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Operational Amplifier Basics

Introduction

It can be mathematically shown that if an amplifier has:

- an infinite high gain,
- an infinite bandwidth,
- an infinite high input impedance, and
- a zero output impedance,

then its characteristics can be totally determined by external components connected in feedback loops. Such an "ideal" operational amplifier (op amp) is shown in Figure 1. Since R_{IN} is infinite, there is no current flow into either input terminal and the differential input voltage is zero. If there is an output voltage E_O present, it can only be the result of an infinitesimal voltage at the input. These are theoretical considerations only.



Figure 1. "Ideal" Operational Amplifier

The op amp is, therefore, an ideal building block for all kinds of analog tasks. In practice, no single characteristic listed above can be fully achieved. The values that can be practically obtained are nevertheless such that the resulting performance is extremely useful.

Op Amp Fundamentals

An op amp is a very high gain dc amplifier and usually has voltage gains in the range of 20,000 to 1,000,000. The detailed schematic symbol of an op amp is shown in Figure 2a with the simplified symbol shown in Figure 2b.

As shown in Figure 2a, the -input of the op amp is called the inverting input, and the + input is called the noninverting input. If an input signal is applied to the - input, with + input grounded, the polarity of the output signal will be opposite to that of the input signal. If an input signal is applied to the + input, with the - input grounded, the polarity of the output signal will be the same as that of the input signal. For an ac signal, this means that the output of the op amp will be 180 degrees out of phase with a signal applied to the - input, but in phase with a signal applied to the + input.

Ideal Op Amps With Negative Feedback

The most common op amp circuit configuration uses two external components: 1) an input component and 2) a feedback component (see Figure 3). When the feedback component is between the op amp output and the -input, the circuit is said to have negative feedback. When the feed-







Figure 3. Op amp circuit with negative feedback

back component is between the op amp output and the +input, the circuit is said to have positive feedback.

In Figure 3, an op amp is shown with negative feedback. E_{IN} is the inut signal, V_{IN} is the differential input to the op amp, and E_O is the op amp output. The open loop gain is defined as the ratio of E_O to V_{IN} :

Open-Loop Gain =
$$\frac{E_O}{V_{IN}}$$

The closed-loop gain is defined as the ratio of E_O to E_{IN} :

Closed-Loop Gain =
$$\frac{E_O}{E_{IN}}$$

The open-loop gain is the gain of the op amp and this gain is independent of the input and feedback components. The closed-loop gain, however, depends only on the values of the input and feedback components when the closed-loop gain of the circuit is much less than the large openloop gain of the op amp.

Input Current and Feedback Current

When an input signal (E_{IN}) is applied to the circuit of Figure 3, a current (I_{IN}) flows through the input component and a voltage (V_{IN}) develops across the input terminals of the op amp. The very high gain op amp amplifies the differential input voltage (V_{IN}) producing an output voltage (E_O) with a polarity opposite to that of V_{IN} . This output is fed back through the feedback component and opposes the input voltage that produced it.

Because the negative feedback signal opposes the input signal, V_{IN} is very small. Therefore, the higher the gain of the op amp, the smaller is V_{IN} . In fact, for some calculations, V_{IN} can be assumed equal to zero and the inverting input at virtually the same potential as the noninverting input.

The relationship between the input current (I_{IN}) and the feedback current (I_{FB}) is most important. Assuming that V_{IN} is equal to zero^{*}, it follows from Ohm's law that no current can flow into the op amp. By applying Kirchhoff's current law (the algebraic

sum of the currents toward any point in a network is zero) to the invertinginput terminal of the op amp, it can be seen that the input current must be equal and opposite to the feedback current. Extensive use will be made of this important result:

$$I_{FB} = -I_{IN}$$

The Equivalent Circuit

The schematic diagram of the op amp with negative feedback may be simplified, using the previous results, to the equivalent circuit shown in Figure 4. The equivalent circuit is obtained by recalling that I_{IN} is equal and opposite to I_{FB} , and so Figure 3 can be relabeled as shown in Figure 4.

Since V_{IN} is nearly zero, the inverting input of the op amp can be considered to be at ground potential. This simplifies the diagram of Figure 4 to the important equivalent circuit shown in Figure 5.

The equivalent circuit shown in Figure 5 shows why op amps with negative feedback are so useful. The input circuit is electrically isolated from the output circuit, yet the current flowing through the input com-

*If V_{IN} were actually equal to zero, the output of the op amp would also be zero and the op amp would be useless. In reality, V_{IN} is a very small voltage (usually less than a millivolt). But, for the purpose of calculating input and feedback currents, very little error is introduced by approximating V_{IN} as equal to zero.



Figure 4. Negative feedback op amp circuit with equivalent feedback current shown



Figure 5. Equivalent circuit of Figure 4

ponent dictates what current must flow through the feedback component. By choosing different input and feedback components, different circuit functions can be performed. The circuit functions listed in Table 1 will be discussed later in this article.



Figure 6. Simple voltage divider

Figures 6 and 7 illustrate a very simple concept in conjunction with an op amp. The potentiometer P is used as a variable voltage divider and the voltage delivered to the load is in a linear ratio with the shaft rotation. This will only be the case if the load is an infinite resistance, or is very high in comparison to the resistance of P. As an example, if P and the load are each 10 kilohms and P is set at 50 percent, the division ratio will not be 1/2 but 2/5, as shown in Figure 7.

If an op amp is connected between the movable contact of P and the load as shown in Figure 8, the op amp will provide perfect isolation. The op amp must have a closed loop gain of one. This is possible with a 100 percent

Table 1. Table of Circuit Functions

Circuit	Input	Feedback	
Function	Components	Component	
Amplification	Resistor	Resistor	
Integration	Resistor	Capacitor	
Differentiation	Capacitor	Resistor	



Figure 7. Simple voltage divider set at 50 percent

feedback. An op amp connected in this way is often referred to as a voltage follower or an impedance converter.

An additional characteristic that is important for an op amp is stability. If the input is zero, the output will also be zero. The op amp must be free of drift or offset voltage.

The Noninverting Op Amp

In the configuration shown in Figure 9, E_{IN} and E_O are in phase. The closed loop gain is controlled by the ratio of R_1 and R_2 . If $R_2 = 0$, then the amplifier is simply a voltage follower and R_1 becomes meaningless. However, in practical applications, it will be made

equal to the E_{IN} source impedance for balancing purposes of the differential input stage.

The Inverting Op Amp

 E_{IN} and E_O are inverted as is indicated by the negative sign for the gain (see Figure 10). R_3 may be included instead of a ground connection of the noninverting input only because of a less than ideal amplifier (less than infinite input impedance). R_3 should be equal to R_1 , again for balancing purposes. If the characteristics are close enough to ideal, then input B is at ground level. By definition, input A must also be at ground level since there shall be no potential difference between inputs A and B. Point A is called a virtual ground.



Figure 8. Simple voltage divider with op amp providing isolation



Figure 9. Simple noninverting op amp







The summing op amp shown in Figure 11 is simply an inverting op amp with multiple inputs. Since input A is a virtual ground, there can be no current flowing over A from one input into another input. The total input current is $I_1 = I_2$. The number of inputs is only limited by practical considerations.

The Differential (Subtracting) Op Amp

If the same signal is applied to both the + input and the – input of the op amp shown in Figure 12, the two amplified output signals will be 180 degrees out of phase and will completely cancel each other. Since the op amp responds only to differences between its two inputs, it is said to be a differential amplifier. The voltage difference between the +input and the - input is called the differential input voltage. Since a differential amplifier amplifies only the differential input voltage and is unaffected by signals common to both inputs, it is said to have common-mode rejection. Common-mode rejection can be very useful, for example, when measuring





small signals in the presence of 60 Hertz noise. The 60 Hertz noise common to both inputs is rejected and the op amp amplifies only the small signal difference between the two inputs.

Note that an ideal differential op amp produces neither a differential-mode nor a common-mode output in response to a common-mode interference input.

The Integrating Op Amp

The simple integrating op amp shown in Figure 13 produces an output signal proportional to the integral of the input signal and time variable of the resistance and capacitance charging rate. The capacitor connected across the input and output of the inverting op amp improves the performance by what is called the Miller Effect. The Miller Effect says that a capacitor connected as shown appears to the input as being multiplied by the gain of the amplifier.

If R = 1 megohm and C = 1 microfarad, according to the formula shown in the figure, E_O will increase at a rate of 1 volt per second, positive or negative, depending on the polar-



Figure 12. Simple differential operational amplifier

ity of E_{IN} . If E_{IN} is a symmetrical square wave, E_O will be a triangle wave. For $E_{IN} = \sin X$ the output is $-\cos X$. The output leads the input by 90°.

The Differentiator Op Amp

As shown in Figure 14, E_O is proportional to the rate of change of the input voltage. A dc input will make $E_O = 0$ because of the blocking capacitor C. A square wave at the input will produce a spike at the output triggered by the leading and trailing edges of the input signal. Because the top of the square wave is a constant value, the differential factor is zero and the integrating capacitor discharges towards zero.

The Logarithmic Converter Op Amp

If a nonlinear element such as a transistor is connected into the feedback path as shown in Figure 15, E_{IN} versus E_O follows a nonlinear function. The base-emitter junction of Q_1 represents the logarithmic element in the feedback loop.

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Figure 13. Simple integrating operational amplifier

The Antilog Op Amp

By placing the nonlinear element in the input path as shown in Figure 16, the circuit becomes an antilog converter. For a logarithmic input E_{IN} , the output E_O will be a linear signal.

Multiplication (Division) With Log-Antilog Operational Amplifiers

Since log A + log B = log AB, and log A - log B = log A/B, a circuit using two log converters and a summing (subtracting) amplifier, followed by an antilog converter, will output the product or the division of the two input signals. Figure 17 illustrates the circuit for multiplication.



Figure 15. Simple logarithmic converter operational amplifier



Figure 14. Simple differentiator operational amplifier



Figure 16. Simple antilog op amp



Figure 17. Multiplication circuit using log and antilog operational amplifiers



Figure 18. Typical op amp application as an impedance converter (U_6) , isolator (U_7) , and summing amplifier (U_4)

Typical Op Amp Circuit

Figure 18 illustrates a typical example op amp application. The task is to produce and mix a tuning voltage with a sweep ramp for a YIG oscillator. U_6 operates as an impedance converter for the zener diode CR_1 . R_{100} produces the tuning voltage that is isolated by U_7 . Finally, U_4 operates as a summing amplifier for the tuning and sweep voltage.

If you desire a more detailed and indepth discussion of op amps, I suggest that you read *Understanding IC Operational Amplifiers*, by Roger Melen and Harry Garland, available from Howard W. Sams & Co.

Safety-Related Service Notes

Service notes from HP relating to personal safety and possible equipment damage are of vital importance to our customers. To make you more aware of these important notes, they are printed on paper with a red border, and the service note number has an "-S" suffix. In order to make you immediately aware of any potential safety problems, we are highlighting safety-related service notes here with a brief description of each problem. Also, in order to draw your attention to safety-related service notes in the service note index, each appropriate safety-related service note is highlighted with a contrasting color.

HP E2500 Frequency Agile Signal Simulator

Safety Service Note E2500A-01-S describes how to safety check the routine dielectric withstand test (HIPOT). This service note applies to instruments within the serial number range of 2845A00101/2935A00114. If you own one of these HP E2500As, you may return it to your nearest HP Customer Service Center for the HIPOT test, which will be conducted free of charge.

HP 5347A & 5348A Counter/ Power Meters

Safety Service Notes 5347A-02-S and 5348A-02-S describe an incorrectly manufactured power transformer. The serial numbers affected are 2924A00101-158 and 160-170 for the HP 5347 and 2924A00101-120 for the HP 5348. The manufacturing error concerns the transformer primary wires being improperly connected inside the transformer. The result is that when*operating on the 220 volt line setting, the line fuse blows. Both the line module and power transformer must be replaced at an HP Customer Service Center. Do *not* attempt to rewire the transformer as this will invalidate the color coding used on the wires.

Safety Service Notes 5347A-03A-S and 5348A-03A-S concern the removal of unsafe carrying straps. This service note applies to all units. The clips on the ends of the carrying straps, which are part of Option 060 are defective and may fail in the normal use of the instrument. Damage to the instrument and/or injury to personnel may occur. The defective straps are identifiable as being *non*-adjustable in length. Also, there is concern over the strength of the front panel casting at the carrying strap holes. *Mis*-application of a customer furnished strap may apply sufficient leverage by weight of the instrument to cause breakage. *No* strap or other carrying appliance should be connected at these points.

Please remove the non-adjustable strap from the unit and return it to the following address. A new adjustable strap will be returned to you, in addition to a pair of plastic plugs to block the front panel carrying strap attachment holes.

Joe Dore 52U/19 Hewlett-Packard Co. 5301 Stevens Creek Blvd. Santa Clara, CA 95052-8059

The solution to making the instrument portable is the design of a soft carrying case. This case has been designed to enhance portability with tough cordura nylon double-stitched construction that is padded on all six sides. The permanently attached adjustable strap provides a much safer means of carrying the product and allows easy measurements with the unit suspended from the shoulder.

If you originally purchased the product with the Option 060 portability option and have returned the strap as previously requested, you will receive this new soft carrying case at no charge.

If you did not order Option 060 and wish to buy the soft carrying case, order HP P/N 05348-60214. The price is \$295 U.S. list.

For those customers ordering the instrument and specifying the portability option, please be advised that Option 060 has been discontinued. The portability option is now the soft carrying case and is listed as Option 070, \$295 U.S. list. HP 8340A/B Synthesized Sweepers HP 8341A/B Synthesized Sweepers HP 8719A Network Analyzers HP 8720A/B Network Analyzers

The following Safety Service Notes describe possible injury from falling instruments due to damaged front panel handles.

8720A-05-S, 8720B-01-S, 8719A-01-S, 8340A-30-S, 8340B-30-S, 8341A-30-S, 8341B-30-S

If damaged handles are used to lift or support the instrument, the handles will break causing the instrument to fall, possibly resulting in personal injury. Exercise caution when using the front handles to lift the instrument. The damaged handles will be replaced at an HP Customer Service Center, or you may order the front handle replacement kit free of charge (HP P/N 5062-3990) and perform the replacement following the instructions in the product's service manual.

HP 8657B Synthesized Signal Generator

Safety Service Note 8657B-01-S describes a possible shock hazard that may exist if the instrument's toroidal mains transformer's insulation has been damaged by the A14 assembly's heatsink. The units affected are within the following serial number ranges:

2935U00470/2935U00543 2948U00544/2948U00576 2949U00577/2949U00585

The HP 8657B may present a shock hazard only if the following conditions exists:

- The mains transformer has not been manufactured by "NUVOTEM" and
- The A14 heatsink has damaged the

mains transformer's insulation.

If your instrument is one of the above, please return the unit to an HP Customer Service Center for repair.

HP 8757C/E Scalar Network Analyzer

Safety Service Note 8757C/E-02-S describes a possible defective lithium battery contained on the CPU board of the instrument. The battery may leak electrolyte and damage the board. Symptoms include yellowing around the battery or in some cases, corrosion of the board and other boards in close proximity. The Safety Service Note contains a list of applicable serial numbers. Please return your unit to the nearest HP Customer Service Center for repair.

HP 85620A Mass Memory Module

Safety Service Note 85620A-03-S describes a possible defective lithium battery contained on the mass memory module of the instrument. The battery may leak electrolyte and damage the board. Symptoms include yellowing around the battery or in some cases, corrosion of the board and other boards in close proximity. The serial number range of affected instruments is 0000A00000/3003A00604. If you possess one of the affected units, please return it to the nearest HP Customer Service Center for repair.

HP 86792A Agile Upconverter

Safety Service Note 86792A-02-S describes how to safety check the routine dielectric withstand test (HIPOT). This service note applies to instruments within the serial number range of 2814A00105/2814A00108. If you own an HP 86792A within the serial number range listed, you may return it to your nearest HP Customer Service Center for the HIPOT test, which will be conducted free of charge.

The Logistics Data Book 1990

The *Logistics Data Book* is designed to assist Hewlett-Packard's government customers with their logistics needs associated with the support of HP products. Information is drawn from HP records and Department of Defense information. It reflects logistics actions through October, 1989.

The types of HP products covered in this publication include electronic components, medical instruments and systems, computer equipment, calculators, instruments and systems for chemical analysis, and electronic test and measuring systems.

The intent of the *Logistics Data Book* is to include HP model numbers and their various options that are purchased by government customers. Additional information on all HP parts having National Stock Numbers may be found in the microfiche NSN/ HP Cross Reference (Publication No. 5957-4171).

Options are usually identified in HP product literature by a three-digit number following the model number (e.g., HP 1234A Option 003). Other options may be designated as a letter

and two numbers after the model number (e.g., HP 1234A Option H02). Modifications and systems may be identified in HP literature with a single alpha and two digits preceding the model number (e.g., E12-1234A).

For procurement purposes, all options and modifications in this book follow the model numbers and are connected to them by hyphens (e.g., 1234-001 and 1234A-E12). Where space is limited, option numbers may be shortened to less than three characters (e.g., 1234A-01-02-EI2).

Model numbers are arranged in typical computer sequence (i.e., left-justified, alpha-numeric sequence) so that all the "1s" (Model 1, 11, 111, 1111, etc.) are listed before the "2s" (Model 2, 22, 2222, etc.).

The Hewlett-Packard office nearest you will be pleased to work with you on procurement matters. Also note that the most currently-produced HP products are on GSA federal supply schedule multi-award contracts, and that GSA catalogs are available from your HP office. To order your *Logistics Data Book*, specify publication no. 5954-7744.



If you have any comments or suggestions about how we can make the *Logistics Data Book* more useful to you, please let us know by writing to:

Hewlett-Packard Co. U.S. Field Operations John Cloutier 19320 Pruneridge Avenue Cupertino, CA 95014

What is Traceability and How Does It Relate to Calibration Uncertainty?

Jim Bechtold Hewlett-Packard

Traceability is proof that measurements can be attributed to a national standard. The process involves the "tracing" of measurement comparisons made between the instrument used and the national laboratory. Technically, traceability involves the ability to quantify the measurement errors present in the test or measurement process in terms of the national standards. Agencies require us to objectively demonstrate that traceability exists. Usually through a series of reports attesting to the calibration of the instruments used in a test situation, the standards used to calibrate these standards, and so on back to the reports issued by national laboratory standards.

Calibration uncertainty is the ability to quantify the error sources present in the calibration measurement. Knowing that the act of making measurements is experimental, there have to be some guidelines that tell the person doing the measurement that the job is being performed satisfactorily. A rule-of-thumb method called the "accuracy ratio" is used in many laboratories to determine whether the uncertainty is reasonable. The desired ratio of accuracy between a source instrument and one to be calibrated is 10:1. If this cannot be met, it is permissible to reduce the ratio downward to a lower limit of 4:1.

This ratio is calculated by dividing the tolerance of the measurement to be made (e.g., 5%) by the tolerance assigned to the instrument used to make the measurement (e.g., 1%). In this example of accuracy, the ratio would be 5:1.

In some of today's instrumentation, the instrument being calibrated is almost as accurate as the source device, and therefore at a ratio of less than 4:1. In this situation, the method of calibration used is called "enhanced accuracy calibration." To achieve enhanced accuracy it is necessary to utilize a higher echelon standard during the calibration process. This will provide as much as another order of magnitude to the source equipment, which widens the ratio between the two instruments to maintain a good confidence level. We recommend that when you get below 4:1, contact your metrologist for advice.

A more specific method of determining uncertainty involves understanding the types of errors that exist during measurement and calibration, and ways of avoiding the most serious ones.

But before we get into the three types of errors, we should define two terms used in measurement that are quite often confused — *accuracy* and *precision*.

Accuracy

The accuracy of a measurement is an expression of the closeness of its result to the true value. A high accuracy indicates a close approach to the true value.

Precision

The precision of a measurement is a measure of its repeatability. A high precision indicates the ability to repeat measurements within narrow limits.

These two definitions form the basis for everything else that follows relative to measurement and calibration. The ultimate goal is to achieve both accuracy and precision in every measurement and in every calibration.

To eliminate the confusion that normally exists between accuracy and precision, the following example will serve to show the difference.

Example

Assume that two meters are used to measure a perfect 100.00 volt power source. Ten readings are taken with

each meter. Figure 1 shows a range from 97.16 to 97.24 volts, and Figure 2 shows a range from 99.1 to 100.7 volts.

When we compare the ranges of the two meters we find that meter #1 has a spread of 0.08 volts and meter #2 has a spread of 1.6 volts. This defines meter #1 as having 20 times the *precision* than meter #2 has (i.e., 1.6/.08 = 20).

But meter #2 is closer in its readings to 100.00 volts, which indicates that precision by itself does not necessarily produce accuracy. Meter #2 is more *accurate* than meter #1.

Now, as shown in Figure 3, if meter #1 were recalibrated, with adjustment, to agree with the "perfect" calibration source, the spread of its readings would likely range from 99.96 volts to 100.04 volts (a spread of 0.08 volts). While this shows that *precision* and *accuracy* combine to provide greater *confidence* in the use of an instrument, the two are truly independent. However, the best of both are desirable in any measurement situation.

Errors in Measurement

There are three types of errors by which you may be confronted. They are: gross errors, systematic errors, and random errors.

Gross Errors

These errors are strictly under the control of the individual, totally separate from the instrumentation. Example of gross errors are:

- Misreading the instruments
- Making an incorrect adjustment
- Applying instruments improperly
- Computational errors
- Recording interpolated data

As can be seen, these are errors that can be avoided by care and attention.



Figure 1. Reading of Meter #1



Figure 2. Reading of Meter #2



Figure 3. Altered Reading of Meter #1

Systematic Errors

Systematic errors relate to the instrumentation or external influences to the instruments. Systematic errors cause the measured value to be offset by a fixed amount, as contrasted to random errors, which are bipolar in nature.

One example may be loading. All measuring systems are intrusive. That is, no matter what type of device is used to make a measurement, it will have some effect on the system being measured.

Consider the fact that no voltage source has the ideal zero impedance, and no current source has the ideal infinite impedance.

A voltage source may have a source resistance of as little as 0.01 ohm, and the ability to supply current of 25 milliamps at 10 volts output. The internal voltage drop could then be 250 microvolts, which represents 25 ppm of the 10 volt output (0.0025%). In addition, assume that the connecting leads have 10 milliohms of resistance. The voltage drop across the leads would add another 250 microvolts of error, or 25 ppm. These two elements alone constitute an error of 50 ppm, or 0.005%. Note that the obvious way to avoid these errors is to use 4-wire remote sensing whenever the load is less than 1 megohm.

Random Errors

These errors are indicated as a scatter about an average when a multiple number of measurements are taken. This can be the result of variations in the measuring system, or changes in the quantity being measured. Random uncertainties are of a type that lend themselves to statistical analysis, since they can be considered to vary either positively or negatively around the measured value. They are due to totally unknown causes, and are detectable when repeated measurements are made with a seemingly constant set-up and consistent technique by the user.

An example may be a variation in impedance matching in high frequency instruments due to changes in characteristics when the lead connectors are removed and replaced. This may be due to the wear, or contamination, of BNC connectors, which are quite susceptible to this behavior.

When a succession of measurements are made which differ one from another, some preliminary observations can be made regarding random errors:

- The closest value to the true value is the mean value
- It is likely that values close to the mean value will be more numerous than values that are far away from the mean value
- There are likely to be as many readings that lie above the mean value as there are those that lie below the mean value

It is the general nature of a measured variable that there will be a distribution of values (both plus and minus) around the true, or actual, value.

Note that it is never possible to know the true value, since there will always be some small error. \Box



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If you want to order a service note, refer to the list of service notes in the index, find the service note number belonging to the product you are interested in, and note the package number. Use the form on the last page of *Bench Briefs* to order the number that appears in the "service note package" column. You will receive a package of service notes that includes the one you ordered.

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Service notes are still available on microfiche. The part numbers are:

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IO	54503A-01	Replacing cabinet requires a new rear panel	02

2ND QUARTER 1990

SN Type	SN e No.		e Note ickage
	60501A-01	Rec. addition of a 0.5 amp fuse to protect the bias transformer circuit	019
0	60501A-02	Oper. inst. & procedures when using the EMPRO current shunts for calibration	019
MR	60502A-01	Rec. addition of a 0.5 amp fuse to protect the bias transformer circuit	019
0	60502A-02	Oper. inst. & procedures when using the EMPRO current shunts for calibration	019
AR	60504A-01	Rec. addition of a 0.5 amp fuse to protect the bias transformer circuit	019
0	60504A-02	Oper. inst. & procedures when using the EMPRO current shunts for calibration	019
I A	70001A-12A	Mod. available to convert Mainline Switch actuator from pull ON to push ON	023
1A	70001A-12	Mainline switch conversion kit changes on/off positions	021
0	70001A-13A	Replacing A1 Pwr. Sup. Assys. in mainframes with pull ON Mainline switches	023
0	70001A-13	Mainline switch conversion kit changes on/off positions	021
0	70004A-02	Rack mount with slides kit is now available	020
0	70004A-03	Replacement of keypad retainer spring	021
I A	70300A-05	Recommended replacement for the RF converter HP P/N 5086-7737	017
0	70300A-07A	Replacement for obsolete attenuators	021
0	70300A-07	Replacement of obsolete attenuators	017
/IR	70300A-08	Modification to update HP 70300A module verification software revision A.03.02	018
0	70600A-01B	Replacement of A12 YIG driver board assy. or A13W1 MSIB cable assy.	021
0	70600A-02A	Replacement for obsolete attenuators	021
0	70600A-02	Replacement of obsolete attenuators	017
ИR	70600A-05	Recommended modification to the module verification software	020
0	70601A-01B	Replacement of A12 YIG driver board assy. or A12W1 MSIB cable assy.	021
0	70601A-02A	Replacement for obsolete attenuators	021
0	70601A-02	Recommended replacement for obsolete attenuators	018
I R	70601A-05	Recommended modification to the module verification software	020
ИR	70700A-02	HP 70700A performance software modification to Rev.A.00.00	019
ИR	70700A-03	Mod. to correct power-up sequence failure	020
0	70700A-04	1 MHz and 10 MHz low pass filters required for performance tests	023
MR	70810A-01A	Firmware upgrade kit	017
МR	70810A-01B	New firmware upgrade kit corrects firmware problems	020
0	70900A-14D	Firmware history showing compatibility	017
0	70900A-14E	70900A compatability requirements when used with diff. spect. anal. systems	019
0	70900A-21	Verification software mod. to prevent ERROR 18 during enhanced inverse video	017
0	70900A-22	Repl. screws for A6A3 idler buffer assembly	019
МR	70900A-23	HP 70900A module verification software modifications to Rev.A.03.00	019
MR	70900A-24	Mod. to correct $+5.2$ volt power supply	020
MR	70900A-25	Mod. to correct FFS handshake error	020
0]	70900B-01	70900B compatability requirements when used with diff. spect. anal. systems	019
MR	70902A-03A	Module verif. S/W changes eliminate errors when performing module gain adj.	018
0	70902A-04	Recommended replacements for A1, A2 and A3 board assemblies	017
0	70902A-05	Procedure for replacing matched crystals and A1 or A2 board assemblies	017
ΜA	70903A-01	Availability of linear AGC video out upgrade kit	018
MR	70903A-02A	Module verif. S/W changes eliminate errors when performing module gain adj.	018
0	70904A-02	Recommended replacement of the RF converter HP P/N 5086-7737	017
0	70904A-03B	Replacement of A4 Power Supply/Cntl. Assy. or W13 MSIB cable assy.	021
0	70904A-05A	Replacement for obsolete attenuators	021
0	70904A-05	Recommended replacement for obsolete attenuators	018
MR	70904A-08	2nd converter phase lock loop align. adj. seq. is modified in S/W rev. A.03.00	018
ΙΟ	70905A-02B	Replacement of A4 Power Supply/Cntl. Assy. or W13 MSIB cable assy.	021

SN	SN	Abstract Se	rvice Note
Туре	No.		Package
	700054 054		
	70905A-05A	Replacement for obsolete attenuators	021
	70905A-05	Recommended replacement for obsolete attenuators	018
	70905A-08	2nd converter phase lock loop align. adj. seq. is modified in S/W rev. A.03.00	018
	70905B-01B	Replacement of A4 Power Supply/Cntl. Assy. or W13 MSIB cable assy.	021
	70905B-06	2nd converter phase lock loop align. adj. seq. is modified in S/W rev. A.03.00	018
	70906A-02B	Replacement of A4 Power Supply/Cntl. Assy. or W13 MSIB cable assy.	021
	70906A-05A	Replacement for obsolete attenuators	021
	70906A-05	Recommended replacement for obsolete attenuators	018
	70906A-08	2nd converter phase lock loop align. adj. seq. is modified in S/W rev. A.03.00	018
	70906B-01B	Replacement of A4 Power Supply/Cntl. Assy. or W13 MSIB cable assy.	021
	70906B-06	2nd converter phase lock loop align. adj. seq. is modified in S/W rev. A.03.00	018
	70907A-03A	Replacement for obsolete attenuators	021
IO	70907A-03	Recomended replacement for obsolete attenuators	018
IO	70908A-01	Recommended replacement of the RF converter HP P/N 5086-7737	017
IO	70908A-02A	Identification of correct connector used with A1 LO leveling amplifier	020
MR	70908A-08	Module verification software modification	021
MR	85081A-01	Instructions for replacing faulty locking knobs	022
MR	85082A-01	Instructions for replacing faulty locking knobs	022
MR	85620A-01	Firmware update kit for the HP 8562A/B	019
MA	85620A-02	Firmware upgrade kit	017
SA	85620A-03-S	A1BT1 batteries may leak	019
MA	85629A-01C	Modification available to upgrade the ROM	023
MA	85629A-07B	Modification to upgrade an HP 85629A to an HP 85629B	019
IO	85629A-08	Rec. method for perf. LO distrib. amp. adj. with 85629A/B Test & Adj. Modul	
MR	85629B-01A	Mod. to update test & adj. modules to test later ver. of port. SA assys.	020
MR	85629B-01	Updating test and adj. modules to Test Later versions of portable spect. assys.	
MA	85629B-02B	Firmware (ROM) upgrade kit	017
	85629B-04	Rec. method for perf. LO distrib. amp. adj. with 85629A/B Test & Adj. Modul	
	85640A-01	Mod. to prevent the tracking gen. from blowing line fuse in 220V/240V oper.	017
	85640A-03	Recommended replacement for the 230 V line fuse	023
	85685A-21	New ROM repl. kit to correct 3 dB amplitude accuracy error at 2 GHz	019
	86790A-01	How to use the internal DSA capabilities to isolate faults	020
	86792A-01A	Firmware history and upgrade procedures	020
	86792A-02-S	Instructions for performing routine dielectric withstand test (HIPOT)	017
	86792A-03	Method of using internal firmware and diag. routines to adj. det. offset & gain	017
	86792A-04	How to adjust 268 MHz step switch assembly for best 4/2 spur response	021

Service Note Types

ΙΟ

Information Only Modification Recommended MR

PR **Priority Safety** MA Modification Available

SA Safety

Extend the Life of Your CRT

Cathode ray tubes are expensive components of your HP instrument; therefore, it is in your best interest to extend the life of your CRT as long as possible.

New Product Turn-On

Before you turn on your product for the first time, turn the intensity to the off position. Allow the CRT to warm up and the product's power supplies to stabilize for at least 30 minutes before turning the intensity up to a comfortable viewing level.

Extending Life

Most CRTs have an expected lifetime of 10,000 hours. At 2,000 hours per year, five years of CRT life could be expected (8 hours/day \times 5 days/week \times 50 weeks). If you leave the CRT on for two shifts per day, 2.5 years could be expected. To maximize the CRT lifetime, HP recommends turning the CRT filament off when not in use.

Service Note Order Form

If you want to order a service note, refer to the list of service notes in the index and find the service note number belonging to the product you are interested in. Using the form on this page, order the number that appears in the "service note package" column. You will receive a package of service notes that includes the one you ordered.

Hewlett-Packard Worldwide Customer Support Operation 100 Mayfield Ave. Mtn. View, California 94043 Attn: Bench Briefs

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