Technical Data	P 1	
Accessories	Z 1	
Operating Instructions		
General Information	М1	
Use of tilt handle	М1	
Safety	М1	
Operating conditions	M 2	
Warranty	M 2	
Maintenance	M 2	
	M 2	
	M 3	
Amplitude Measurements	M 3	
Time Measurements	M 4	
	M 4	
Connection of Test Signal		
	M 6	
	M 7	
	M 7	
	M 7	
	M 8	
Operating Modes of the Y Amplifier	M 9	
X-Y Operation	MI 0	
X-Y Phase Measurements	MI 0	
Dual Trace Phase Difference Measurements M	10	
Measurement of an amplitude modulation	MI 1	
Triggering and Timebase	MI 1	
Triggering of video signals	MI 2	
Function of variable HOLD OFF control	MI3	
Sweep Delay /After Delay Triggeringe	MI 3	
Delay Mode Indication	MI 5	
Component Tester	MI 5	
Miscellaneous	MI 7	
Test Patterns	MI 8	
Short Instruction	Κ1,	
Front Panel Elements		
Folder with Front View	Κ2	
Test Instructions		
General	T 1	
Cathode-Ray Tube: Brightness, Focus,		
Linearity, Raster Distortions .	т 1	
Astigmatismus Check		
Symmetry and Drift of the Vertical Amplifier		
Calibration of the Vertical Amplifier		
-	1 1	
Transmission Performance of the Vertical Amplifier	т о	
	12	
Operating Modes: CH I/II-TRIG. I/II, DUAL, ADD, CHOP., INV. I/II and XY-Betrieb	Τク	
Triggering Checks		
Trace Alignment		
Miscellaneous	Τ4	

Oscilloscope HM 604

Service Instructions

General	S 1
Instrument Case Removal	S 1
Operating Voltages	S 1
Minimum Brightness	S 1
Astigmatismus control	S 1
Trouble Shooting the Instrument	S 2
Replacement of Components and Parts	S 2
Replacement of the Power Transformer	S 2
Adjustments	S 3

Circuit Diagrams

Block Diagram	I
Wiring Diagram	2
Identification of Components D	3
Y Input, Attenuator, Preamplifier CH. I/II D	4
Y intermediate Amplifiers, Trigger Pre-Amplifiers,	
Component Tester	5
Y Final Amplifier	6
Post Trigger, Field Selector	7
Timebase (analog)	8
Timebase (digital)	9
Timebase Generator	0
X Final Amplifier, Calibrator	1
CRT and HV circuit	2
Power Supply	3

Component Locations

XY Board				•							D14
TB Board				•						٠	D15
PTFS Board			,					•			D16
TBG, CAL, YF Boards					,						D17
CO, EY, Z Boards											D18

_

Specification

Vertical Deflection

Operating modes: Channel I or Ch. II separate, Channel I and II: alternate or chopped. (Chopper frequency approx. 0.5MHz). Sum or difference of Ch. I and Ch. II, (with invert buttons for both Channels). XY-Mode: via Channel I and Channel II. Frequency range: 2x DC to 60MHz(-3dB). Risetime: approx. 5.8ns. Overshoot: ≤1%. Deflection coefficients: 12 calibrated steps from 5 mV/div. to 20V/div in I-Z-5 sequence, variable 2.5: 1 to min. 50V/cm. Accuracy in calibrated position: ±3%. Y-Magnification x5 (calibrated) to 1 mV/div. (Frequency range DC to 20MHz, – 3dB. input impedance: 1 MΩ II 30 pF. Input coupling: DC-AC-GD (Ground) Input voltage: max. 400V (DC + peak AC). Y-output from CH Ior CH II, ≈ 50 mV_{pp}/div. (50 Ω) Delay Line: approx. 90 ns.

Trigger System

With automatic 10Hz - 100 MHz (≥ 5 mm height) normal with level control from DC- 100 MHz. LED indication for trigger action. Slope: positive or negative. Sources: Ch. I, Ch. II, line, external. Coupling: AC (≥ 10 Hz to approx. 20 MHz), DC LF (DC - ≤ 50 kHz), HF (≥ 50 kHz - 100 MHz). Threshold: external ≥ 50 mV. Active TV-Sync-Separator for line and frame. Slope positive or negative. 2nd. Triggering (Del. Trig.): autom. or slope controlled (independent from slope direction). \pm selection for TV mode. Threshold: 1 div; typical0.5div. Trigger bandwidth: ≥ 25 Hz to 60 MHz.

Horizontal Deflection

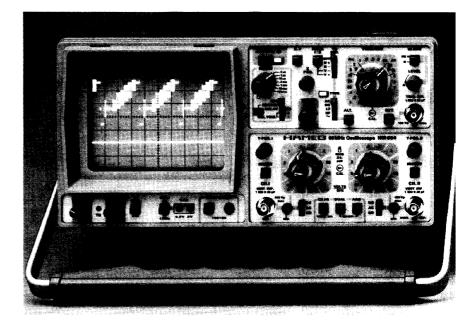
Time coefficients: 23 calibrated steps from 50 ns/div. to 1 s/div in I-2-5 sequence, variable 2.5: 1 to min. 2.5 s/div, accuracy in calibrated position: $\pm 3\%$. with X-Magnifier x10($\pm 5\%$) to ≈ 5 ns/div.. Hold-Off time: variable ($\geq 5:1$). Delay: 7 decade steps from 1 00 ns to 0.1 s, variable approx 10: 1 to 1 s. Bandwidth X-Amplifier: DC-5MHz(-3dB). Input X-Amplifier via Channel II, sensitivity see Ch. II specification. X-Y phase shift: <3° below 120 kHz. Ramp output: approx 5V, positive going

Component Tester

Test voltage: max. 8.5 V_{rms} (open circuit). Test current: max. 8mA_{rms} (shorted). Test frequency: 50 - 60 Hz (line frequency).

General Information

Cathode-ray tube: 150CTB31P43/123, rectangular screen, internal graticule 8x10 cm. Total acceleration voltage: 12 kV. Trace rotation: adjustable on front panel. Calibrator: square-wave generator switchable from \approx 1 kHz to 1 MHz (t_r approx. 3 ns). Output voltage: 0.2V and 2V ±1%. Protective system Safety Class I (IEC 348). Linevoltage: 110. 125, 220, 240V ~± 10%. Line frequency: 50 to 60 Hz. Power consumption: \approx 40 Watt. Weight: approx. 8kg. Colour: techno-brown. Cabinet: W 285, H 145, D 380mm. Lockable tilt handle.



60 MHz Universal Oscilloscope

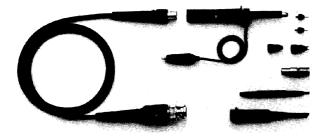
2 Channels, 1 mV/div. Sensitivity, Delay Line, Component Tester Timebase: 2.5s/div. to 5ns/div. including x10 Magnifier&Sweep Delay Triggering: DC-100MHz, TV Sync Separator, After-Delay Trigger

With its variety of operating and trigger modes, the **HM604** is a new innovative **general purpose oscilloscope** satisfying a wide range of exacting requirements in laboratory, production, and service. The dual-channel measurement amplifier ensures highly faithful waveform transfer characteristics, which can be readily checked on the built-in fast-risetime **1 MHz Calibrator** – from probe tip to **CRT** screen! Using Y-axis magnification, the instrument's high sensitivity enables stable displays of very small signals as low as 0.5 mV. An analog output is provided for connecting multimeters or counters. Another important feature is the internal **delay line** for observations of the leading edge of a signal. As in dual-time base oscilloscopes, the **HM604** features a calibrated sweep delay mode, allowing smallest waveform sections to be expanded up to **1000 times.**

The **HM604**'s most outstanding feature, however, is the unique, newly developed automatic **After-Delay Trigger** mode to ensure extremely stable displays and jitter-free measurements of asynchronous signal sections and bursts or pulse trains, independent of amplitude fluctuations. An active **TV-Sync-Separator** further enhances trigger quality of video frame and line signals. In the alternate trigger mode, two signals of different frequencies can be compared.

With this state-of-the-art oscilloscope, **HAMEG** again sets a new price/ performance standard which is not likely to be met by others in this category. Users will be particularly impressed by the instrument's outstanding versatility and ease of operation. These features are possible in the HM604 due to **HAMEG**'s meticulous attention to detail and many decades of successful design experience.

Accessories supplied Two 10:1 probes, Line cord, Operators Manual.



Modular Probes

The clear advantage over ordinary probes are field replaceable parts and the HF-compensation feature on the 10: 1 attenuator probes For the first time, probes in this price range allow adjustments of their HF-characteristics to match individually the input impedance of each scope. This is particularly important for scopes with higher bandwidths (>50MHz), as otherwise strong overshoot or rounding may occur, when measuring fast-rising square waves. An exact HF-compensation, however, is only possible with square-wave generators having a risetime <5ns. Most HAMEG scopes already feature such a calibration generator. For other oscillo-scopes, it is available as accessory item HZ60-2. At present the following Modular Probes are available (HZ36 without HF-compensation):

Туре	HZ36 selectable	HZ51	HZ52	HZ53	HZ54 selectable			
Attenuation Ratio	1:1/10:1	10:1	10:1 (HF)	100:1	1:1/10:1			
Bandwidth min. (MHz)	10/ 100	150	250	150	10 / 150			
Risetime (ns)	3513.5	<2	tl.4	<2	35/<2			
Inp. Capacitance (pF)	47/18	16	16	6,5	40/18			
Inp. Resistance (MR)	1/10	10	10	100	1/10			
Inp. Voltage max. (V _p)	600	600	600	1200	600			
Cable Length (m)	1.5	1.2	1.5	1.5	1.2			
Spare Cable for HZ36 HZ39								
Spare Cable for HZ51,	HZ54				HZ57			
Sparepart Kit (2 sprung	hooks, 2 sc	crew tip	os, 1 ground	cable)	HZ46			

D	emo	du	lator	· Pro	be	

Special probe for AM-demodulation and wobbulator measurements. HF-Bandwidth 100kHz - 500MHz (± 1 dB). AC Input Voltage 250mV - $50V_{rms}$. DC isolation Voltage 200V DC including peak AC. Cable length 1.2m.

High Voltage Probe

HZ58

HZ55

For measurement of voltages up to $15 kV_{pp}.$ Input resistance approx. 500 m $\Omega.$ Recommended load resistance1 $M\Omega/10\,M\Omega$ (switchable). Attenuation ratio 1000 : 1. Bandwidth 1 MHz. Cable length 1.5 m. BNC connector.

Test Cable Banana - BNC

HZ32

Coaxial test cable; length 1.15 m, characteristic impedance $50\Omega.$ Cable capacitance 120 pF. Input voltage max. $500 V_{p}.$

Test Cable BNC-BNC HZ34

Coaxial test cable; length 1 m, characteristic impedance $50\Omega.$ Cable capacitance 126 pF. Input voltage max. $500V_p.$

Adapter Banana - BNC HZ20

Two 4mm binding posts (19mm between centers) to standard BNC male plug. input voltage max. $500V_{o}$.

50 Ω Through-Termination HZ22

For terminating systems with $50\,\Omega$ characteristic impedance. Maximum load 2 W. Max. voltage 1 $0V_{rms}$.

Carrying Cases

For HM 103	HZ95
For HM203, HM204. HM205, HM208, HM408. HM604,	
HM605 and HM 1005	HZ96

Viewing Hood

HZ47

HZ60-2

For HM203, HM204, HM205, HM208, HM408, HM604, HM605 and HM 1005

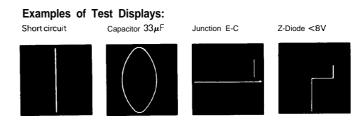
Scope-Tester

For Checking the Y amplifier, timebase, and compensation of all probes, the HZ 60-2 is a crystal-controlled, fast rising (typ. 3ns) square-wave generator with switchable frequencies of DC, 1 -10-100 Hz, I-IO-I 00 kHz, and 1 MHz. Three BNC outputs provide signals of 25 mV_{pp} into 50 Ω , 0.25 V_{pp} and 2.5 V_{pp} (open circuit for 1 Ox and 100x probes); accuracy ±1%. Battery-powered.

Component-Tester

HZ65

Indispensable for trouble-shooting in electronic circuits. Single component and in-circuit tests are both possible. The HZ65 operates with all scopes, which can be switched to X-Y operation (ext. horizontal deflection). Non-destructive tests can be carried out on almost all semiconductors, resistors, capacitors, and coils. Two sockets provide for quick testing of the 3 junction areas in any small power transistor. Other components are connected by using 2 banana jacks. Test leads supplied.



General Information

This oscilloscope is easy to operate. The logical arrangement of the controls allows anyone to become familiar with the operation of the instrument after a short time, however, experienced users are also advised to read through these instructions so that all functions are understood.

Immediately after unpacking, the instrument should be checked for mechanical damage and loose parts in the interior. If there is transport damage, the supplier must be informed immediately. The instrument must then not be put into operation.

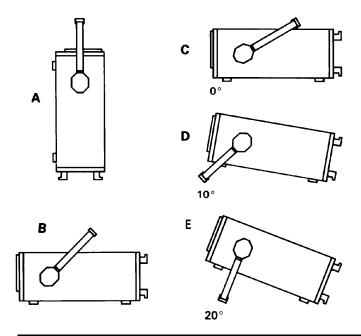
Check that the instrument is set to the correct mains/line voltage. If not, refer to instructions on page M2.

Use of tilt handle

To view the screen from the best angle, there are three different positions (C, D, E) for setting up the instrument. If the instrument is set down on the floor after being carried, the handle remains automatically in the upright carrying position (A).

In order to place the instrument onto a horizontal surface, the handle should be turned to the upper side of the oscilloscope (C). For the D position $(10^{\circ} \text{ inclination})$, the handle should be turned in the opposite direction out of the carry ing position until it locks in place automatically underneath the instrument. For the E position (20" inclination), the handle should be pulled to release it from the D position and swing backwards until it locks once more.

The handle may also be set to a position for horizontal carrying by turning it to the upper side to lock in the B position. At the same time, the instrument must be moved upwards, because otherwise the handle will jump back.



Safety

This instrument has been designed and tested in accordance with *IEC Publication 348, Safety Requirements for Electronic Measuring Apparatus,* and has left the factory in a safe condition. The present instruction manual contains important information and warnings which have to be followed by the user to ensure safe operation and to retain the oscilloscope in safe condition. The case, chassis and all measuring terminals are connected to the protective earth contact of the appliance inlet. The instrument operates according to *Safety C/ass I* (three-conductor power cord with protective earthing conductor and a plug with earthing contact). The mains/line plug shall only be inserted in a socket outlet provided with a protective earth contact. The protective action must not be negated by the use of an extension cord without a protective conductor.

Warning! Any interruption of the protective conductor inside or outside the instrument or disconnection of the protective earth terminal is likely to make the instrument dangerous. Intentional interruption of the protective earth connection is prohibited. The mains/line plug should be inserted before connections are made to measuring circuits.

The grounded accessible metal parts (case, sockets, jacks) and the mains/line supply contacts (line, neutral) of the instrument have been tested against insulation breakdown with 2000 *Vr.m.s.* (50 Hz).

Under certain conditions, 50 Hz or 60 Hz hum voltages can occur in the measuring circuit due to the interconnection with other mains/line powered equipment or instruments. This can be avoided by using an isolation transformer (Safety Class II) between the mains/line outlet and the power plug of the instrument. When displaying waveforms where the "low-level" side of the signal is at a high potential, even with the use of a protective isolation transformer, it should be noted that this potential is connected to the oscilloscope's case and other accessible metal parts. High voltages are dangerous. In this case, special safety precautions are to be taken, which must be supervised by qualified personnel if the voltage is higher than 42V.

Most cathode-ray tubes develop X-rays. However, the dose equivalent rate falls far below the maximum permissible value of 36pA/kg (0.5mR/h).

Whenever it is likely that protection has been impaired, the instrument shall be made inoperative and be secured against any unintended operation. The protection is likely to be impaired if, for example, the instrument

- shows visible damage,
- fails to perform the intended measurements,
- has been subjected to prolonged storage under unfavourable conditions (e.g. in the open or in moist environments),
- has been subject to severe transport stress (e.g. in poor packaging).

Operating conditions

The instrument has been designed for indoor use. The permissible ambient temperature range during operation is + 15°C... + 30°C. It may occasionally be subjected to temperatures between + 10°C and - 10°C without degrading its safety. The permissible ambient temperature range for storage or transportation is -40°C .+70°C.

The maximum operating altitude is up to 2200m (nonoperating 15000m). The maximum relative humidity is up to 80%.

If condensed water exists in the instrument it should be acclimatized before switching on. In some cases (e.g. extremely cold oscilloscope) two hours should be allowed before the instrument is put into operation. The instrument should be kept in a clean and dry room and must not be operated in explosive, corrosive, dusty, or moist environments The oscilloscope can be operated in any position, but the convection cooling must not be impaired. The **ven***tilation holes may not be covered.* For continuous operation the instrument should be used in the horizontal position, preferably tilted upwards, resting on the tilt handle.

The specifications stating tolerances are only valid if the instrument has warmed up for 30 minutes at an ambient temperature between $+15C^{\circ}$ and $+30C^{\circ}$. Values not stating tolerances are typical for an average instrument.

Warranty

Each instrument runs through a quality test with 10 hour burn-in before leaving the production. Practically every early failure is detected in intermittent operation by this method. However, it is possible that a component fails only after a lengthy operating period. Therefore a *functional guarantee of 2 years* is given for all units. The condition for this is that no modifications have been made in the instrument. In the case of shipments by post, rail or carrier it is recommended that the original packing is carefully preserved. Transport damages and damage due to gross negligence are not covered by the guarantee.

In the case of a complaint, a label should be attached to the housing of the instrument which describes briefly the faults observed. If at the same time the name and telephone number (dialing code and telephone or direct number or department designation) is stated for possible queries, this helps towards speeding up the processing of guarantee claims.

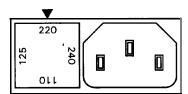
Maintenance

Various important properties of the oscilloscope should be carefully checked at certain intervals. Only in this way is it largely certain that all signals are displayed with the accuracy on which the technical data are based. The test methods described in the test plan of this manual can be performed without great expenditure on measuring instruments. However, purchase of the new HAMEG scope tester HZ 60, which despite its low price is highly suitable for tasks of this type, is very much recommended.

The exterior of the oscilloscope should be cleaned regularly with a dusting brush. Dirt which is difficult to remove on the casing and handle, the plastic and aluminium parts, can be removed with a moistened cloth (99% water +1 % mild detergent). Spirit or washing benzine (petroleum ether) can be used to remove greasy dirt. The screen may be cleaned with water or washing benzine (but not with spirit (alcohol) or solvents), it must then be wiped with a dry clean lint-free cloth. Under no circumstances may the cleaning fluid get into the instrument. The use of other cleaning agents can attack the plastic and paint surfaces.

Switching over the mains/line voltage

The instrument is set for 220V (240V U.K.) line voltage on delivery. It can be switched over to other voltages at the fuse holder combined with the 3-pole appliance inlet at the rear of the instrument. Firstly the fuse holder printed with the voltage values is removed using a small screw driver and - if required - provided with another fuse. Refer to the table below for the prescribed value of the fuse. Then replace the fuse holder so that the impressed white triangle points to the desired voltage. Here pay attention that the cover plate is also correctly engaged. The use of repaired fuses or short circuiting the fuse holder is not allowed. Damage arising because of this is not covered by the guarantee.



Fuse type: Size 5 x 20 mm; 250 V-, C; IEC 127, Sheet III; DIN 41 662 (possibly DIN 41 571 sheet 3).

Cutoff: time lag (T).

Line voltage	Fuse rating
110 V $\sim \pm 10$ %	TO.63 A
125 V ~ ±10 %	TO.63 A
220 V ~ ±10 %	T 0.315 A
240 V ~ ±10 %	T 0.315 A

Type of Signal

All types of signals with a frequency spectrum below 60 MHz can be displayed on the HM 604. The display of simple electrical processes such as sinusoidal RF and AF signals or ripple poses no problems. However, when square or pulse-shaped signals are displayed it must be remembered that their harmonic content must also be transmitted. In this case, the bandwidth of the vertical amplifier must be considerably higher than the repetition frequency of the signal. In view of this, accurate evaluation of such signals with the HM 604 is only possible up to a maximum repetition rate of 6MHz. Operating problems can sometimes occur when composite signals are to be displayed, especially if they do not contain any suitable level components and repetition frequency which can be used for triggering. This occurs, for example, with burst signals. To obtain a stably triggered display in these cases, it may be necessary to use Normal Triggering, HOLD OFF time control, and/or TIME/DIV. variable control.

Video signals are easily triggerable by the aid of the active TV sync separator (TV SEP. switch).

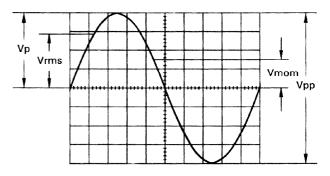
For optional operation as a DC or AC voltage amplifier, each channel is provided with a **DC-AC** coupling switch. The **DC** position should only be used with an attenuator probe or at very low frequencies or if the determination of DC voltage content of the signal is absolutely necessary.

However, when investigating very low-frequency pulses, misleading ramp-offs may occur with AC coupling. In this case, DC operation is to be preferred if the signal voltage is not superimposed on a too high DC voltage level. Otherwise, a capacitor of adequate capacitance must be connected before the input of the vertical amplifier (switched to DC coupling). It should be remembered that this capacitor must have a sufficiently high breakdown voltage. DC operation is also recommended for the display of logic and pulse signals, particularly if their pulse duty factor changes permanently during operation. Otherwise, the display will move up and down with any change. DC voltages can only be measured in the **DC** position.

Amplitude Measurements

In general electrical engineering, alternating voltage data normally refers to effective values (rms = root-mean-square value). However, for signal magnitudes and voltage designations in oscilloscope measurements, the peak-to-peakvoltage (V_{pp}) value is applied. The latter corresponds to the real potential difference between the most positive and most negative points of a signal waveform.

If a sinusoidal waveform, displayed on the oscilloscope screen, is to be converted into an effective (rms) value, the resulting peak-to-peak value must be divided by $2x\sqrt{2} = 2.83$. Conversely, it should be observed that sinusoidal voltages indicated in V_{rms} (V_{eff}) have 2.83 times the potential difference in V,,. The relationship between the different voltage magnitudes can be seen from the following figure.



The minimum signal voltage required at the vertical amplifier input for a display of **1** cm is approximately $1 m V_{pp}$. This is achieved with the attenuator control set at 5 m V/cm, its variable control in the fully clockwise position and pulled out. However, smaller signals than this may also be displayed. The *deflection coefficients* on the input attenuators are indicated in m V/cm or V/cm (peak-to-peak value).

The magnitude of the applied voltage is ascertained by multiplying the selected deflection coefficient by the vertical display height in cm.

If an attenuator probe x 70 is used, a further multiplication by a factor of 70 is required to ascertain the correct voltage value.

For exact amplitude measurements the variable control on the attenuator switch must be set to its calibrated detent CAL. When turning the variable control ccw the sensitivity will be decreased by a factor of 2.5.

Therefore every intermediate value is possible within the 7-2-5 sequence.

With direct connection to the vertical input, signals up to $400 V_{pp}$ may be displayed (attenuator set to 20 V/cm, variable control ccw).

When pulling the variable control knob (MAG x5), the sensitivity is increased by a factor of 5. Hence follows a min. deflection coefficient of 1 mV/cm (reduced bandwidth).

With the designations

U

- H = display height in cm,
- U = signal voltage in V_{pp} at the vertical input,

D = **deflection coefficient in V/cm** at attenuator switch, the required quantity can be calculated from the two given quantities:

$$= D \cdot H \qquad H = \frac{U}{D} \qquad D = \frac{U}{H}$$

However, these three values are not freely selectable. They have to be within the following limits (trigger threshold, accuracy of reading):

H between 0.5 and 8cm, if possible 3.2 to 8cm,

U between 1 mV_{pp} and $160V_{pp}$,

D between 5mV/cm and 20V/cm in I-2-5 sequence.

D between 1 mV/cm and 4V/cm in I-2-5 sequence

(with pulled MAG x5 knob).

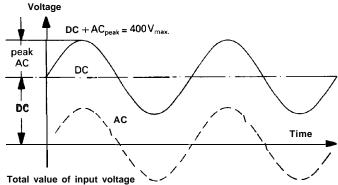
Examples:

Set deflection coefficient **D** = 50 mV/cm \triangleq 0.05 V/cm, observed display height **H** = 4.6 cm, required voltage **U** = 0.05 \cdot 4.6 = 0.23 V,,.

Input voltage $U = 5 V_{pp}$, set deflection coefficient D = 1 V/cm, required display height H = 5: 1 = 5 cm

Signal voltage U = $220V_{rms} \cdot 2 \cdot \sqrt{2} = 622 V_{pp}$ (voltage >160V_{pp}, with probe X 10 : U = 62.2 V_{pp}), desired display height H = min. 3.2 cm, max. 8 cm, max. deflection coefficient D = 62.2 : 3.2 = 19.4 V/cm, min. deflection coefficient D = 62.2 : 8 = 7.8 V/cm, adjusted deflection coefficient D = 10 V/cm

If the applied signal is superimposed on a DC (direct voltage) level the total value (DC + peak value of the alternating voltage) of the signal across the Y-input must **not exceed** \pm **400 V** (see figure). This same limit applies to normal x 10 attenuator probes, the attenuation ratio of which allows signal voltages up to approximately $\mathbf{1}$,000 V_{pp} to be evaluated. Voltages of up to approximately $2,400 V_{pp}$ may be measured by using the HZ53 high voltage probe which has an attenuation ratio of 100: 1. It should be noted that its AC_{peak} value is derated at higher frequencies. If a normal x 10 probe is used to measure high voltages there is the risk that the compensation trimmer bridging the attenuator series resistor will break down causing damage to the input of the oscilloscope. However, if for example only the residual ripple of a high voltage is to be displayed on the oscilloscope, a normal x 10 probe is sufficient. In this case, an appropriate high voltage capacitor (approx. 22-68nF) must be connected in series with the input tip of the probe.



The dotted line shows a voltage alternating at zero volt level. When superimposed a DC level, the addition of the positive peak and the DC voltage results in the max. voltage (DC + AC,...).

It is very important that the oscilloscope input coupling is set to **DC**, if an attenuator probe is used for voltages higher than 400V (see page M6: Connection of Test Signal).

Time Measurements

As a rule, all signals to be displayed are periodically repeating processes and can also be designated as periods. **The** number of periods per second is the recurrence frequency or repetition rate. One or more signal periods or even part of a period may be shown as a function of the adjustment of the **TIME/DIV**. switch. The time coefficients on the **TIME/ DIV**. switch are indicated in **s/cm**, **ms/cm**, and µ**s/cm**. Accordingly, the dial is subdivided into three sectors. **The du***ration of a signal period or a portion of the waveform is* **ascertained by multiplying the relevant time (horizon***tal distance in cm) by the time coefficient selected on the* **TIME/DIV**. *switch. The time variable control (small knob on the* **TIME/DIV**. *switch) must be in its calibrated* **detent CAL. for accurate measurement** (arrow horizontal and pointing to the right).

With the designations

- L = displayed wave length in cm of one period,
- T = time in seconds for one period,
- F = recurrence frequency in Hz of the signal,

T_c = time coefficient in s/cm on timebase switch

and the relation $\mathbf{F} = \mathbf{1}/\mathbf{T}$, the following equations can be stated :

$$T = L \cdot T_c$$
 $L = \frac{T}{T_c}$ $T_c = \frac{T}{L}$

$$F = \frac{1}{L \cdot T_c} \qquad L = \frac{1}{F \cdot T_c} \qquad T_c = \frac{1}{L \cdot F}$$

With X-MAG. x10 button depressed the T_c value must be divided by 10.

However, these four values are not freely selectable. They have to be within the following'limits:

- L between 0.2 and 1 Ocm, if possible 4 to 1 Ocm,
- T between 5 ns and 1 Os,
- F between 0.1 Hz and 60 MHz,
- T_c between 50 ns/cm and 1 s/cm in I-2-5 sequence (with X MAG. x **10** in out position), and
- T_c between 5 ns/cm and 100 ms/cm in I-2-5 sequence (with pushed X MAG. x10 button).

Examples:

Displayed wavelength L = 7 cm, set time coefficient $T_c = 0.5 \ \mu s/cm$, required period T = $7 \cdot 0.5 \cdot 10^{-6} = 3.5 \ \mu s$ required rec. freq. F = $1:(3.5 \cdot 10^{-6}) = 286 \ \text{kHz}$.

Signal period T = 0.5 s, set time coefficient $T_c = 0.2$ s/cm, required wavelength L = 0.5 : 0.2 = 2.5 cm. Displayed ripple wavelength L = 1 cm, set time coefficient $T_c = 10 \text{ ms/cm}$, required ripple freq. F = 1 : $(1 \cdot 10 \cdot 10^{-3}) = 100 \text{ Hz}$.

TV-line frequency F = 15 625 Hz, set time coefficient $T_c = 10$ @cm, required wavelength L = 1: (15 625.1 0⁻⁵) = 6.4 cm.

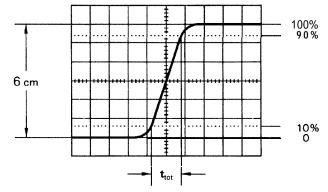
Sine wavelength L = min. 4 cm, max. 1 Ocm, Frequency F = 1 kHz, max. time coefficient $T_c = 1 : (4 \cdot 10^3) = 0.25 \text{ ms/cm}$, min. time coefficient $T_c = 1 : (10 \cdot 10^3) = 0.1 \text{ ms/cm}$, set time coefficient $T_c = 0.2 \text{ ms/cm}$, required wavelength L = 1: (1 $0^3 \cdot 0.2 \cdot 1 \ 0^{-3}$) = 5cm.

Displayed wavelength L = 0.8 cm, set time coefficient T_c = $0.5 \text{ }\mu\text{s/cm}$, *pressed* MAG X 10 *button:* T_c = $0.05 \text{ }\mu\text{s/cm}$, required rec. freq. F = 1: $(0.8 \cdot 0.05 \cdot 10^{-6}) = 25 \text{ MHz}$, required period T = 1: $(25.1 \ 0^6) = 40 \text{ ns.}$

If the time is relatively short as compared with the complete signal period, an expanded time scale should always be applied (X MAG x10 button pushed). In this case, the ascertained time values have to be divided by 10. Very small time intervals at optional points of the signal can be measured more exactly with the aid of the sweep delay. With it, the display and measurement of time intervals, which are smaller than 1 % of the full signal period, are possible. The smallest measurable time interval is, on the whole, dependent on the obtainable brightness of the CRT. The limit is an expansion of approximately 1000 times. Using a Viewing Hood HZ47, more expansion is possible, provided that the time coefficient set on the TIME/DIV. switch is greater than 5 µs/cm (and using the X MAG x 10 facility) for the signal's basic period. Otherwise, the fastest sweep speed determines the greatest possible expansion.

When investigating pulse or square waveforms, the critical feature is the *risetime of the voltage step*. To ensure that transients, ramp-offs, and bandwidth limits do not unduly influence the measuring accuracy, the risetime is generally measured between 10% and 90% of the vertical pulse height. For peak-to-peak signal amplitude of 6cm height, which are symmetrically adjusted to the horizontal center line, the internal graticule of the CRT has two horizontal dotted lines ± 2.4 cm from the center line. Adjust the Y attenuator switch with its variable control together with the Y-POS. control so that the pulse height is precisely aligned with the 0 and 100 % lines. The 10 % and 90 % points of the signal will now coincide with the two lines, which have a distance of ±2.4 cm from the horizontal center line and an additional subdivision of 0.2 cm. The risetime is given by the product of the horizontal distance in cm between these two coincidence points and the time coefficient setting.

If magnification is used, this product must be divided by 10. The *fall time* of a pulse can also be measured by using this method.



The above figure shows correct positioning of the oscilloscope trace for accurate risetime measurement.

With a time coefficient of 0.05μ s/cm and pushed X MAG x10 button the example shown in the above figure results in a measured total risetime of

When very fast risetimes are being measured, the risetimes of the oscilloscope amplifier and the attenuator probe have to be deducted from the measured time value. The risetime of the signal can be calculated using the following formula.

$$t_{r} = \sqrt{t_{tot}^{2} - t_{osc}^{2} - t_{p}^{2}}$$

In this \mathbf{t}_{tot} is the total measured risetime, \mathbf{t}_{osc} is the risetime of the oscilloscope amplifier (approx. 5.8ns), and \mathbf{t}_p the risetime of the probe (e.g. = 2 ns). If \mathbf{t}_{tot} is greater than 42 ns, then \mathbf{t}_{tot} can be taken as the risetime of the pulse, and calculation is unnecessary.

Calculation of the example in the figure above results in a signal risetime

$$\mathbf{t}_{\rm r} = \sqrt{8^2 - 5.8^2 - 2^2} = 5.1$$
 ns

Connection of Test Signal

Caution: When connecting unknown signals to the oscilloscope input, always use automatic triggering and set the **DC-AC** input coupling switch to **AC**. The attenuator switch should initially be set to **20V/cm**.

Sometimes the trace will disappear after an input signal has been applied. The attenuator switch must then be turned back to the left, until thevertical signal height is only $3-8 \,\mathrm{cm}$. With a signal amplitude greater than $160 \,\mathrm{V_{pp}}$, an attenuator probe must be inserted before the oscilloscope's vertical input. If, after applying the signal, the trace is nearly blanked, the period of the signal is probably substantially

longer than the set value on the **TIME/DIV.** switch. It should be turned to the left to an adequately greater time coefficient.

The signal to be displayed should be fed to the vertical input of the oscilloscope by means of a shielded test cable, e.g. the HZ32 or HZ34, or by a x 10 or x 100 attenuator probe. The use of these shielded cables with high impedance circuits is only recommended for relatively low frequencies (up to approx. 50kHz). For higher frequencies, and when the signal source is of low impedance, a cable of matched characteristic impedance (usually $50\,\Omega$) is recommended. In addition, and especially when investigating square or pulse waveforms, a resistor equivalent to the characteristic impedance of the cable must also be connected to the cable directly at the input of the oscilloscope. When using a 509 cable, such as the HZ34, a 50 Ω through-termination type HZ22 is available from HAMEG. When investigating square or pulse waveforms with fast risetimes, transient phenomena on both the edge and top of the signal may become visible if the correct termination is not used. It must be remembered that the 50Ω through-termination will only dissipate a maximum of 2 watts. This power consumption is reached with 1 $0\,V_{rms}\,\text{or}$ with $28\,V_{pp}$ sine signal.

If a x 10 or x 100 attenuator probe is used, no termination is necessary. In this case, the connecting cable is matched directly to the high impedance input of the oscilloscope. When using attenuator probes even high internal impedance sources are only slightly loaded by approximately 10 M Ω I I 16 pF or 100 M Ω I I7 pF respectively. Therefore, when the voltage loss due to the attenuation of the probe can be compensated by a higher sensitivity setting on the HM 604, the probe should always be used. Also it should be remembered that the series impedance of the probe provides a certain amount of protection for the input of the oscilloscope amplifier. It should be noted that all attenuator probes must be compensated in conjunction with the oscilloscope (see: Probe Adjustment, page M8).

If a x IO or x 100 attenuator probe is used at voltages higher than 400 V, the DC input coupling must always be set. With AC coupling, the attenuation is frequency-dependent, the pulses displayed can exhibit ramp-off, DC-voltage contents are suppressed — but loads the respective input coupling capacitor of the oscilloscope. The electric strength of which is maximum 400V (DC + peak AC). For the suppression of unwanted DC voltages, a *capacitor* of adequate capacitance and electric strength *may be connected before the input tip of the probe (e.g.* for ripple measurements).

It is important to remember that when low voltage signals are being investigated the position of the ground point on the test circuit can be critical. This ground point should always be located as close as possible to the measuring point. If this is not done, serious signal deformation may result from any spurious currents through the ground leads or test chassis parts. This comment also applies to the ground leads on attenuator probes which ideally should be as short and as thick as possible. For connection of a probe to a BNC socket, a BNC-adapter should be used. It forms often a part of the probe accessory. Grounding and matching problems are then eliminated.

Hum or interference voltage appearing in the measuring circuit (especially with a small deflection coefficient) is possibly caused by multiple grounding, because equalizing currents can flow in the shielding of the measuring cables (voltage drop between non-fused earthed conductors of other line powered devices, which are connected to the oscilloscope or test object, e.g. signal generators with anti-interference capacitors).

Operating

For a better understanding of these Operating Instructions the front panel picture at the end of these instructions can be unfolded for reference alongside the text.

The front panel is subdivided into three sections according to the various functions. The **INTENS**., FOCUS and TR (trace rotation) controls are arranged on the left directly below the screen of the cathode-ray tube (CRT). Continuing towards the right are the horizontal magnification button (X MAG. **x10**), the switch for calibrator frequency selection (1 kHz/1 MHz) and calibrator output sockets 0.2 V/2V (CAL.). The COMPONENT TESTER pushbutton and its measuring socket are located on the right side.

The X-Section, located on the upper right, next to the screen, contains the red POWER pushbutton and indicating LED, all controls for timebase (**TIME/DIV**.), triggering (**TRIG.**), horizontal trace position (**X-POS.**), sweep delay (DELAY), TV separator (TV SEP.) together with the field select button (FIELD **I/II**), the XYmode button (XV), and the knob for hol-doff adjustment (HOLD OFF).

The lower Y-Section contains the controls for the vertical deflection system. On the right and left in this section are located: vertical input connector, DC-AC-GD input coupling slide switch, Y-POS. control, INVERT pushbutton, attenuator switch with variable control, and ground jack. All these controls and connectors exist in duplicate for each of the Channels I and II. Three pushbuttons for selecting the operating mode are arranged below the attenuator switches: CH **I/II -TRIG I/II, DUAL** and ADD.

These are explained later.

The instrument is so designed that even incorrect operation will not cause serious damage. The pushbuttons control only minor functions, and it is recommended that before commencement of operation all pushbuttons are in the "out" position. After this the pushbuttons can be operated depending upon the mode of operation required. The HM 604 accepts all signals from DC (direct voltage) up to a frequency of at least 60MHz(-3dB). For sinewave voltages the upper frequency limit will be 80MHz. However, in this higher frequency range the vertical display height on the screen is limited to approx. 6cm. The time resolution poses no problem. For example, with 100 MHz and the fastest adjustable sweep rate (5ns/cm), one cycle will be displayed every 2cm. The tolerance on indicated values amounts to $\pm 3\%$ in both deflection directions. All values to be measured can therefore be determined relatively accurately. However, from approximately 25 MHz upwards the measuring error will increase as a result of loss of gain. At 40MHz this reduction is about 10%. Thus, approximately 11 % should be added to the measured voltage at this frequency. As the bandwidth of the amplifiers differ (normally between 65 and 70 MHz), the measured values in the upper limit range cannot be defined exactly. Additionally, as already mentioned, for frequencies above 60MHz the dynamic range of the display height steadily decreases. The vertical amplifier is designed so that the transmission performance is not affected by its own overshoot.

First Time Operation

Check that the instrument is set to the correct mains/ line voltage. (Refer to page M2).

Before applying power to the oscilloscope it is recommended that the following simple procedures are performed:

- Check that all pushbuttons are in the *out* position, i.e. released.
- Rotate the three variable controls with arrows to their calibrated detent.
- Set the variable controls with marker lines to their midrange position (marker lines pointing vertically).
- The LEVEL control knob should be on its left stop (AT).
- The three lever switches in the X-Section should be set to their uppermost position.
- Both input coupling slide switches for CH.1 and CH.II in the Y-Section should be set to the GD position.

Switch on the oscilloscope by depressing the red POWER pushbutton. An LED will illuminate to indicate working order. The trace, displaying one baseline, should be visible after a short warm-up period of 10 seconds. Adjust **Y-POS.I** and **X-POS.** controls to center the baseline. Adjust **IN-TENS.** (intensity) and **FOCUS** controls for medium brightness and optimum sharpness of the trace. The oscilloscope is now ready for use.

If only a spot appears *(CAUTION!* CRT phosphor can be damaged.), reduce the intensity immediately and check that the X-Y pushbutton is in the released (out) position. If the trace is not visible, check the correct positions of all knobs and switches (particularly LEVEL knob in AT position and DELAY MODE lever switch to OFF).

To obtain the maximum life from the cathode-ray tube, the minimum intensity setting necessary for the measurement in hand and the ambient light conditions should be used. *Particular care is required* when a single spot is displayed, as a very high intensity setting may cause damage to the fluorescent screen of the CRT. Switching the oscillo-scope off and on at short intervals stresses the cathode of the CRT and should therefore be avoided.

Trace Rotation TR

In spite of Mumetal-shielding of the CRT, effects of the earth's magnetic field on the horizontal trace position cannot be completely avoided. This is dependent upon the orientation of the oscilloscope on the place of work. A centred trace may not align exactly with the horizontal center line of the graticule. A few degrees of misalignment can be corrected by a potentiometer **acessi**ble through an opening on the front panel marked **TR**.

DC Balance Adjustment

The vertical preamplifiers for **CH.I** and **CH.II** contain matched dual FETs connected as input source followers. After long periods of use the FET characteristics may change which can alter the DC balance of the vertical amplifier. A quick check of DC Balance can be made on each channel by pulling the fine amplitude control MAG x5 and pushing it back. If the trace moves from the vertical position (up or down) more than 1 mm, the DC Balance will require readjustment. This check should be made after a 20-minute warm-up period.

Adjustment procedure

The following instructions should be performed to obtain the correct DC balance adjustment of both channels.

- Remove all input cables and adjust oscilloscope controls to display the baseline.
- Center the baseline using Y-POS. and X-POS. controls.
- Set attenuator switches to 5mV/cm and input coupling switches to GD.
- Release all pushbuttons in the Y-Section.
- Place the oscilloscope so that it rests firmly on its back (upright position) and locate DC balance adjustment potentiometer access holes marked CH.I DC-BALANCE
 CH.II which are found underneath the instrument.
- Insert a screwdriver (blade approx. 3mm, length min.
 20 mm) in CH.I hole. A plastic guide with slotted bottom is located behind the hole.
- Pull and push the CH.I variable control MAG x5 and adjust balance pot so that the baseline no longer moves up or down. When the trace remains steady, correction of CH.I is completed.
- Depress CH I/II-TRIG. I/II button. Repeat adjustment procedure for CH.II.

Use and Compensation of Probes

To display an undistorted waveform on an oscilloscope, the probe must be matched to the individual input impedance of the vertical amplifier.

The HM 604's built-in calibration generator provides a squarewave signal with a very low risetime (<5ns), and switch-selectable frequencies of approx. 1 kHz and 1 MHz at two output sockets below the CRT screen. One output provides $0.2V_{pp}\pm1\%$ for 10: 1 probes, and $2V_{pp}\pm1\%$ are present at the other, for 100: 1 probes.

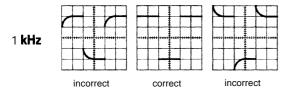
When the attenuator switches are set to **5 mV/cm** vertical deflection coefficient, these calibration voltages correspond to a screen amplitude of **4 cm**.

The output sockets have an internal diameter of 4.9mm to accommodate the internationally accepted shielding tube diameter of modern **Modular Probes** and **F-series** slimline probes. Only this type of construction ensures the extremely short ground connections which are essential for an undistorted waveform reproduction of non-sinusoidal high frequency signals.

Adjustment at 1 kHz

The C-trimmer adjustment compensates the capacitive loading on the oscilloscope input (approx. 30 pF with the HM604). By this adjustment, the capacitive division assumes the same division ratio as the ohmic voltage divider to ensure an equal division ratio for high and low frequencies, as for DC. (For 1: 1 probes or switchable probes set to 1: 1, this adjustment is neither required nor possible). A baseline exactly parallel to the horizontal graticule lines is a major condition for accurate probe adjustments. (See also 'Trace Rotation TR', page M7.)

Connect the probes (Types HZ51, 52, 53, 54, or HZ37) to CH.1 input. All pushbuttons should be released (in the 'out' position), and all push-pull knobs pushed 'in'. Set the input coupling switch to DC, the attenuator switch to **5mV/cm**, and the **TIME/DIV**. switch to **0.2ms/cm**, and all variable controls to CAL. position. Plug the probe tip into the appropriate calibrator output socket, i.e. 10:1 probes into the **0.2V** socket, 100: 1 probes into the **2.0V** socket.



Approximately 2 complete waveform periods are displayed on the CRT screen. Now the compensation trimmer has to be adjusted. Normally, this trimmer is located in the probe head. On the 100: 1 probe HZ53, however, it is located in the connecting box at the other end of the cable. Using a small insulated non-metallic screwdriver or trimming tool, the trimmer has to be adjusted slowly until the tops of the squarewave signal are exactly parallel to the horizontal graticule lines. (See Fig. above for **1** kHz.) The signal amplitude shown should be 4cm \pm **1.2** mm (= 3 %). During this adjustment, the signal edges will remain invisible.

Adjustment at 1 MHz

Probes HZ51, 52, and 54 will also allow for HF-adjustments. They incorporate resonance deemphasizing networks (Rtrimmer in conjunction with inductances and capacitors) which permit – for the first time – probe compensation in the range of the upper frequency limit of the vertical oscilloscope amplifier. Only this compensative adjustment ensures optimum utilisation of the full bandwidth, together with constant group delay at the high frequency end, thereby reducing characteristic transient-distortion near the leading signal edge (e.g. overshoot, rounding, ringing, holes or bumps) to an absolute minimum.

Using the probes HZ51, 52, and 54, the full bandwidth of the HM 604 can be utilized without risk of unwanted waveform distortion.

Prerequisite for this HF-adjustment is a squarewave generator with fast risetime (typical 4ns), and low output impedance (approx. 50Ω), providing 0.2V and 2V at a frequency of approx. **1** MHz. The calibrator output of the HM604 meets these requirements when the pushbutton 1 MHz is depressed.

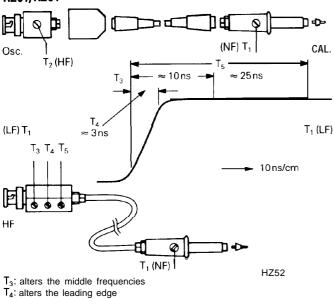
Connect the probe (HZ51, 52, or 54) to CH.1 input. Depress the calibrator pushbutton **1MHz**. All other pushbuttons should be released ('out' position). Set the input coupling switch to DC, attenuator switch to **5mV/cm**, and TIME/ DIV. switch to 0.1 μ **s/cm**. Set all variable controls to CAL. position.

Insert the probe tip into the output socket marked 0.2V. A waveform will be displayed on the CRT screen, with leading and trailing edges clearly visible. For the HF-adjustment now to be performed, it will be necessary to observe the rising edge as well as the upper left corner of the pulse top. To gain access to the HF-compensation trimmer, the plastic cover of the probe connecting box has to be slid off after unscrewing the probe cable. The connecting boxes of the HZ51 and HZ54 contain one R-trimmer screw, each, while that of the HZ52 provides three. These R-trimmers have to be adjusted in such a manner that the beginning of the pulse top is as straight as possible. Overshoot or excessive rounding are unacceptable. This is relatively easy on the HZ51 and HZ54, but slightly more difficult on the HZ52. The rising edge should be as steep as possible, with the pulse top remaining as straight and horizontal as possible.

On the HZ52, each of the three trimmers has a clearly

defined area of influence on the waveform shape (see Fig.), offering the added advantage of being able to 'straighten out' waveform aberrations near the leading edge.

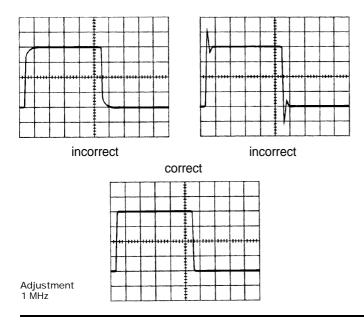
Adjustment points of the probes **HZ51**, **HZ54**



T₄, alters the lower frequencies

After completion of the HF-adjustment, the signal amplitude displayed on the CRT screen should have the same value as during the 1 kHz adjustment.

Probes other than those mentioned above, normally have a larger tip diameter and may not fit into the calibrator outputs Whilst it is not difficult for an experienced operator to build a suitable adapter, it should be pointed out that most of these probes have a slower risetime with the effect that the total bandwidth of scope together with probe may fall far below that of the HM604. Furthermore, the HF-adjustment feature is nearly always missing so that waveform distortion can not be entirely excluded.



The adjustment sequence must be followed in the order described, i.e. first at 1 kHz, then at 1 MHz. The calibrator frequencies should not be used for timebase calibrations. The pulse duty cycle deviates from 1:1 ratio.

Prerequisites for precise and easy probe adjustments, as well as checks of deflection coefficients, are straight horizontal pulse tops, calibrated pulse amplitude, and zeropotential at the pulse base. Frequency and duty cycle are relatively uncritical. For interpretations of transient response, fast pulse risetimes and low-impedance generator outputs are of particular importance.

Providing these essential features, as well as switch-selectable output-frequencies, the calibrator of the HM 604 can, under certain conditions, replace expensive squarewave generators when testing or compensating widebandattenuators or -amplifiers. In such a case, the input of an appropriate circuit will be connected to one of the **CAL**.-outputs via a suitable probe.

The voltage provided at a high-impedance input (I M Ω II15-50pF) will correspond to the division ratio of the probe used (10: **1** = 20mV_{pp}, 100: **1** = also 20mV_{pp} from 2V output). Suitable probes are HZ51, 52, 53, and 54.

For low-impedance inputs (e.g. 50 Ω), a 1: 1 probe can be employed which, however, must be fully terminated with a 50 Ω through-termination. Suitable probe types are HZ50 and HZ54. The latter must be switched to the 1: 1 position, and the HF-trimmer in the connecting box turned fully counterclockwise.

When connected to the **0.2V CAL.** socket, and using the HZ50, this arrangement will provide approx. 40 mV_{pp} at 50Ω circuit input, and approx. 24 mV_{pp} if the HZ54 is used.

The voltages given here will have larger tolerances than 1 % since operation of a 1: 1 probe together with a 50 Ω load is very uncommon.

Using the 2V CAL. socket under similar conditions is only possible with the HZ54 probe. The potential obtained at the 50 Ω input will then be approx. 190 mV_{pp}, but with almost twice the risetime. Accurate readings of the available input voltage can be shown directly on the HM604 when connecting a 50 Ω through-termination between the BNC plug of the probe and the input of the oscilloscope.

Operating Modes of the Y Amplifier

The required operating modes are selected on three pushbuttons located in the Y-Section. For *Mono* operation all pushbuttons should be in the out position, the instrument is then operating on **Channel/only.**

T, allers the lower nequencies

For Mono operation with *Channel II*, the CH I/II -TRIG. I/II pushbutton has to be pressed. When the DUAL button is depressed, the HM604 is in *Dual* channel operation. In this mode, the channels are displayed consecutively (alternate mode). This mode is not suitable for the display of very low frequency signals (<1kHz), as the tracewill appear to flicker or jump. Under these conditions, the **ADD** button should be depressed additionally selecting chopped mode. In this position, both channels then share the trace during each sweep period. For the display of high frequency signals, the type of channel switching selected is less important.

To select the *add mode* only the **ADD** button should be depressed. The signals on both channels are then added together. If in this mode one channel is *inverted* (pushbutton **INVERT** depressed), then the *difference* between the two channels is displayed. For both of these operating modes, the vertical position of the trace depends on the setting of the **Y-POS.** controls of *both* channels.

Differential measurements techniques allow direct measurement of the voltage drop across floating components (both ends above ground). Two identical probes should be used for both vertical inputs. Using a separate ground connection and notconnecting the probe or cable shields to the circuit under test avoid ground loops (hum, common-mode disturbances).

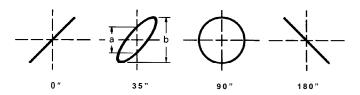
X-Y Operation

For X-Yoperation, the pushbutton in the X-Section marked X-Y must be depressed. The X signal is then derived from the Channel II (HOR. INP.). The calibration of the X signal during X-Y operation is determined by the setting of the Channel II input attenuator and variable control. This means that the sensitivity ranges and input impedances are identical for both the X and Y axes. However, the Y-POS.II control is disconnected in this mode. Its function is taken over by the X-POS. control. It is important to note that the X MAG. x10 facility, normally used for expanding the sweep, should not be operated in the X-Y mode. It should also be noted that the bandwidth of the X amplifier is approximately 5MHz(-3dB), and therefore an increase in phase difference between both axes is noticeable from 50 kHz upwards.

The Y-Input signal may be inverted by using the **INVERT** (channel I) facility.

X-Y Phase Measurements

The X-Y phase measurement method can be used to measure the phase difference between two signals of the same frequency. This provides a method of measurement for signal frequencies up to 120 kHz. However, above this frequency the inherent phase difference between the vertical and horizontal system makes accurate measurements difficult. In this mode, one of the sinewave signals provides horizontal deflection (X) while the other signal provides the vertical deflection (Y).



The phase angle between the two signals can be determined from the Lissajous pattern as follows:

$$\sin \varphi = \frac{a}{b}$$
$$\cos \varphi = \sqrt{1 - \left(\frac{a}{b}\right)^2}$$
$$\varphi = \arcsin \frac{a}{b}$$

This simple formula works for angles less than 90° , it is *independent* from both deflection amplitudes on the screen.

Caution!

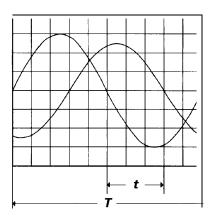
If a single spot appears (both deflection voltages are missing) reduce the intensity immediately, as a high intensity setting may cause damage to the fluorescent screen of the CRT.

Dual-Trace Phase Difference Measurements

Phase comparison between two signals of the same frequency can be made using the dual-trace feature **(DUAL** button depressed). This method of phase difference measurement can be used up to the frequency limit of the vertical system. To make the comparison, use the following procedure:

Set the Input Coupling switches to the same position, and the **CH. I/II-TRIG. I/II** pushbutton to the channel where the reference signal (Phase 0") is connected. Select **ALT.** channel switching for frequencies above **1** kHz, and **CHOP.** for frequencies below **1** kHz. Use probes which have equal time delay to connect the signals to the input connectors. Set the Input Attenuator switches and the CH I and CH II variable controls so that the displays are approximately equal and about five divisions in amplitude. Set the **TIME/DIV.** switch to a sweep rate which displays about one cycle of the waveform. Move the waveforms to the center of the graticule with the **Y-POS.II** and **Y-POS.II** controls.

Turn the Variable Time Control until one cycle of the reference signal occupies exactly 10 divisions (see next figure). Each division represents 36" of the cycle.



Dual-Trace Phase Difference Measurements T = Horizontal distance *for one period* (cm).

t = Horizontal distance of zero-crossing points (cm).

Assume a horizontal difference of 3 divisions (t = 3cm) and a period of 10 divisions (T = 1 Ocm), the phase difference ϕ can be calculated using the following formula:

or

 $\varphi^{\circ} = \frac{t}{T} \cdot 360^{\circ} = \frac{3}{10} \cdot 360^{\circ} = 108^{\circ}$ $\operatorname{arc} \varphi = \frac{t}{T} \cdot 2\pi = \frac{3}{10} \cdot 2\pi = 1.885 \, \mathrm{rad}$ ectively.

respectively.

Measurement of an amplitude modulation

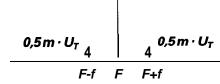
The momentary amplitude \boldsymbol{u} at time \boldsymbol{t} of a HF-carrier voltage, which is amplitude modulated without distortion by a sinusoidal AF voltage, is in accordance with the equation

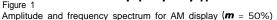
 $u = U_{T} \cdot \sin \Omega t + 0.5 m \cdot U_{T} \cdot \cos (\Omega - \omega) t - 0.5 m \cdot U_{T} \cdot \cos (\Omega + \omega) t$

where U_T = unmodulated carrier amplitude $\Omega = 2\pi F$ = angular carrier frequency $\omega = 2\pi f$ = modulation angular frequency

m = modulation factor ($\leq 1 \leq 100$ %).

The lower side frequency F-fand the upper side frequency F+f arise because of the modulation apart from the carrier frequency F.





The display of the amplitude-modulated HF oscillation can be evaluated with the oscilloscope provided the frequency spectrum is inside the oscilloscope bandwidth. The time base is set so that several cycles of the modulation frequency are visible. Strictly speaking, triggering should be external with modulation frequency (from the AF generator or a demodulator). However, internal triggering is frequently possible with normal triggering using a suitable **LEVEL** setting and possibly also using the time variable adjustment.

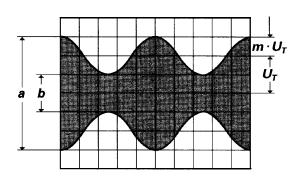


Figure 2 Amplitude modulated oscillation: F = 1 MHz; f = 1 kHz; m = 50 %; $U_T = 28.3$ mV_{rms}.

Oscilloscope setting for a signal according to figure 2:

Depress no buttons. **Y**: CH. I; **20mV/div**; AC. TIM E/DIV.: 0.2 ms/div. Triggering: NORMAL with LEVEL-setting; *internal (or external) triggering.*

If the two values \boldsymbol{a} and bare read from the screen, the modulation factor is calculated from

$$m = \frac{a-b}{a+b}$$
 resp. $m = \frac{a-b}{a+b} \cdot 100$ [%]

where $a = U_T (l+m)$ and $b = U_T (l-m)$.

The variable controls for amplitude and time can be set arbitrarily in the modulation factor measurement. Their position does not influence the result.

Triggering and Timebase

With the **LEVEL** knob in locked position (turned ccw to **AT** position = automatic triggering), a baseline is displayed continuously even when no signal is present. In this position it is possible to obtain stable displays of virtually all uncomplicated, periodically repeating signals above 30 Hz. Adjustments of the timebase then are limited to timebase setting.

With normal triggering **(LEVEL** knob not in **AT** position) and LEVEL adjustment, triggering of time/div. deflection can be set in any point of a given signal. The triggering range which can be set with the **LEVEL** control depends greatly on the amplitude of the displayed signal. If it is less than 1 div, then the range is quite small and performance of settings requires a delicate touch.

If the **LEVEL** control is incorrectly set, no trace will be visible.

In order to obtain a satisfactory stable display, the timebase must be triggered synchronously with the test signal. The trigger signal can be derived from the test signal itself, when internal triggering is selected, or from a frequency related signal applied to the external trigger input. Triggering can be selected on either the rising or falling edge of the trigger signal depending on whether the SLOPE +/pushbutton (next to **LEVEL**) is in the out or in position. In the out position, triggering from the positive-going edge is selected. The correct slope setting is important in obtaining a display when only a portion of a cycle is being displayed.

With internal *triggering* in the *Mono* channel mode on the Y amplifier, the trigger signal is derived from the respective channel in use. In the *Dualchannelmode*, the internal trigger signal may be selected from either *Channel I* or *Channel//using* the CHI/II-TRIG.I/II button; in the out position, the trigger signal is derived from Channel I. However, it is always preferable to trigger from the less complicated signal.

With *internal alternate triggering* (ALT pushbutton in the X-Section depressed) in the DUAL channel alternate mode of the Y amplifier, the trigger voltage is derived *alternately from Channel I and Channel II*. This trigger mode is particularly useful when two *asynchronous signals* are being investigated. Normal triggering should be preferable in this mode. The display of one signal only is not possible on the alternate trigger mode.

For **external triggering**, the **EXT.** pushbutton in the X-Section must be depressed. The sync. signal $(0.05 V_{pp}-0.5 V_{pp})$ must then be fed to the **TRIG. INP.** input socket.

Coupling mode and frequency range of the trigger signal are selected with the **TRIG.** lever switch in the X-Section for internal and external triggering, provided that the **TV** SEP. switch is in off position. The HM 604 has 4 coupling modes: AC, DC, **LF, HF.** The AC coupling mode is mainly used. DC trigger coupling is only recommended, when very low frequency signals are being investigated and triggering at a particular value is necessary, or when pulses, which significantly change in duty cycle during observation time, have to be displayed. If DC coupling is selected, it is advisable to use the *normal triggering* mode. In the **HF** coupling mode, a *high pass filter* is switched into the trigger amplifier. This filter cuts off the DC content of the trigger signal and the lower frequency range.

In the LF coupling mode, a *low-pass filter* is switched into the trigger amplifier. This filter cuts off any amplifier noise and the frequency range of the trigger signal above 50 kHz.

For the purpose of *line triggering* (TRIG. lever switch in the X-Section) to \sim , a (divided) secondary voltage of the power transformer is used as a trigger signal. This trigger mode is independent of the signal amplitude or display height and allows a display below the (internal) trigger threshold. Line triggering is recommended for all signals which are time-related (multiple or submultiple) to the mains/line frequency or when it is desirable to provide a stable display of a line-

frequency component in complex waveforms. Therefore it is especially suited for the measurement of small ripple voltages from power supply rectifiers or of magnetic or static leakage fields in a circuit.

In some countries, the standard power plug has symmetrically arranged plugs (interchanging of Line and Neutral is possible). In such cases, the **SLOPE** +/- pushbutton may indicate the wrong polarity compared with the display (triggering with falling edge instead of rising edge). For correction, the power plug of the instrument has to be turned.

Triggering of video signals

The built-in active **TV-Sync-Separator** separates the sync pulses from the video signal, permitting the display of distorted video signals either in line (H = horizontal) or in frame (**V** = vertical) trigger mode. The TV lever switch has five positions: the **OFF** position is for normal operation.

The TV: H+ and H- positions (horizontal= line) and the TV: V+ and V- (vertical = frame) positions are used for video triggering. In these four positions the **TRIG.** coupling switch and the LEVEL control (in NORM. trigger mode) are inoperative. In the TV: V+ and V- positions (frame triggering), a low-pass filter or integrating network is connected into circuit, which forms a trigger pulse sequence with frame frequency from thevertical sync pulses (incl. pre-and postequalizing pulses).

When in V mode, it is possible to select field I or II by releasing or depressing FIELD **I/II** pushbutton.

For correct video triggering, the + and – positions at V and H must be selected corresponding to the video input signal. If the sync pulses are placed above the picture content, H+ or V+ should be in use. For sync pulses below the picture content of the input signal, correct triggering, without any influence from changing picture contents, will be possible only in V- or **H**– setting. The INVERT pushbutton only changes the display on the CRT, not the input signal.

In TV: H trigger mode, the trigger point lies on the starting edge of a sync pulse if SLOPE button is in + position. As mentioned before, in TV: V mode an integrating network is additionally added to the sync separator which delays the formed trigger pulse by about $50\,\mu$ s.

Video signals are triggered in the automatic mode. Therefore the adjustment of the trigger point is superfluous. The internal triggering is virtually independent of the display height, which may differ from 0.8 to 8div. As opposed to **AT** mode, when in **normal** mode, the screen is blanked without signal at the input (turning the LEVEL knob is ineffectual).

Function of var. HOLD-OFF control

If it is found that a trigger point cannot be located on extremely complex signals even after repeated and careful adjustment of the LEVEL control in the **Norma/ Triggering** mode, a stable display may be obtained using the **HOLD**-OFF control (in the X-Section). This facility varies the holdoff time between two sweep periods up to the ratio >5:1. Pulses or other signal waveforms appearing during this off period cannot trigger the timebase. Particularly with burst signals or aperiodic pulse trains of the same amplitude, the start of the sweep can be shifted to the optimum or required moment. After specific use the HOLD-OFF control should be re-set into its calibration detent **min.**, otherwise the brightness of the display is reduced drastically.

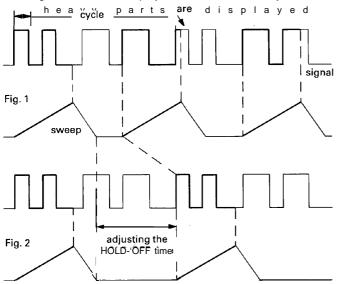


Fig. 1 shows a case where the HOLD-OFF knob is in the **XI** position and various different waveforms are overlapped on the screen, making the signal observation unsuccessful.

Fig. 2 shows a case where only the desired parts of the signal are stably displayed.

Sweep Delay / After Delay Triggering

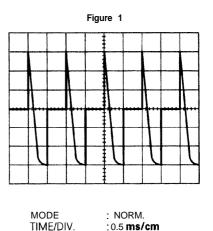
With the sweep delay, the start of the sweep can be delayed from the trigger point by a selectable time (100 ns to maximum 1 s). It is therefore possible to start the sweep at practically any point of a waveform. The interval, which follows the start of the sweep, can be greatly expanded by the increase of the sweep speed. From the 5 p/cm TIME/DIV. range downwards to slower sweep speeds, an expansion of at least 100 times, and with the aid of the X MAG. x10 expansion of 1000 times, is possible. With time coefficients higher than 5µs/cm, the maximum expansion increases proportionally. However, with increasing the expansion, the display brightness decreases. Under very high ambient light conditions a Viewing Hood like HZ47 can overcome this problem. It should be noted that there are some difficulties with higher expansions, if the test signal has inherent jitter. To reduce or eliminate this jitter, expanded parts of a signal can be triggered again "after delay" provided there is another suitable edge (DEL. TRIG.).

Subject to change without notice

In DEL. TRIG. mode **and TV** SEP. switch in H or V position, after delay triggering to the next following line is possible. Therefore discrete lines are representable. The slope is appointed by the TV+ or N- position of the delay switch. When not in TV-trig. mode the susceptibility to interference **(sense)** can be influenced (\triangleq = normal, \checkmark = reduced).

Operation of the sweep delay is relatively easy, as normally only 3 controls in the X-Section need to be used: the DELAY operating mode lever switch (OFF-SEARCH-DELAY-DEL. TRIG.), the DELAY rotary switch (delay time range), and its variable control VAR. **10:1** (small knob on the DELAY switch). The latter, a twenty-turn precision potentiometer with overwind protection, can increase the delay time range tenfold. An LED near the DELAY mode switch indicates the operating mode.

For reliable operation of the sweep delay, it is recommended that the following procedure should always be adopted; also reference to the accompanying figures will be of assistance.

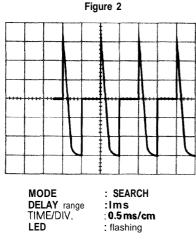


: off

LED

Initially, the sweep delay mode lever switch should be set in the OFF position. In this mode, the complete waveform under investigation will be displayed as for normal oscilloscope operation. The mode indicator LED is not illuminated in OFF mode. The time coefficient on the **TIME/DIV**, switch is selected so that 1 to 3 basic periods of the signal are displayed. A larger number unnecessarily decreases the brightness and maximum expansion. The display of only a portion of a period limits the choice of the expanded time interval and possibly complicates the triggering. On the other hand, the range of 1 to 3 basic periods can always be set unconstrainedly with the TIME/DIV. switch. In doing so, the x10 expansion must be switched off temporarily (X MAG. x10 button is in out position). In the X-Section, the HOLD-OFF control should be set to min. and the variable control to CAL. The LEVEL control is adjusted so that stable triggering is ensured (TRIG. LED is illuminated).

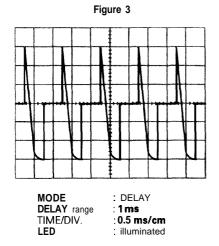
The mode switch should now be set to the SEARCH position; it will be seen that the start of the display will shift to the right. The amount of shift indicates the exact delay time. If a display is not obtained in this mode, then a lower delay time range should be selected. For example, when investigating the waveform shown in the figures, a display could not be obtained with a delay time setting of **10 ms**, as the display is completely blanked. However, as a result of setting the **DELAY** rotary switch to **0.1** μ **s**, the shifting is not visible. The **DELAY** range switch should then be rotated clockwise until the display starts just prior to the short time interval to be investigated. The precise adjustment to the start is done with the VAR. **10:1** delay time control. The rotating range of the latter has no stop. On the range limits a certain snapping noise is audible. Initially, this control should be set in the left start position. In the SEARCH mode, the LED indicator will flash.



Delay time = $2.5 \text{ cm} \cdot 0.5 \text{ ms/cm} = 1.25 \text{ ms}$

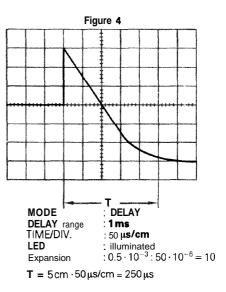
In figure 2 it can be seen that the delay time is also measurable as the blanked portion or apparent shift of the start of the trace. This time can be determined by multiplication of (the horizontal shifting in cm) by (the time coefficient set on the **TIME/DIV**. switch).

Now the mode switch can be set to DELAY. In this mode, the LED is permanently illuminated. The display will now shift to the left and the trace will commence in the same position as for a normal display; however, the short time interval under investigation now starts on the first or left vertical graticule line.



If the timebase sweep speed is increased (rotate TIME/ DIV. switch clockwise), then the short time interval will be expanded. It may be found that, as the amount of expansion is increased, the trace will tend to shift. If this happens, the VAR. delay time control can be readjusted — also subsequently at any time — to enable the exact point of interest to be displayed.

In the example shown in figure 4, it can be seen that an expansion of **x10** was obtained by increasing the timebase sweep speed *from 0.5 ms/cm to 50 µs/cm*. Also the precise measurement for the delayed portion of the waveform is possible. In the example, this was found to be $250 \mu s$ on multiplication of the horizontal length in cm (of an optional signal section) by the time coefficient just adjusted.



Operation of the sweep delay requires a constant trigger point. All signals, which have a constant phase shift between the expanded section and trigger point, pose no problems. This means all electrical signal shapes, which contain signal edges of the same polarity and with triggerable level values, which are constantly repeated with the recurring frequency.

If there is no constant phase shift, the triggering may fail after switching from the SEARCH to DELAY position or with changing of the time coefficient. It is best to attempt to find a trigger point, which has a constant phase shift up to the signal section to be expanded in the OFF mode. With complicated composite signals, the display of the basic period could become superimposed by other signal portions. These dissappear as a rule when the sweep is increased. Otherwise, a stable expanded display is obtained by adjusting the LEVEL and the variable sweep control or by means of the HOLD-OFF control.

Using the X MAG. **x10** button, a *tenfold expansion* of the desired signal section is possible without any change of triggering or timebase. This can be of assistance with complicated or difficult-to-trigger signals.

Operation of the sweep delay needs some experience, particularly with composite signals. However, the display of sections from simple signal waveforms is easily possible. It is recommended to operate only the sequence **OFF-SEARCH-DELAY**, because otherwise location of the short time interval to be investigated will be relatively difficult. The sweep delay facility can be used in the following modes: *Mono, Dual*, and *Algebraic Addition* (±**I**+**II**).

Delay Mode Indication

Both operating modes of the sweep delay are indicated with an LED, located to the right of the DELAY mode lever switch. In **SEARCH** position, the LED will flash. This is an indication of the temporary operating state. The **DELAY** position is indicated by constant lighting of the LED. However, should this be noted, and normal operating mode is required then the change-over of the lever switch to its **OFF** position has been overlooked. This could cause errors in displaying a signal by complete or partial blanking. This indication, therefore, should be closely observed.

Component Tester

General

The HM 604 has a built-in electronic Component Tester (abbreviated CT), which is used for instant display of a test pattern to indicate whether or not components are faulty. The CT can be used for quick checks of semiconductors (e.g. diodes and transistors), resistors, capacitors, and inductors. Certain tests can also be made to integrated circuits. All these components can be tested in and out of circuit.

The test principle is fascinatingly simple. The power transformer of the oscilloscope delivers a sine voltage, which is applied across the component under test and a built-in fixed resistor. The sine voltage across the test object is used for the horizontal deflection, and the voltage drop across the resistor (i.e. current through test object) is used for vertical deflection of the oscilloscope. The test pattern shows a current-voltage characteristic of the test object.

Since this circuit operates with mains/line frequency (50 or 60 Hz) and a voltage of 8.5V max. (open circuit), the indicating range of the **CT** is limited. The impedance of the component under test is limited to a range from 20Ω to 4.7 k Ω . Below and above these values, the test pattern shows only short-circuit or open-circuit. For the interpretation of the displayed test pattern, these limits should always be borne in mind. However, most electronic components can normally be tested without any restriction.

Using the Component Tester

The *CT* is switched on by depressing the **COMPONENT TESTER** pushbutton. This makes the vertical preamplifier

and the timebase generator inoperative. A shortened horizontal trace will be observed. It is not necessary to disconnect scope input cables unless in-circuit measurements are to be carried out. In the *CT* mode, the only controls which can be operated are **INTENS.**, **FOCUS**, and **X-POS.**. All other controls and settings have no influence on the test operation.

For the component connection, two simple test leads with $4mm \emptyset$ banana plugs, and with test prod, alligator clip or sprung hook, are required. The test leads are connected to the insulated *CT* socket and the adjacent ground socket in the Y-Section. The component can be connected to the test leads either way round.

After use, to return the oscilloscope to normal operation, release the **COMPONENT TESTER** pushbutton.

Test Procedure

Caution! Do not testanycomponent in live circuitry – remove all grounds, power and signals connected to the component under test. Set up Component Tester as stated above. Connect test leads **acoss** component to be tested. Observe oscilloscope display.

Only discharged capacitors should be tested!

A built-in quick-acting fuse protects the *CT* and the oscilloscope against mis-operation, e.g. device under test not disconnected from mains/line supply. In that case the fuse will blow. For fuse replacement the oscilloscope has to be opened (see service instruction page S1 "Instrument Case Removal"). The fuse is located on the bottom side of the instrument (close to the CT pushbutton). Make sure that only fuses of the specified type are used for replacement: **5x20mm**, **quick-acting**, **250V**, **C**, **50mA** (IEC 127/II or DIN 41661).

Test Pattern Displays

Page M 18 shows typical test patterns displayed by the various components under test.

- Open circuit is indicated by a straight horizontal line.

- Short circuit is shown by a straight vertical line.

Testing Resistors

1

If the test object has a linear ohmic resistance, both deflecting voltages are in the same phase. The test pattern expected from a resistor is therefore a sloping straight line. The angle of slope is determined by the resistance of the resistor under test. With high values of resistance, the slope will tend towards the horizontal axis, and with low values, the slope will move towards the vertical axis.

Values of resistance from 20Ω to $4.7k\Omega$ can be approximately evaluated. The determination of actual values will come with experience, or by direct comparison with a component of a known value.

Testing Capacitors and Inductors

Capacitors and inductors cause a phase difference between current and voltage, and therefore between the X and Y deflection, giving an ellipse-shaped display. The position and opening width of the ellipse will vary according to the impedance value (at 50 or 60Hz) of the component under test.

A horizontal ellipse indicates a high impedance or a relatively small capacitance or a relatively high inductance.

A vertical ellipse indicates a small impedance or a relatively large capacitance or a relatively small inductance.

A sloping ellipse means that the component has a considerable ohmic resistance in addition to its reactance.

The values of capacitance of normal or electrolytic capacitors from **0.1** μ **F** to **1000** μ **F** can be displayed and approximate values obtained. More precise measurement can be obtained in a smaller range by comparing the capacitor under test with a capacitor of known value. Inductive components (coils, transformers) can also be tested. The determination of the value of inductance needs some experience, because inductors have usually a higher ohmic series resistance. However, the impedance value (at 50 or 60 Hz) of an inductor in the range from 20 S2 to 4.7 k Ω can easily be obtained or compared.

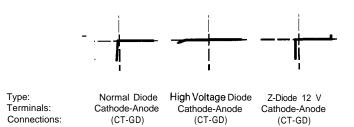
Testing Semiconductors

Most semiconductor devices, such as diodes, Z-diodes, transistors, FETs can be tested. The test pattern displays vary according to the component type as shown in the figures below.

The main characteristic displayed during semiconductor testing is the voltage dependent knee caused by the junction changing from the conducting state to the non conducting state. It should be noted that both the forward and the reverse characteristic are displayed simultaneously. This is always a two-terminal test, therefore testing of transistor amplification is not possible, but testing of a single junction is easily and quickly possible. Since the CT test voltage applied is only very low (max. $8.5V_{rms}$), all sections of most semiconductors can be tested without damage. However, checking the breakdown or reverse voltage of high voltage semiconductors is not possible. More important is testing components for open or short-circuit, which from experience is most frequently needed.

Testing Diodes

Diodes normally show at least their knee in the forward characteristic. This is not valid for some high voltage diode types, because they contain a series connection of several diodes. Possibly only a small portion of the knee is visible. Zdiodes always show their forward knee and, up to approx. 1 OV, their Z-breakdown, forms a second knee in the opposite direction. A Z-breakdown voltage of more than 12V can not be displayed.

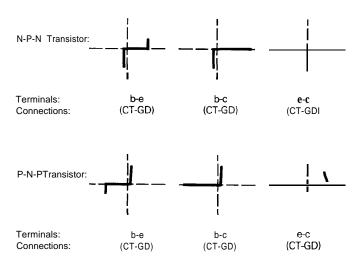


The polarity of an unknown diode can be identified by comparison with a known diode.

Testing Transistors

Three different tests can be made to transistors: base-emitter, base-collector and emitter-collector. The resulting test patterns are shown below.

The basic equivalent circuit of a transistor is a Z-diode between base and emitter and a normal diode with reverse polarity between base and collector in series connection. There are three different test patterns:



For a transistor the figures b-e and b-c are important. The figure e-c can vary; but a vertical line only shows short circuit condition.

These transistor test patterns are valid in most cases, but there are exceptions to the rule (e.g. Darlington, FETs). With the *CT*, the distinction between a P-N-P to a N-P-N transistor is discernible. In case of doubt, comparison with a known type is helpful. It should be noted that the same socket connection (*CT* or ground) for the same terminal is then absolutely necessary. A connection inversion effects a rotation of the test pattern by 180 degrees round about the center point of the scope graticule. Pay attention to the usual caution with single **MOS-com**ponents relating to static charge or frictional electricity!

In-Circuit Tests

Caution! During in-circuittests make sure the circuit is dead. No power from mains/line or battery and no signal inputs are permitted. Remove all ground connections including Safety Earth (pull out power plug from outlet). Remove all measuring cables including probes between oscilloscope and circuit under test. Otherwise the connection of both CT test leads is not recommended.

In-circuit tests are possible in many cases. However, they are not so well-defined. This is caused by a shunt connection of real or complex impedances – especially if they are of relatively low impedance at 50 or 60 Hz – to the component under test, often results differ greatly when compared with single components. In case of doubt, one component terminal may be unsoldered. This terminal should then be connected to the insulated *CT* socket avoiding hum distortion of the test pattern.

Another way is a test pattern comparison to an identical circuit which is known to be operational (likewise without power and any external connections). Using the test prods, identical test points in each circuit can be checked, and a defect can be determined quickly and easily. Possibly the device itself under test contains a reference circuit (e.g. a second stereo channel, push-pull amplifier, symmetrical bridge circuit), which is not defective.

The test patterns on page M 18 show some typical displays for in-circuit tests.

Miscellaneous

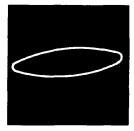
A posifive-going **sawtooth voltage** of approximately **5** V_{pp} coincident with display's sweep time is available at a BNC output connector on the rear panel. This ramp output is marked with \mathcal{W} . The load impedance should not be less than 10 k Ω II 47 pF. If the DC potential of the ramp output is not required, a capacitor should be connected in series with the output. The ramp output can be used for different measuring tasks in combination with the oscilloscope and other instruments, triggering of signal sources, swept-frequency signal generators and so on.

The oscilloscope also contains a vertical output with BNC connector marked Y on the rear panel. The output voltage is \geq **50 mV**_{pp}/cm display height (into 50 Ω). It is picked off from the vertical amplifier like the trigger signal and it is similarly switchable. Channel or II is selected with the CHI/II-TRIGI/II pushbutton. With alternate channel switching (DUAL button in the Y-Section depressed) and alternate triggering (ALT. button in the X-Section depressed), the vertical output is consecutively driven (in time with the sweep period) from Channel I and Channel II. The vertical output is not dependent on the vertical trace position. It does not respond to the adjustment of the Y-POS.I and Y-POS.II controls and to the depressing one of the INVERT buttons. The vertical output is DC coupled and has approximately zero level to ground. The bandwidth of the output is approx. 60 MHz (with 50 Ω termination).

Single Components



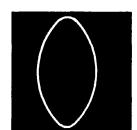
Short circuit



Mains transformer prim.



Resistor 510 Ω



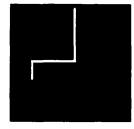
Capacitor 33 µF

Single Transistors



Junction B-C

Junction E-C

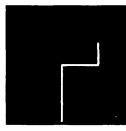


Junction B-E

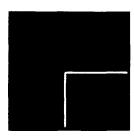


FET

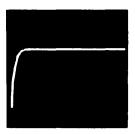
Single Diodes



Z-diode under 8V



Silicon diode

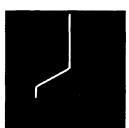


Rectifier

Z-diode beyond 12V

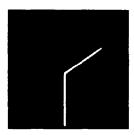


Germanium diode

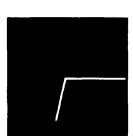


Thyristor G + A together

In-circuit Semiconductors



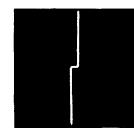
Diode paralleled by $\textbf{680}\,\Omega$



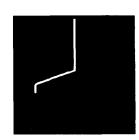
Diode in series with 51 Ω



B-E with 1 $\mu \textbf{F}$ + 68052



2 Diodes antiparallel



B-E paralleled by6808



Si-diode with 10 μF

First Time Operation

Connect the instrument to power outlet. Switch on POWER pushbutton. No other button is depressed. LED indicates operating condition.

Case, chassis, and all measuring connectors are connected to

the Safety Earth Conductor (Safety Class I).

TRIG. selector switch to AC, TV SEP. switch to OFF, LEVEL knob in AT position (Automatic Triggering) DELAY lever switch to OFF. and HOLD-OFF control min.

Adjust INTENS. control for average brightness.

Center trace on screen using X-POS. and Y-POS.I controls. Then focus trace using FOCUS control.

Operating Modes of the Vertical System

Channel I: All pushbuttons in out position.

Channel II: CH I/II -TRIG. I/II button depressed.

Channel I and Channel II: DUAL button depressed.

Alternate channel switching: ADD. button in out position.

Chopped channel switching: DUAL **and ADD** buttons depressed. Signals <1kHz with CHOP.

Channel I+ II (sum): ADD button depressed only.

Channel - I + II (difference): ADD and INVERT (CH. I) buttons depressed.

Channel + I - II (difference): ADD and INVERT (CH. II) buttons depressed.

Trigger Modes

Automatic Triggering: LEVEL knob in ATposition. Trace always visible. Normal Triggering: LEVEL turnded cw. Trace visible when triggered. Triggering from positive-going signal edge: SLOPE +/- button is in out position. Triggering from negative-going signal edge: SLOPE +/- button depressed. This facility is important when only a portion of a cycle is being displayed. Internal triggering: select Channel with button CH. I/II-TRIG. I/II. Internal alternate triggering: ALT button is depressed (only with alternate channel switching). External triggering from TRIG. INP. connector: EXT. button depressed. External trigger signal: $50 \text{ mV}-0.5 \text{ V}_{pp}$, time-related to vertical input signal. Line triggering: **TRIG.** selector switch in \sim position. Trigger coupling selected with TRIG. switch AC-DC-LF-HF. Trig. freg. range: AC and DC to 20 MHz, HF above 20 MHz, LF below 50 kHz. Video signals with line freq .: TV SEP. switch to H+ or H-. Video signals with frame freq.: TV SEP. switch to V+ or V-. First or second half of frame selection with FIELDI/II switch. Pay attention to TRIG. indication LED!

Measuring

Connect test signal to CH.I and/or CH.II vertical input connector.

Compensate attenuator probe using CAL. square-wave signal.

Select AC or DC input coupling. GD: Y amplifier is disconnected from input and grounded.

Adjust required display height of signal with attenuator switch and variable control.

Select sweep speed with TIME/DIV. switch and variable control.

Adjust trigger point with LEVEL control.

Calibrated amplitude measurement with attenuator variable control to CAL.

Calibrated time measurement with TIME/DIV. variable control to CAL.

Trigger complex or aperiodic signals using **HOLD-OFF** control in normal trigger mode.

Horizontal xI 0 expansion: X MAG. x10 button depressed.

External horizontal deflection: (X-Y operation) with X-Y button depressed (X input via CH.II).

Sweep Delay Operation

OFF:	normal oscilloscope operation. DELAY LED not illuminated.
SEARCH:	use DELAY range switch and VAR. 10:1 fine control to select point of interest on
	displayed wave-form. DELAY LED flashing.
DELAY:	delayed signal now displayed. Expansion obtained by rotating TIME/DIV. switch
	clockwise. Press X MAG. x10 button if necessary. DELAY LED illuminated.
DEL TRIG.:	After Delay Triggering; together with TV SEP .: selection of trigger slope.

Component Tester

Press **COMPONENT TESTER** button. Connect both component terminals to **CT** and ground jacks. *In-circuit test:* Circuit under test must be disconnected to battery or power (pull out power plug), signals and ground (earth). Remove all signal connections to HM 604 (cable, probe), then start testing.

General

These Test Instructions are intended as an aid for checking the most important characteristics of the HM 604 at regular intervals without the need for expensive test equipment. Resulting corrections and readjustments inside the instrument, detected by the following tests, are described in the Service Instructions or on the Adjusting Plan. They should only be undertaken by qualified personnel.

As with the First Time Operation instructions, care should be taken that all knobs with arrows are set to their calibrated positions. None of the pushbuttons should be depressed. LEVEL knob out in AT position, TRIG. selector switch to AC, DELAY slide switch to OFF. It is recommended to switch on the instrument for about 30 minutes prior to the commencement of any check.

Cathode-Ray Tube: Brightness and Focus, Linearity, Raster Distortions

Normally, the CRT of the HM 604 has very good brightness. Any reduction of this brightness can only be judged visually. However, decreased brightness may be the result of reduced high voltage. This is easily recognized by the greatly increased sensitivity of the vertical amplifier. The control range for maximum and minimum brightness (intensity) must be such that the beam just disappears before reaching the left hand stop of the **INTENS.** control (particularly when the X-Y button is depressed), while with the control at the right hand stop the focus and the line width are just acceptable.

With maximum intensity the timebase fly-back must onnoaccountbe visible. Visible display fault without input signal: Bright dot on the left side - or - decreasing brightness from left to right or shortening of the baseline. (Cause: incorrect Unblanking Pulse.

It should be noted that with wide variations in brightness, refocusing is always necessary. Moreover, with maximum brightness, no "pumping" of the display must occur. If pumping does occur, it is normally due to a fault in the regulation circuitry for the high voltage supply. The presetting pots for the high voltage circuit, minimum and maximum intensity, are only accessible inside the instrument (see Adjusting Plan and Service Instructions).

A certain out-of-focus condition in the edge zone of the screen must be accepted. It is limited by standards of the CRT manufacturer. The same is valid for tolerances of the orthogonality, the undeflected spot position, the non-linearity and the raster distortion in the marginal zone of the screen in accordance with international standards (see CRT data book). These limit values are strictly supervised by HAMEG. The selection of a cathode-ray tube without any tolerances is practically impossible.

Astigmatism Check

Check whether the horizontal and vertical sharpness of the display are equal. This is best seen by displaying a square-wave signal with the repetition rate of approximately 1 MHz. Focus the horizontal tops of the square-wave signal at normal intensity, then check the sharpness of the vertical edges. If it is possible to improve this vertical sharpness by turning the FOCUS control, then an adjustment of the astigmatism control is necessary. A potentiometer of 50 k Ω (see Adjusting Plan) is provided inside the instrument for the correction of astigmatism (see Service Instructions). A certain loss of marginal sharpness of the CRT is unavoidable; this is due to the manufacturing process of the CRT.

Symmetry and Drift of the Vertical Amplifier

Both of these characteristics are substantially determined by the input stages of the amplifiers. *The checking and* **correction** of the DC balance for the amplifiers should be carried out as already described in the Operating Instructions (page M 7).

The symmetry of Channel I and the vertical final amplifier can be checked by inverting Channel I, depress INVERT (CH I pushbutton). The vertical position of the trace should not change by more than 5mm. However, a change of **1** cm is just permissible. Larger deviations indicate that changes have occurred in the amplifier.

A further check of the vertical amplifier symmetry is possible by checking the control range of the **Y-POS**. controls. A sine-wave signal of 1 O-I 00 kHz is applied to the amplifier input. When the Y-POS. control is then turned fully in both directions from stop to stop with a display height of approximately **8***cm*, the upper and lower positions of the trace that are visible should be approximately of the same height. Differences of up to **1** cm are permissible (input coupling should be set to **AC**).

Checking the drift is relatively simple. **Ten minutes after switching on the instrument**, set the baseline exactly on the horizontal center line of the graticule. The beam position must not change by more than *5mm* during the following hour. Larger deviations generally result from different characteristics of the dual FETs in both channel inputs to the Y amplifier. To some extent, fluctuations in drift are caused by offset current on the gate. The drift is too high, if the vertical trace position drifts by more than **0.5mm** on turning the appropriate attenuator switch through all 12 steps. Sometimes such effects occur after long periods of operation.

Calibration of the Vertical Amplifier

Two square-wave voltages of $\textit{0.2mV}_{pp}$ and 2 $\textit{V}_{pp}\pm$ 1 % are present at the output sockets of the calibrator (CAL.). If a

direct connection is made between the 0.2 mV output and the input of the vertical amplifier, the displayed signal in the **50 mV/cm** position (variable control to **CAL.)** should be **4 cm** high (DC input coupling). Maximum deviations of 1.2 mm (3%) are permissible. If a x **10 probe** is connected between the 2V-output socket and Y input, the same display height should result. With higher tolerances it should first be investigated whether the cause lies, within the amplifier or in the amplitude of the square-wave signal. On occasions it is possible that the probe is faulty or incorrectly compensated. If necessary the measuring amplifier can be calibrated with an accurately known DC voltage (DC input coupling). The trace position should then vary in accordance with the deflection coefficient set.

With variable control at the attenuator switch fully counterclockwise, the input sensitivity is decreased at least by the factor 2.5 in each position. In the **50mV/cm** position, the displayed calibrator signal height should **vary** from 4cm to at least 1.6 cm.

When pulling the Y-expansion x5 knob (MAG x5), the sensitivity is increased by the factor 5. In the 0.2V/cm position the displayed signal should change from 1 cm to 5cm by pulling the MAG x5 knob.

Transmission Performance of the Vertical Amplifier

The transient response and the delay distortion correction can only be checked with the aid of a square-wave generator with a fast risetime (max. 5ns). The signal coaxial cable (e.g. HZ34) must be terminated at the vertical input of the oscilloscope with a resistor equal to the characteristic impedance of the cable (e.g. with HZ22). Checks should be made at 1 00 Hz, 1 kHz, 10 kHz, 100 kHz and 1 MHz, the deflection coefficient should be set at 5mV/cm with DC input coupling (Y variable control in CAL. position). In so doing, the square pulses must have a flat top without rampoff, spikes and glitches; no overshoot is permitted, especially at 1 MHz and a display height of 4-5 cm. At the same time, the leading top corner of the pulse must not be rounded. In general, no great changes occur after the instrument has left the factory, and it is left to the operator's discretion whether this test is undertaken or not.

Of course, the quality of the transmission performance is not only dependent on the vertical amplifier. **The input attenuators,** located in the front of the amplifier, **are frequency-compensated in each position.** Even small capacitive changes can reduce the transmission performance. Faults of this kind are as a rule most easily detected with a square-wave signal with a low repetition rate (e.g. 1 kHz). If a suitable generator with max. output of $40V_{pp}$ is available, it is advisable to check at regular intervals the deflection coefficients on all positions of the input attenuators and readjust them as necessary. A compensated 2:1 series attenuator is also necessary, and this must be matched to the input impedance of the oscilloscope. This attenuator can be made up locally. It is important that this attenuator is shielded. For local manufacture, the electrical components required are a 1 M $\Omega \pm 1\%$ resistor and, in parallel with it, a trimmer 3-I 5 pF in parallel with approx. 20 pF. One side of this parallel circuit is connected directly to the input connector of the vertical amplifier and the other side is connected to the generator, if possible via a low-capacitance coaxial cable. The series attenuator must be matched to the input impedance of the oscilloscope in the 5mV/cm position (variable control to CAL., DC input coupling; square tops exactly horizontal; no ramp-off is permitted). This is achieved by adjusting the trimmer located in the 2: 1 attenuator. Theshapeofthesguare-waveshould then be the same in each input attenuator position.

Operating Modes: CH.I/II - TRIG.I/II, DUAL, ADD, CHOP., INV.I/II and X-Y Operation

On depressing the **DUAL** pushbutton, two traces must appear immediately. On actuation of the Y-POS. controls, the trace positions should have no effect on each other. Nevertheless, this cannot be entirely avoided, even in fully serviceable instruments. When one trace is shifted vertically across the entire screen, the position of the other trace must not vary by more than 0.5 mm.

A criterion in chopped operation is trace widening and shadowing around and within the two traces in the upper or lower region of the screen. Set **TIME/DIV**. switch to **1** μ **s**/**cm**, depress the DUAL and CHOP. pushbutton, set input coupling of both channels to GD and advance the **INTENS**. control fully clockwise. Adjust FOCUS for a sharp display. With the Y-POS. controls shift one of the traces to a +2 cm, the other to a -2cm vertical position from the horizontal center line of the graticule. Do not try to synchronize the chop frequency (500 kHz)! Then alternately release and depress the **CHOP**. pushbutton. Check for negligible trace widening and periodic shadowing in the chopped mode.

It is important to note that in the **I+II** add mode (only **ADD** depressed) or the **-I+II** difference mode **INVERT** (CH I) button depressed in addition) the vertical position of the trace can be adjusted by using **both** the Channel I and Channel II Y-POS. controls. If a trace is not visible in either these modes, the overscanning LEDs will indicate the position of the trace.

In X-Y Operation (X-Y pushbutton depressed), the sensitivity in both deflection directions will be the same. When the signal from the built-in square-wave generator is applied to the input of Channel II, then, as with Channel I in the vertical direction, there must be a horizontal deflection of **4***cm* when the deflection coefficient is set to **50 mV/cm** position (variable control set to its CAL. position, X MAG. $\mathbf{x10}$ released). The check of the mono channel display with the CH **I/II** button is unnecessary; it is contained indirectly in the tests above stated.

Triggering Checks

The internal trigger threshold is important as it determines the display height from which a signal will be stably displayed. It should be approx. 5mm for the HM 604. An increased trigger sensitivity creates the risk of response to the noise level in the trigger circuit, especially when the sensitivity of the vertical input is increased by pulling the MAG. x5 knob. This can produce double-triggering with two out-of-phase traces. Alteration of the trigger threshold is only possible internally. Checks can be made with any **sine-wave voltage** between 50 Hz and 1 MHz. The LEVEL knob should be in AT position. Following this it should be ascertained whether the same trigger sensitivity is also present with Normal Triggering (LEVEL knob not in AT position).

On depressing the SLOPE +/- button, the start of the sweep changes from the positive-going to the negative-going edge of the trigger signal.

Internally the HM 604 should trigger perfectly with sinosoidal signals up to $100\,MHz$ at a display height of approx. 5mm , when the HF trigger coupling mode is set.

For external triggering (EXT. button depressed), the EXT. TRIG. input connector requires a signal voltage of at least $50 \,\text{mV}_{pp}$, which is in synchronism with the Y input signal. Checking of TV triggering is possible with a video signal of switchable polarity. A check of both polarities in V and H mode should be made.

The display should not shift horizontally during a change of the trigger coupling from AC to DC with a sine-wave signal without DC offset. The basic requirement for this is a correct DC Balance *Adjustment* on the input of the vertical amplifier (see Operating Instructions, page M7).

In the **dual channel mode** (DUAL button depressed) with alternate channel switching and with alternate triggering (ALT. button in the X-Section depressed), two non-frequency related signals (e.g. mains/line frequency signal and calibrator signal) should reliably be triggered internally dependent on the positions of the **CHI/II-TRIG.I/II** pushbuttons. In the dual channel mode with chop channel switching and depressed ALT. button, only triggering from Channel I should be possible. Periodical signal blanks (due to the chopper frequency 0.5 MHz) should not be visible.

If both vertical inputs are AC coupled to the same signal and both traces are brought to coincide exactly on the screen, when working in the *alternate dual channel mode*, then no change in display should be noticeable, when the ALT. button is depressed or released. Checking of the line/mains frequency triggering (50-60 Hz) is possible, when the input signal is time-related (multiple or submultiple) to the power line frequency (TRIG. selector switch to LINE). In this trigger mode, there is no trigger threshold. Even very small input signals are triggered stably (e.g. ripple voltage). For this check, use an input of approx. 1 V. The displayed signal height can then be varied by turning the respective input attenuator switch and its variable control.

Timebase

Before checking the timebase it should be ascertained that the *trace length is* **10***cm*. If not, it can be corrected with the potentiometer for sweep amplitude (see Adjusting Plan). This adjustment should be made with the **TIME/DIV**. switch in a mid position (i.e. 5μ s/cm). Prior to the commencement of any check set the time variable control to CAL. and the HOLD-OFF time control to min. The X MAG. **x10** button should be released. This condition should be maintained until the variation ranges of these controls are checked.

If a precise marker signal is not available for checking the **Timebase** time coefficients, then an accurate sine-wave generator may be used. Its frequency tolerance should not be greater than ± 1 %. The timebase accuracy of the HM 604 is given as ± 3 %, but as a rule it is considerably better than this. For the simultaneous checking of timebase linearity and accuracy at least 10 oscillations, i.e. **1** cycle **everycm**, should always be displayed. For precise determination, set the peak of the first marker or cycle peak exactly behind the first vertical graticule line using the X-POS. control. Deviation tendencies can be noted after some of the marker or cycle peaks.

The 20 and **10ms/cm** ranges of the **TIME/DIV.** switch can be checked very precisely with line frequency (*50Hz* only). On the **20ms/cm** range a cycle will be displayed every cm, while on the **10ms/cm** range it will be every 2 cm.

The following table shows which frequencies are required for the particular ranges.

1 s/cm – 1	Hz	0.1 ms/cm – 10 kHz
0.5 s/cm - 2	Hz	50 @cm — 20 kHz
0.2 s/cm - 5	Hz	20 μ s/cm – 50 kHz
0.1 s/cm – 10	Hz	10 μ s/cm – 100 kHz
50 ms/cm – 20	Hz	5 μs/cm – 200 kHz
20 ms/cm - 50 I	Ηz	2 yslcm – 500 kHz
10 ms/cm – 100	Hz	1 μs/cm – 1 MHz
5 ms/cm – 200	Hz	0.5 μs/cm – 2MHz
2 ms/cm - 500	Hz	0.2 μs/cm – 5MHz
1 ms/cm – 1	kHz	0.1 μs/cm — 10MHz
0.5 ms/cm – 2	kHz	0.05
0.2 ms/cm – 5	kHz	

The *time variable control* range can also be checked. The sweep speed becomes slower by turning this variable control counter-clockwise to its left stop. 2.5 cycles at least every cm should be displayed (with X **MAG. x10** button released; measurement in the $5\mu s/cm$ range).

When the X **MAG. x10** button is depressed, a marker or cycle peak will be displayed every **10** cm \pm 5 % (with variable control in **CAL.** position; measurement in the **5µs/cm** range). The tolerance is better measureable in the **50µs/cm** range (one cycle every 1 cm).

Check the ramp output voltage on rear panel (BNC connector marked by M) with a Test Oscilloscope. Test Scope setting: 1 V/cm; Timebase to one step slower sweep speed than on the HM 604 under test. The latter must have no input and no trigger voltage (free-running sweep; input coupling switch to GD). The sawtooth voltage is applied with a BNC cable without termination from the ramp output connector to the input of the Test Scope. The Test Scope should show a positive-going linear sawtooth with an amplitude of approx. $5V_{pp}$. At the same time the function of the HOLD-OFF control min.-max. can be checked. The hold-off time variation cannot be measured precisely with this method, because the unblanking pulse of the timebase generator is smaller than the ramp width.

Sweep Delay

When the Sweep DELAY mode lever switch is set to the OFF mode, the delay should have no effect on the display of the **1kHz** calibration signal. When the Sweep Delay is set in the SEARCH mode (refer to Sweep Delay Operating Instructions), it is possible to check the delay time by means of a distance measurement of the blanked baseline. For this, the DELAY VAR. **10:1** control must be set to xl (rotate counterclockwise until a snap noise is audible). When DELAY mode is selected, the trace reverts to the full 10 cm display without any blanking.

Over the *full range of adjustment of the DELAY* VAR. 10: 1 *control, the displayed waveform of the calibration signal should be shifted without any jitter, jumping or intermittent blanking.*

Control settings: Connect calibrator socket (0.2 V/1kHz) to CH.1 input connector, DC input coupling, deflection coefficient 50mV/cm, TRIG. selector switch to AC, time coefficient 1 ms/cm, no pushbutton depressed, all controls in calibrated position, DELAY mode switch in OFF position, LEVEL knob in AT position. Now the calibrator signal is displayed with a signal height of 4cm and approx. 1 cycle per cm. After switching to SEARCH, the mode indication lamp blinks. Set the DELAY range switch to 1 ms. Rotate the DELAY VAR. control until the left half of the display is blanked. The delay time is now 5ms. After switching to DELAY, the display is again fully visible. The DELAY mode LED is illuminated continuously. The displayed signal can now be expanded. For this purpose turn the TIME/DR/. switch clockwise to 5µs/ cm. The expansion is now x200. With the DELAY VAR. control, the nearest edge of the calibration signal can be brought in the screen center and checked on the above-mentioned criteria. With x200 expansion, the display brightness normally needs increasing (with **INTENS.** and FOCUS control). However, larger expansions than x200 are quite possible, but the decrease of brightness and the jitter restricts the evaluation.

When switching to DEL. TRIG. positions, every slope is accepted for triggering after the delay time has elapsed. In **TVSEP.** mode condition, the slope can be chosen (+/-).

Component Tester

After pressing the **COMPONENT TESTER** button, a horizontal straight line has to appear immediately, when the **CT** socket is open. The length of this trace should be approx. **8cm**. With connection of the **CT** socket to one of the ground jacks in the Y-Section, a vertical straight line with approx. 6cm height should be displayed. The above stated measurements have some tolerances. They are dependent among other things on the mains/line voltage.

Trace Alignment

The CRT has an admissible angular deviation $\pm 5^{\circ}$ between the X deflection plane DI-D2 and the horizontal center line of the internal graticule. This deviation, due to tube producion tolerances (and only important after changing the CRT), and also the influence of the earth's magnetic field, which is dependent on the instrument's North orientation, are corrected by means of the TR potentiometer. In general, the trace rotation range is asymmetric. It should be checked, whether the baseline can be adjusted somewhat sloping to **both sides** round about the horizontal center line of the graticule. With the HM 604 in its closed case, an angle of rotation $\pm 0.57^{\circ}$ (1 mm difference in elevation per 10 cm graticule length) is sufficient for the compensation of the earth's magnetic field.

Miscellaneous

Y output

A check of the Y output (rear panel) is possible on the screen using the dual channel mode by means of the calibrator signal. To this a connection is made from the calibrator socket (0.2V/1kHz) to the CH.I input connector and - using a BNC-BNC cable and a 50 Ω through-termination – a second connection from the Y output to CH.II input connector. Settings: CH.I attenuator switch to 50mV/cm, CH.II attenuator switch to 0.1 V/cm, CH.I input coupling to DC, CH.II input coupling to GD, TIME/DIV. switch to 0.5 ms/cm, automatic triggering (LEVEL knob in AT position), TRIG. selector to AC, no button depressed. Now the square-wave signal is visible with 4cm display height. With Y-POS.I control, the tops of the square-wave are adjusted to ± 2 cm from the horizontal center line of the graticule. Then the DUAL button has to be pressed. The appearing second trace (without signal) is adjusted to -2cm using the Y-POS.II control. Then the CH.II input coupling is set to DC. Now the signal across the Y output appears with the same phase position as the calibrator signal via Channel I. As well as the DC offset (e.g. +0.8cm ≙ +80 mV) the amplitude (e.g. 2cm $\triangleq 0.2 V_{pp}$) of the Y output can be measured. In the example, the sensivity of the Y output can be calculated as 0.2V:4cm = 50mV/cm.

tor. Settings: CH.1 attenuator switch to 50 mV/cm, CH.II attenuator switch to 0.1 V/cm, CH.1 input coupling to DC, CH.II input coupling to GD, TIME/DIV. switch to 0.5 ms/ cm, automatic triggering (LEVEL knob in AT position), TRIG. selector to AC, no button depressed. Now the square-wave signal is visible with **4***cm* display height. With **Y-POS.I** control, the tops of the square-wave are adjusted to $\pm 2 \text{ cm}$ from the horizontal center

Y-POS.IICH.IIrolinT pantthecoupling is set to DC. Now the signal across the Y output appears with the same phase position as the calibrator signal via Channel I. As well as the DC offset (e.g. $+0.8 \text{ cm} \triangleq +80 \text{ mV}$) the amplitude (e.g. $2\text{ cm} \triangleq 0.2 \text{ V}_{pp}$) of the Y output can be measured. In the example, the sensivity of the Y output can be calculated as 0.2 V:4 cm = 50 mV/cm.

Z modulation (optional)

Checking the blanking facility on the **Z** modulation connector (rear panel) is possible with a sine- or square-wave generator. The sine-wave generator requires an output voltage control. The square-wave generator must deliver positive pulses to ground (chassis). Alternatively a small adjustable sine voltage from a (separate) power transformer may be used. For the latter set the **TIME/DIV**. range to e.g. **10 ms/cm**. With the TIMEBASE variable control, the gap in the baseline can be brought acceptably to a standstill. Line frequency triggering is better **(TRIG.** selector switch to \sim . The length ratio from bright to dark lines on the display is dependent on the sine voltage amplitude. Without a modulating generator, the function of the Z modulation can be checked coarsely by short-circuiting the **Z** connector. Then the baseline is blanked fully.

General

The following instructions are intended as an aid for the electronic technician, who is carrying out readjustments on the HM 604, if the nominal values do not meet the specifications. These instructions primarily refer to those faults, which were found after using the Test Instructions. However, this work should only be carried out by properly qualified personnel. For any further technical information call or write to HAMEG. Addresses are provided at the back of the manual. It is recommended to use only the original packing material, should the instrument be shipped to HAMEG for service or repair (see also Warranty, page M2).

Instrument Case Removal

The rear cover can be taken off after unplugging the power cord's triple-contact connector and after two cross recessed pan head screws ($M4x30\,mm$) with two washers on it have been removed. While the instrument case is firmly held, the entire chassis with its front panel can withdrawn forward. When the chassis is inserted into the case later on, it should be noticed that the case has to fit under the flange of the front panel. The same applies for the rear of the case, on which the rear cover is put.

Caution

During opening or closing of the case, the instrument must be disconnected from all power sources for maintenance work or a change of parts or components. If a measurement, trouble-shooting, or an adjustment is unavoidable, this work must be done by a specialist, who is familiar with the risk involved.

When the instrument is set into operation after the case has been removed, attention must be paid to the acceleration voltage for the CRT - 12.5kV – and to the operating voltages for both final amplifier stages – together 140V. Potentials of these voltages are on the CRT socket, on the upper and the lower horizontal *PCBs*, and on the lateral *PCB* directly beside the CRTneck. High voltages of max. 2000 V are at the **INTENS**. and FOCUS potentiometers. They are highly dangerous and therefore precautions must be taken. It should be noted furthermore that shorts occuring on different points of the CRT high voltage and unblanking circuitry will definitely damage some semiconductors. For the same reason it is very risky to connect capacitors to these points while the instrument is on.

Capacitors in the instrument **may** still be charged, even when the instrument is disconnected from all voltage sources. Normally, the capacitors are discharged 6 seconds after switching **off**. However, with a defective instrument an interruption of the load is possible. Therefore, after switching off, it is recommended to connect one by one all terminals of the check strip across $1 \ k\Omega$ to ground (chassis) for a period of 1 second.

Handling of the CRT needs utmost caution. The glass bulb must not be allowed – under any circumstances – to come into contact with hardened tools, nor should it undergo local superheating (e.g. by soldering iron) or local undercooling (e.g. by cryogenic-spray). We recommend the wearing of safety goggles (implosion danger).

Operating Voltages

Besides the two AC voltages for the CRT heating (6.3V) and graticule illumination, Component Tester and line triggering (12V) there are eight electronically regulated DC operating voltages generated (+12V, +5V, -5V, -12V, +68V, +140V, -1800V, and + **10.4** kV). These different operating voltages are fixed voltages, except the + 12V, which can be adjusted. The variation of the fixed voltages greater than ± 2 % from the nominal value indicates a fault. This voltage is measured on the checkpoint strip (located on XY Board) with reference to ground. Measurements of the high voltage may only be accomplished by the use of a sufficient highly resistive voltmeter (>10M\Omega). You must make absolutely sure that the electric strength of the voltmeter is sufficiently high.

Minimum Brightness

The variable resistor VR601, located on the Z-PCB, is used for this adjustment procedure. It may only be touched by a properly insulating screwdriver (Caution! High voltage!). Correct adjustment is achieved, when the trace can be blanked while X-Y pushbutton is depressed and, in addition, when the requirement described in the Test Instructions are met.

Astigmatism control

The ratio of vertical and horizontal sharpness can be adjusted by the variable resistor VR603, located on the Z-PCB (see Adjusting Plan). As a precaution however, the voltage for the vertical deflecting plates (approx. +41.5V) should firstly be checked, because this voltage will affect the astigmastism correction. Use the **1** MHz square-wave signal from the built-in calibrator via 10: **1** probe for this correction. Firstly adjust the sharpness of the *horizontal* square-wave lines using the FOCUS control. The sharpness of the Astigm. pot VR626. Repeat the correction several times in this sequence. Adjustment has been correctly carried out when, on using FOCUS control *only*, the sharpness is not improved in eitherdirection.

Trouble-Shooting the Instrument

For this job, at least an isolating variable mains/line transformer (protection class II), a signal generator, an adequate precise multimeter, and, if possible, an oscilloscope are needed. This last item is required for complex faults, which can be traced by the display of signal or ripple voltages. As noted before, the regulated high voltage and the supply voltage for the final stages are highly dangerous. Therefore it is recommended to use **totally insulated extended probe tips**, when trouble-shooting the instrument. Accidental contact with dangerous voltage potentials is then unlikely. Of course, these instructions cannot thoroughly cover all kinds of faults. Some common-sense will certainly be required, when a complex fault has to be investigated.

If trouble is suspected, visually inspect the instrument thoroughly after removal of the case. Look for loose or badly contacted or discolored components (caused by overheating). Check to see that all circuit board connections are making good contact and are not shorting to an adjacent circuit, Especially inspect the connections between the PCBs, to the power transformer, to front chassis parts, to CRT socket, to trace rotation coil (inside of CRT's shielding), to the 3 BNC connectors at the rear chassis, and to the control potentiometers and switches on top of and beneath both main-PCBs. Furthermore, the soldering connections of the transistors and Fixed Three-Terminal Regulators resp. on the rear chassis. This visual inspection can lead to success much more quickly than a systematic fault location using measuring instruments. Prior to any extensive troubleshooting, also check the external power source.

If the instrument fails completely, the first and most important step – *after checking the mains/line voltage and power fuse* – will be to measure the deflecting plate voltages of the CRT. In almost any case, the faulty section can be located. The sections represent:

- 1. Vertical deflection.
- 2. Horizontal Deflection.
- 3. CRT circuit.
- 4. Power supply.

While the measurement takes place, the position controls of both deflection devices must be in mid-position. When the deflection devices are operating properly, the separate voltages of each plate pair are almost equal then $(Y \approx 42V)$ and $X \approx 67 V$.

If the separate voltages of a plate pair are very different, the associated circuit must be faulty.

An absent trace in spite of correct plate voltages means a fault in the CRT circuit. Missing deflection plate voltages is probably caused by a defect in the power supply.

Replacement of Components and Parts

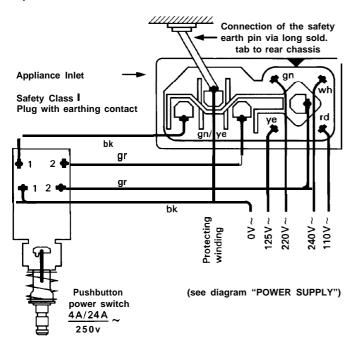
For the replacement of parts and components use only parts of the same or equivalent type. Resistors without specific data in the diagrams have a power dissipation of 0.25 Watt and a tolerance of 2 %. Resistors in the high voltage circuit must have sufficient electric strength. Capacitors without a voltage value must be rated for an operating voltage of 63V. The capacitance tolerance should not exceed 20%. Many semiconductors are selected, especially the gate-diodes 1 N4154, and all amplifier transistors, which are contained in push-pull circuits (including the FETs). If a selected semiconductor is defective, all gatediodes or both push-pull transistors of a stage should be replaced by selected components, because otherwise there are possibly deviations of the specified data or functions. The HAMEG Service Department can give you advice for troubleshooting and replaceable parts. Replacement parts can be ordered by letter or telephone from the nearest HAMEG Service Office. Please supply the following information: Instrument type and serial number, description of the part (including function and location in the instrument), quantity desired.

Replacement of the Power Transformer

Should it be necessary to replace the mains/line transformer, the correct terminal sequence (color identification) for primary and secondary windings must be followed (see diagram "Power Supply" and the figure below). In addition, the relevant Safety Regulations must be observed. Here, we refer only to those requirements relative to the parts conductively connected to the supply mains:

- The construction of the instrument shall be such as to prevent any short-circuiting or bridging of the insulation, clearances or Creepage distances between those parts connected to the supply mains and any accessible conductive parts due to accidental loosening or freeing of the wiring, screws, etc.
- The rigidity of the mains wiring connections, which may be subject to mechanical stresses, shall not be dependent upon the soldering alone. To meet this requirement, the bare ends of the wires must be pushed through the holes in the respective soldering tab, bent over with a pair of pliers, and subsequently fixed by soldering.
- The minimum cross section of the protective earth connection between the instrument's power inlet and the connecting soldering tab on the rear chassis must be 0.81 mm² in North America and 0.75 mm² in Western Europe. The connecting soldering tab on the rear chassis has to be secured mechanically against loosening (e.g. with lock washer).

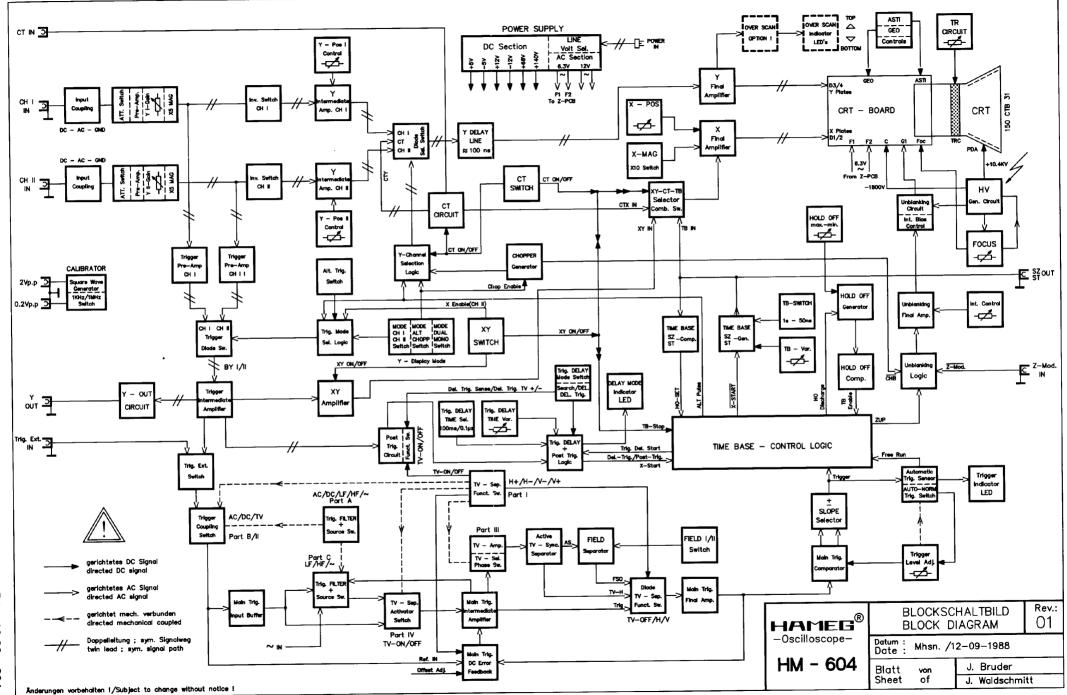
After replacing the power transformer, all remaining bits of wire, solder and other foreign matter must be removed from the PCBs, the vicinity of the power transformer and from within the insulating connecting box by shaking, brushing and blowing. Finally, the top plate of the insulating connecting box has to be replaced. Before connecting the instrument to the power supply, replace the possibly defective fuse, press the **POWER** button and make sure that there is an adequate insulation state between chassis (= safety earth conductor) on the one hand, and the live/line pin as well as the neutral pin, on the other. Only after proper insulation has been established may further function tests with open chassis follow, but with appropriate precautionary measures.



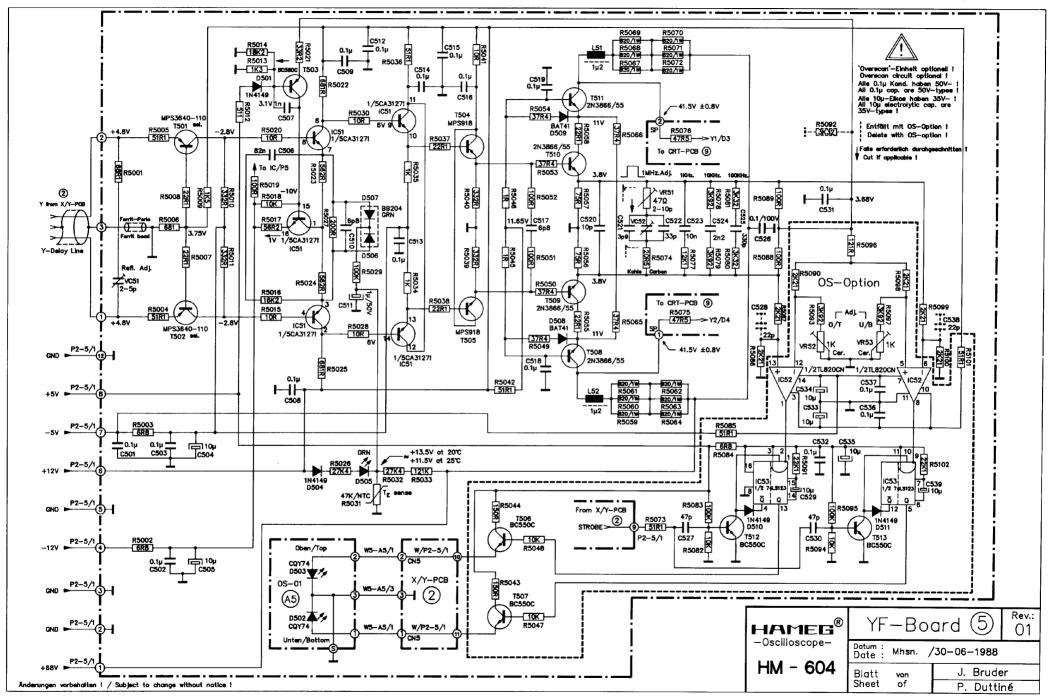
Rear View of Power Switch and Appliance Inlet with Voltage Selector and Fuse

Adjustments

As advised in the Operating, Test and Service Instructions, small corrections and adjustments are easily carried out with the aid of the Circuit Diagrams and *Adjusting P/an*. However, a complete recalibration of the scope should not be attempted by an inexperienced operator, but only someone with sufficient expertise. Several precision measuring instruments with cables and adapters are required, and only then should the pots and trimmers be readjusted, provided that the result of each adjustment can be exactly determined. Thus for each operating mode and switch position, a signal with the appropriate sine or square waveform, frequency, amplitude, risetime and duty cycle is required.



D1 - 10.88 - 604



•	Bezeichnung der Bauteile		HM 604						Iden t	tificati	on of	electr	ical C	ompo	n
-	Die elektrischen Bautelle sind so gekennzeichnet, daß di Nummer mit der Baugruppen-Nummer übereinstimmt:	e erste	Electrical components on certain parts of the HM604 are ma such that the first numeral is on:	rked	Farbkennzeic					lor-Abbr ellow	eviations	for insu = grau		re grev	
-	Chassis Y-Eingang, ext. Triggereingang, Gerätestecker, Netschalter, Netztransformator usw.	0	Chassis Y-inputs, Trigext. input, Appliance inlet, Power switch, Power transformer	0	bn = braun rd = rot or = orange	/ bro / red	wn	gn = g bl = b	rün / g	reen lue	wh trp gn/ye	= weiß = trans	oarent /	white transpare green-yel	
	EY-Board I+II Eingangsteiler, Y-Vorverstärker (Kanal I + II)	1	EY-Board I+II Attenuator and Preamplifier (Channel I + II)	1				······	1						T
	X/Y-Board Y-Zwischenverstärker, Kanalumschaltungs-Flip-Flop, Dioden-Schalttogik, Chopper-Generator, Triggerverstärker, Niederspannungsversorgung, Testleiste, X-Endverstärker, Component-Tester	2	X / Y-Board Y Intermediate amplifiers, Channel selection flip-flop, Y-Gate driver stages, Chopper generator, Trig. and ext. Trigger amplifier, Trig. gate driver stages, LV-Power, Check point strip, Component tester, X-Final amplifier	2	Anschlußfolge der Transistoren Terminals of Transistors	BC550C BC560C BF414	BF 199 BF 311 BF 959	BF422 BF423	MPS918 MPS3640	BF 256 B-1	BSX 19 (TO 18) 2N3866 (TO 39) 2N2369A (TO 18)	∪440	BF 458 BF 472 BD 433 BD 434 BD 675 BD 676	BD 651	T
	Teseard Triggerschaltung, Zeitbasis (passiv), Ablenkverzögerung, Hold-off Schaltung, Heiltastung, TV-SyncSep., Y-Out Verstärk	3	TB-Board Trigger circuit, Timebase circuit (passive), Unblanking circuit, Delay circuit, Hold-off circuit, TV sync. separator, Y-Out amplifier	3									BD 677 BD 681		
2	Calibrator Board	4	Calibrator Board	4							E	G1 S2	Ø		
	YF-Board Y-Endverstärker, Overscan (Bereichsüberschreitung; Option)	5	YF-Board Y-Final amplifier, overscan circuit (optional)	5	Ansicht von unten			B C C			(°•)B				C A Ret
	Z-Board Beschaltung der Kasthodenstrahlröhre, Rücklaufaustastung, Hochspannungs-Netzteil	6	Z-Board CRT-circuit, Unblanking, HV-supply	6	Bottom view						°	sīL ₆₂	BCE	E C B	L
	CO-Board Potentiometer für horizontale Strahllage, Power LED, DEL u. TRIGLED, Potentiometer für Hold-off	7	CO-Board X-pos. pot., Power LED, DEL,- and TRIGLED, Hold-off pot.	7	Ansicht von oben	o (j) o E B C	O D B M C	O D O BCE	(lo Can		E B (°	52 G1 02 0 0 000			(
	TR-Board Potentiometer für Strahldrehung, Spule für Strahldrehung, Raster-Beleuchtungsschalter (Option)	8	TR-Board Trace rotation pot., Trace rotation coil, illumination switch (optional)	8	Top view	:	ě		Ĩ	Ğ	C C	62 ¹ S1	ECB	BCE	
	CRT-Board CRT-Fassung	9	CRT-Board CRT socket	9	· · · · · · · · · · · · · · · · · · ·	1		L	J	L	I				
	TBG-Board Zeitbasis-Generator (aktiv), Zeitbasis-Trennverstärker	A3	TBG-Board Timebase generator (active), Timebase buffer amplifier	A3			Bei	spiel Kab	elverbin	dung: P2	- 3/1- 5	bzw. W2	-3/1-5		
	PTFS-Board "After-Delay"-Trigger, Halbbild-Auswerter	B3	PTFS-Board "After-Delay" trigger (post trigger), Field selector	B3	TB-Board P2-3/1	3	P W		:hkabelste :hkabelve		Board) : eine Seite	e verlötet,	andere S	eite Buch	ıse
	OS-Board (Option) "Overscan"-Anzeigen (LED's)	A5	OS-Board (Option) "Overscan"-indicators (LED's)	A 5	ہلینینی ا	ւ	2-3 1				Board 2 ur ng zwische				
	IL-Board (Option) Rasterbeleuchtung	A8	IL-Board (Option) Graticule illumination	A8			5		ht-Numm		•				
	Abkürzungen / Abbreviations					Steck	er								
	Al Gerätestecker / Appliance inlet					Plug	Exa	mple, ca	ble conn	ection: P	2-3/1-5	or W2-3	8/1- 6 re	spective	viv
	C Kondensator / Capacitor				\Box								0	•	•
	ChP Testpunkt / Check point CN Steckverbinder / Connector		13 · +68V (XY-Board)				P		•	-	ed on boar				
	D Brückengleichrichter / Bridge rectifier				(- - -	Buchs	_{ie} W				t soldered			cket (mov	/ab
	D Diode / Diode		11 · NC		ι	2 Socke	et 2-3	= Con	nection b	etween E	Board 2 an	d Board 3			
	D Leuchtdiode / Light emitting diode		10 · NC		1 6		1	= Firs	t flat cable	e connect	ion betwe	en Board	2 and 3		
	F Sicherung / Fuse		9 · +140V				(5)	= Seri	al numbe	r of the w	vire (in the	flat cable)		
	IC Integr. Schaltung / Integrated circuit				· · · · · · · · · · · · · · · · · · ·	L	~								
	L Spule, Drossel / Inductor, Coil		8 - NC 7 - GND												
D	P Stecker / Plug P Lötöse / Eyelet				L I										
ĩ	P Lötöse / Eyelet R Widerstand / Resistor					Draht									
- 0	S Schalter / Switch		5 + 12V			Wire									
	T Transistor / Transistor		4 • NC	-			Wid	erstand-	/Resisto	r identifi	cation				
88	TR Transformator / Transformer		3 · +5V				1		Midarata	nd / Donia	HOT DELA	1 1 1 2 \ 0/	oarbon fil	~	
1	VC Trimmkondensator / Variable capacitor		2 · -5V						vvidersta	nu / Resis	stor; 0.6W	1, 1(2)%,	carpon fil	111	

X/Y-Board 2..

Al	Geralestecker		Appliance mici
C	Kondensator	1	Capacitor
ChP	Testpunkt	1	Check point
CN	Steckverbinder	1	Connector
D	Brückengleichrichter	1	Bridge rectifier
D	Diode	1	Diode
D	Leuchtdiode	1	Light emitting diode
F	Sicherung	1	Fuse
IC	Integr. Schaltung	1	Integrated circuit
L	Spule, Drossel	1	Inductor, Coil
Ρ	Stecker	1	Plug
P	Lötöse	1	Eyelet
R	Widerstand	1	Resistor
S	Schalter	1	Switch
Τ	Transistor	1	Transistor
TR	Transformator	1	Fransformer
VC	Trimmkondensator	1	Variable capacitor
W	Draht	1	Wire
Z	Zenerdiode	1	Z-Diode

2 . -5V -12V

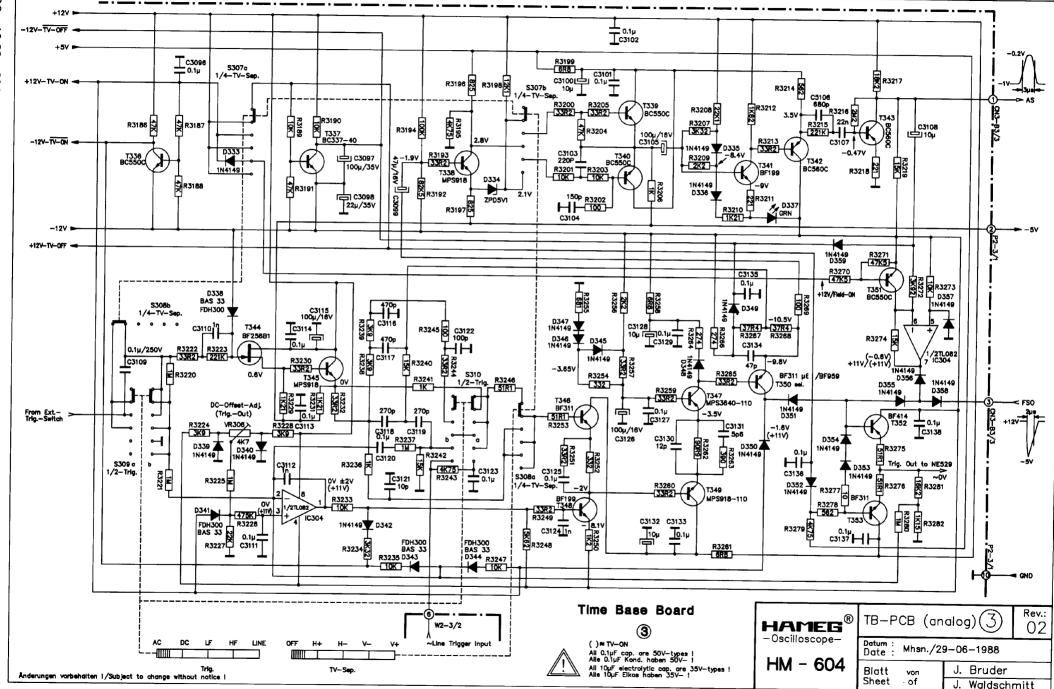
• 1

Identification of electrical Components

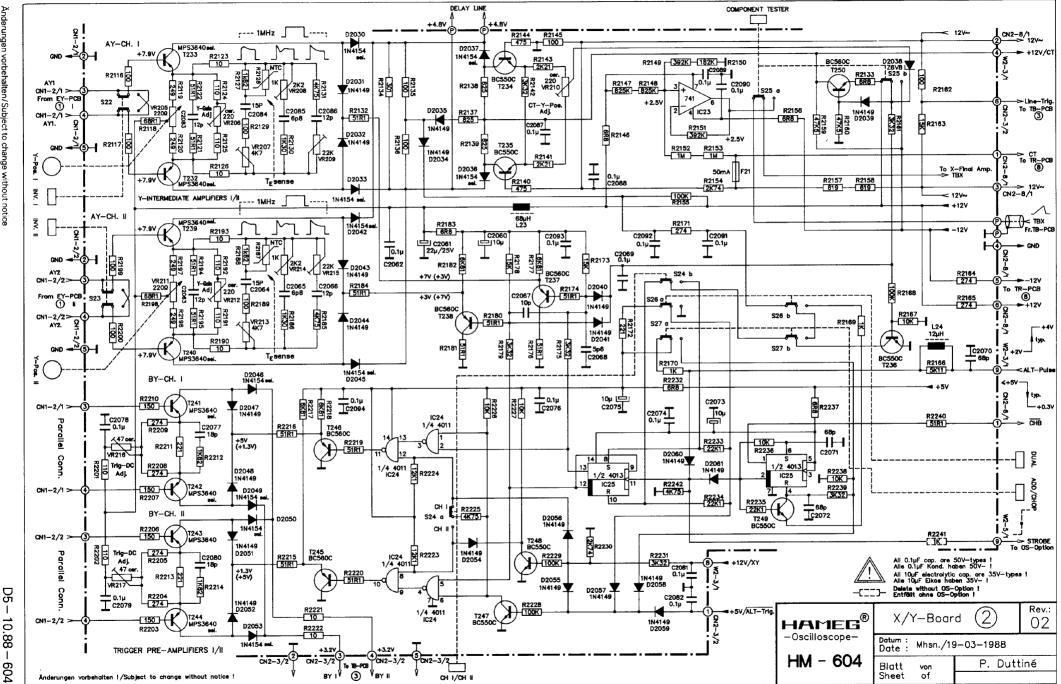
on = braun d = rot or = orange	/ brov / red e / orar	1	bl = b	rün /g lau /b iolett /v		wh trp gn/ye			white transpare green-ye	ent Ilow stripe	
Anschlußfolge der Transistoren Terminals of Transistors	BC550C BC560C BF414	BF 199 BF 311 BF 959	BF 422 BF 423	MPS918 MPS3640	BF 256 B-1	BSX 19 (TO 18) 2N 3866 (TO 39) 2N 2369 A (TO 18)	U 440	BF 458 BF 472 BD 433 BD 434 BD 675 BD 676 BD 677 BD 681	BD 651	TL431 C	
Ansicht von unten Bottom view	Common Co	olo Buc	BCW	C Bu		E O C		BCE	E C B	C A Ref	
Ansicht von oben Top view	() Uner	o [] o Buto	O D O	(lo) CBr	D S G	B C C	S2 G1 D2(○○○)D1 G2□S1	E C B	BCE	C A A Ref	
		Beis	apiel Kab	elverbing	dung: P2	-3/1-5	bzw. W2	-3/1-5			
TB-Board P2-3/1	3 Stecke		 W = Flachkabelverbindung: eine Seite verlötet, andere Seite Buchsenleiste 2-3 = Verbindung zwischen Board 2 und Board 3 = 1. Flachkabelverbindung zwischen Board 2 und 3 5 = Draht-Nummer des Flachkabels 								
	Buchs Socke	P e W	W= Flat cable wiring (direct soldered on board) with socket (movable)2-3= Connection between Board 2 and Board 31= First flat cable connection between Board 2 and 3								
1 6 W2-3/1	Draht Wire	Wid —(Widersta		tor; 0.6V	V, 1(2)%,		lm		

Widerstand / Resistor: 0.25W, 0.5%, T_c=50·10⁻⁶/K, metal film

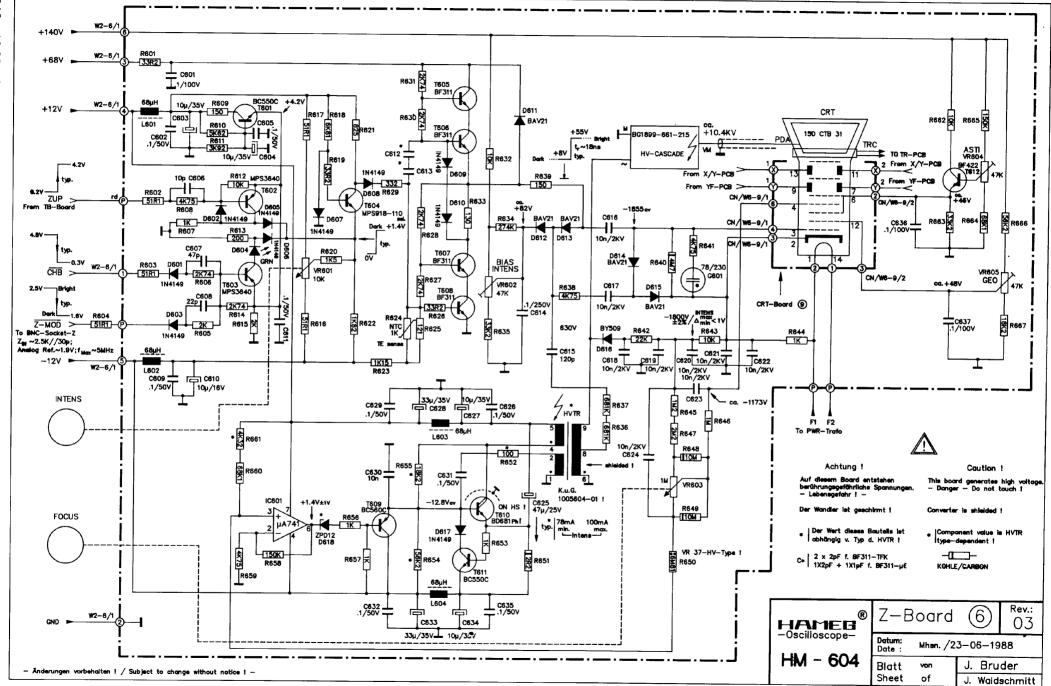
D3-10.88-604 Z...

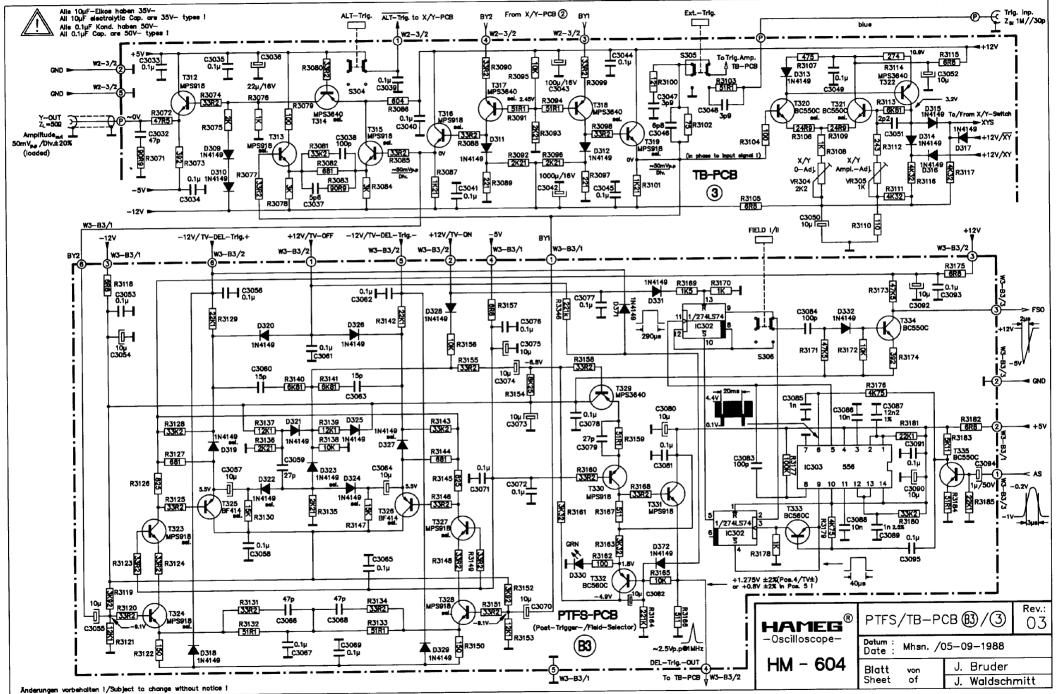


Anderungen vorbehalten/Subject to change without notice





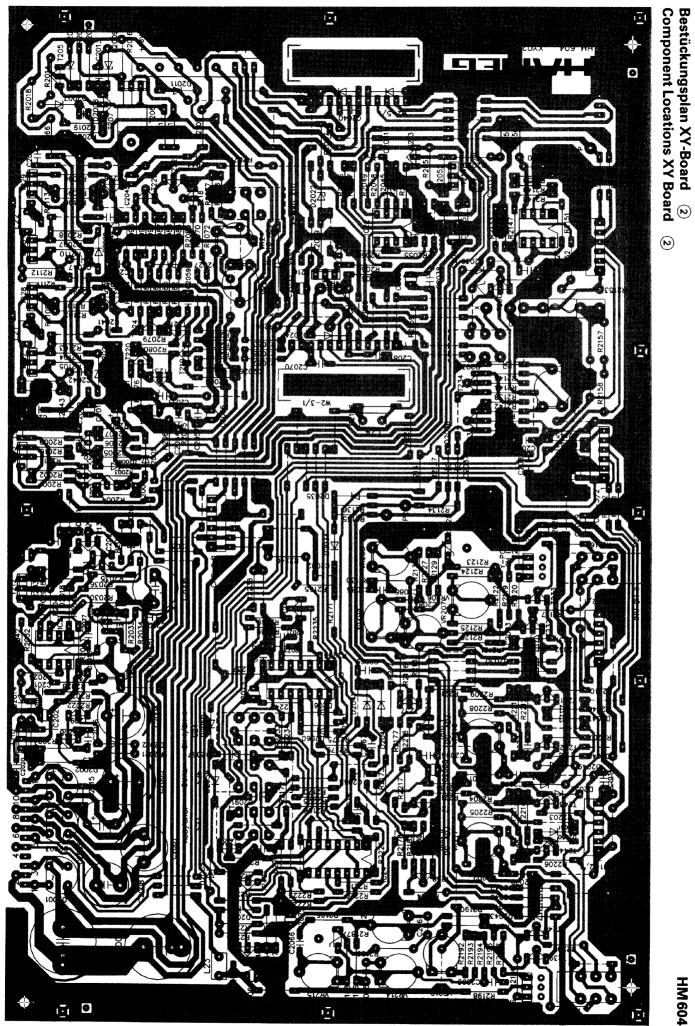




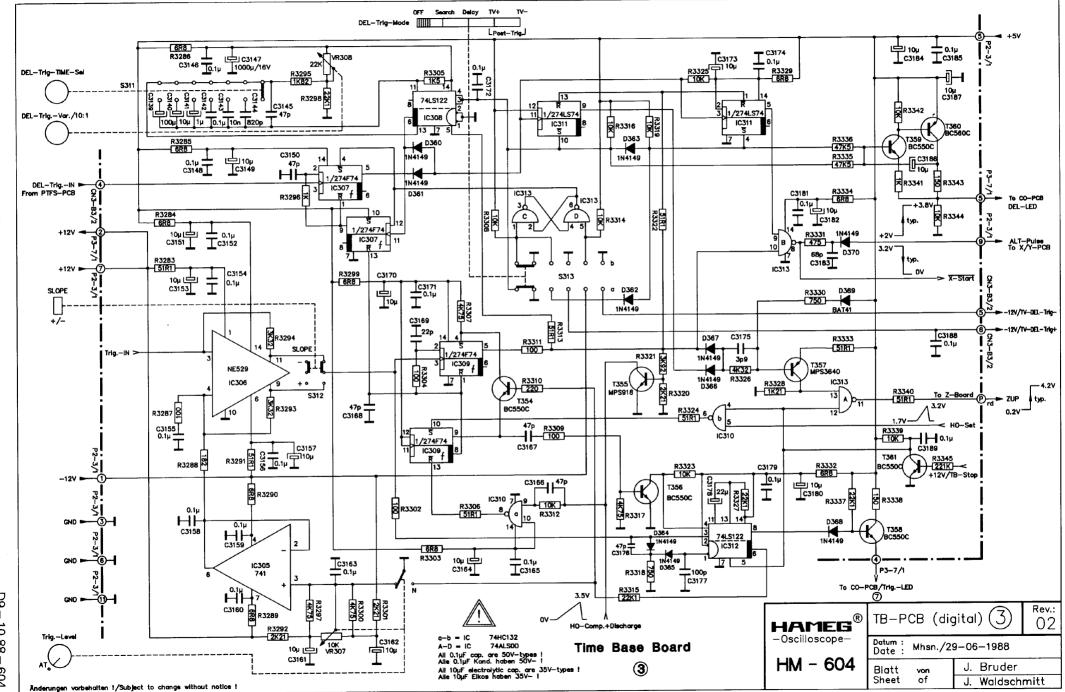
notice

D7-10.88-604



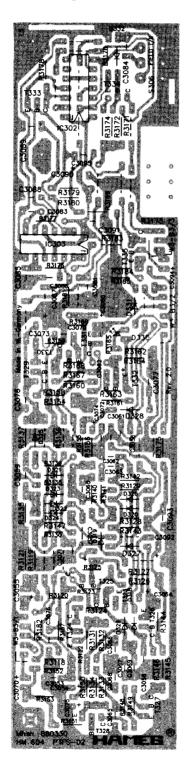


HM 604

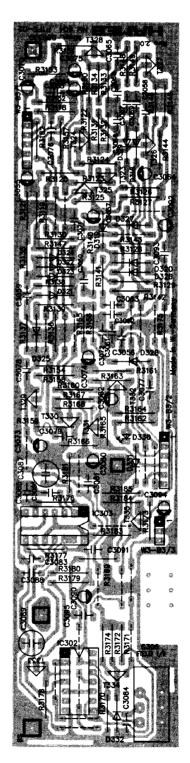


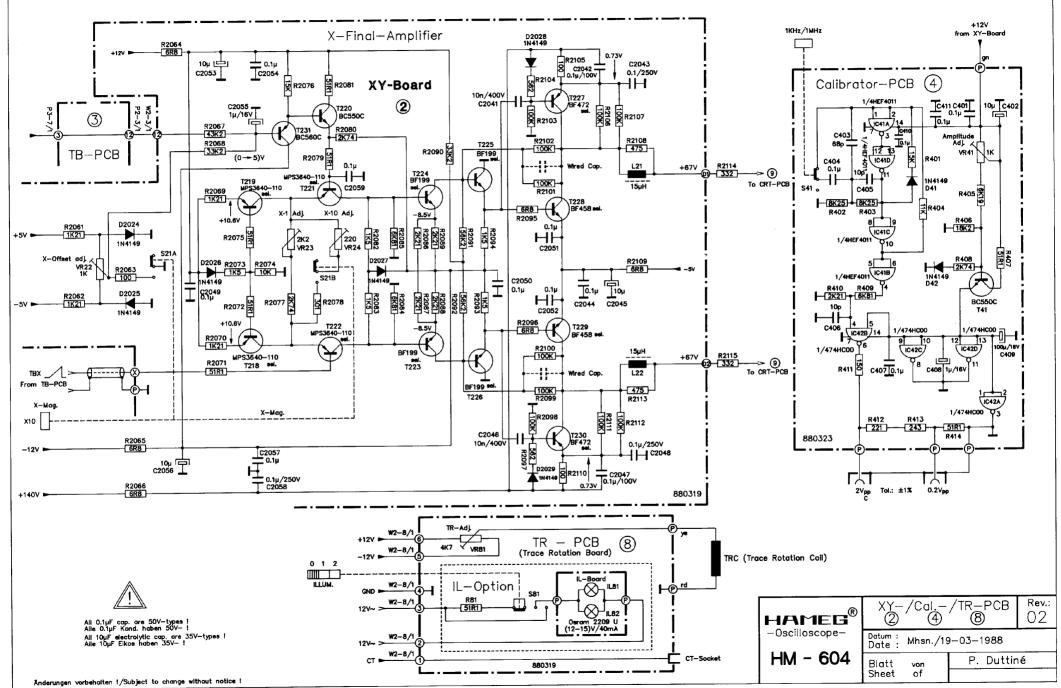
Bestückungsplan PTFS-Board (B) Component Locations PTFS Board (B)

Lötseite Soldering side



Bestückungsseite Component side

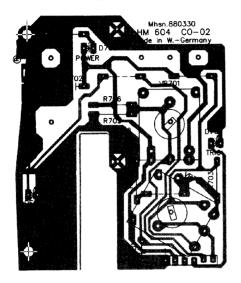




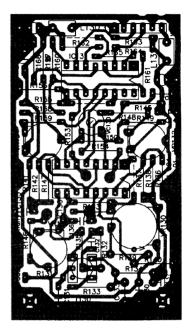
D11 - 10.88 - 604

Bestückungsplan Component Locations

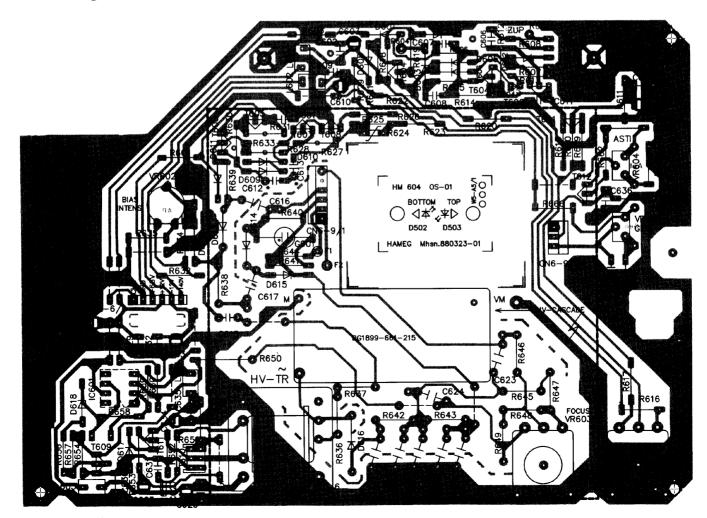
CO-Board (7)



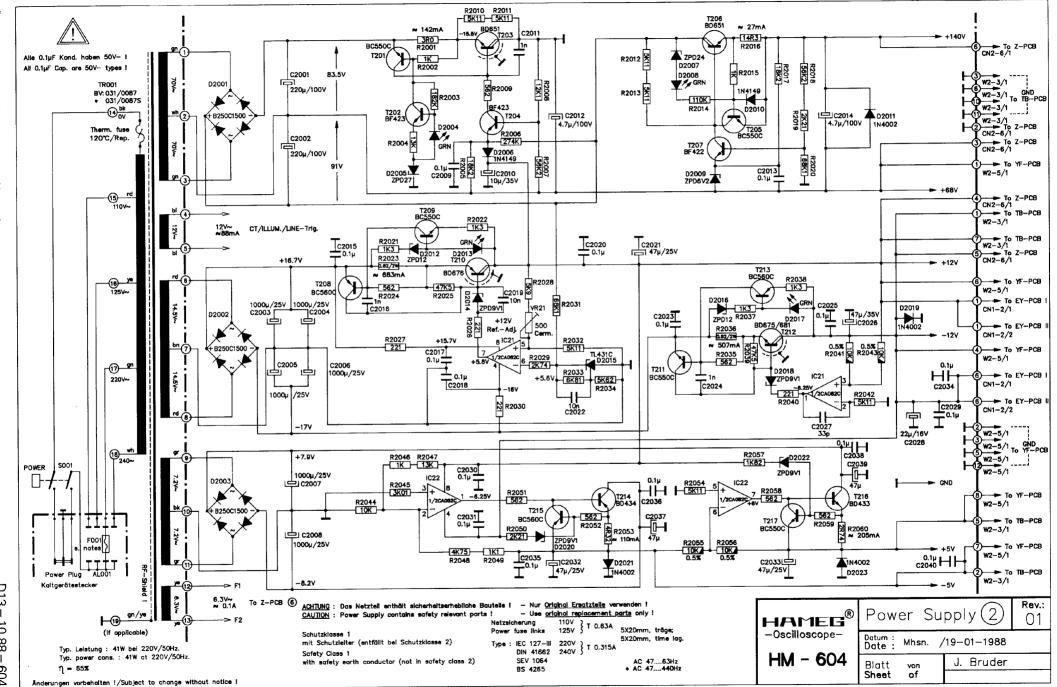
EYBoard ①



Z-Board (7)

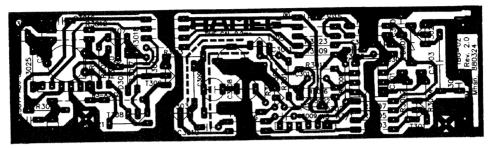


D18-10.88-604

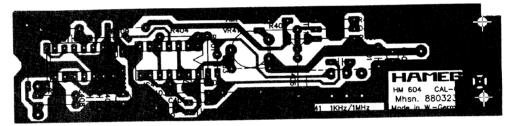


HM 604

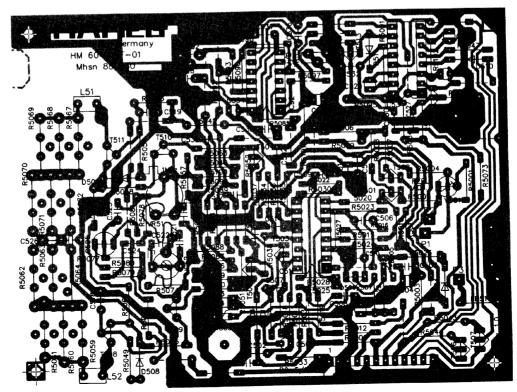
TBG-Board A3



CAL-Board ④



YF-Board (5)



Bestückungsplan TB-Board ③ Component Locations TB Board ③

