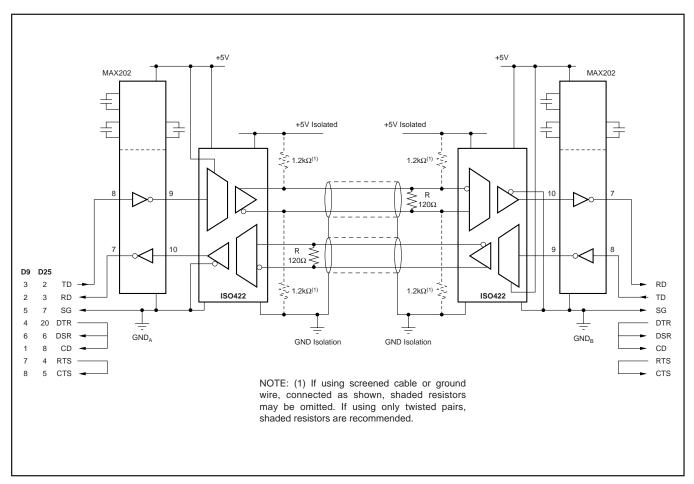


FIGURE 8. Isolated RS232 to RS422. Null Modem Configuration.